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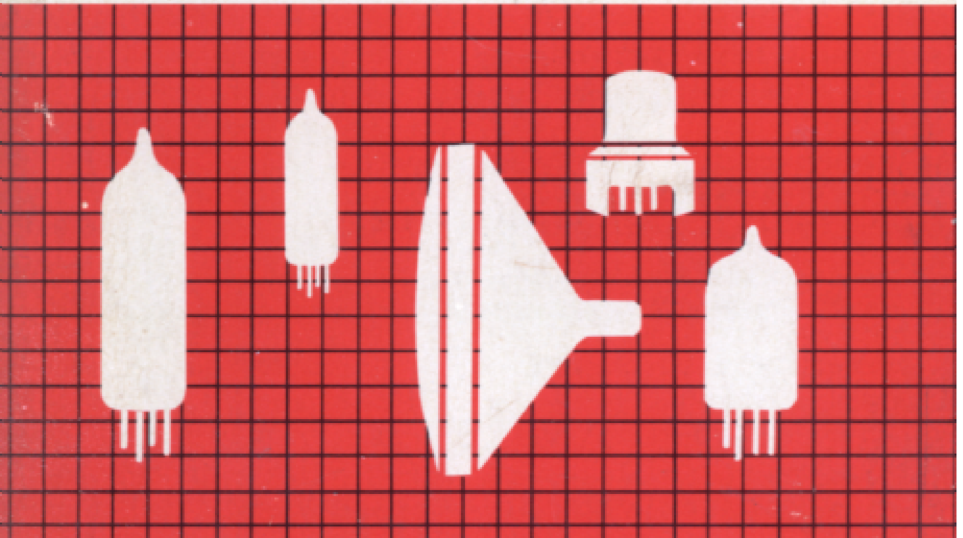
RCA Receiving Tube Manual



RCA

Contents

RCA Receiving Tube Manual

THIS MANUAL, like its preceding editions, has been prepared to assist those who work or experiment with home-entertainment-type electron tubes and circuits. It will be found valuable by engineers, service technicians, educators, experimenters, radio amateurs, hobbyists, students, and many others technically interested in electron tubes.

The material in this edition has been augmented and revised to include the recent technological advances in the electronics field. For more convenient referencing of the latest tube types, the **Technical Data** Section has been restricted to coverage of active RCA types; basic data for replacement and discontinued RCA tubes are given in the **RCA Types for Replacement Use** table.

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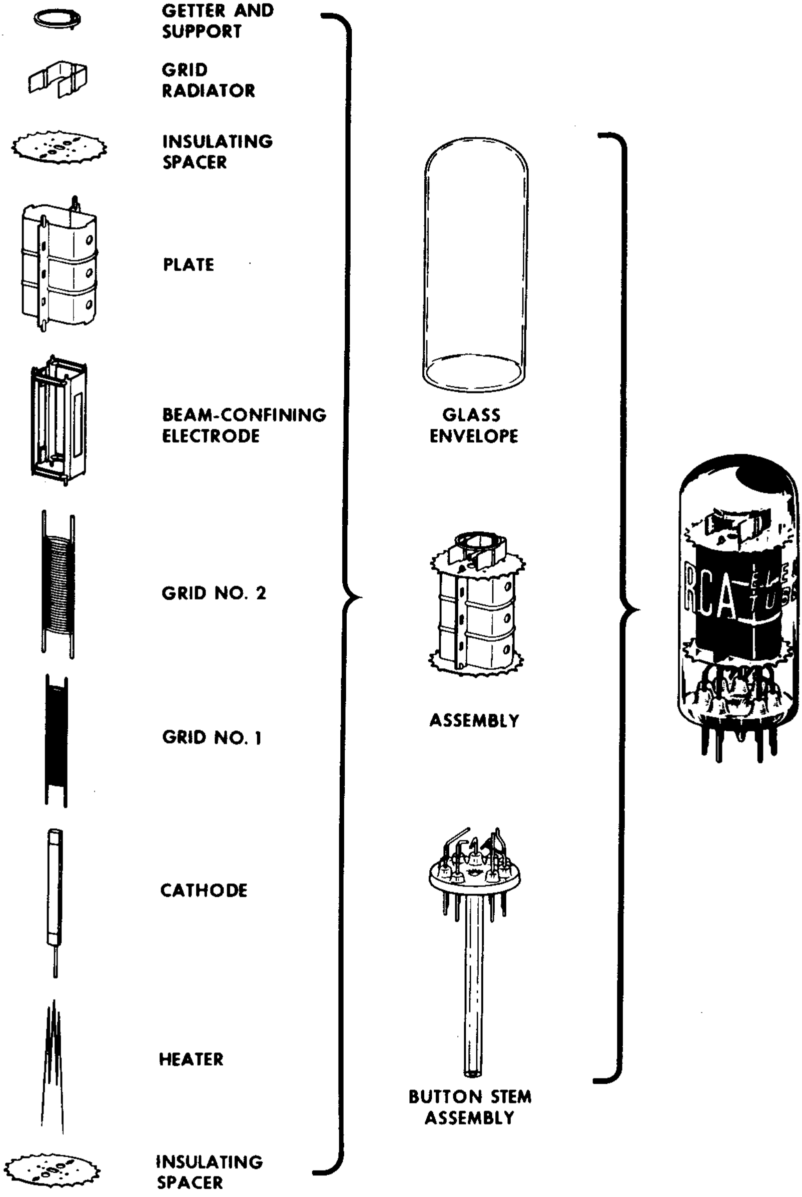

GETTER AND SUPPORT
GRID RADIATOR
INSULATING SPACER
PLATE
BEAM-CONFINING ELECTRODE
GRID NO. 2
GRID NO. 1
CATHODE
HEATER
INSULATING SPACER
GLASS ENVELOPE
ASSEMBLY
BUTTON STEM ASSEMBLY
RCA

Electrons, Electrodes and Electron Tubes

THE electron tube is a marvelous device. It makes possible the performing of operations, amazing in conception, with a precision and a certainty that are astounding. It is an exceedingly sensitive and accurate instrument—the product of coordinated efforts of engineers and craftsmen. Its construction requires materials from every corner of the earth. Its use is world-wide. Its future possibilities, even in the light of present-day accomplishments, are but dimly foreseen, for each development opens new fields of design and application.

The importance of the electron tube lies in its ability to control almost instantly the flight of the millions of electrons supplied by the cathode. It accomplishes this control with a minimum of energy. Because it is almost instantaneous in its action, the electron tube can operate efficiently and accurately at electrical frequencies much higher than those attainable with rotating machines.

Electrons

All matter exists in the solid, liquid, or gaseous state. These three forms consist entirely of minute divisions known as molecules, which, in turn, are composed of atoms. Atoms have a nucleus which is a positive charge of electricity, around which revolve tiny charges of negative electricity known as **electrons.** Scientists have estimated that electrons weigh only 1/30-billion, billion, billion, billionths of an ounce, and that they may travel at speeds of thousands of miles per second.

Electron movement may be accelerated by the addition of energy. Heat is one form of energy which can be conveniently used to speed up the electron. For example, if the temperature of a metal is gradually raised, the electrons in the metal gain velocity. When the metal becomes hot enough, some electrons may acquire sufficient speed to break away from the surface of the metal. This action, which is accelerated when the metal is heated in a vacuum, is utilized in most electron tubes to produce the necessary electron supply.

An electron tube consists of a cathode, which supplies electrons, and one or more additional electrodes, which control and collect these electrons, mounted in an evacuated envelope. The envelope may be made of glass, metal, ceramic, or a combination of these materials.

Cathodes

A cathode is an essential part of an electron tube because it supplies the electrons necessary for tube operation. When energy in some form is applied to the cathode, electrons are released. Heat is the form of energy generally used. The method of heating the cathode may be used to distinguish between the different forms of cathodes. For example, a directly heated cathode, or filament-cathode, is a wire heated by the passage of an electric current. An indirectly heated cathode, or heater-cathode, consists of a filament, or heater, enclosed in a metal sleeve. The sleeve carries the electron-emitting material on its outside surface and is heated by radiation and conduction from the heater.

A **filament, or directly heated cathode,** such as that shown in Fig. 1 may be further classified by identifying the filament or electron-emitting material. The materials in regular use are tungsten, thoriated tungsten, and metals which have been coated with alkaline-earth oxides. Tungsten filaments are made from the pure metal. Because they must operate at high temperatures (a dazzling white) to emit sufficient electrons, a relatively large amount of filament power is required.

Thoriated-tungsten filaments are made from tungsten impregnated with thorium oxide. Due to the presence of thorium, these filaments liberate electrons at a more moderate temperature of about 1700°C (a bright yellow) and are, therefore, much more economical of filament power than are pure tungsten filaments.

Alkaline earths are usually applied as a coating on a nickel-alloy wire or ribbon. This coating, which is dried in a relatively thick layer on the filament, requires only a relatively low temperature of about 700-750°C (a dull red) to produce a copious supply of electrons. Coated filaments operate very efficiently and require relatively little filament power. However, each of these cathode materials has special advantages which determine the choice for a particular application.

Directly heated filament-cathodes require comparatively little heating power. They are used in tube types designed for battery operation because it is, of course, desirable to impose as small a drain as possible on the batteries. They are also used in rectifiers such as the 1G3GT/1B3GT and the 5Y3GT.

An **indirectly heated cathode,** or **heater-cathode,** consists of a thin metal sleeve coated with electron-emitting material such as alkaline-earth oxides. The emissive surface of the cathode is maintained at the required temperature (approximately 1050°K) by resistance-heating of a tungsten or tungsten-alloy wire which is placed inside the cathode sleeve and electrically insulated from it, as shown in Fig. 2. The heater is used only for the purpose of heating the cathode sleeve and sleeve coating to an electron-emitting temperature. Useful emission does not take place from the heater wire.

A new dark heater insulating coating developed by RCA has better heat transfer than earlier aluminum-oxide coatings, and makes it possible to operate heaters at lower temperatures for given power inputs. Because the tensile strength of the heater wire increases at the lower operating temperatures, tubes using **dark heaters** have increased reliability, stability, and life.

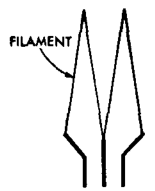


Fig. 1—Filament or directly heated cathode.

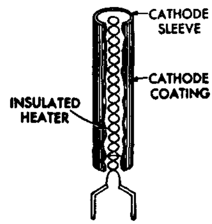


Fig. 2—Indirectly heated cathode or heater-cathode.

The heater-cathode construction is well adapted for use in electron tubes intended for operation from ac power lines and from storage batteries. The use of separate parts for emitter and heater functions, the electrical insulation of the heater from the emitter, and the shielding effect of the sleeve may all be utilized in the design of the tube to minimize the introduction of hum from the ac heater supply and to minimize electrical interference which might enter the tube circuit through the heater-supply line. From the viewpoint of circuit design, the heater-cathode construction offers advantages in connection flexibility because of the electrical separation of the heater from the cathode.

Another advantage of the heater-cathode construction is that it makes practical the design of a rectifier tube having close spacing between its cathode and plate, and of an amplifier tube having close spacing between its cathode and grid. In a close-spaced rectifier tube, the voltage drop in the tube is low, and, therefore, the regulation is improved. In an amplifier tube, the close spacing increases the gain obtainable from the tube. Because of the advantages of the heater-cathode construction, almost all present-day receiving tubes designed for ac operation have heater-cathodes.

Generic Tube Types

Electrons are of no value in an electron tube unless they can be put to work. Therefore, a tube is designed with the parts necessary to utilize electrons as well as those required to produce them. These parts consist of a cathode and one or more supplementary electrodes. The electrodes are enclosed in an evacuated envelope having the necessary connections brought out through air-tight seals. The air is removed from the envelope to allow free movement of the electrons and to prevent injury to the emitting surface of the cathode.

When the cathode is heated, electrons leave the cathode surface and form an invisible cloud in the space around it. Any positive electric potential within the evacuated envelope offers a strong attraction to the electrons (unlike electric charges attract; like charges repel). Such a positive electric potential can be supplied by an **anode** (positive electrode) located within the tube in proximity to the cathode.

Diodes

The simplest form of electron tube contains two electrodes, a cathode and an anode (plate), and is often called a diode, the family name for a two-electrode tube. In a diode, the positive potential is supplied by a suitable electrical source connected between the plate terminal and a cathode terminal, as shown in Fig. 3. Under the influence of the positive plate potential, electrons

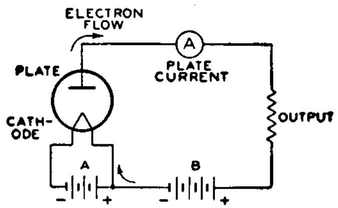


Fig. 3—Basic diode circuit.

flow from the cathode to the plate and return through the external plate-battery circuit to the cathode, thus completing the circuit. This flow of electrons is known as the **plate current.**

If a negative potential is applied to the plate, the free electrons in the space surrounding the cathode will be forced back to the cathode and no plate current will flow. If an alternating voltage is applied to the plate, the plate is alternately made positive and negative. Because plate current flows only during the time when the plate is positive, current flows through the tube in only one direction and is said to be rectified. Fig. 4 shows the rectified output current produced by an alternating input voltage.

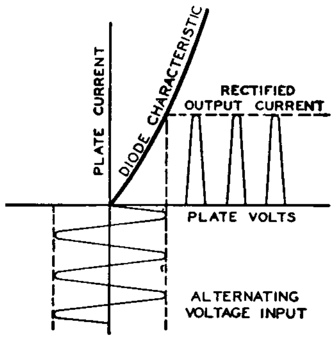


Fig. 4—Current characteristics of rectifier circuit.

Diode rectifiers are used in ac receivers to convert the ac supply voltage to dc voltage for the electrodes of the other tubes in the receiver. Rectifier tubes having only one plate and one cathode, such as the 35W4, are called **half-wave rectifiers,** because current can flow only during one-half of the alternating-current cycle. When two plates and one or more cathodes are used in the same tube, current may be obtained on both halves of the ac cycle. The 6X4, 5Y3GT, and 5U4GB are examples of this type and are called **full-wave rectifiers.**

Not all of the electrons emitted by the cathode reach the plate. Some return to the cathode, while others remain in the space between the cathode and plate for a brief period to produce an effect known as **space charge.** This charge has a repelling action on other electrons which leave the cathode surface and impedes their passage to the plate. The extent of this action and the amount of space charge depend on the cathode temperature, the distance between the cathode and the plate, and the plate potential. The higher the plate potential, the less is the tendency for electrons to remain in the space-charge region and repel other electrons. This effect may be noted by applying increasingly higher plate voltages to a tube operating at a fixed heater or filament voltage. Under these conditions, the maximum number of available electrons is fixed, but increasingly higher plate voltages will succeed in attracting a greater proportion of the free electrons.

Beyond a certain plate voltage, however, additional plate voltage has little effect in increasing the plate current because all of the electrons emitted by the cathode are already being drawn to the plate. This maximum current, illustrated in Fig. 5, is called **saturation current.** Because it is an indication of the total number of electrons emitted, it is also known as **emission current** or simply **emission.**

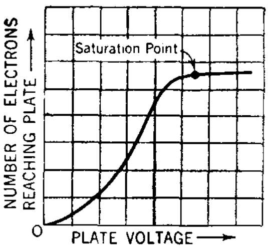


Fig. 5—Current characteristic of diode tube.

Although tubes are sometimes tested by measurement of their emission current, it is generally not advisable to measure the full value of emission because this value would be sufficiently large to cause change in the tube characteristics or even to damage the tube. Consequently, while the test value of emission current is somewhat larger than the maximum current which will be required from the cathode in the use of the tube, it is ordinarily less than the full emission current. The emission test, therefore, is used to indicate whether the cathode can supply a sufficient number of electrons for satisfactory operation of the tube.

If space charge were not present to repel electrons coming from the cathode, the same plate current could be produced at a lower plate voltage. One way to make the effect of space charge small is to make the distance between plate and cathode small. This method is used in rectifier types having heater-cathodes, such as the 5V4GA and the 6AX5GT. In these types the radial distance between cathode and plate is only about two hundredths of an inch.

Another method of reducing space-charge effect is utilized in **mercury-vapor rectifier tubes.** When such tubes are operated, a small amount of mercury contained in the tube is partially vaporized, filling the space inside the bulb with mercury atoms. These atoms are bombarded by electrons on their way to the plate. If the electrons are moving at a sufficiently high speed, the collisions tear off electrons from the mercury atoms. The mercury atom is then said to be **"ionized,"** *i.e.,* it has lost one or more electrons and, therefore, has a positive charge. Ionization is evidenced by a bluish-green glow between the cathode and plate. When ionization occurs, the space charge is neutralized by the positive mercury atoms so that increased numbers of electrons are made available. Mercury-vapor tubes are used primarily for power rectifiers.

Ionic-heated-cathode rectifiers depend on gas ionization for their operation. These tubes are of the full-wave design and contain two anodes and a coated cathode sealed in a bulb containing a reduced pressure of inert gas. The cathode becomes hot during tube operation, but the heating effect is caused by bombardment of the cathode by ions within the tube rather than by heater or filament current from an external source.

The internal structure of an ionic-heated-cathode tube is designed so that when sufficient voltage is applied to the tube, ionization of the gas occurs between the anode which is instantaneously positive and the cathode. Under normal operating voltages, ionization does not take place between the anode that is negative and the cathode, so that the requirements for rectification are satisfied. The initial small flow of current through the tube is sufficient to raise the cathode temperature quickly to incandescence, whereupon the cathode emits electrons. The voltage drop in such tubes is slightly higher than that of the usual hot-cathode gas rectifiers because energy is taken from the ionization discharge to keep the cathode at operating temperature. Proper operation of these rectifiers requires a minimum flow of load current at all times to maintain the cathode at the temperature required to supply sufficient emission.

Triodes

When a third electrode, called the **grid,** is placed between the cathode and plate, the tube is known as a triode, the family name for a three-electrode tube. The grid usually consists of relatively fine wire wound on two support rods (siderods) and extending the length of the cathode. The spacing between turns of wire is large compared with the size of the wire so that the passage of electrons from cathode to plate is practically unobstructed by the grid. In some types, a **frame grid** is used. The frame consists of two siderods supported by four metal straps. Extremely fine lateral wire (diameter of 0.5 mil or less) is wound under tension around the frame. This type of grid permits the use of closer spacings between grid wires and between tube electrodes, and thus improves tube performance.

The purpose of the grid is to control the flow of plate current. When a tube is used as an amplifier, a negative dc voltage is usually applied to the grid. Under this conditon the grid does not draw appreciable current.

The number of electrons attracted to the plate depends on the combined effect of the grid and plate polarities, as shown in Fig. 6. When the plate is positive, as is normal, and the dc grid voltage is made more and more negative, the plate is less able to attract electrons to it and plate current decreases. When the grid is made less and less negative (more and more positive), the plate more readily attracts electrons to it and plate current increases. Hence, when the voltage on the grid is varied in accordance with a signal, the plate current varies with the signal. Because a small voltage applied to the grid can control a comparatively large amount of plate current, the signal is amplified by the tube. Typical three-electrode tube types are the 6C4 and 6AF4A.

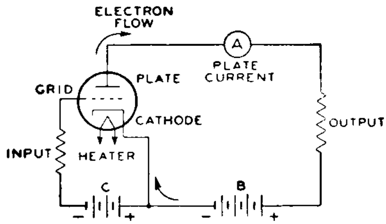


Fig. 6—Basic triode circuit.

The grid, plate, and cathode of a triode form an electrostatic system, each electrode acting as one plate of a small capacitor. The capacitances are those existing between grid and plate, plate and cathode, and grid and cathode. These capacitances are known as **interelectrode capacitances.** Generally, the capacitance between grid and plate is of the most importance. In high-gain radio-frequency amplifier circuits, this capacitance may act to produce undesired coupling between the **input circuit,** the circuit between grid and cathode, and the **output circuit,** the circuit between plate and cathode. This coupling is undesirable in an amplifier because it may cause instability and unsatisfactory performance.

Tetrodes

The capacitance between grid and plate can be made small by mounting an additional electrode, called the **screen grid** (grid No. 2), in the tube. With the addition of the grid No. 2, the tube has four electrodes and is, accordingly, called a tetrode. The screen grid or grid No. 2 is mounted between the grid No. 1 **(control grid)** and the plate, as shown in Fig. 7, and acts as an electrostatic shield between them, thus reducing the grid-to-plate capacitance. The effectiveness of this shielding action is increased by a bypass

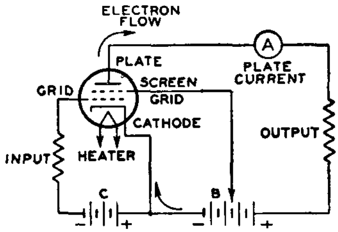


Fig. 7—Basic tetrode circuit.

capacitor connected between screen grid and cathode. By means of the screen grid and this bypass capacitor, the grid-plate capacitance of a tetrode is made very small. In practice, the grid-plate capacitance is reduced from several picofarads (pF) for a triode to 0.01 pF or less for a screen-grid tube.

The screen grid has another desirable effect in that it makes plate current practically independent of plate voltage over a certain range. The screen grid is operated at a positive voltage and, therefore, attracts electrons from the cathode. However, because of the comparatively large space between wires of the screen grid, most of the electrons drawn to the screen grid pass through it to the plate. Hence the screen grid supplies an electrostatic force pulling electrons from the cathode to the plate. At the same time the screen grid shields the electrons between cathode and screen grid from the plate so that the plate exerts very little electrostatic force on electrons near the cathode.

So long as the plate voltage is higher than the screen-grid voltage, plate current in a screen-grid tube depends to a great degree on the screen-grid voltage and very little on the plate voltage. The fact that plate current in a screen-grid tube is largely independent of plate voltage makes it possible to obtain much higher amplification with a tetrode than with a triode. The low grid-plate capacitance makes it possible to obtain this high amplification without plate-to-grid feedback and resultant instability. In receiving-tube applications, the tetrode has been replaced to a considerable degree by the pentode.

Pentodes

In all electron tubes, electrons striking the plate may, if moving at sufficient speed, dislodge other electrons. In two- and three-electrode types, these dislodged electrons usually do not cause trouble because no positive electrode other than the plate itself is present to attract them. These electrons, therefore, are drawn back to the plate. Emission caused by bombardment of an electrode by electrons from the cathode is called **secondary emission** because the effect is secondary to the original cathode emission.

In the case of screen-grid tubes, the proximity of the positive screen grid to the plate offers a strong attraction to these secondary electrons, and particularly so if the plate voltage swings lower than the screen-grid voltage. This effect reduces the plate current and limits the useful plate-voltage swing for tetrodes.

The effects of secondary emission are minimized when a fifth electrode is placed within the tube between the screen grid and plate. This fifth electrode is known as the **suppressor grid** (grid No. 3) and is usually connected to the cathode, as shown in Fig. 8. Because of its negative potential with respect to the plate, the suppressor grid retards the flight of secondary electrons and diverts them back to the plate.

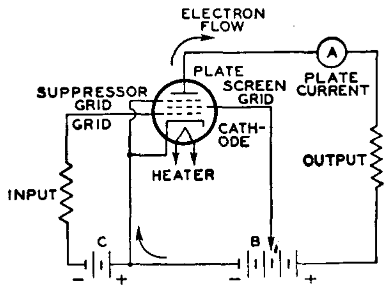


Fig. 8—Basic pentode circuit.

The *family name* for a five-electrode tube is "pentode." In power-output pentodes, the suppressor grid makes possible higher power output with lower grid-driving voltage; in radio-frequency amplifier pentodes, the suppressor grid makes possible high voltage amplification at moderate values of plate voltage. These desirable features result from the fact that the plate-voltage swing can be made very large. In fact, the plate voltage may be as low as, or lower than, the screen-grid voltage without serious loss in signal-gain capability. Representative pentodes used for power amplification are the 6CL6 and 6K6GT; representative pentodes used for voltage amplification are the 6AU6A, 6BA6, and 5879.

Beam Power Tubes

A beam power tube is a tetrode or pentode in which directed electron beams are used to increase substantially the power-handling capability of the tube. Such a tube contains a cathode, a control grid (grid No. 1), a screen grid (grid No. 2), a plate, and, optionally, a suppressor grid (grid No. 3). When a beam power tube is designed without an actual suppressor grid, the electrodes are so spaced that secondary emission from the plate is suppressed by space-charge effects between screen grid and plate. The space charge is produced by the slowing up of electrons traveling from a high-potential screen grid to a lower-potential plate. In this low-velocity region, the space charge produced is sufficient to repel secondary electrons emitted from the plate and to cause them to return to the plate.

Beam power tubes of this design employ beam-confining electrodes at cathode potential to assist in producing the desired beam effects and to prevent stray electrons from the plate from returning to the screen grid outside of the beam. A feature of a beam power tube is its low screen-grid current. The screen grid and the control grid are spiral wires wound so that each turn of the screen grid is shaded from the cathode by a grid turn. This alignment of the screen grid and control grid causes the electrons to travel in sheets between the turns of the screen grid so that very few of them strike the screen grid. Because of the effective suppressor action provided by space charge and because of the low current drawn by the screen grid, the beam power tube has the advantages of high power output, high power sensitivity, and high efficiency.

Fig. 9 shows the structure of a beam power tube employing space-charge suppression and illustrates how

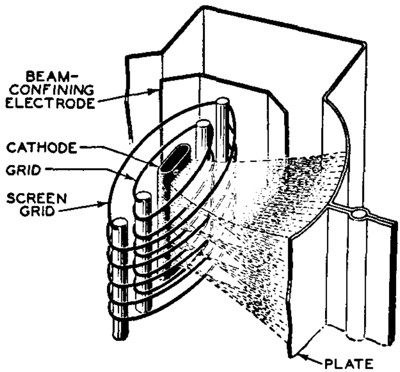


Fig. 9—Structure of beam power tube showing beam-confining action.

the electrons are confined to beams. The beam condition illustrated is that for a plate potential less than the screen-grid potential. The high-density space-charge region is indicated by the heavily dashed lines in the beam. Note that the edges of the beam-confining electrodes coincide with the dashed portion of the beam. In this way the space-charge potential region is extended beyond the beam boundaries and stray secondary electrons are prevented from returning to the screen grid outside of the beam. The space-charge effect may also be obtained by use of an actual suppressor grid. Examples of beam power tubes are 6AQ5A, 6L6GC, 6V6GTA, and 50C5.

Multi-Electrode and Multi-Unit Tubes

Early in the history of tube development and application, tubes were designed for a general service; that is, a single tube type—a triode—was used as a radio-frequency amplifier, an intermediate-frequency amplifier, an audio-frequency amplifier, an oscillator, or a detector. Obviously, with this diversity of application, one tube did not meet all requirements to the best advantage.

Later and present trends of tube design are the development of "specialty" types. These types are intended either to give optimum performance in a particular application or to combine in one bulb functions which formerly required two or more tubes. The first class of tubes includes such examples of specialty types as the 6CB6A and 6BY6. Types of this class generally require more than three electrodes to obtain the desired special characteristics and may be broadly classed as multi-electrode types. The 6BY6 is an especially interesting type in this class. This tube has an unusually large number of electrodes, namely seven, exclusive of the heater. Plate current in the tube is varied at two different frequencies at the same time. The tube is designed primarily for use as a combined sync separator and sync clipper in television receivers.

The second class includes multi-unit tubes such as the twin-diode triodes 6CN7 and 6AV6, as well as triode-pentodes such as the 6U8A and 6X8. This class also includes class A twin triodes such as the 6CG7 and 12AX7A, and types such as the 6CM7 containing dissimilar triode units used primarily as combined vertical oscillators and vertical deflection amplifiers in television receivers. Full-wave rectifiers are also multi-unit types.

A third class of tubes combines features of each of the other two classes. Typical of this third class are the pentagrid-converter types 6BE6 and 6SA7. These tubes are similar to the multi-electrode types in that they have seven electrodes, all of which affect the electron stream; and they are similar to the multi-unit tubes in that they perform simultaneously the double function of oscillator and mixer in superheterodyne receivers.

Receiving Tube Structure

Receiving tubes generally utilize a glass or metal envelope and a **base.** Originally, the base was made of metal or molded phenolic material. Types having a glass envelope and a molded phenolic base include the **"octal"** types such as the 5U4GB and the 6SN7GTB. Types having a metal envelope and molded phenolic octal base include the 6F6 and the 6L6. Many modern types utilize integral glass bases. Present-day conventional tube designs utilizing glass envelopes and integral glass bases include the seven-pin and nine-pin **miniature** types, the nine-pin **novar** and **neonoval** types, and the twelve-pin **duodecar** types. Examples of the seven-pin miniature types are the 6AU6A and 6BN6. Examples of the nine-pin miniature types are the 12AU7A and 6EA8. Examples of the novar types are the 6BH3 and 7868. The nine-pin base for the novar types has a relatively large pin-circle diameter and long pins to insure firm retention of the tube in its socket.

The **nuvistor** concept provided a new approach to electron tube design. Nuvistor tubes utilize a light-weight cantilever-supported cyclindrical electrode structure housed in a ceramic-metal envelope. These tubes combine new materials, processes, and fabrication techniques. Examples of the nuvistor are the 6CW4 and the 6DV4.

Television Picture Tubes

The picture tube, or kinescope, is a multi-electrode tube used principally in television receivers for picture display. It consists essentially of an electron gun, a glass or metal-and-glass envelope and faceplate combination, and a fluorescent screen.

The electron gun includes a cathode for the production of free electrons, one or more control electrodes for accelerating the electrons in the beam, and, optionally, a device for "trapping" unwanted ions out of the electron beam.

Focusing of the beam is accomplished either electromagnetically by means of a focusing coil placed on the neck of the tube, or electrostatically, as shown in Fig. 10, by means of a focusing electrode (grid No. 4) within the envelope of the tube. The screen is a white-fluorescing phosphor P4 of either the silicate or the sulfide type.

Deflection of the beam is accomplished either electrostatically by means of deflecting electrodes within the envelope of the tube, or electromagnetically by means of a deflecting yoke placed on the neck of the tube. Fig. 10 shows the structure of the gun section of a picture tube and illustrates how the electron beam is formed and how the beam is deflected by means of an electromagnetic deflecting yoke. In this type of tube, ions in the beam are prevented from damaging the fluorescent screen by an aluminum film on the gun side of the screen. This film not only "traps" unwanted ions, but also improves picture contrast. In many types of non-aluminized tubes, ions are separated from the electron beam by means of a tilted-gun and ion-trap-magnet arrangement.

Color television picture tubes are similar to black-and-white picture tubes, but differ in three major ways: (1) The light-emitting screen is made up of trios of phosphor dots deposited in an interlaced pattern. Each dot of a trio is capable of emitting light in one of the three primary colors (red, green, or blue). (2) A shadow mask mounted near the screen of the tube contains over 300,000 apertures, one for each of the phosphor dot trios. This mask provides color separation by shadowing two of the three phosphor dots of each trio. (3) Three closely spaced electron guns, built as a unit, provide separate beams for excitation of the three different color-phosphor-dot arrays. Thus it is possible to control the brightness of each of the three colors independently of the other two. Fig. 11 shows a cutaway view of a color television picture tube.

The three electron guns are mounted with their axes tilted toward the central axis of the envelope, and are spaced 120 degrees with respect to each other. The focusing electrodes of the three guns are interconnected internally, and their potential is adjusted to cause the separate beams to focus at the phosphor-dot screen. All three beams must be made to converge at the screen while they are simultaneously being deflected. Convergence is accomplished by the action of static and

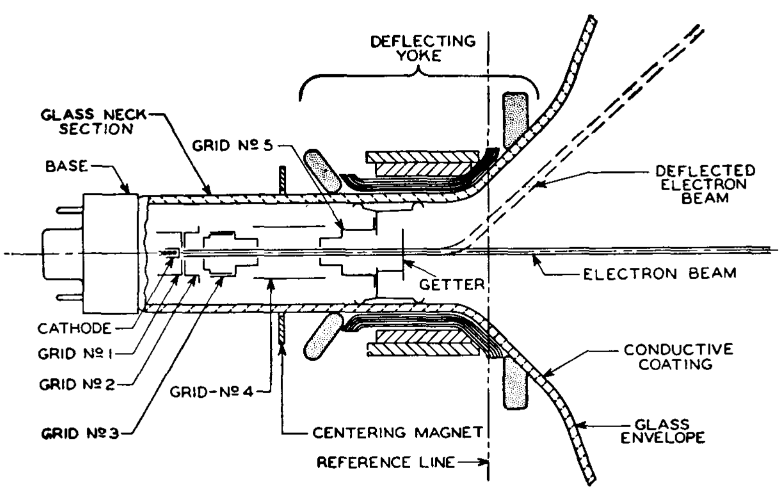


Fig. 10—Structure of television-picture-tube electron gun

dynamic magnetic fields set up by the radial-converging magnet assembly mounted on the neck of the tube. These fields are coupled into the radial-converging pole pieces within the tube. Another pair of pole pieces in the tube is activated by the lateral-converging magnet also mounted on the neck of the tube. These pole pieces permit lateral shift in position of the blue beam in opposition to the lateral shift of the green and red beams.

A purifying magnet is used with color picture tubes to provide a magnetic field, adjustable in magnitude and direction, to effect register over the entire area of the screen. A magnetic shield is used to minimize the effects of the earth's magnetic field.

Deflection of the three beams is accomplished simultaneously by a deflecting yoke using four electromagnetic coils similar to the deflecting yoke used for black-and-white picture tubes.

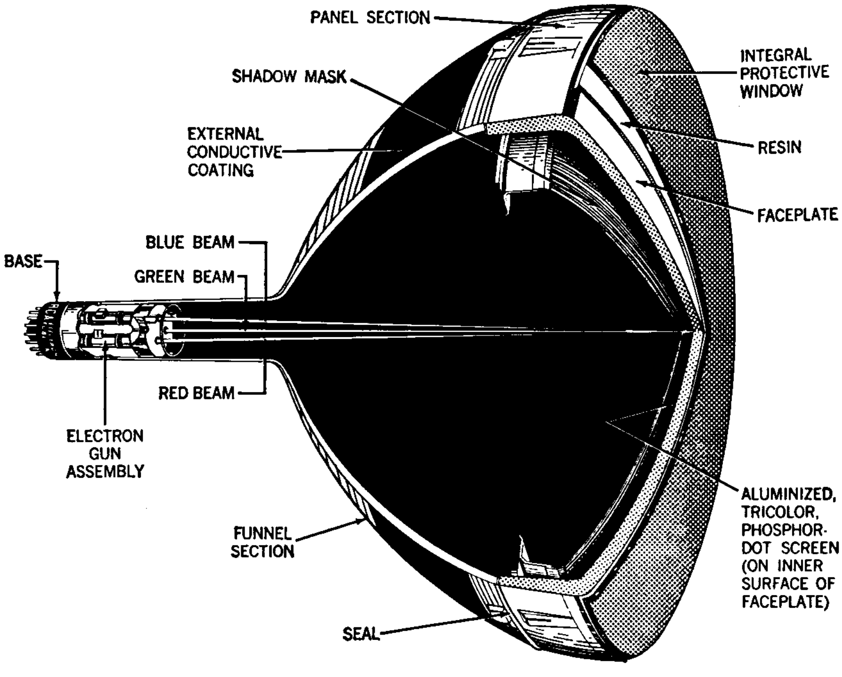


Fig. 11—Cutaway view of color television picture tube.

Electron Tube Characteristics

THE term "characteristics" is used to identify the distinguishing electrical features and values of an electron tube. These values may be shown in curve form or they may be tabulated. When the characteristics values are given in curve form, the curves may be used for the determination of tube performance and the calculation of additional tube factors.

Tube characteristics are obtained from electrical measurements of a tube in various circuits under certain definite conditions of voltages. Characteristics may be further described by denoting the conditions of measurements. For example, Static Characteristics are the values obtained with different dc potentials applied to the tube electrodes, while Dynamic Characteristics are the values obtained with an ac voltage on a control grid under various conditions of dc potentials on the electrodes. The dynamic characteristics, therefore, are indicative of the performance capabilities of a tube under actual working conditions.

Static characteristics may be shown by plate characteristics curves and transfer (mutual) characteristics curves. These curves present the same information, but in two different forms to increase its usefulness. The plate characteristic curve is obtained by varying plate voltage and measuring plate current for different grid-bias voltages, while the transfer-characteristic curve is obtained by varying grid-bias voltage and measuring plate current for different plate voltages. A plate-characteristic family of curves is shown in Fig. 12. Fig. 13 gives the transfer-characteristic family of curves for the same tube.

Dynamic characteristics include amplification factor, plate resistance, control-grid—plate transconductance, and certain detector characteristics, and may be shown in curve form for variations in tube operating conditions.

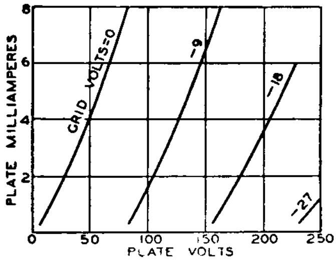


Fig. 12—Family of plate-characteristics curves.

The **amplification factor,** or μ, is the ratio of the change in plate voltage to a change in control-electrode voltage in the opposite direction, under the condition that the plate current remains

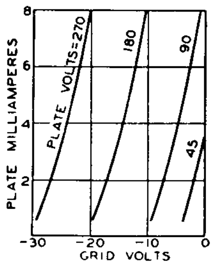


Fig. 13—Family of transfer-characteristics curves.

unchanged and that all other electrode voltages are maintained constant. For example, if, when the plate voltage is made 1 volt more positive, the control-electrode (grid-No. 1) voltage must be made 0.1 volt more negative to hold plate current unchanged, the amplification factor is 1 divided by 0.1, or 10. In other words, a small voltage variation in the grid circuit of a tube has the same effect on the plate current as a large plate-voltage change—the latter equal to the product of the grid-voltage change and amplification factor. The μ of a tube is often useful for calculating stage gain. This use is discussed in the **Electron Tube Applications** section.

Plate resistance (r_p) of an electron tube is the resistance of the path between cathode and plate to the flow of alternating current. It is the quotient of a small change in plate voltage divided by the corresponding change in plate current and is expressed in ohms, the unit of resistance. Thus, if a change of 0.1 milliampere (0.0001 ampere) is produced by a plate-voltage variation of 1 volt, the plate resistance is 1 divided by 0.0001, or 10000 ohms.

Control grid-to-plate transconductance, or simply **transconductance** (g_m), is a factor which combines in one term the amplification factor and the plate resistance, and is the quotient of the first divided by the second. This term has also been known as mutual conductance. Transconductance may be more strictly defined as the quotient of a small change in plate current (amperes) divided by the small change in the control-grid voltage producing it, under the condition that all other voltages remain unchanged. Thus, if a grid-voltage change of 0.5 volt causes a plate-current change of 1 milliampere (0.001 ampere), with all other voltages constant, the transconductance is 0.001 divided by 0.5, or 0.002 mho. A "mho" is the unit of conductance and was named by spelling ohm backwards. For convenience, a millionth of a mho, or a micromho (μmho), is used to express transconductance. Thus, in the example, 0.002 mho is 2000 micromhos.

Conversion transconductance (g_c) is a characteristic associated with the mixer (first detector) function of tubes and may be defined as the quotient of the intermediate-frequency (if) current in the primary of the if transformer divided by the applied radio-frequency (rf) voltage producing it; more precisely, it is the limiting value of this quotient as the rf voltage and if current approach zero. When the performance of a frequency converter is determined, conversion transconductance is used in the same way as control grid-to-plate transconductance is used in single-frequency amplifier computations.

The **plate efficiency** of a power amplifier tube is the ratio of the ac power output (P_o) to the product of the average dc plate voltage (E_b) and dc plate current (I_b) at full signal, or

$$\text{Plate efficiency } \% = \frac{P_o \text{ watts}}{E_b \text{ volts} \times I_b \text{ amperes}} \times 100$$

The **power sensitivity** of a tube is the ratio of the power output to the square of the input signal voltage (E_{in}), and is expressed in mhos as follows:

$$\text{Power sensitivity (mhos)} = \frac{P_o \text{ watts}}{(E_{in,}\ \text{rms})^2}$$

Electron Tube Applications

THE diversified applications of an electron receiving tube have, within the scope of this section, been treated under seven headings: Rectification; Detection; Amplification; TV Scanning, Sync, and Deflection; Oscillation; Frequency Conversion; and Tuning Indication with Electron-Ray Tubes. Although these operations may take place at either radio or audio frequencies and may involve the use of different circuits and different supplemental parts, the general considerations of each kind of operation are basic.

General System Functions

When speech, music, or video information is transmitted from a radio or television station, the station radiates a modulated radio-frequency (rf) carrier. The function of a radio or television receiver is simply to reproduce the modulating wave from the modulated carrier.

As shown in Fig. 14, a superheterodyne radio receiver picks up the transmitted modulated rf signal, amplifies it and converts it to a modulated intermediate-frequency (if) signal, amplifies the modulated if signal, separates the modulating signal from the basic carrier wave, and amplifies the resulting audio signal to a level sufficient to produce the desired volume in a speaker. In addition, the receiver usually includes some means of producing automatic gain control (agc) of the modulated signal before the audio information is separated from the carrier.

The transmitted rf signal picked up by the radio receiver may contain either amplitude modulation (AM) or frequency modulation (FM). (These modulation techniques are described later in the section on **Detection.**) In either case, amplification prior to the detector stage is performed by tuned amplifier circuits designed for the proper frequency and bandwidth. Frequency conversion is performed by mixer and oscillator circuits or by a single converter stage which performs both mixer and oscillator functions. Separation of the modulating signal is normally accomplished by one or more diodes in a detector or discriminator circuit. Amplification of the audio signal is then performed by one or more audio amplifier stages.

Audio-amplifier systems for phonograph or tape recordings are similar to the stages after detection in a radio receiver. The input to the amplifier is a low-power-level audio signal from the

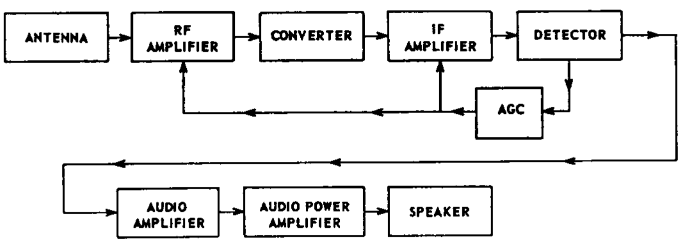


Fig. 14—Simplified block diagram for a broadcast-band receiver.

phonograph or magnetic-tape pickup head. This signal is usually amplified through a preamplifier stage, one or more low-level (pre-driver or driver) audio stages, and an audio power amplifier. The system may also include frequency-selective circuits which act as equalization networks and/or tone controls.

The operation of a television receiver is more complex than that of a radio receiver, as shown by the simplified block diagram in Fig. 15. The tuner section of the receiver selects the proper rf signals for the desired channel frequency, amplifies them, and converts them to a lower intermediate frequency. As in a radio receiver, these functions are accomplished in rf-amplifier, mixer, and local-oscillator stages. The if signal is then amplified in if-amplifier stages which provide the additional gain required to bring the signal level to an amplitude suitable for detection.

After if amplification, the detected signal is separated into sound and picture information. The sound signal is amplified and processed to provide an audio signal which is fed to an audio amplifier system similar to those described above. The picture (video) signal is passed through a video amplifier stage which conveys beam-intensity information to the television picture tube and thus controls instantaneous "spot" brightness. At the same time, deflection circuits cause the electron beam of the picture tube to move the "spot" across the faceplate horizontally and vertically. Special "sync" signals derived from the video signal assure that the horizontal and vertical scanning are timed so that the picture produced on the receiver exactly duplicates the picture being viewed by the camera or pickup tube.

A communications transceiver contains transmitting circuits, as well as receiving circuits similar to those of a radio receiver. The transmitter portion of such a system consists of two sections.

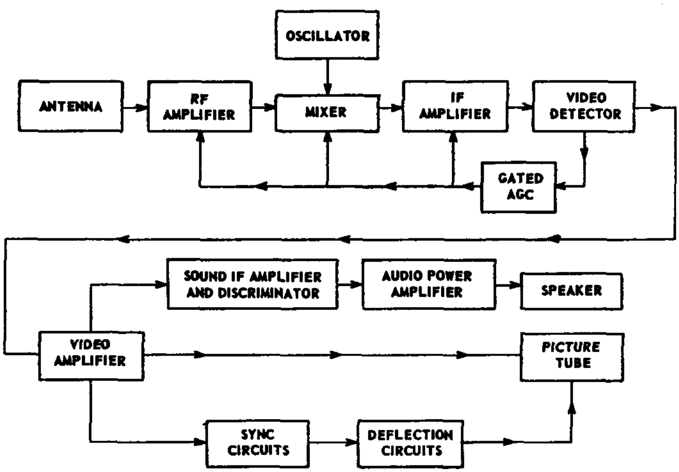


Fig. 15—Simplified block diagram for a black-and-white television receiver.

In one section, the desired intelligence (voice, code, or the like) is picked up and amplified through one or more amplifier stages (which are usually common to the receiver portion) to a high-level stage called a modulator. In the other section, an rf signal of the desired frequency is developed in an oscillator stage and amplified in one or more rf-amplifier stages. The audio-frequency (af) modulating signal is impressed on the rf carrier in the final rf-power-amplifier stage (high-level modulation), in the rf low-level stage (low-level modulation), or in both. Fig. 16 shows a simplified block diagram of the transmitter portion of a citizens-band trans-

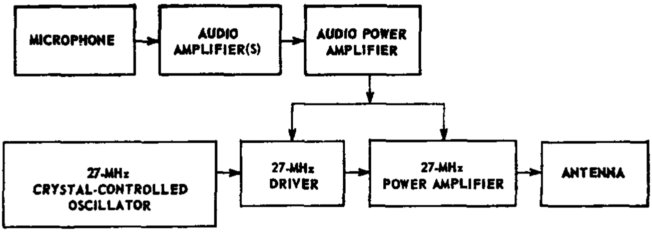


Fig. 16—Simplified block diagram for the transmitter portion of a 27-MHz communications receiver.

ceiver that operates at a frequency of 27 MHz (megacycles per second). The transmitting section of a communications system may also include frequency-multiplier circuits which raise the frequency of the developed rf signal as required.

Rectification

The rectifying action of a diode finds important applications in supplying a receiver with dc power from an ac line and in supplying high dc voltage from a high-voltage pulse. A typical arrangement for converting ac to dc includes a rectifier tube, a filter, and a voltage divider. The rectifying action of the tube is explained briefly under **Diodes,** in the **Electrons, Electrodes, and Electron Tubes** section. High-voltage pulse rectification is described later under **Horizontal Output Circuits.**

The function of a filter is to smooth out the ripple of the tube output, as indicated in Fig. 17, and to increase rectifier efficiency. The action of the filter is explained in the **Electron Tube Installation** section under **Filters.** The voltage divider is used to cut down the output voltage to the values required by the plates and the other electrodes of the tubes in the receiver.

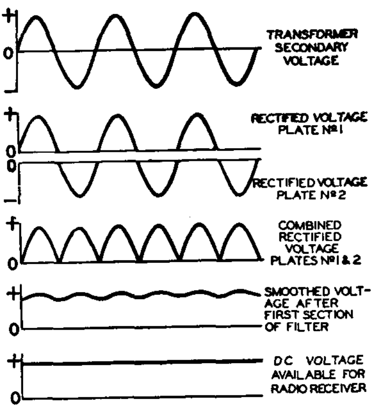


Fig. 17—Voltage waveforms of full-wave rectifier circuit.

A **half-wave rectifier** and a **full-wave rectifier** circuit are shown in Fig. 18. In the half-wave circuit, current flows through the rectifier tube to the filter on every other half-cycle of the ac input voltage when the plate is positive with respect to the cathode. In the full-wave circuit, current flows to the filter on every half-cycle, through plate No. 1 on one half-cycle when plate No. 1 is positive with respect to the cathode, and through plate No. 2 on the next half-cycle when plate No. 2 is positive with respect to the cathode.

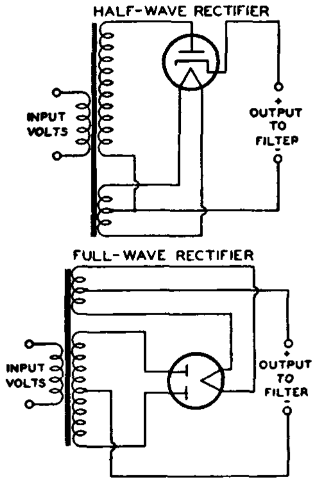


Fig. 18—Half-wave and full-wave rectifier circuits.

Because the current flow to the filter is more uniform in the full-wave circuit than in the half-wave circuit, the output of the full-wave circuit requires less filtering. Rectifier operating information and circuits are given under each rectifier tube type and in the **Circuits** section, respectively.

Parallel operation of rectifier tubes furnishes an output current greater than that obtainable with the use of one tube. For example, when two full-wave rectifier tubes are connected in parallel, the plates of each tube are connected together and each tube acts as a half-wave rectifier. The permissible voltage and load conditions per tube are the same as for full-wave service but the total load-handling capability of the complete rectifier is approximately doubled.

When mercury-vapor rectifier tubes are connected in parallel, a stabilizing resistor of 50 to 100 ohms should be connected in series with each plate lead in order that each tube will carry an equal share of the load. The value of the resistor to be used will depend on the amount of plate current that passes through the rectifier. Low plate current requires a high value; high plate current, a low value. When the plates of mercury-vapor rectifier tubes are connected in parallel, the corresponding filament leads should be similarly connected. Otherwise, the tube drops will be considerably unbalanced and larger stabilizing resistors will be required.

Two or more vacuum rectifier tubes can also be connected in parallel to give correspondingly higher output current and, as a result of paralleling their internal resistances, give somewhat increased voltage output. With vacuum types, stabilizing resistors may or may not be necessary depending on the tube type and the circuit.

A **voltage-doubler** circuit of simple form is shown in Fig. 19. The circuit derives its name from the fact that its dc voltage output can be as high as twice the peak value of ac input. Basically, a voltage doubler is a rectifier circuit arranged so that the output voltages of two half-wave rectifiers are in series.

The action of a voltage doubler can be described briefly as follows. On the positive half-cycle of the ac input, that is, when the upper side of the ac input line is positive with respect to the lower side, the upper diode passes current and feeds a positive charge into the upper capacitor. As positive charge accumulates on the upper plate of the capacitor, a positive voltage builds up across the capacitor. On the next half-cycle of the ac input, when the upper side of the line is negative with respect to the lower side, the lower diode passes current so that

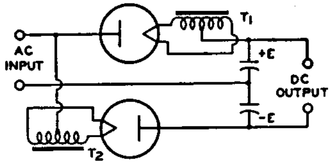


Fig. 19—Full-wave voltage-doubler circuit.

a negative voltage builds up across the lower capacitor.

So long as no current is drawn at the output terminals from the capacitor, each capacitor can charge up to a voltage of magnitude E, the peak value of the ac input. It can be seen from the diagram that with a voltage of $+E$ on one capacitor and $-E$ on the other, the total voltage across the capacitors is 2E. Thus the voltage doubler supplies a no-load dc output voltage twice as large as the peak ac input voltage. When current is drawn at the output terminals by the load, the output voltage drops below 2E by an amount that depends on the magnitude of the load current and the capacitance of the capacitors. The arrangement shown in Fig. 19 is called a full-wave voltage doubler because each rectifier passes current to the load on each halt of the ac input cycle.

Two rectifier types especially designed for use as voltage doublers are the 25Z6GT and 117Z6GT. These tubes combine two separate diodes in one tube. As voltage doublers, the tubes are used in "transformerless" receivers. In these receivers, the heaters of all tubes

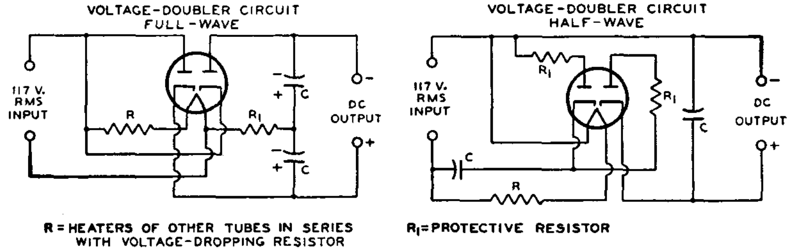


Fig. 20—Full-wave and half-wave voltage-doubler circuits showing heater-supply connections.

in the set are connected in series with a voltage-dropping resistor across the line. The connections for the heater supply and the voltage-doubling circuit are shown in Fig. 20.

With the full-wave voltage-doubler circuit in Fig. 20, it will be noted that the dc load circuit can not be connected to ground or to one side of the ac supply line. This circuit presents certain disadvantages when the heaters of all the tubes in the set are connected in series with a resistance across the ac line. Such a circuit arrangement may cause hum because of the high ac potential between the heaters and cathodes of the tubes.

The half-wave voltage-doubler circuit in Fig. 20 overcomes this difficulty by making one side of the ac line common with the negative side of the dc load circuit. In this circuit, one half of the tube is used to charge a capacitor which, on the following half cycle, discharges in series with the line voltage through the other half of the tube. This circuit is called a half-wave voltage doubler because rectified current flows to the load only on alternate halves of the ac input cycle. The voltage regulation of this arrangement is somewhat poorer than that of the full-wave voltage doubler.

Detection

When speech, music, or video information is transmitted from a radio or television station, the station radiates a radio-frequency (rf) wave which is of either of two general types. In one type, the wave is said to be amplitude modulated when its frequency remains constant and the amplitude is varied. In the other type, the wave is said to be frequency modulated when its amplitude remains essentially constant but its frequency is varied.

The function of the receiver is to reproduce the original modulating wave from the modulated rf wave. The receiver stage in which this function is performed is called the **demodulator** or detector stage.

AM Detection

The effect of **amplitude modulation** on the waveform of the rf wave is shown in Fig. 21. There are three different basic circuits used for the detection of amplitude-modulated waves: the diode detector, the grid-bias detector, and the grid-resistor detector. These circuits are alike in that they eliminate, either partially or completely, alternate half-cycles of the rf wave. With alternate half-cycles removed, the audio variations of the other half-cycles can be

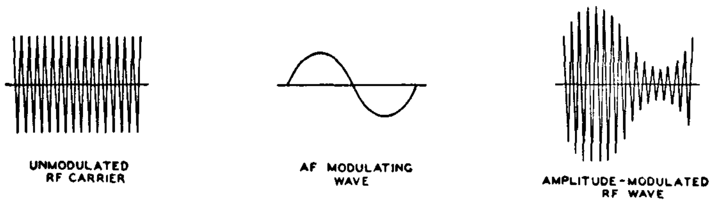


Fig. 21—Waveforms showing effect of amplitude modulation on an rf wave.

amplified to drive headphones or a loudspeaker.

A **diode-detector** circuit is shown in Fig. 22. The action of this circuit when a modulated rf wave is applied is

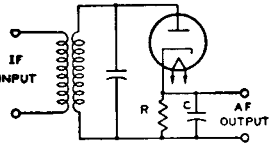


Fig. 22—Basic diode-detector circuit.

illustrated by Fig. 23. The rf voltage applied to the circuit is shown in light line; the output voltage across capacitor C is shown in heavy line.

Between points (a) and (b) on the first positive half-cycle of the applied rf voltage, capacitor C charges up to the peak value of the rf voltage. Then as the applied rf voltage falls away from its peak value, the capacitor holds the cathode at a potential more positive than the voltage applied to the anode.

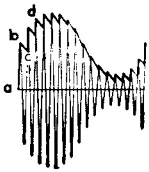


Fig. 23—Waveforms showing modulated rf input (light line) and output voltage (heavy line) of diode-detector circuit.

The capacitor thus temporarily cuts off current through the diode. While the diode current is cut off, the capacitor discharges from (b) to (c) through the diode load resistor R.

When the rf voltage on the anode rises high enough to exceed the potential at which the capacitor holds the cathode, current flows again and the capacitor charges up to the peak value of the second positive half-cycle at (d). In this way, the voltage across the capacitor follows the peak value of the applied rf voltage and reproduces the af modulation.

The curve for voltage across the capacitor, as shown in Fig. 23, is somewhat jagged. However, this jaggedness, which represents an rf component in the voltage across the capacitor, is exaggerated in the drawing. In an actual circuit the rf component of the voltage across the capacitor is negligible. Hence, when the voltage across the capacitor is amplified, the output of the amplifier reproduces the speech or music originating at the transmitting station.

Another way to describe the action of a diode detector is to consider the circuit as a half-wave rectifier. When the rf signal on the plate swings positive, the tube conducts and the rectified current flows through the load resistance R. Because the dc output voltage of a rectifier depends on the voltage of the ac input, the dc voltage across C varies in accordance with the amplitude of the rf carrier and thus reproduces the af signal. Capacitor C should be large enough to smooth out rf or if variations, but should not be so large as to affect the audio variations. Two diodes can be connected in a circuit similar to a full-wave rectifier to provide full-wave detection. However, in practice, the advantages of this connection generally do not justify the extra circuit complication.

The diode method of detection produces less distortion than other methods because the dynamic characteristics of a diode can be made more linear than those of other detectors. The disadvantages of a diode are that it does not amplify the signal, and that it draws current from the input circuit and therefore reduces the selectivity of the input circuit. However, because the diode method of detection produces less distortion and because it permits the use of simple avc circuits without the necessity for an additional voltage supply, the diode method of detection is most widely used in broadcast receivers.

A typical diode-detector circuit using a twin-diode—triode tube is shown in Fig. 24. Both diodes are connected together. R_1 is the diode load resistor. A portion of the af voltage developed across this resistor is applied to the triode grid through the volume control R_3. In a typical circuit, resistor R_1 may be tapped so that five-sixths of the total af voltage across R_1 is applied to the volume control. This tapped connection reduces the af voltage output of the detector circuit slightly, but it

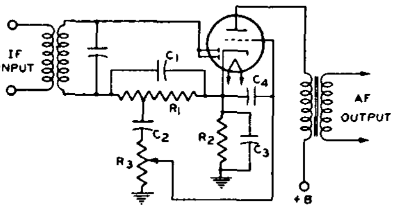


Fig. 24—Typical diode-detector circuit using a twin diode—triode tube.

reduces audio distortion and improves the rf filtering.

DC bias for the triode section is provided by the cathode-bias resistor R_2 and the audio bypass capacitor C_3. The function of capacitor C_2 is to block the dc bias of the cathode from the grid. The function of capacitor C_4 is to bypass any rf voltage on the grid to cathode. A twin-diode—pentode may also be used in this circuit. With a pentode, the af output should be resistance-coupled rather than transformer-coupled.

Another diode-detector circuit, called a diode-biased circuit, is shown in Fig. 25. In this circuit, the triode grid

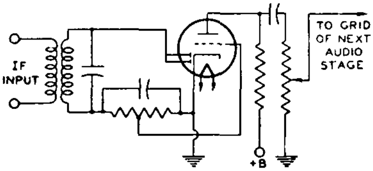


Fig. 25—Diode-biased detector circuit.

is connected directly to a tap on the diode load resistor. When an rf signal voltage is applied to the diode, the dc voltage at the tap supplies bias to the triode grid. When the rf signal is modulated, the af voltage at the tap is applied to the grid and is amplified by the triode.

The advantage of the circuit shown in Fig. 25 over the self-biased arrangement shown in Fig. 24 is that the diode-biased circuit does not employ a capacitor between the grid and the diode load resistor, and consequently does not produce as much distortion of a signal having a high percentage of modulation.

However, there are restrictions on the use of the diode-biased circuit. Because the bias voltage on the triode depends on the average amplitude of the rf voltage applied to the diode, the average amplitude of the voltage applied to the diode should be constant for all values of signal strength at the antenna. Otherwise there will be different values of bias on the triode grid for different signal strengths and the triode will produce distortion. Because there is no bias applied to the diode-biased triode when no rf voltage is applied to the diode, sufficient resistance should be included in the plate circuit of the triode to limit its zero-bias plate current to a safe value.

These restrictions mean, in practice, that the receiver should have a separate-channel automatic-volume-control (avc) system. With such an avc system, the average amplitude of the signal voltage applied to the diode can be held within very close limits for all values of signal strength at the antenna.

The tube used in a diode-biased circuit should be one which operates at a fairly large value of bias voltage. The variations in bias voltage are then a small percentage of the total bias and hence produce small distortion. Tubes taking a fairly large bias voltage are types such as the 6BF6 or 6SR7 having a medium-mu triode. Tube types having a high-mu triode or a pentode should not be used in a diode-biased circuit.

A **grid-bias detector** circuit is shown in Fig. 26. In this circuit, the grid is biased almost to cutoff, *i.e.*, operated so that the plate current with zero signal is practically zero. The bias voltage can be obtained from a cathode-bias resistor, a C-battery, or a bleeder tap. Because of the high negative bias, only the positive half-cycles of the rf signal are amplified by the tube. The signal is, therefore, detected in the

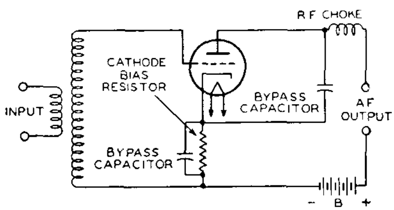


Fig. 26—Grid-bias detector circuit.

plate circuit. The advantages of this method of detection are that it amplifies the signal, besides detecting it, and that it does not draw current from the input circuit and therefore does not reduce the selectivity of the input circuit.

The **grid-resistor-and-capacitor method,** illustrated in Fig. 27, is somewhat more sensitive than the grid-bias method and gives its best results on weak signals. In this circuit, there is no negative dc bias voltage applied to the grid. Hence, on the positive half-cycles of the rf signal, current flows from grid to cathode. The grid and cathode thus act as a diode detector, with the grid resistor as the diode load resistor and the grid capacitor as the rf bypass capacitor. The voltage across the capacitor then reproduces the af modulation in the same manner as has been explained for the diode detector. This voltage appears between the grid and cathode and is therefore amplified in

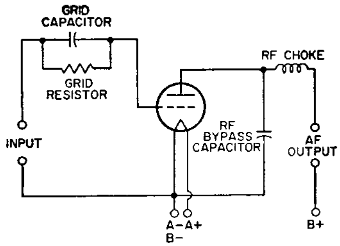


Fig. 27—Detector circuit using grid-resistor-and-capacitor bias.

the plate circuit. The output voltage thus reproduces the original af signal.

In this detector circuit, the use of a high-resistance grid resistor increases selectivity and sensitivity. However, improved af response and stability are obtained with lower values of grid-circuit resistance. This detector circuit amplifies the signal, but draws current from the input circuit and therefore reduces the selectivity of the input circuit.

FM Detection

The effect of **frequency modulation** on the waveform of the rf wave is shown in Fig. 28. In this type of transmission, the frequency of the rf wave deviates from a mean value, at an audio-frequency rate depending on the modu-

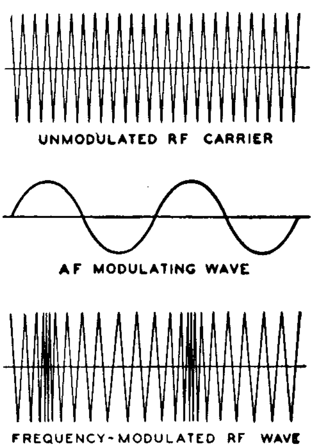


Fig. 28—Waveforms showing effect of frequency modulation on an rf wave.

lation, by an amount that is determined in the transmitter and is proportional to the amplitude of the af modulation signal.

For this type of modulation, a detector is required to discriminate between deviations above and below the mean frequency and to translate those deviations into a voltage whose amplitude varies at audio frequencies. Since the deviations occur at an audio frequency, the process is one of demodulation, and the degree of frequency deviation determines the amplitude of the demodulated (af) voltage.

A simple circuit for converting frequency variations to amplitude variations is a circuit which is tuned so that the mean radio frequency is on one slope of its resonance characteristic, as at A of Fig. 29. With modulation, the

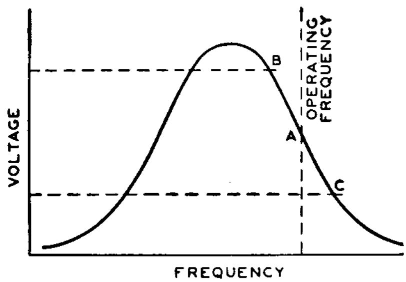


Fig. 29—Resonance curve showing desired operating range for frequency-modulation converter.

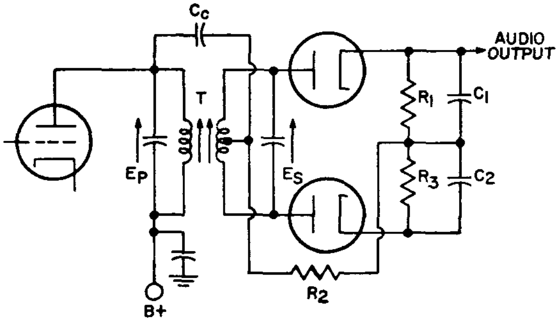


Fig. 30—Balanced phase-shift discriminator circuit.

frequency swings between B and C, and the voltage developed across the circuit varies at the modulating rate. In order that no distortion will be introduced in this circuit, the frequency swing must be restricted to the portion of the slope which is effectively straight. Since this portion is very short, the voltage developed is low. Because of these limitations, this circuit is not commonly used but it serves to illustrate the principle.

The faults of the simple circuit are overcome in a push-pull arrangement, such as that shown in Fig. 30, called a **balanced phase-shift discriminator.** In this detector, the mutually coupled tuned circuits in the primary and secondary windings of the transformer T are tuned to the center frequency. A characteristic of a double-tuned transformer is that the voltages in the primary and secondary windings are 90 degrees out of phase at resonance, and that the phase shift changes as the frequency changes from resonance. Therefore, the signal applied to the diodes and the RC combinations for peak detection also changes with frequency.

Because the secondary winding of the transformer T is center-tapped, the applied primary voltage E_p is added to one-half the secondary voltage E_s through the capacitor C_c. The addition of these voltages at resonance can be represented by the diagram in Fig. 31(a); the resultant voltage E_1 is the signal applied to one peak-detector network consisting of one diode and its RC load. When the signal frequency decreases (from resonance), the phase shift of $E_s/2$ becomes greater than 90 degrees, as shown at (b) in Fig. 31, and E_1 becomes smaller. When the signal frequency increases (above resonance), the phase shift of $E_s/2$ is less than 90 degrees as shown at (c), and E_1 becomes

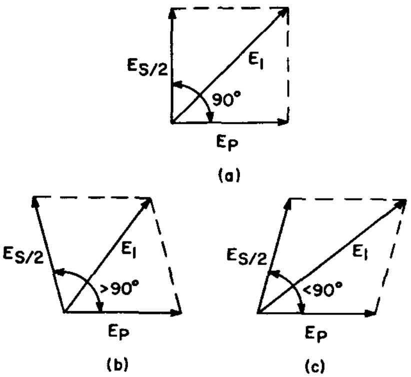


Fig. 31—Diagram illustrating phase shift in double-tuned transformer (a) at resonance, (b) below resonance, and (c) above resonance.

larger. The curve of E_1 as a function of frequency in Fig. 32 is readily identified as the response curve of an FM detector.

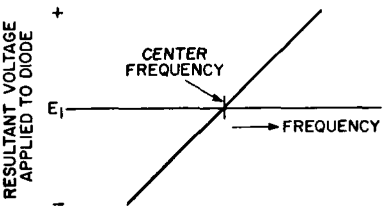


Fig. 32—Diagram showing resultant voltage E_1 in Fig. 31 as a function of frequency.

Because the discriminator circuit shown in Fig. 30 uses a push-pull configuration, the diodes conduct on alternate half-cycles of the signal frequency and produce a plus-and-minus output with respect to zero rather than with respect to E_1. The primary advantage of this arrangement is that there is no output at resonance. When an FM signal is applied to the input, the audio output voltage varies above and below zero as the instantaneous frequency varies above and below resonance. The frequency of this audio voltage is determined by the modulation frequency of the FM signal, and the amplitude of the voltage is proportional to the frequency excursion from resonance. (The resistor R_2 in the circuit provides a dc return for the diodes, and also maintains a load impedance across the primary winding of the transformer.)

One disadvantage of the balanced phase-shift discriminator shown in Fig. 30 is that it detects audio modulation (AM) as well as frequency modulation (FM) in the if signal because the circuit is balanced only at the center frequency. At frequencies off resonance, any variation in amplitude of the if signal is reproduced to some extent in the audio output.

The **ratio-detector** circuit shown in Fig. 33 is a discriminator circuit which has the advantage of being relatively insensitive to amplitude variations in the FM signal. In this circuit, E_p is added to $E_s/2$ through the mutual coupling M_2 (this voltage addition may be made by either mutual or capacitive coupling). Because of the phase-shift relationship of these voltages, the resultant detected signals vary with frequency variations in the same manner as described for the phase-shift discriminator circuit shown in Fig. 30. However, the diodes in the ratio detector are placed "back-to-back" (in series, rather than in push-pull) so that both halves of the circuit operate simultaneously during one-half of the signal frequency cycle (and are cut off on the other half-cycle). As a result, the detected voltages E_1 and E_2 are in series, as shown for the instantaneous polarities that occur during the conduction half-cycle. When the audio output is taken between the equal capacitors C_1 and C_2, therefore, the output voltage is equal to $(E_2 - E_1)/2$ (for equal resistors R_1 and R_2).

The dc circuit of the ratio detector consists of a path through the secondary winding of the transformer, both diodes (which are in series), and resistors R_1 and R_2. The value of the electrolytic capacitor C_3 is selected so that the time constant of R_1, R_2, and C_3 is very long compared to the detected audio signal. As a result, the sum of the detected voltages $(E_1 + E_2)$ is a constant and the AM components on the signal frequency are suppressed. This feature of the ratio detector provides improved AM rejection as compared to the phase-shift discriminator circuit shown in Fig. 30.

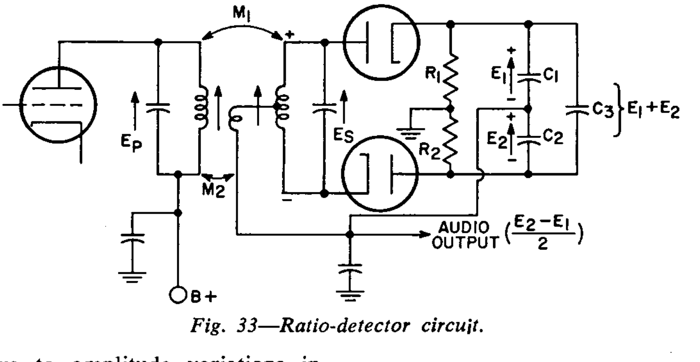


Fig. 33—Ratio-detector circuit.

Amplification

The amplifying action of an electron tube was mentioned under **Triodes** in the section on **Electrons, Electrodes, and Electron Tubes.** This action can be utilized in electronic circuits in a number of ways, depending upon the results desired. Four classes of amplifier service recognized by engineers are covered by definitions standardized by the Institute of Electrical and

Electronics Engineers. This classification depends primarily on the fraction of input cycle during which plate current is expected to flow under rated full-load conditions. The classes are class A, class AB, class B, and class C. The term "cutoff bias" used in these definitions is the value of grid bias at which plate current is very small (i.e., approaches zero).

Classes of Service

A **class A amplifier** is an amplifier in which the grid bias and alternating grid voltages are such that plate current in a specific tube flows at all times.

A **class AB amplifier** is an amplifier in which the grid bias and alternating grid voltages are such that plate current in a specific tube flows for appreciably more than half but less than the entire electrical cycle.

A **class B amplifier** is an amplifier in which the grid bias is approximately equal to the cutoff value, so that the plate current is approximately zero when no exciting grid voltage is applied, and so that plate current in a specific tube flows for approximately one-half of each cycle when an alternating grid voltage is applied.

A **class C amplifier** is an amplifier in which the grid bias is appreciably greater than the cutoff value, so that the plate current in each tube is zero when no alternating grid voltage is applied, and so that plate current flows in a specific tube for appreciably less than one-half of each cycle when an alternating grid voltage is applied.

The suffix 1 may be added to the letter or letters of the class identification to denote that grid current does not flow during any part of the input cycle. The suffix 2 may be used to denote that grid current flows during part of the cycle.

For radio-frequency (rf) amplifiers which operate into a selective tuned circuit, as in radio transmitter applications, or under requirements where distortion is not an important factor, any of the above classes of amplifiers may be used, either with a single tube or with a push-pull stage. For audio-frequency (af) amplifiers in which distortion is an important factor, only class A amplifiers permit single-tube operation. In this case, operating conditions are usually chosen so that distortion is kept below the conventional 5 per cent for triodes and the conventional 7 to 10 per cent for tetrodes or pentodes. Distortion can be reduced below these figures by means of special circuit arrangements such as that discussed under **inverse feedback.** With class A amplifiers, reduced distortion with improved power performance can be obtained by using a push-pull stage for audio service. With class AB and class B amplifiers, a balanced stage using two tubes is required for audio service.

Class A Voltage Amplifiers

As a class A voltage amplifier, an electron tube is used to reproduce grid-voltage variations across an impedance or a resistance in the plate circuit. These variations are essentially of the same form as the input signal voltage impressed on the grid, but their amplitude is increased. This increase is accomplished by operation of the tube at a suitable grid bias so that the applied grid input voltage produces plate-current variations proportional to the signal swings. Because the voltage variation obtained in the plate circuit is much larger than that required to swing the grid, amplification of the signal is obtained.

Fig. 34 gives a graphical illustration of this method of amplication and shows, by means of the grid-voltage vs. plate-current characteristics curve, the effect of an input signal (S) applied to

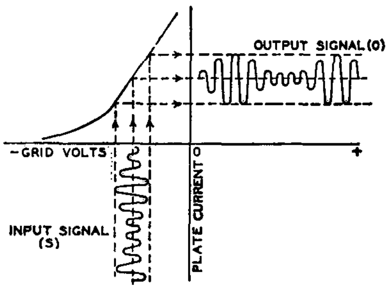


Fig. 34—Current characteristics of class A amplifier.

the grid of a tube. The output signal (O) is the resulting amplified plate-current variation.

The plate current flowing through the load resistance (R) of Fig. 35 causes a voltage drop which varies directly with the plate current. The ratio of this voltage variation produced in the load resistance to the input signal volt-

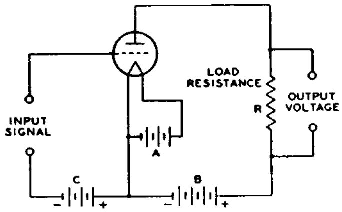


Fig. 35—Triode amplifier circuit.

age is the voltage amplification, or **gain**, provided by the tube. The voltage amplification due to the tube is expressed by the following convenient formulas:

$$\text{Voltage amplification} = \frac{\mu \times R_L}{R_L + r_p}$$

$$\text{or } \frac{g_m \times r_p \times R_L}{1000000 \times (r_p + R_L)}$$

where μ is the amplification factor of the tube, R_L is the load resistance in ohms, r_p is the plate resistance in ohms, and g_m is the transconductance in micromhos.

From the first formula, it can be seen that the gain actually obtainable from the tube is less than the tube amplification factor, but that the gain approaches the amplification factor when the load resistance is large compared to the tube plate resistance. Fig. 36 shows graphically how the gain approaches the amplification factor of the tube as the load resistance is increased. From the curve it can be seen that a high value of load resistance should be used to obtain high gain in a voltage amplifier.

In a **resistance-coupled amplifier,** the load resistance of the tube is approximately equal to the resistance of the plate resistor in parallel with the grid resistor of the following stage. Hence, to obtain a large value of load resistance, it is necessary to use a plate resistor and a grid resistor of large resistance. However, the plate resistor should not be too large because the flow of plate current through the plate resistor produces a voltage drop which reduces the plate voltage applied to the tube. If the plate resistor is too large, this drop will be too large, the plate voltage on the tube will be too small, and the voltage output of the tube will be too small. Also, the grid resistor of the following stage should not be too large, the actual maximum value being dependent on the particular tube type. This precaution is necessary because all tubes contain minute amounts of residual gas which cause a minute flow of current through the grid resistor. If the grid resistor is too large, the positive bias developed by the flow of this current through the resistor decreases the normal negative bias and produces an increase in the plate current. This increased current may overheat the tube and cause liberation of more gas which, in turn, will cause further decrease in bias. The action is cumulative and results in a runaway condition which can destroy the tube.

A higher value of grid resistance is permissible when cathode-resistor bias is used than when fixed bias is used.

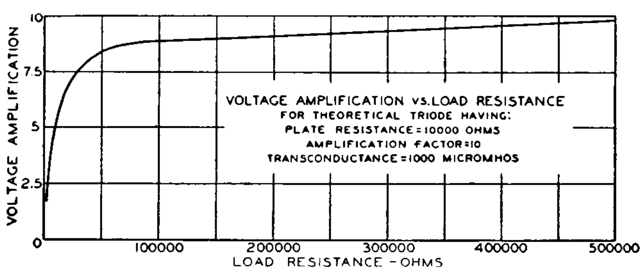


Fig. 36—Gain curve for triode amplifier circuit.

When cathode-resistor bias is used, a loss in bias due to gas or grid-emission effects is almost completely offset by an increase in bias due to the voltage drop across the cathode resistor. Typical values of plate resistor and grid resistor for tube types used in resistance-coupled circuits, and the values of gain obtainable, are shown in the **Resistance-Coupled Amplifier** section.

The **input impedance** of an electron tube (that is, the impedance between grid and cathode) consists of (1) a reactive component due to the capacitance between grid and cathode, (2) a resistive component resulting from the time of transit of electrons between cathode and grid, and (3) a resistive component developed by the part of the cathode lead inductance which is common to both the input and output circuits. These components are dependent on the frequency of the incoming signal. The input impedance is very high at audio frequencies when a tube is operated with its grid biased negative. In a class A_1 or AB_1 transformer-coupled audio amplifier, therefore, the loading imposed by the grid on the input transformer is negligible. As a result, the secondary impedance of a class A_1 or class AB_1 input transformer can be made very high because the choice is not limited by the input impedance of the tube; however, transformer design considerations may limit the choice.

At the higher radio frequencies, the input impedance may become very low even when the grid is negative, due to the finite time of passage of electrons between cathode and grid and to the appreciable lead reactance. This impedance drops very rapidly as the frequency is raised, and increases input-circuit loading. In fact, the input impedance may become low enough at very high radio frequencies to affect the gain and selectivity of a preceding stage appreciably. Tubes such as the "acorn" and "pencil" types and the high-frequency miniatures have been developed to have low input capacitances, low electron-transit time, and low lead inductance so that their input impedance is high even at the ultra-high radio frequencies. **Input admittance** is the reciprocal of input impedance.

A **remote-cutoff amplifier** tube is a modified construction of a pentode or a tetrode type designed to reduce modulation-distortion and cross-modulation in radio-frequency stages. **Cross-modulation** is the effect produced in a radio or television receiver by an interfering station "riding through" on the carrier of the station to which the receiver is tuned. **Modulation-distortion** is a distortion of the modulated carrier and appears as audio-frequency distortion in the output. This effect is produced by a radio-frequency amplifier stage operating on an excessively curved characteristic when the grid bias has been increased to reduce volume. The offending stage for cross-modulation is usually the first radio-frequency amplifier, while for modulation-distortion the cause is usually the last intermediate-frequency stage. The characteristics of remote-cutoff types are such as to enable them to handle both large and small input signals with minimum distortion over a wide range of signal strength.

Fig. 37 illustrates the construction of the grid No. 1 (control grid) in a remote-cutoff tube. The remote-cutoff

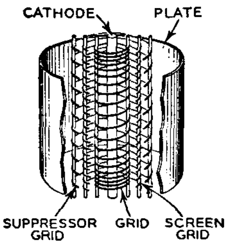


Fig. 37—Structure of remote-cutoff grid.

action is due to the structure of the grid which provides a variation in amplification factor with change in grid bias. The grid No. 1 is wound with open spacing at the middle and with close spacing at the ends. When weak signals and low grid bias are applied to the tube, the effect of the non-uniform turn spacing of the grid on cathode emission and tube characteristics is essentially the same as for uniform spacing. As the grid bias is made more negative to handle larger input signals, the electron flow from the sections of the cathode enclosed by the ends of the grid is cut off. The plate current and other tube characteristics are then dependent on the electron flow through the open section of the grid. This action changes the gain of the tube so that large signals may be handled with minimum distortion due to cross-modulation and modulation-distortion.

Fig. 38 shows a typical plate-current vs. grid-voltage curve for a remote-cutoff type compared with the curve

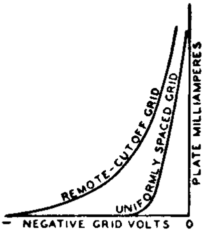


Fig. 38—Plate-current curves for triodes having remote-cutoff and uniformly spaced grids.

for a type having a uniformly spaced grid. It will be noted that while the curves are similar at small grid-bias voltages, the plate current of the remote-cutoff tube drops quite slowly with large values of bias voltage. This slow change makes it possible for the tube to handle large signals satisfactorily. Because remote-cutoff types can accommodate large and small signals, they are particularly suitable for use in sets having automatic volume control. Remote-cutoff tubes also are known as **variable-mu** types.

Class A Power Amplifiers

As a class A power amplifier, an electron tube is used in the output stage of a radio or television receiver to supply a relatively large amount of power to the loudspeaker. For this application, large power output is of more importance than high voltage amplification; therefore, gain possibilities are sacrificed in the design of power tubes to obtain power-handling capability.

Triodes, pentodes, and beam power tubes designed for power amplifier service have certain inherent features for each structure. Power tubes of the triode type for class A service are characterized by low power sensitivity, low plate-power efficiency, and low distortion. Power tubes of the pentode type are characterized by high power sensitivity, high plate-power efficiency and, usually, somewhat higher distortion than class A triodes. Beam power tubes have higher power sensitivity and efficiency than triode or conventional pentode types.

A class A power amplifier is also used as a driver to supply power to a class AB_2 or a class B stage. It is usually advisable to use a triode, rather than a pentode, in a driver stage because of the lower plate impedance of the triode.

Power tubes connected in either **parallel** or **push-pull** may be employed as class A amplifiers to obtain increased output. The parallel connection (Fig. 39) provides twice the output of a single tube with the same value of grid-signal voltage. With this connection, the effective transconductance of the stage is doubled, and the effective plate

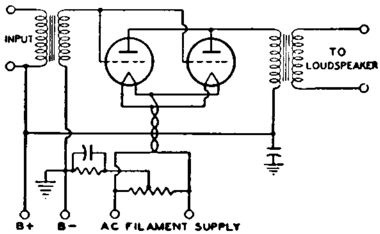


Fig. 39—Power amplifier with tubes connected in parallel.

resistance and the load resistance required are halved as compared with single-tube values.

The push-pull connection (Fig. 40), although it requires twice the grid-signal voltage, provides increased power and has other important advantages over single-tube operation. Distortion caused by even-order harmonics and hum caused by plate-voltage-supply fluctuations are either eliminated or decidedly reduced through cancellation

Because distortion for push-pull operation is less than for single-tube operation, appreciably more than twice single-tube output can be obtained with triodes by decreasing the load resistance for the stage to a value approaching the load resistance for a single tube.

For either parallel or push-pull class A operation of two tubes, all electrode currents are doubled while all dc electrode voltages remain the same as for single-tube operation. If a cathode resistor is used, its value should be about one-half that for a single tube.

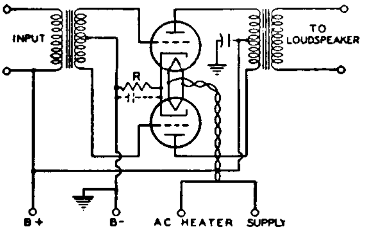


Fig. 40—Power amplifier with tubes connected in push-pull.

If oscillations occur with either type of connection, they can often be eliminated by the use of a non-inductive resistor of approximately 100 ohms connected in series with each grid at the socket terminal.

Operation of power tubes so that the grids run positive is inadvisable except under conditions such as those discussed in this section for class AB and class B amplifiers.

Power-Output Calculations

Calculation of the power output of a triode used as a class A amplifier with either an output transformer or a choke having low dc resistance can be made without serious error from the plate family of curves by assuming a resistance load. The proper plate current, grid bias, optimum load resistance, and per-cent second-harmonic distortion can also be determined. The calculations are made graphically and are illustrated in Fig. 41 for given conditions. The procedure is as follows:

(1) Locate the zero-signal bias point P by determining the zero-signal bias Ec_o from the formula:

$$\text{Zero-signal bias } (Ec_o) = -(0.68 \times E_b)/\mu$$

where E_b is the chosen value in volts of dc plate voltage at which the tube is to be operated, and μ is the amplification factor of the tube. This quantity is shown as negative to indicate that a negative bias is used.

(2) Locate the value of zero-signal plate current, I_o, corresponding to point P.

(3) Locate the point $2I_o$, which is twice the value of I_o and corresponds to the value of the maximum-signal plate current I_{max}.

(4) Locate the point X on the dc bias curve at zero volts, $E_c = 0$, corresponding to the value of I_{max}.

(5) Draw a straight line XY through X and P.

Line XY is known as the load resistance line. Its slope corresponds to the value of the load resistance. The load resistance in ohms is equal to $(E_{max} - E_{min})$ divided by $(I_{max} - I_{min})$, where E is in volts and I is in amperes.

It should be noted that in the case of filament types of tubes, the calculations are given on the basis of a dc-operated filament. When the filament is ac-operated, the calculated value of dc

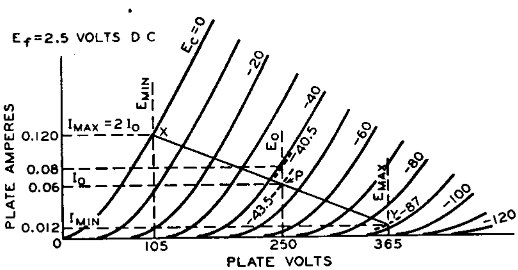


Fig. 41—Graphic calculations for class A amplifier using a power triode.

bias should be increased by approximately one-half the filament voltage rating of the tube.

The value of zero-signal plate current I_o should be used to determine the plate dissipation, an important factor influencing tube life. In a class A amplifier under zero-signal conditions, the plate dissipation is equal to the power input, *i.e.,* the product of the dc plate voltage E_o and the zero-signal dc plate current I_o. If it is found that the plate-dissipation rating of the tube is exceeded with the zero-signal bias Ec_o calculated above, it will be necessary to increase the bias by a sufficient amount so that the actual plate dissipation does not exceed the rating before proceeding further with the remaining calculations.

For power-output calculations, it is assumed that the peak alternating grid voltage is sufficient (1) to swing the grid from the zero-signal bias value Ec_o to zero bias ($E_c = 0$) on the positive swing and (2) to swing the grid to a value twice the zero-signal bias value on the negative swing. During the negative swing, the plate voltage and plate current reach values of E_{max} and I_{min}; during the positive swing, they reach values of E_{min} and I_{max}. Because power is the product of voltage and current, the power output P_o as shown by a watt-meter is given by

$$P_o = \frac{(I_{max} - I_{min}) \times (E_{max} - E_{min})}{8}$$

where E is in volts, I is in amperes, and P_o is in watts.

In the output of power-amplifier triodes, some distortion is present. This distortion is due predominantly to second harmonics in single-tube amplifiers. The percentage of second-harmonic distortion may be calculated by the following formula:

$$\%\ \text{distortion} = \frac{\dfrac{I_{max} + I_{min}}{2} - I_o}{I_{max} - I_{min}} \times 100$$

where I_o is the zero-signal plate current in amperes. If the distortion is excessive, the load resistance should be increased or, occasionally, decreased slightly and the calculations repeated.

Example: Determine the load resistance, power output, and distortion of a triode having an amplification factor of 4.2, a plate-dissipation rating of 15 watts, and plate-characteristics curves as shown in Fig. 41. The tube is to be operated at 250 volts on the plate.

Procedure: For a first approximation, determine the operating point P from the zero-signal bias formula, $Ec_o = -(0.68 \times 250)/4.2 = -40.5$ volts. From the curve for this voltage, it is found that the zero-signal plate current is 0.08 ampere and, therefore, the plate-dissipation rating is exceeded ($0.08 \times 250 = 20$ watts). Consequently, it is necessary to reduce the zero-signal plate current to 0.06 ampere at 250 volts. The grid bias is then −43.5 volts. Note that the curve was taken with a dc filament supply; if the filament is to be operated on an ac supply, the bias must be increased by about one-half the filament voltage, or to −45 volts, and the circuit returns made to the mid-point of the filament circuit.

Point X can then be determined. Point X is at the intersection of the dc bias curve at zero volts with I_{max}, where $I_{max} = 2I_o = 2 \times 0.06 = 0.12$ ampere. Line XY is drawn through points P and X. E_{max}, E_{min}, and I_{min} are then found from the curves. When these values are substituted in the power-output formula, the following result is obtained:

$$P_o = \frac{(0.12 - 0.012) \times (365 - 105)}{8} = 3.52 \text{ watts}$$

The resistance represented by load line XY is

$$\frac{(365 - 105)}{(0.12 - 0.012)} = 2410 \text{ ohms}$$

When the values from the curves are substituted in the distortion formula, the following result is obtained:

$$\%\ \text{distortion} = \frac{\dfrac{0.12 + 0.012}{2} - 0.06}{0.12 - 0.012} \times 100 = 5.5\%$$

It is customary to select the load resistance so that the distortion does not exceed five per cent. When the method shown is used to determine the slope of the load-resistance line, the second-harmonic distortion generally does not exceed five per cent. In the example, however, the distortion is excessive and it is desirable, therefore, to use a slightly higher load resistance. A load resistance of 2500 ohms will provide a distortion of about 4.9 per cent. The power output is reduced only slightly to 3.5 watts.

Operating conditions for **triodes in push-pull** depend on the type of operation desired. Under class A conditions, distortion, power output, and efficiency are all relatively low. The operating bias can be anywhere between that specified for single-tube operation and that equal to one-half the grid-bias voltage required to produce plate-current cutoff at a plate voltage of $1.4E_o$, where E_o is the operating plate voltage. Higher bias than this value requires higher grid-signal voltage and results in class AB_1 operation, which is discussed later.

The method for calculating maximum power output for **triodes in push-pull class A operation** is as follows: Erect a vertical line at 0.6 E_o (see Fig. 42), intersecting the $E_c = 0$ curve at the point I_{max}. Then, I_{max} is determined from the curve for use in the formula

$$P_o = (I_{max} \times E_o)/5$$

If I_{max} is expressed in amperes and E_o in volts, power output is in watts.

Example: Assume that the plate voltage (E_o) is to be 300 volts, and the plate-dissipation rating of the tube is 15 watts. Then, for class A operation, the operating bias can be equal to, but not more than, one-half the grid bias for cutoff with a plate voltage of $1.4 \times 300 = 420$ volts. (Since cutoff bias is approximately -115 volts at a plate voltage of 420 volts, one-half of this value is -57.5 volts bias.) At this bias, the plate current is found from the plate family to be 0.054 ampere and, therefore, the plate dissipation is 0.054×300 or 16.2 watts. Since -57.5 volts is the limit of bias for class A operation of these tubes at a plate voltage of 300 volts, the dissipation cannot be reduced by increasing the bias and it becomes necessary to reduce the plate voltage.

If the plate voltage is reduced to 250 volts, the bias will be found to be -43.5 volts. For this value, the plate current is 0.06 ampere, and the plate dissipation is 15 watts. Then, following the method for calculating power output, erect a vertical line at $0.6E_o = 150$

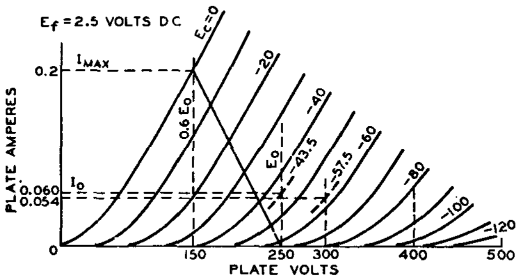


Fig. 42—Graphic calculations for push-pull class A amplifier using a power triode.

volts. The intersection of the line with the curve $E_c = 0$ is I_{max} or 0.2 ampere. When this value is substituted in the power formula, the power output is $(0.2 \times 250)/5 = 10$ watts. The load resistance is determined from the load formula: Plate-to-plate load (R_{pp}) $= 4 \times (250 - 150)/0.2 = 2000$ ohms.

The method for determining the proper load resistance for triodes in push-pull is as follows: Draw a load line through I_{max} on the zero-bias curve and through the E_o point on the zero-current axis. Four times the resistance represented by this load line is the plate-to-plate load (R_{pp}) for two triodes in a class A push-pull amplifier. Expressed as a formula,

$$R_{pp} = 4 \times (E_o - 0.6E_o)/I_{max}$$

where E_o is expressed in volts, I_{max} in amperes, and R_{pp} in ohms.

Power output for a pentode or a beam power tube as a class A amplifier can be calculated in much the same way as for triodes. Calculations can be made graphically from a special plate family of curves, as shown in Fig. 43.

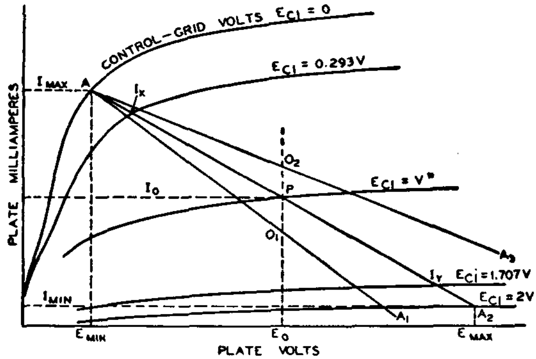


Fig. 43—Graphic calculations for class A amplifier using a pentode or beam power tube.

From a point A at or just below the knee of the zero-bias curve, draw arbitrarily selected load lines to intersect the zero-plate-current axis. These lines should be on both sides of the operating point P, whose position is determined by the desired operating plate voltage, E_o, and one-half the maximum-signal plate current. Along any load line, say AA_1, measure the distance AO_1. On the same line, lay off an equal distance, O_1A_1. For optimum operation, the change in bias from A to O_1 should be nearly equal to the change in bias from O_1 to A_1. If this condition can not be met with one line, as is the case for the line first chosen, then another should be chosen. When the most satisfactory line has been selected, its resistance may be determined by the following formula:

$$\text{Load resistance } (R_L) = \frac{E_{max} - E_{min}}{I_{max} - I_{min}}$$

The value of R_L may then be substituted in the following formula for calculating power output.

$$P_o = \frac{[I_{max} - I_{min} + 1.41\ (I_x - I_y)]^2\ R_L}{32}$$

In both of these formulas, I is in amperes, E is in volts, R_L is in ohms, and P_o is in watts. I_x and I_y are the current values on the load line at bias voltages of $E_{c1} = V - 0.707V = 0.293V$ and $E_{c1} = V + 0.707V = 1.707V$, respectively.

Calculations for distortion mav be made by means of the following formulas. The terms used have already been defined.

% 2nd-harmonic distortion =

$$\frac{I_{max} + I_{min} - 2\ I_o}{I_{max} - I_{min} + 1.41\ (I_x - I_y)} \times 100$$

% 3rd-harmonic distortion =

$$\frac{I_{max} - I_{min} - 1.41\ (I_x - I_y)}{I_{max} - I_{min} + 1.41\ (I_x - I_y)} \times 100$$

% total (2nd and 3rd) harmonic distortion =

$$\sqrt{(\%\ 2nd)^2 + (\%\ 3rd)^2}$$

Conversion Factors

Operating conditions for voltage values other than those shown in the published data can be obtained by use of the **nomograph** shown in Fig. 44 when all electrode voltages are changed simultaneously in the same ratio. The nomograph includes conversion factors for current (F_i), power output (F_p), plate resistance or load resistance (F_r), and transconductance (F_{gm}) for voltage ratios between 0.5 and 2.0. These factors are expressed as functions of the ratio between the desired or new voltage for any electrode (E_{des}) and the published or original value of that voltage (E_{pub}). The relations shown are applicable to triodes and multigrid tubes in all classes of service.

To use the nomograph, simply place a straight-edge across the page so that it intersects the scales for E_{des} and E_{pub} at the desired values. The desired conversion factor may then be read directly or estimated at the point where the straight-edge intersects the F_i, F_p F_r, or F_{gm} scale.

For example, suppose it is desired to operate two 6L6GC's in class A_1 push-pull, fixed bias, with a plate voltage of 200 volts. The nearest published operating conditions for this class of service are for a plate voltage of 250 volts. The operating conditions for the new plate voltage can be determined as follows:

The voltage conversion factor, F_e, is equal to 200/250 or 0.8. The dashed lines on the nomograph of Fig. 44 indicate that for this voltage ratio F_i is approximately 0.72, F_p is approximately 0.57, F_r is 1.12, and F_{gm} is approximately 0.892. These factors may be applied directly to operating values shown in the tube data, or to values calculated by the methods described previously.

Because this method for conversion of characteristics is necessarily an approximation, the accuracy of the nomograph decreases progressively as the ratio E_{des}/E_{pub} departs from unity. In general, results are substantially correct when the value of the ratio E_{des}/E_{pub} is between 0.7 and 1.5. Beyond these limits, the accuracy decreases rapidly, and the results obtained must be considered rough approximations.

The nomograph does not take into consideration the effects of contact potential or secondary emission in tubes. Because contact-potential effects become noticeable only at very small dc grid-No. 1 (bias) voltages, they are generally negligible in power tubes. Secondary emission may occur in conventional tetrodes, however, if the plate voltage swings below the grid-No. 2 voltage. Consequently, the conversion

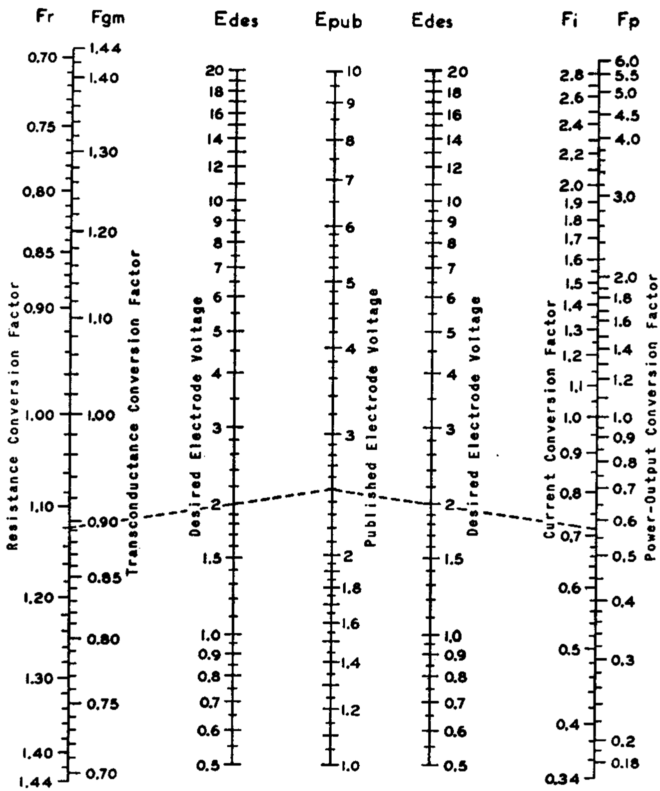


Fig. 44—Nomograph of tube conversion factors.

factors shown in the nomograph apply to such tubes only when the plate voltage is greater than the grid-No. 2 voltage. Because secondary emission may also occur in certain beam power tubes at very low values of plate current and plate voltage, the conversion factors shown in the nomograph do not apply when these tubes are operated under such conditions.

Class AB Power Amplifiers

A class AB power amplifier employs two tubes connected in push-pull with a higher negative grid bias than is used in a class A stage. With this higher negative bias, the plate and screen-grid voltages can usually be made higher than for class A amplifiers because the increased negative bias holds plate current within the limit of the tube plate-dissipation rating. As a result of these higher voltages, more power output can be obtained from class AB operation.

Class AB amplifiers are subdivided into class AB_1 and class AB_2. In class AB_1, there is no flow of grid current. That is, the peak signal voltage applied to each grid is not greater than the negative grid-bias voltage. The grids therefore are not driven to a positive potential and do not draw current. In class AB_2, the peak signal voltage is greater than the bias so that the grids are driven positive and draw current.

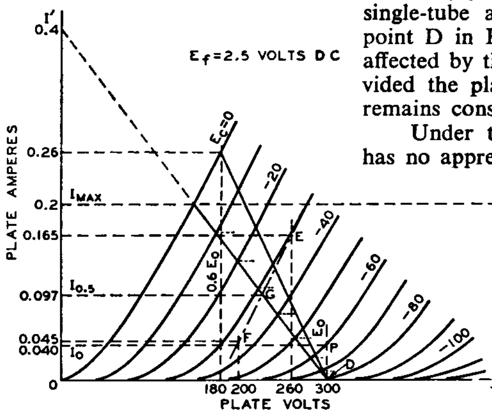
Because of the flow of grid current in a class AB_2 stage, there is a loss of power in the grid circuit. The sum of this loss and the loss in the input transformer is the total driving power required by the grid circuit. The driver stage should be capable of a power output considerably larger than this required power in order that distortion introduced in the grid circuit be kept low. The input transformer used in a class AB_2 amplifier usually has a step-down turns ratio.

Because of the large fluctuations of plate current in a class AB_2 stage, it is important that the plate power supply have good regulation. Otherwise the fluctuations in plate current cause fluctuations in the voltage output of the power supply, with the result that power output is decreased and distortion is increased. To obtain satisfactory regulation, it is usually advisable to use a low-drop rectifier, such as the 5V4GA, with a choke-input filter. In all cases, the resistance of the choke and transformers should be as low as possible

Class AB_1 Power Amplifiers

In class AB_1 push-pull amplifier service using triodes, the operating conditions may be determined graphically by means of the plate family if E_o, the desired operating plate voltage, is given. In this service, the dynamic load line does not pass through the operating point P as in the case of the single-tube amplifier, but through the point D in Fig. 45. Its position is not affected by the operating grid bias provided the plate-to-plate load resistance remains constant.

Under these conditions, grid bias has no appreciable effect on the power



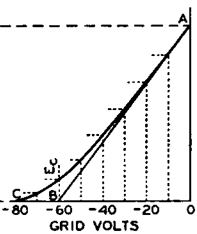


Fig. 45—Graphic calculations for class AB_1 amplifier using a power triode.

Fig. 46—Instantaneous curve for class AB_1 amplifier.

output. Grid bias cannot be neglected, however, since it is used to find the zero-signal plate current and, from it, the zero-signal plate dissipation. Because the grid bias is higher in class AB_1 than in class A service for the same plate voltage, a higher signal voltage may be used without grid current being drawn and, therefore, higher power output is obtained.

In general, for any load line through point D, Fig. 45, the plate-to-plate load resistance in ohms of a push-pull amplifier is $R_{pp} = 4E_o/I'$, where I' *is the plate-current value in amperes at which the load line as projected intersects the plate-current axis, and* E_o is in volts. This formula is another form of the one given under push-pull class A amplifiers, $R_{pp} = 4(E_o - 0.6E_o)/I_{max}$, but is more general. Power output $= (I_{max}/\sqrt{2})^2 \times R_{pp}/4$, where I_{max} is the peak plate current at zero grid volts for the load chosen. This formula *simplified is* $(I_{max})^2 \times R_{pp}/8$. The maximum-signal average plate current is $2I_{max}/\pi$ or 0.636 I_{max}; the maximum-signal average power input is 0.636 $I_{max} \times E_o$.

It is desirable to simplify these formulas for a first approximation. This simplification can be made if it is assumed that the peak plate current, I_{max}, occurs at the point of the zero-bias curve corresponding approximately to 0.6 E_o, the condition for maximum power output. The simplified formulas are:

$$P_o \text{ (for two tubes)} = (I_{max} \times E_o)/5$$
$$R_{pp} = 1.6E_o/I_{max}$$

where E_o *is in volts,* I_{max} *is in amperes,* R_{pp} is in ohms, and P_o is in watts.

It may be found during subsequent calculations that the distortion or the plate dissipation is excessive for this approximation; in that case, a different load resistance must be selected, using the first approximation as a guide, and the process repeated to obtain satisfactory operating conditions.

Example: Fig. 45 illustrates the application of this method to a pair of power triodes operated at $E_o = 300$ volts. Each tube has a plate-dissipation rating of 15 watts. The method is to erect a vertical line at $0.6E_o$, or at 180 volts, which intersects the $E_c = 0$ curve at the point $I_{max} = 0.26$ ampere. Using the simplified formulas, the following values are obtained:

$$R_{pp} = (1.6 \times 300)/0.26 = 1845 \text{ ohms}$$
$$P_o = (0.26 \times 300)/5 = 15.6 \text{ watts}$$

At this point, it is well to determine the plate dissipation and to compare it with the maximum rated value. From the average-plate-current formula (0.636 I_{max}) mentioned previously, the maximum-signal average plate current is 0.166 ampere. The product of this current and the operating plate voltage is 49.8 watts, the average input to the two tubes. From this value, subtract the power output of 15.6 watts to obtain the total dissipation for both tubes, which is 34.2 watts. Half of this value, 17 watts, is in excess of the 15-watt rating of the tube and it is necessary, therefore, to assume another and higher load resistance so that the plate-dissipation rating will not be exceeded.

It will be found that at an operating plate voltage of 300 volts the tubes require a plate-to-plate load resistance of 3000 ohms. From the formula for R_{pp}, the value of I' is found to be 0.4 ampere. The load line for the 3000-ohm load resistance is then represented by a straight line from the point $I' = 0.4$ ampere on the plate-current ordinate to the point $E_o = 300$ volts on the plate-voltage abscissa. At the intersection of the load line with the zero-bias curve, the peak plate current, I_{max}, can be read at 0.2 ampere. Then

$$\begin{aligned} P_o &= (I_{max}/\sqrt{2})^2 \times R_{pp}/4 \\ &= (0.2/1.41)^2 \times 3000/4 \\ &= 15 \text{ watts} \end{aligned}$$

Proceeding as in the first approximation, it is found that the maximum-signal average plate current, $0.636I_{max}$, is 0.127 ampere, and the maximum-signal average power input is 38.1 watts. This input minus the power output is $38.1 - 15 = 23.1$ watts. This value is the dissipation for two tubes; the value per tube is 11.6 watts, a value well within the rating of this tube type.

The operating bias and the zero-signal plate current may then be found by use of a curve which is derived from the plate family and the load line. Fig. 46 is a curve of instantaneous values of plate current and dc grid-bias voltages taken from Fig. 45. Values of grid bias are read from each of the grid-bias curves of Fig. 45 along the load line and are transferred to Fig. 46 to produce the curved line from A to C. A tangent to this curve, starting at A, is drawn to intersect the grid-voltage abscissa. The point of intersection, B, is the operating grid bias for fixed-bias operation. In the example, the bias is −60 volts. Refer back to the plate family at the operating conditions of plate volts = 300 and grid bias = −60 volts; the zero-signal plate current per tube is seen to be 0.04 ampere.

This procedure locates the operating point for each tube at P. The plate current must be doubled, of course, to obtain the zero-signal plate current for both tubes. Under maximum-signal conditions, the signal voltage swings from zero-signal bias voltage to zero bias for each tube on alternate half cycles. Hence, in the example, the peak of signal voltage per tube is 60 volts, or the grid-to-grid value is 120 volts.

As in the case of the push-pull class A amplifier, the second-harmonic distortion in a class AB_1 amplifier using triodes is very small and is largely canceled by virtue of the push-pull connection. Third-harmonic distortion, however, which may be larger than permissible, can be found by means of composite characteristic curves. A complete family of curves can be plotted, but for the present purpose only the one corresponding to a grid bias of one-half the peak grid-voltage swing is needed. In the example, the peak grid voltage per tube is 60 volts, and the half value is 30 volts. The composite curve, since it is nearly a straight line, can be constructed with only two points (see Fig. 45). These two points are obtained from deviations above and below the operating grid and plate voltages.

In order to find the curve for a bias of −30 volts, a deviation of 30 volts from the operating grid voltage of −60 volts is assumed. Next assume a deviation from the operating plate voltage of, say, 40 volts. Then at 300 − 40 = 260 volts, erect a vertical line to intersect the (−60) − (−30) = −30-volt bias curve and read the plate current at this intersection, which is 0.167 ampere; likewise, at the intersection of a vertical line at 300 + 40 = 340 volts and the (−60) + (−30) = −90-volt bias curve, read the plate current. In this example, the plate current is estimated to be 0.002 ampere. The difference of 0.165 ampere between these two currents determines the point E on the 300 − 40 = 260-volt vertical. Similarly, another point F on the same composite curve is found by assuming the same grid-bias deviation but a larger plate-voltage deviation, say, 100 volts.

These steps provide points at 260 volts and 0.165 ampere (E), and at 200 volts and 0.045 ampere (F). A straight line through these points is the composite curve for a bias of −30 volts, shown as a long-short dash line in Fig. 45. At the intersection of the composite curve and the load line, G, the instantaneous composite plate current at the point of one-half the peak signal swing is determined. This current value, designated $I_{0.5}$ and the peak plate current, I_{max}, are used in the following formula to find the peak value of the third-harmonic component of plate current.

$$I_{h3} = (2I_{0.5} - I_{max})/3$$

In the example, where $I_{0.5}$ is 0.097 ampere and I_{max} is 0.2 ampere, $I_{h3} = (2 \times 0.097 - 0.2)/3 = (0.194 - 0.2)/3 = -0.006/3 = -0.002$ ampere. (The fact that I_{h3} is negative indicates that the phase relation of the fundamental (first-harmonic) and third-harmonic components of the plate current is such as to result in a slightly peaked wave form. I_{h3} is positive in some cases, indicating a flattening of the wave form.)

The peak value of the fundamental or first-harmonic component of the plate current is found by the following formula:

$$I_{h1} = 2/3 \times (I_{max} + I_{0.5})$$

In the example, $I_{h1} = 2/3 \times (0.2 + 0.097) = 0.198$ ampere. Thus, the percentage of third-harmonic distortion is $(I_{h3}/I_{h1}) \times 100 = (0.002/0.198) \times 100 = 1$ per cent approx.

Class AB_2 Power Amplifiers

A class AB_2 amplifier employs two tubes connected in push-pull as in the case of class AB_1 amplifiers. It differs in that it is biased so that plate current flows for somewhat more than half the electrical cycle but less than the full cycle, the peak signal voltage is greater than the dc bias voltage, grid current is drawn, and, consequently, power is consumed in the grid circuit. These conditions permit high power output to be obtained without excessive plate dissipation.

The sum of the power used in the grid circuit and the losses in the input transformer is the total driving power required by the grid circuit. The driver stage should be capable of a power output considerably larger than this required power in order that distortion introduced in the grid circuit be kept low. In addition, the internal impedance of the driver stage as reflected into or as effective in the grid circuit of the power stage should always be as low as possible in order that distortion may be kept low. The input transformer used in a class AB_2 stage usually has a step-down ratio adjusted for this condition.

Load resistance, plate dissipation, power output, and distortion determinations are similar to those for class AB_1. These quantities are interdependent with peak grid-voltage swing and driving power; a satisfactory set of operating conditions involves a series of approximations. The load resistance and signal swing are limited by the permissible grid current and power and the distortion. If the load resistance is too high or the signal swing is excessive, the plate-dissipation rating will be exceeded, distortion will be high, and the driving power will be unnecessarily high.

Class B Power Amplifiers

A **class B amplifier** employs two tubes connected in push-pull, so biased that plate current is almost zero when no signal voltage is applied to the grids. Because of this low value of no-signal plate current, class B amplification has the same advantage as class AB_2, *i.e.*, large power output can be obtained without excessive plate dissipation. Class B operation differs from class AB_2 in that plate current is cut off for a larger portion of the negative grid swing, and the signal swing is usually larger than in class AB_2 operation.

Because certain triodes used as class B amplifiers are designed to operate very close to zero bias, the grid of each tube is at a positive potential during all or most of the positive half-cycle of its signal swing. In this type of triode operation, considerable grid current is drawn and there is a loss of power in the grid circuit. This condition imposes the same requirement in the driver stage as in a class AB_2 stage; *i.e.*, the driver should be capable of delivering considerably more power output than the power required for the grid circuit of the class B amplifier so that distortion will be low. Similarly, the interstage transformer between the driver and the class B stage usually has a step-down turns ratio. Because of the high dissipations involved in class B operation at zero bias, it is not feasible to use tetrodes or pentodes in this type of class B operation.

Determination of load resistance, plate dissipation, power output, and distortion is similar to that for a class AB_2 stage.

Power amplifier tubes designed for class A operation can be used in class AB_2 and class B service under suitable operating conditions. There are several tube types designed especially for class B service. The characteristic common to all of these types is a high amplification factor. With a high amplification factor, plate current is small even when the grid bias is zero. These tubes, therefore, can be operated in class B service at a bias of zero volts so that no bias supply is required. A number of class B amplifier tubes consist of two triode units mounted in one tube. The two units can be connected in push-pull so that only one tube is required for a class B stage.

Cathode-Drive Circuits

The preceding text has discussed the use of tubes in the conventional **grid-drive** type of amplifier—that is, where the cathode is common to both the input and output circuits. Tubes may also be employed as amplifiers in circuit arrangements which utilize the grid or plate as the common terminal. Probably the most important of these amplifiers are the cathode-drive circuit, which is discussed below, and the cathode-follower circuit, which will be discussed later in connection with inverse feedback.

A typical **cathode-drive** circuit is shown in Fig. 47. The load is placed in

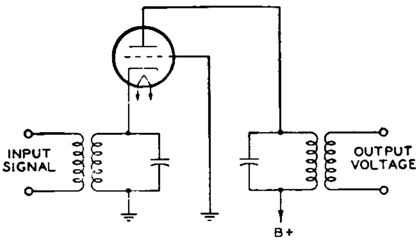


Fig. 47—Cathode-drive circuit.

the plate circuit and the output voltage is taken off between the plate and ground as in the grid-drive method of operation. The grid is grounded, and the input voltage is applied across an appropriate impedance in the cathode circuit. The cathode-drive circuit is particularly useful for vhf and uhf applications, in which it is necessary to obtain the low-noise performance usually associated with a triode, but where a conventional grid-drive circuit would be unstable because of feedback through the grid-to-plate capacitance of the tube. In the cathode-drive circuit, the grounded grid serves as a capacitive shield between plate and cathode and permits stable operation at frequencies higher than those in which conventional circuits can be used.

The input impedance of a cathode-drive circuit is approximately equal to $1/g_m$ when the load resistance is small compared to the r_p of the tube. A certain amount of power is required, therefore, to drive such a circuit. However, in the type of service in which cathode-drive circuits are normally used, the advantages of the grounded-grid connection usually outweigh this disadvantage.

Inverse Feedback

An inverse-feedback circuit, sometimes called a **degenerative** circuit, is one in which a portion of the output voltage of a tube is applied to the input of the same or a preceding tube in opposite phase to the signal applied to the tube. Two important advantages of feedback are (1) reduced distortion from each stage included in the feedback circuit and (2) reduction in the variations in gain due to changes in line voltage, possible differences between tubes of the same type, or variations in the values of circuit constants included in the feedback circuit.

Inverse feedback is used in audio amplifiers to reduce distortion in the output stage where the load impedance on the tube is a loudspeaker. Because the impedance of a loudspeaker is not constant for all audio frequencies, the load impedance on the output tube varies with frequency. When the output tube is a pentode or beam power tube having high plate resistance, this variation in plate load impedance can, if not corrected, produce considerable frequency distortion. Such frequency distortion can be reduced by means of inverse feedback. Inverse-feedback circuits are of the **constant-voltage** type and the **constant-current** type.

The application of the **constant-voltage** type of inverse feedback to a power-output stage using a single beam power tube is illustrated in Fig. 48. In this circuit, R_1, R_2, and C are connected as a voltage divider across the output of the tube. The secondary winding of the grid-input transformer is returned to a point on this voltage divider. Capacitor

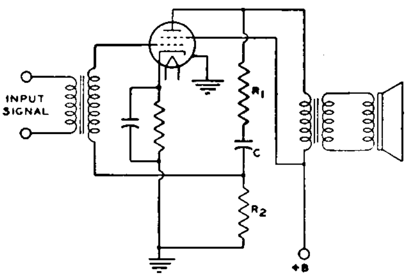


Fig. 48—Power-output stage using constant voltage inverse feedback.

C blocks the dc plate voltage from the grid. However, a portion of the tube af output voltage, approximately equal to the output voltage multiplied by the fraction $R_2/(R_1 + R_2)$, is applied to the grid. This voltage reduces the source impedance of the circuit and a decrease in distortion results which is explained in the curves of Fig. 49.

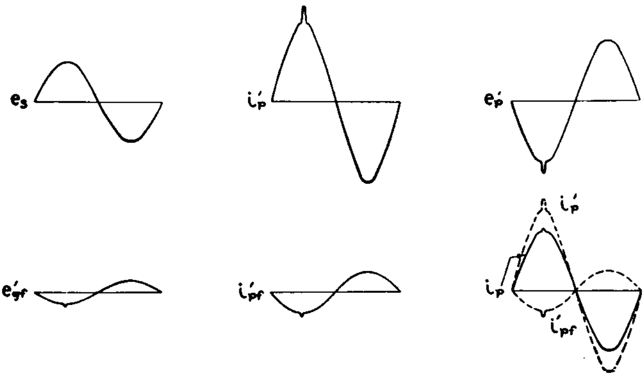


Fig. 49—Voltage and current waveforms showing effect of inverse feedback.

Consider first the amplifier without the use of inverse feedback. Suppose that when a signal voltage e_s is applied to the grid the af plate current i'_p has an irregularity in its positive half-cycle. This irregularity represents a departure from the waveform of the input signal and is, therefore, distortion. For this plate-current waveform, the af plate voltage has a waveform shown by e'_p. The plate-voltage waveform is inverted compared to the plate-current waveform because a plate-current increase produces an increase in the drop across the plate load. The voltage at the plate is the difference between the drop across the load and the supply voltage; thus, when plate current goes up, plate voltage goes down; when plate current goes down, plate voltage goes up.

Now suppose that inverse feedback is applied to the amplifier. The voltage fed back to the grid has the same waveform and phase as the plate voltage, but is smaller in magnitude. Hence, with a plate voltage of waveform shown by e'_p, the feedback voltage appearing on the grid is as shown by e'_{gf}. This voltage applied to the grid produces a component of plate current i'_{pf}. It is evident that the irregularity of the waveform of this component of plate current would act to cancel the original irregularity and thus reduce distortion.

After inverse feedback has been applied, the relations are as shown in the curve for i_p. The dotted curve shown by i'_{pf} is the component of plate current due to the feedback voltage on the grid. The dotted curve shown by i'_p is the component of plate current due to the signal voltage on the grid. The algebraic sum of these two components gives the resultant plate current shown by the solid curve of i_p. Since i'_p is the plate current that would flow without inverse feedback, it can be seen that the application of inverse feedback has reduced the irregularity in the output current. In this manner inverse feedback acts to correct any component of plate current that does not correspond to the input signal voltage, and thus reduces distortion.

From the curve for i_p, it can be seen that, besides reducing distortion, inverse feedback also reduces the amplitude of the output current. Consequently, when inverse feedback is applied to an amplifier there is a decrease in gain or power sensitivity as well as a decrease in distortion. Hence, the application of inverse feedback to an amplifier requires that more driving voltage be applied to obtain full power output, but this output is obtained with less distortion.

Inverse feedback may also be applied to resistance-coupled stages, as shown in Fig. 50. The circuit is conventional except that a feedback resistor,

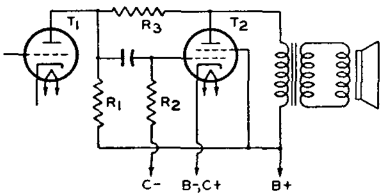


Fig. 50—Resistance-coupled stages using feedback resistor.

R_3, is connected between the plates of tubes T_1 and T_2. The output signal voltage of T_1 and a portion of the output signal voltage of T_2 appear across R_2. Because the distortion generated in the plate circuit of T_2 is applied to its grid out of phase with the input signal, the distortion in the output of T_2 is comparatively low. With sufficient inverse feedback of the constant-voltage type in a power-output stage, it is not necessary to employ a network of resistance and capacitance in the output circuit to reduce response at high audio frequencies. Inverse-feedback circuits can also be applied to push-pull class A and class AB_1 amplifiers.

Constant-current inverse feedback is usually obtained by omitting the by-pass capacitor across a cathode resistor. This method decreases the gain and the distortion but increases the source impedance of the circuit. Consequently, the output voltage rises at the resonant frequency of the loudspeaker and accentuates hangover effects.

Inverse feedback is not generally applied to a triode power amplifier because the variation in speaker impedance with frequency does not produce much distortion in a triode stage having low plate resistance. It is sometimes applied in a pentode stage, but is not always convenient. As has been shown, when inverse feedback is used in an amplifier, the driving voltage must be increased in order to provide full power output. When inverse feedback is used with a pentode, the total driving voltage required for full power output may be inconveniently large, although still less than that required for a triode. Because a beam power tube gives full power output on a comparatively small driving voltage, inverse feedback is especially applicable to beam power tubes. By means of inverse feedback, the high efficiency and high power output of beam power tubes can be combined with freedom from the effects of varying speaker impedance.

Cathode-Follower Circuits

Another important application of inverse feedback is in the cathode-follower circuit, an example of which is shown in Fig. 51. In this application, the load has been transferred from the plate circuit to the cathode circuit of the tube. The input voltage is applied between the grid and ground, and the output voltage is obtained between the cathode and ground. The voltage amplification (V.A.) of this circuit is always less than unity and may be expressed by the following convenient formulas.

For a triode:

$$\text{V. A.} = \frac{\mu \times R_L}{r_p + [R_L \times (\mu + 1)]}$$

For a pentode:

$$\text{V. A.} = \frac{g_m \times R_L}{1 + (g_m \times R_L)}$$

In these formulas, μ is the amplification factor, R_L is the load resistance in ohms, r_p is the plate resistance in ohms, and g_m is the transconductance in mhos.

The use of the cathode follower permits the design of circuits which have high input resistance and high output voltage. The output impedance is

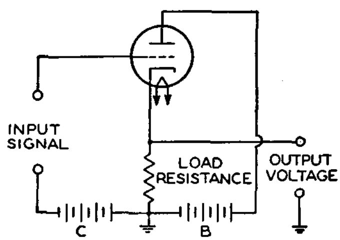


Fig. 51—Cathode-follower circuit.

quite low and very low distortion may be obtained. Cathode-follower circuits may be used for power amplifiers or as impedance transformers designed either to match a transmission line or to produce a relatively high output voltage at a low impedance level.

In a power amplifier which is transformer coupled to the load, the same output power can be obtained from the tube as would be obtained in a conventional grid-drive type of amplifier. The output impedance is very low and provides excellent damping to the load, with the result that very low distortion can be obtained. The peak-to-peak signal voltage, however, approaches 1½ times the plate supply voltage if maximum power output is required from the tube. Some problems may be encountered, therefore, in the design of an adequate driver stage for a cathode-follower output system.

When a cathode-follower circuit is used as an impedance transformer, the load is usually a simple resistance in the cathode circuit of the tube. With relatively low values of cathode resistor, the circuit may be designed to supply significant amounts of power and to match the impedance of the device to a transmission line. With somewhat higher values of cathode resistor, the circuit may be used to decrease the output impedance sufficiently to permit the transmission of audio signals along a line in which appreciable capacitance is present.

The cathode follower may also be used as an isolation device to provide extremely high input resistance and low input capacitance as might be required in the probe of an oscilloscope or vacuum-tube voltmeter. Such circuits can be designed to provide effective impedance transformation with no significant loss of voltage.

Selection of a suitable tube and its operating conditions for use in a cathode-follower circuit having a specified output impedance (Z_o) can be made, in most practical cases, by the use of the following formula to determine the approximate value of the required tube transconductance.

$$\text{Required } g_m\ (\mu\text{mhos}) = \frac{1{,}000{,}000}{Z_o\ (\text{ohms})}$$

Once the required transconductance is obtained, a suitable tube and its operating conditions may be determined from the technical data given in the **Technical Data** section. The tube selected should have a value of transconductance slightly lower than that obtained from the above expression to allow for the shunting effect of the cathode load resistance. The conversion nomograph given in Fig. 44 may be used for calculation of operating conditions for values of transconductance not included in the tabulated data. After the operating conditions have been determined, the approximate value of the required cathode load resistance may be calculated from the following formulas. For a triode:

$$\text{Cathode } R_L = \frac{Z_o \times r_p}{r_p - [Z_o \times (1 + \mu)]}$$

For a pentode:

$$\text{Cathode } R_L = \frac{Z_o}{1 - (g_m \times Z_o)}$$

Resistance and impedance values are in ohms; transconductance values are in mhos.

If the value of the cathode load resistance calculated to provide the required output impedance does not provide the required operating bias, the basic cathode-follower circuit can be modified in a number of ways. Two of the more common modifications are shown in Figs. 52 and 53.

In Fig. 52 the bias is increased by adding a bypassed resistance between

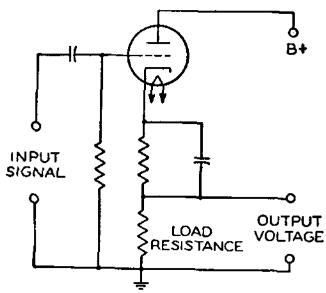


Fig. 52—Cathode-follower circuit modified for increased bias.

the cathode and the unbypassed load resistance and returning the grid to the low end of the load resistance. In Fig. 53 the bias is reduced by adding a bypassed resistance between the cathode and the unbypassed load resistance but, in this case, the grid is returned to the junction of the two cathode resistors so that the bias voltage is only the dc voltage drop across the added resistance The size of the bypass capacitor should be large enough so that it has negligible reactance at the lowest frequency to be handled. In both cases the B-supply should be increased to make up for the voltage taken for biasing.

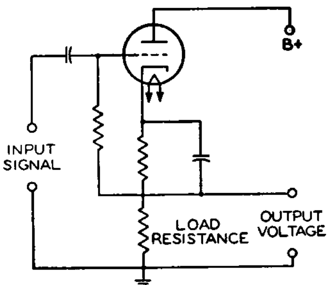


Fig. 53—Cathode-follower circuit modified for reduced bias.

Example: Select a suitable tube and determine the operating conditions and circuit components for a cathode-follower circuit having an output impedance that will match a 500-ohm transmission line.

Procedure: First, determine the approximate transconductance required.

$$\text{Required } g_m = \frac{1{,}000{,}000}{500} = 2000\ \mu\text{mhos}$$

A survey of the tubes that have a transconductance in this order of magnitude shows that type 12AX7A is among the tubes to be considered. Referring to the characteristics given in the technical data section for one triode unit of high-mu twin triode 12AX7, we find that for a plate voltage of 250 volts and a bias of −2 volts, the transconductance is 1600 micromhos, the plate resistance is 62500 ohms, the amplification factor is 100, and the plate current is 0.0012 ampere. When these values are used in the expression for determining the cathode load resistance, the following result is obtained:

$$\text{Cathode } R_L = \frac{500 \times 62500}{62500 - 500 \times (100+1)} = 2600 \text{ ohms}$$

The voltage across this resistor for a plate current of 0.0012 ampere is 2600 × 0.0012 = 3.12 volts. Because the required bias voltage is only −2 volts, the circuit arrangement given in Fig. 53 is employed. The bias is furnished by a resistance that will have a voltage drop of 2 volts when it carries a current of 0.0012 ampere. The required bias resistance, therefore, is 2/0.0012 = 1670 ohms. If 60 Hz is the lowest frequency to be passed, 20 microfarads is a suitable value for the bypass capacitor. The B-supply, of course, is increased by the voltage drop across the cathode resistance which, in this example, is approximately 5 volts. The B-supply, therefore, is 250 + 5 = 255 volts.

Because it is desirable to eliminate, if possible, the bias resistor and bypass capacitor, it is worthwhile to try other tubes and other operating conditions to obtain a value of cathode load resistance which will also provide the required bias. If the triode section of twin diode—high-mu triode 6AT6 is operated under the conditions given in the technical data section with a plate voltage of 100 volts and a bias of −1 volt, it will have an amplification factor of 70, a plate resistance of 54000 ohms, a transconductance of 1300 micromhos, and a plate current of 0.0008 ampere Then,

$$\text{Cathode } R_L = \frac{500 \times 54000}{54000 - 500 \times (70 + 1)} = 1460 \text{ ohms}$$

The bias voltage obtained across this resistance is 1460 × 0.0008 = 1.17 volts. Since this value is for all practical purposes close enough to the required bias, no addition bias resistance will be required and the grid may be returned directly to ground. There is no need to adjust the B-supply voltage to make up for the drop in the cathode resistor. The voltage amplification (V.A.) for the cathode-follower circuit utilizing the triode section of type 6AT6 is

$$\text{V.A.} = \frac{70 \times 1460}{54000 + 1460 \times (70 + 1)} = 0.65$$

For applications in which the cathode follower is used to isolate two circuits—for example, when it is used between a circuit being tested and the input stage of an oscilloscope or a vacuum-tube voltmeter—voltage output and not impedance matching is the primary consideration. In such applications it is desirable to use a relatively high value of cathode load resistance, such as 50,000 ohms, in order to get the maximum voltage output. In order to obtain proper bias, a circuit such as that of Fig. 53 should be used. With a high value of cathode resistance, the voltage amplification will approximate unity.

Corrective Filters

A corrective filter can be used to improve the frequency characteristic of an output stage using a beam power tube or a pentode when inverse feedback is not applicable. The filter consists of a resistor and a capacitor connected in series across the primary of the output transformer. Connected in this way, the filter is in parallel with the plate load impedance reflected from the voicecoil by the output transformer. The magnitude of this reflected impedance increases with increasing frequency in the middle and upper audio range. The impedance of the filter, however, decreases with increasing frequency. It follows that, by use of the proper values for the resistance and the capacitance in the filter, the effective load impedance on the output tubes can be made practically constant for all frequencies in the middle and upper audio range. The result is an improvement in the frequency characteristic of the output stage.

The resistance to be used in the filter for a push-pull stage is 1.3 times the recommended plate-to-plate load resistance; or, for a single-tube stage, is 1.3 times the recommended plate load resistance. The capacitance in the filter should have a value such that the voltage gain of the output stage at a frequency of 1000 Hz or higher is equal to the voltage gain at 400 Hz.

A method of determining the proper value of capacitance for the filter is to make two measurements of the output voltage across the primary of the output transformer: first, when a 400-Hz signal is applied to the input, and second, when a 1000-Hz signal of the same voltage as the 400-Hz signal is applied to the input. The correct value of capacitance is the one which gives equal output voltages for the two signal inputs. In practice, this value is usually found to be in the order of 0.05 microfarad.

Phonograph and Tape Preamplifiers

The frequency range and dynamic range* which can be recorded on a phonograph record or on magnetic tape depend on several factors, including the composition, mechanical characteristics, and speed of the record or tape, and the electrical and mechanical characteristics of the recording equipment. To achieve wide frequency and dynamic ranges, manufacturers of commercial recordings use equipment which introduces a nonuniform relationship between amplitude and frequency. This relationship is known as a "recording characteristic." To assure proper reproduction of a high-fidelity recording, therefore, some part of the reproducing system must have a frequency-response characteristic which is the inverse of the recording characteristic. Most manufacturers of high-fidelity recordings use the RCA "New Orthophonic" (RIAA) characteristic for discs and the NARTB characteristic for magnetic tape.

The simplest type of equalization network is shown in Fig. 54. Because the capacitor C is effectively an open circuit at low frequencies, the low frequencies must be passed through the resistor R and are attenuated. The capacitor has a lower reactance at high

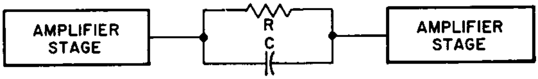


Fig. 54—Simple RC frequency-compensation network.

* The dynamic range of an amplifier is a measure of its signal-handling capability. The dynamic range expresses in dB the ratio of the maximum usable output signal (generally for a distortion of about 10 per cent) to the minimum usable output signal (generally for a signal-to-noise ratio of about 20 dB). A dynamic range of 40 dB is usually acceptable; a value of 70 dB is exceptional for any audio system.

frequencies, however, and bypasses high-frequency components around R so that they receive negligible attenuation. Thus the network effectively "boosts" the high frequencies. This type of equalization is called "attenuative."

Some typical preamplifier stages are shown in the **Circuits** section. The location of the frequency-compensating network or "equalizer" in the reproducing system will depend on the types of recordings which are to be reproduced and on the pickup devices used.

A ceramic high-fidelity phonograph pickup is usually designed to provide proper compensation for the RIAA recording characteristic when the pickup is operated into the load resistance specified by its manufacturer. Because this type of pickup also has relatively high output (0.5 to 1.5 volts), it does not require the use of either an equalizer network or a preamplifier, and can be connected directly to the input of a tone-control amplifier and/or power amplifier.

A magnetic high-fidelity phonograph pickup, on the other hand, usually has an essentially flat frequency-response characteristic and very low output (1 to 10 millivolts). Because a pickup of this type merely reproduces the recording characteristic, it must be followed by an equalizer network, as well as by a preamplifier having sufficient voltage gain to provide the input voltage required by the tone-control amplifier and/or power amplifier. Many designs include both the equalizing and amplifying circuits in a single unit.

A high-fidelity magnetic-tape pickup head, like a magnetic phonograph pickup, reproduces the recording characteristic and has an output of only a few millivolts. This type of pickup device, therefore, must also be followed by an equalizing network and preamplifier, or by a preamplifier which provides "built-in" equalization for the NARTB characteristic.

Feedback networks may also be used for frequency compensation and for reduction of distortion. Basically, a feedback network returns a portion of the output signal to the input circuit of an amplifier. The feedback signal may be returned in phase with the input signal (**positive** or **regenerative** feedback) or 180 degrees out of phase with the input signal (**negative, inverse,** or **degenerative** feedback). In either case, the feedback can be made proportional to either the output voltage or the output current, and can be applied to either the input voltage or the input current. A negative feedback signal proportional to the output current raises the output impedance of the amplifier; negative feedback proportional to the output voltage reduces the output impedance. A negative feedback signal applied to the input current decreases the input impedance; negative feedback applied to the input voltage increases the input impedance. Opposite effects are produced by positive feedback.

A simple negative or inverse feedback network which provides high-frequency boost is shown in Fig. 55. This network provides equalization comparable to that obtained with Fig. 54, but is more suitable for low-level amplifier stages because it does not require the first amplifier stage to provide high-level low frequencies. In addition, the inverse feedback improves the distortion characteristics of the amplifier.

Some preamplifier or low-level audio amplifier circuits include variable resistors or potentiometers which function as **volume** or **tone controls.** Such circuits should be designed to minimize the flow of dc currents through these controls so that little or no noise will be developed by the movable contact during the life of the circuit. Volume controls and their associated circuits should permit variation of gain from zero to maximum, and should attenuate all frequencies equally for all positions

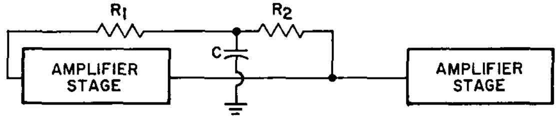


Fig. 55—Negative-feedback frequency-compensation network.

of the variable arm of the control. Several examples of volume controls and tone controls are shown in the **Circuits** section.

Tone Controls

A tone control is a variable filter (or one in which at least one element is adjustable) by means of which the user may vary the frequency response of an amplifier to suit his own taste. In radio receivers and home amplifiers, the tone control usually consists of a resistance-capacitance network in which the resistance is the variable element.

The simplest form of tone control is a fixed tone-compensating or "equalizing" network such as that shown in Fig. 56. This type of network is often

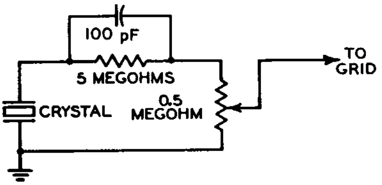


Fig. 56—Tone-control circuit for fixed tone compensation or "equalizing".

used to equalize the low- and high-frequency response of a crystal phonograph pickup. At low frequencies the attenuation of this network is 20.8 dB. As the frequency is increased, the 100-picofarad capacitor serves as a bypass for the 5-megohm resistor, and the combined impedance of the resistor-capacitor network is reduced. Thus, more of the crystal output appears across the 0.5-megohm resistor at high frequencies than at low frequencies, and the frequency response at the grid is reasonably flat over a wide frequency range. Fig. 57 shows a comparison between the output of the crystal (curve A) and the output of the equalizing network (curve B). The response curve

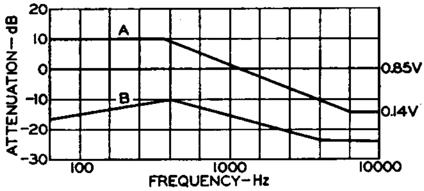


Fig. 57—Curve showing output from crystal phonograph pickup (A) and from equalizing network (B).

can be "flattened" still more if the attenuation at low frequencies is increased by changing the 0.5-megohm resistor to 0.125 megohm.

The tone-control network shown in Fig. 58 has two stages with completely separate bass and treble controls. Fig. 59 shows simplified representations of the bass control of this circuit when the potentiometer is turned to its extreme variations (usually labeled "Boost" and "Cut"). In this network, as in the crystal-equalizing network shown in Fig. 56, the parallel RC combination is the controlling factor. For bass "boost," the capacitor C_2 bypasses resistor R_3 so that less impedance is placed across the output to grid B at high frequencies than at low frequencies. For bass "cut," the parallel combination is shifted so that C_1 bypasses R_3, causing more high-frequency than low-frequency output. Essentially, the network is a variable-

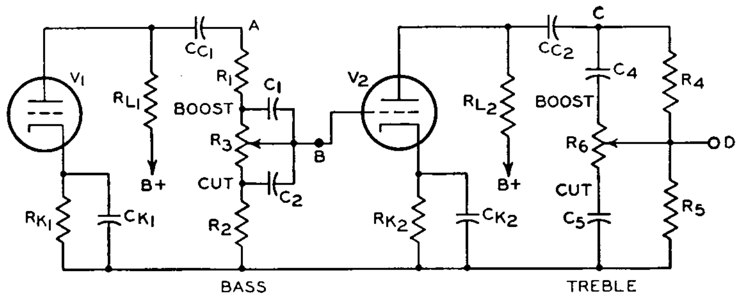


Fig. 58—Two-stage tone-control circuit incorporating separate bass and treble controls.

frequency voltage divider. With proper values for the components, it may be made to respond to changes in the R_3 potentiometer setting for only low frequencies (below 1000 Hz).

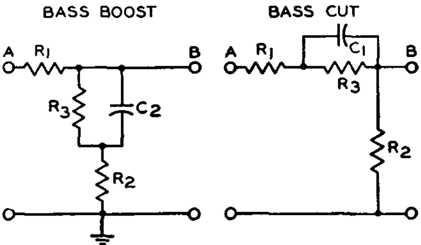


Fig. 59—Simplified representations of bass-control circuit at extreme ends of potentiometer.

Fig. 60 shows extreme positions of the treble control. The attenuation of the two circuits is approximately the same at 1000 Hz. The treble "boost" circuit is similar to the crystal-equalizing network shown in Fig. 56. In the treble "cut" circuit, the parallel RC elements serve to attenuate the signal voltage further because the capacitor bypasses the resistance across the output.

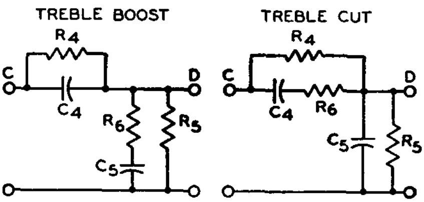


Fig. 60—Simplified representations of treble-control circuit at extreme ends of potentiometer.

The effect of the capacitor is negligible at low frequencies; beyond 1000 Hz, the signal voltage is attenuated at a maximum rate of 6 dB per octave.

The location of a tone-control network is of considerable importance. In a typical radio receiver, it may be inserted in the plate circuit of the power tube, the coupling circuit between the first af amplifier tube and the power tube, or the grid circuit of the first tube. In an amplifier using a beam power tube or pentode power amplifier without negative feedback, it is desirable to connect a resistance-capacitance filter across the primary of the output transformer. This filter may be fixed, with a supplementary tone control elsewhere, or it may form the tone control itself. If the amplifier incorporates negative feedback, the tone control may be inserted in the feedback network or else should be connected to a part of the amplifier which is external to the feedback loop. The overall gain of a well designed tone-control network should be approximately unity.

Automatic Volume or Gain Control

The chief purpose of automatic volume control **(avc)** or automatic gain control **(agc)** in a radio or television receiver is to prevent fluctuations in loudspeaker volume or picture brightness when the audio or video signal at the antenna is fading in and out.

An **automatic volume control** circuit regulates the receiver rf and if gain so that this gain is less for a strong signal than for a weak signal. In this way, when the signal strength at the antenna changes, the avc circuit reduces the resultant change in the voltage output of the last if stage and consequently reduces the change in the speaker output volume.

The avc circuit reduces the rf and if gain for a strong signal usually by increasing the negative bias of the rf, if, and frequency-mixer stage when the signal increases. A simple avc circuit is shown in Fig. 61. On each positive half-cycle of the signal voltage, when the diode plate is positive with respect to the cathode, the diode passes current.

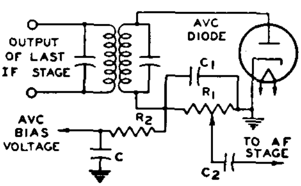


Fig. 61—Automatic-volume-control (avc) circuit.

Because of the flow of diode current through R_1, there is a voltage drop across R_1 which makes the left end of R_1 negative with respect to ground. This voltage drop across R_1 is applied, through the filter R_2 and C, as negative bias on the grids of the preceding stages. When the signal strength at the antenna increases, therefore, the signal applied to the avc diode increases, the voltage drop across R_1 increases, the negative bias voltage applied to the rf and if stages increases, and the gain of the rf and if stages is decreased. Thus the increase in signal strength at the antenna does not produce as much increase in the output of the last if stage as it would produce without avc.

When the signal strength at the antenna decreases from a previous steady value, the avc circuit acts, of course, in the reverse direction, applying less negative bias, permitting the rf and if gain to increase, and thus reducing the decrease in the signal output of the last if stage. In this way, when the signal strength at the antenna changes, the avc circuit acts to reduce change in the output of the last if stage, and thus acts to reduce change in loudspeaker volume.

The filter, C and R_2, prevents the avc voltage from varying at audio frequency. The filter is necessary because the voltage drop across R_1 varies with the modulation of the carrier being received. If avc voltage were taken directly from R_1 without filtering, the audio variations in avc voltage would vary the receiver gain so as to smooth out the modulation of the carrier. To avoid this effect, the avc voltage is taken from the capacitor C. Because of the resistance R_2 in series with C, the capacitor C can charge and discharge at only a comparatively slow rate. The avc voltage therefore cannot vary at frequencies as high as the audio range but can vary at frequencies high enough to compensate for most fading. Thus the filter permits the avc circuit to smooth out variations in signal due to fading, but prevents the circuit from smoothing out audio modulation.

It will be seen that an avc circuit and a diode-detector circuit are much alike. It is therefore convenient in a receiver to combine the detector and the avc diode in a single stage. Examples of how these functions are combined in receivers are shown in **Circuits** section.

In the circuit shown in Fig. 61, a certain amount of avc negative bias is applied to the preceding stages on a weak signal. Because it may be desirable to maintain the receiver rf and if gain at the maximum possible value for a weak signal, avc circuits are designed in some cases to apply no avc bias until the signal strength exceeds a certain value. These avc circuits are known as **delayed avc** or **davc** circuits.

A davc circuit is shown in Fig. 62. In this circuit, the diode section D_1 of the 6AL5 acts as detector and avc diode.

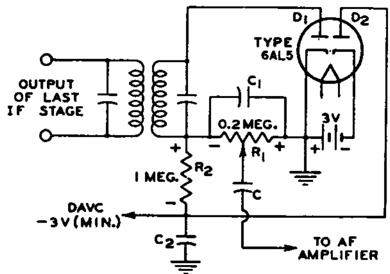


Fig. 62—Delayed avc (davc) circuit.

R_1 is the diode load resistor and R_2 and C_2 are the avc filter. Because the cathode of diode D_2 is returned through a fixed supply of −3 volts to the cathode of D_1, a dc current flows through R_1 and R_2 in series with D_2. The voltage drop caused by this current places the avc lead at approximately −3 volts (less the negligible drop through D_2). When the average amplitude of the rectified signal developed across R_1 does not exceed 3 volts, the avc lead remains at −3 volts. Hence, for signals not strong enough to develop 3 volts across R_1, the bias applied to the controlled tubes stays constant at a value giving high sensitivity.

However, when the average amplitude of rectified signal voltage across R_1 exceeds 3 volts, the plate of diode D_2 becomes more negative than the cathode of D_2 and current flow in diode D_2 ceases. The potential of the avc lead is then controlled by the voltage developed across R_1. Therefore, with further increase in signal strength, the avc circuit applies an increasing avc bias voltage to the controlled stages. In this way, the circuit regulates the receiver gain for strong signals, but permits the gain to stay constant at a maximum value for weak signals.

It can be seen in Fig. 62 that a portion of the —3 volts delay voltage is applied to the plate of the detector diode D_1, this portion being approximately equal to $R_1/(R_1 + R_2)$ times —3 volts. Hence, with the circuit constants as shown, the detector plate is made negative with respect to its cathode by approximately one-half volt. However, this voltage does not interfere with detection because it is not large enough to prevent current flow in the tube.

Automatic gain control (agc) compensates for fluctuations in rf picture carrier amplitude. The peak carrier level rather than the average carrier level is controlled by the agc voltage because the peaks of the sync pulses are fixed when inserted on a fixed carrier level. The peak carrier level may be determined by measurement of the peaks of the sync pulses at the output of the video detector.

A conventional agc circuit, such as that shown in Fig. 63, consists of a diode

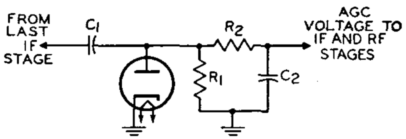


Fig. 63—Automatic-gain control (agc) circuit.

detector circuit and an RC filter. The time constant of the detector circuit is made large enough to prevent the picture content from influencing the magnitude of the agc voltage. The output voltage (agc voltage) is equal to the peak value of the incoming signal.

The diode detector receives the incoming signal from the last if stage of the television receiver through the capacitor C_1. The resistor R_1 provides the load for the diode. The diode conducts only when its plate is driven positive with respect to its cathode. Electrons then flow from the cathode to the plate and thence into capacitor C_1, where the negative charge is stored. Because of the low impedance offered by the diode during conduction, C_1 charges up to the value of the peak applied voltage.

During the negative excursion of the signal, the diode does not conduct, and C_1 discharges through resistor R_1. Because of the large time constant of R_1C_1, however, only a small percentage of the voltage across C_1 is lost during the interval between horizontal sync pulses. During succeeding positive cycles, the incoming signal must overcome the negative charge stored in C_1 before the diode conducts, and plate current flows only at the peak of each positive cycle. The voltage across C_1, therefore, is determined by the level of the peaks of the positive cycles, or the sync pulses.

The negative voltage developed across resistor R_1 by the sync pulses is filtered by resistor R_2 and capacitor C_2 to remove the 15,750-cycle ripple of the horizontal sync pulse. The dc output is then fed to the if and rf amplifiers as an agc voltage.

This agc system may be expanded to include amplification of the agc signal before detection of the peak level, or amplification of the dc output, or both. A direct-coupled amplifier must be used for amplification of the dc signal. The addition of amplification makes the system more sensitive to changes in carrier level.

A "keyed" agc system such as that shown in Fig. 64 is used to eliminate flutter and to improve noise immunity in weak signal areas. This system provides more rapid action than the conventional agc circuits because the filter circuit can employ lower capacitance and resistance values.

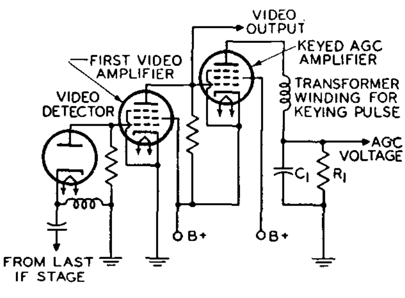


Fig. 64—"Keyed" agc circuit.

In the keyed agc system, the negative output of the video detector is fed directly to the grid No. 1 of the first video amplifier. The positive output of the video amplifier is, in turn, fed directly to the grid No. 1 of the keyed agc amplifier. The video stage increases the gain of the agc system and, in addition, provides noise clipping. The plate voltage for the agc amplifier is a positive pulse obtained from a small winding on the horizontal output transformer which is in phase with the horizontal sync pulse obtained from the video amplifier. The polarity of this pulse is such that the plate of the agc amplifier tube is positive during the retrace time. The tube is biased so that current flows only when the grid No. 1 and the plate are driven positive simultaneously. The amount of current flow depends on the grid-No. 1 potential during the pulse. These pulses are smoothed out in the RC network in the plate circuit (R_1C_1). Because the dc voltage developed across R_1 is negative, it is suitable for application to the grids of the rf and if tubes as an agc voltage.

High-Fidelity Amplifiers

Several high-fidelity amplifiers are shown in the **Circuits** section. The performance capabilities of such amplifiers are usually given in terms of frequency response, total harmonic distortion, maximum power output, and noise level.

To provide high-fidelity reproduction of audio program material, an amplifier should have a frequency response which does not vary more than 1 dB over the entire audio spectrum. General practice is to design the amplifier so that its frequency response is flat within 1 dB from a frequency below the lowest to be reproduced to one well above the upper limit of the audible region.

Harmonic distortion and intermodulation distortion produce changes in program material which may have adverse effects on the quality of the reproduced sound. **Harmonic distortion** causes a change in the character of an individual tone by the introduction of harmonics which were not originally present in the program material. For high-fidelity reproduction, total harmonic distortion (expressed as a percentage of the output power) should not be greater than about 1 per cent at the desired listening level. Types such as the 6973, 7027A and 7868 are designed to provide extremely low harmonic distortion in suitably designed push-pull amplifier circuits.

Intermodulation distortion is a change in the waveform of an individual tone as a result of interaction with another tone present at the same time in the program material. This type of distortion not only alters the character of the modulated tone, but may also result in the generation of spurious signals at frequencies equal to the sum and difference of the interacting frequencies. Intermodulation distortion should be less than 2 per cent at the desired listening level. In general, any amplifier which has low intermodulation distortion will have very low harmonic distortion.

The maximum power output which a high-fidelity amplifier should deliver depends upon a complex relation of several factors, including the size and acoustical characteristics of the listening area, the desired listening level, and the efficiency of the loudspeaker system. Practically, however, it is possible to determine amplifier requirements in terms of room size and loudspeaker efficiency.

The acoustic power required to reproduce the loudest passages of orchestral music at concert-hall level in the average-size living room is about 0.4 watt. Because high-fidelity loudspeakers of the type generally available for home use have an efficiency of only about 5 per cent, the output stage of the amplifier should therefore be able to deliver a power output of at least 8 watts. Because many wide-range loudspeaker systems, particularly those using frequency-divider networks, have efficiencies of less than 5 per cent, output tubes used with such systems must have correspondingly larger power outputs. The 6973, 7027A, 7189, and 7868 can provide ample output for most systems when used in suitable push-pull circuits.

The noise level of a high-fidelity

amplifier determines the range of volume the amplifier is able to reproduce, *i.e.*, the difference (usually expressed in decibels) between the loudest and softest sounds in program material. Because the greatest volume range utilized in electrical program material at the present time is about 60 dB, the noise level of a high-fidelity amplifier should be at least 60 dB below the signal level at the desired listening level.

Limiters

An amplifier may also be used as a limiter. One use of a limiter is in receivers designed for the reception of frequency-modulated signals. The limiter in FM receivers has the function of eliminating amplitude variations from the input to the detector. Because in an FM system amplitude variations are primarily the result of noise disturbances, the use of a limiter prevents such disturbances from being reproduced in the audio output. The limiter usually follows the last if stage so that it can minimize the effects of disturbances coming in on the rf carrier and those produced locally.

The limiter is essentially an if voltage amplifier designed for saturated operation. Saturated operation means that an increase in signal voltage above a certain value produces very little increase in plate current. A signal voltage which is never less than sufficient to cause saturation of the limiter, even on weak signals, is supplied to the limiter input by the preceding stages. Any change in amplitude, therefore, such as might be produced by noise voltage fluctuation, is not reproduced in the limiter output. The limiting action, of course, does not interfere with the reproduction of frequency variations.

Plate-current saturation of the limiter may be obtained by the use of grid-No. 1 resistor-and-capacitor bias with plate and grid-No. 2 voltages which are low compared with customary if-amplifier operating conditions.

As a result of these design features, the limiter is able to maintain its output voltage at a constant amplitude over a wide range of input-signal voltage variations. The output of the limiter is frequency-modulated if voltage, the mean frequency of which is that of the if amplifier. This voltage is impressed on the input of the detector.

The reception of FM signals without serious distortion requires that the response of the receiver be such that satisfactory amplification of the signal is provided over the entire range of frequency deviation from the mean frequency. Since the frequency at any instant depends on the modulation at that instant, it follows that excessive attenuation toward the edges of the band, in the rf or if stages, will cause distortion. In a high-fidelity receiver, therefore, the amplifiers must be capable of amplifying, for the maximum permissible frequency deviation of 75 kHz, a band 150 kHz wide. Suitable tubes for this purpose are the 6BA6 and 6BJ6.

Volume Compressors and Expanders

Volume compression and expansion are used in FM transmitters and receivers and in recording devices and amplifiers to make more natural the reproduction of music which has a very large volume range. For example, in the music of a symphony orchestra the sound intensity of the soft passages is very much lower than that of the loud passages. When this low volume level is raised above the background noise for transmitting or recording, the peak level of the program material may be raised to an excessively high volume level. It is often necessary, therefore, to compress the volume range of the program content within the maximum capabilities of the FM transmitter or the recording device. Exceeding a maximum peak volume level for FM modulation corresponds to exceeding the allowed bandwidth for transmission. In some recording devices, excessive peak volume levels may cause overloading and distortion.

Volume compression may be accomplished by either manual or automatic control. The types of compression used include peak limiters, volume limiters, and volume compressors. A peak limiter limits the peak power to some predetermined level. A volume limiter provides gain reduction based on an average signal level above a predetermined level. A volume compressor provides gain reduction for only the sustained loud portions of the sound level. Only volume compressors can be correctly compensated for with volume expanders.

For faithful reproduction of the original sound, the volume expander used in the FM receiver or audio amplifier should have the reverse characteristic of the volume compressor used in the FM transmitter or recording device. In general, the basic requirements for either a volume compressor or expander are shown in the block diagram of Fig. 65. In a volume compressor, the

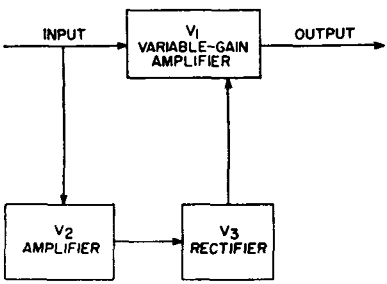


Fig. 65—Block diagram of volume compressor or expander circuit.

variable-gain amplifier V_1 has greater gain for a low-amplitude signal than for a high-amplitude signal; therefore, soft passages are amplified more than loud ones. In an expander, the gain is greater for high-amplitude signals than for low-amplitude signals; therefore, loud passages are amplified more than soft ones and the original amplitude ratio is restored.

In the diagram shown in Fig. 65, the signal to be amplified is applied to V_1, and a portion of the signal is also applied to V_2. The amplified output from V_2 is then rectified by V_3, and applied as a negative (for compressors) or positive (for expanders) bias voltage to V_1. As this bias voltage varies with variations in signal amplitude, the gain of V_1 also varies to produce the desired compression or expansion of the signal.

Tubes having a large dynamic range provide the best results in volume compressor or expander applications. Examples of such types are the 6BJ6 and 6BE6. Push-pull operation is generally desired for the variable-gain amplifier to prevent high distortion and other undesirable effects which may occur in volume compressors and expanders.

Phase Inverters

A phase inverter is a circuit used to provide resistance coupling between the output of a single-tube stage and the input of a push-pull stage. The necessity for a phase inverter arises because the signal-voltage inputs to the grids of a push-pull stage must be 180 degrees out of phase and approximately equal in amplitude with respect to each other. Thus, when the signal voltage input to a push-pull stage swings the grid of one tube in a positive direction, it should swing the grid of the other tube in a negative direction by a similar amount. With transformer coupling between stages, the out-of-phase input voltage to the push-pull stage is supplied by means of the center-tapped secondary. With resistance coupling, the out-of-phase input voltage is obtained by means of the inverter action of a tube.

Fig. 66 shows a push-pull power amplifier, resistance-coupled by means of a phase-inverter circuit to a single-stage triode T_1. Phase inversion in this circuit is provided by triode T_2. The output voltage of T_1 is applied to the grid No. 1 of tetrode T_3. A portion of the output voltage of T_1 is also applied through the resistors R_3 and R_5 to the

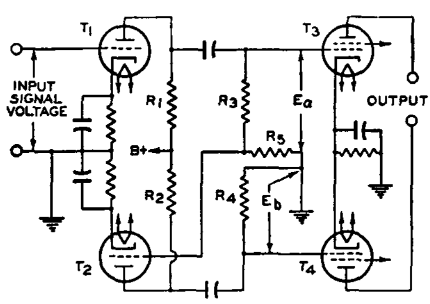


Fig. 66—Push-pull power amplifier resistance-coupled to triode by means of phase inverter.

grid of T_3. The output voltage of T_3 is applied to the grid No. 1 of tetrode T_4.

When the output voltage of T_1 swings in the positive direction, the plate current of T_2 increases. This action increases the voltage drop across the plate resistor R_4 and swings the plate of T_2 in the negative direction. Thus, when the output voltage of T_1 swings positive, the output voltage of T_2 swings negative and is, therefore, 180 degrees out of phase with the output voltage of T_1.

In order to obtain equal voltages at E_a and E_b, $(R_3 + R_5)/R_5$ should equal the voltage gain of T_2. Under the condition where a twin-type tube or two tubes having the same characteristics are used as T_1 and T_2, R_4 should be equal to the sum of R_3 and R_5. The ratio of $R_3 + R_5$ to R_5 should be the same as the voltage gain ratio of T_2 in order to apply the correct value of signal voltage to T_2. The value of R_5 is, therefore, equal to R_4 divided by the voltage gain of T_2; R_3 is equal to R_4 minus R_5. Values of R_1, R_2, R_3 plus R_5, and R_4 may be taken from the chart in the **Resistance-Coupled Amplifiers** section. In the practical application of this circuit, it is convenient to use a twin-triode tube combining T_1 and T_2.

Tuned Amplifiers

In radio-frequency **(rf)** and intermediate-frequency **(if) amplifiers,** the bandwidth of frequencies to be amplified is usually only a small percentage of the center frequency. Tuned amplifiers are used in these applications to select the desired bandwidth of frequencies and to suppress unwanted frequencies. The selectivity of the amplifier is obtained by means of tuned interstage coupling networks.

The properties of tuned amplifiers depend upon the characteristics of **resonant circuits.** A simple parallel resonant circuit (sometimes called a "tank" because it stores energy) is shown in Fig. 67. For practical purposes the resonant frequency of such a circuit may be considered independent of the resistance R, provided R is small compared to the inductive reactance X_L. The resonant frequency f_r is then given by

$$f_r = \frac{1}{2\pi\sqrt{LC}}$$

For any given resonant frequency, the product of L and C is a constant; at low frequencies LC is large; at high frequencies it is small.

The **Q (selectivity)** of a parallel resonant circuit alone is the ratio of the current in the tank (I_L or I_C) to the current in the line (I). This unloaded Q, or Q_o, may be expressed in various ways, for example:

$$Q_L = \frac{I_C}{I} = \frac{X_L}{R} = \frac{R_p}{X_C}$$

where X_L is the inductive reactance ($= 2\pi fL$), X_c is the capacitive reactance ($= 1/[2\pi fC]$), and R_p is the total impedance of the parallel resonant circuit

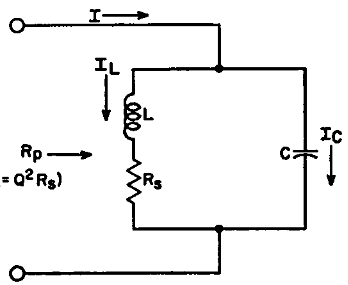


Fig. 67—Simple parallel resonant circuit.

(tank) at resonance. The Q varies inversely with the resistance of the inductor. The lower the resistance, the higher the Q and the greater the difference between the tank impedance at frequencies off resonance compared to the tank impedance at the resonant frequency.

The Q of a tuned interstage coupling network also depends upon the impedances of the preceding and following stages. The output impedance of a tube can be considered as consisting of a resistance R_o in parallel with a capacitance C_o, as shown in Fig. 68. Similarly, the input impedance can be considered as consisting of a resistance R_i in parallel with a capacitance C_i. Because the tuned circuit is shunted by both the output impedance of the preceding tube and the input impedance of the following tube, the effective selectivity of the circuit is the loaded Q (or

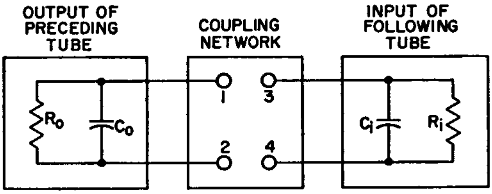


Fig. 68—Equivalent output and input circuits of tubes connected by a coupling network.

Q_L) based upon the total impedance of the coupled network, as follows:

$$Q_L = \frac{\begin{Bmatrix}\text{total loading on}\\ \text{coil at resonance}\end{Bmatrix}}{X_L \text{ or } X_C}$$

The capacitances C_o and C_i in Fig. 68 are usually considered as part of the coupling network. For example, if the required capacitance between terminals 1 and 2 of the coupling network is calculated to be 500 picofarads and the value of C_o is 10 picofarads, a capacitor of 490 picofarads is used between terminals 1 and 2 so that the total capacitance is 500 picofarads. The same method is used to allow for the capacitance C_i at terminals 3 and 4.

When a tuned resonant circuit in the primary winding of a transformer is coupled to the nonresonant secondary winding of the transformer, as shown in Fig. 69, the effect of the input impedance of the following stage on the Q of the tuned circuit can be determined by considering the values reflected (or referred) to the primary circuit by

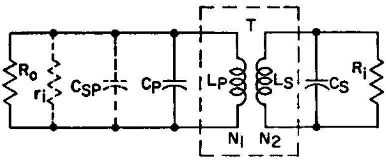


Fig. 69—Equivalent circuit for transformer-coupling network having tuned primary winding.

transformer action. The reflected resistance r_i is equal to the resistance R_i in the secondary circuit times the square of the effective turns ratio between the primary and secondary windings of the transformer T:

$$r_i = R_i\ (N_1/N_2)^2$$

where N_1/N_2 represents the electrical turns ratio between the primary winding and the secondary winding of T. If there is capacitance in the secondary circuit (C_s), it is reflected to the primary circuit as a capacitance C_{sp}, and is given by

$$C_{sp} = C_p \div (N_1/N_2)^2$$

The loaded Q, or Q_L, is then calculated on the basis of the inductance L_p, the total shunt resistance (R_o plus r_i plus the tuned-circuit impedance $Z_t = Q_oX_c = Q_oX_L$), and the total capacitance ($C_p + C_{sp}$) in the tuned circuit.

Fig. 70 shows a coupling network which consists of a single-tuned circuit using mutual inductive coupling. The

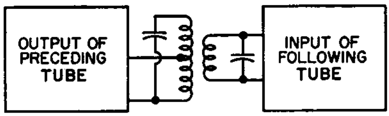


Fig. 70—Equivalent circuit for transformer-coupling network using inductive coupling.

capacitance C_t includes the effects of both the output capacitance of the preceding tube and the input capacitance of the following tube (referred to the primary of transformer T_1). The bandwidth of a single-tuned transformer is determined by the half-power points on the resonance curve (—3 dB or 0.707 down from the maximum). Under these conditions, the band pass Δf is equal to the ratio of the center or resonant frequency f_r divided by the loaded (effective) Q of the circuit, as follows:

$$\Delta f = f_r/Q_L$$

In high-frequency tuned amplifiers, where the input impedance is typically low, mutual inductive coupling may be impracticable because of the small number of turns in the secondary winding. It is extremely difficult in practice to construct a fractional part of a turn. In such cases, capacitance coupling may

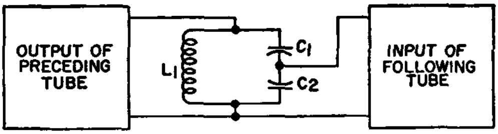


Fig. 71—Single-tuned coupling network using capacitive division.

be used, as shown in Fig. 71. This arrangement, which is also called **capacitive division,** is similar to tapping down on a coil at or near resonance. Impedance transformation in this network is determined by the ratio between capacitors C_1 and C_2. Capacitor C_1 is normally much smaller than C_2; thus the capacitive reactance X_{C1} is normally much larger than X_{C2}. Provided the input resistance of the following tube is much greater than X_{C2}, the effective turns ratio from the top of the coil to the input of the following tube is $(C_1 + C_2)/C_1$. The total capacitance C_t across the inductance L is given by

$$C_t = \frac{C_1C_2}{C_1 + C_2}$$

The resonant frequency f_r is then given by

$$f_r = \frac{1}{2\pi\sqrt{L_1C_t}}$$

Double-tuned interstage coupling networks are often used in preference to single-tuned networks to provide flatter frequency response within the pass band, a sharper drop in response immediately adjacent to the ends of the pass band, or more attenuation at frequencies far removed from resonance. In synchronous double-tuned networks, both the resonant circuit in the input of the coupling network and the resonant circuit in the output are tuned to the same resonant frequency. In "stagger-tuned" networks, the two resonant circuits are tuned to slightly different resonant frequencies to provide a more rectangular band pass with sharper selectivity at the ends of the pass band. Double-tuned or stagger-tuned networks may use capacitive, inductive, or mutual inductance coupling, or any combination of the three.

Television Tuners

The vhf tuner of a television receiver selects the desired frequency channel in the range from 55 to 216 MHz, amplifies it, and converts it to a lower intermediate frequency. These functions are accomplished in rf-amplifier, mixer, and local-oscillator stages employing tube types that are designed specifically for these applications. The rf-amplifier stage uses a high-transconductance tube that has small dimensions to maintain low interelectrode capacitances, particularly between grid and plate. The mixer and oscillator stages usually employ a dual-unit triode-pentode unit and a medium-mu triode unit.

Fig. 72 shows a simplified schematic diagram of a typical vhf television tuner. The balun converts the 300-ohm balanced antenna impedance to an unbalanced impedance of 75 ohms. The high-pass filter eliminates lower-frequency interference signals. The tuner is set to the desired frequency by simultaneous adjustment of the inductances indicated by the several sets of arrows in Fig. 72. The inductances are either replaced completely or incremental amounts of inductance are added as the tuner is switched from high frequencies to lower frequencies. Some tuners use a combination of the two methods.

Because **noise** generated in the first amplifier stage is often the controlling factor in determining the over-all sensitivity of a radio or television receiver, the "front end" is designed with special attention to both gain and noise characteristics. The input circuit of an amplifier inherently contains some thermal noise contributed by the resistive elements in the input device. When an input signal is amplified, therefore, the thermal noise generated in the input circuit is also amplified. If the ratio of signal power to noise power **(signal-to-noise ratio, S/N)** is the same in the output circuit as in the input circuit, the amplifier is considered to be "noise-

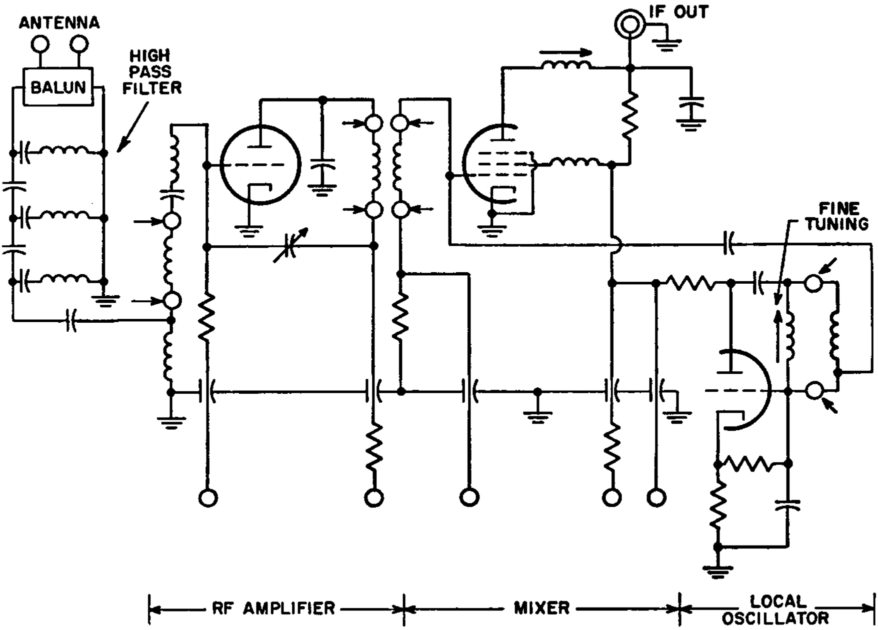


Fig. 72—Simplified schematic of typical vhf television tuner.

less," and is said to have a noise figure of unity, or zero dB.

In practical circuits, however, all amplifier stages generate a certain amount of noise as a result of thermal agitation of electrons in resistors and other components, minute variations in the cathode emission of tubes' (shot effect), and minute grid currents in the amplifier tubes. As a result, the ratio of signal power to noise power is inevitably impaired during amplification. A measure of the degree of impairment is called the **noise figure** (NF) of the amplifier, and is expressed as the ratio of signal power to noise power at the input (S_i/N_i) divided by the ratio of signal power to noise power at the output (S_o/N_o), as follows:

$$NF = \frac{(S_i/N_i)}{(S_o/N_o)}$$

The noise figure in decibels (dB) is equal to ten times the logarithm of this power ratio. For example, a one-dB noise figure in an amplifier decreases the signal-to-noise ratio by a factor of 1.26, a 3-dB noise figure by a factor of 2, a 10-dB noise figure by a factor of 10, and a 20-dB noise figure by a factor of 100.

The over-all noise figure of a receiver is affected by the total number of stages, as shown by the following relationship:

$$NF_{receiver} = NF_1 + \frac{(NF_2 + 1)}{G_1} + \frac{(NF_3 + 1)}{G_1 G_2} \ldots$$

where G represents power gain and the subscripts indicate the number of each stage. This relationship indicates that the contribution of the second-stage noise factor to that of the over-all receiver is reduced by the gain of the first stage. Therefore, it is important that the rf amplifier have enough gain to make the effect of the second stage negligible. The third stage will then have even less effect. The maximum available power gain G of an rf stage is given by

$$G = \frac{g_m^2 R_{in} R_{out}}{4}$$

For maximum gain, therefore, the rf-amplifier tube should have high transconductance and high input and output impedances. At frequencies in the vhf television band, the input resistance is small enough to affect the gain. As mentioned previously, the rf tube is designed to have low interelectrode capacitances, small interelectrode spacings, and low lead inductances (particularly the cathode lead).

The gain of the rf stage must be reduced as the incoming-signal amplitude changes to prevent overload distortion in the following stages. As the signal amplitude increases, an automatic-gain-control (agc) circuit biases the rf tube to decrease its gain. The rf tube usually employs a *semiremote-cutoff* grid to reduce cross-modulation distortion.

Either a triode or a pentode can be used in the **rf-amplifier** stage of tuner input circuits of vhf television receivers. Such stages are required to amplify signals ranging from 55 to 216 MHz and having a bandwidth of 4.5 MHz (the tuner is usually aligned for a bandwidth of 6 MHz to assure complete coverage of the band). In early rf tuners, pentodes rather than triodes were used because the grid-plate capacitance of triodes created stability problems. However, the use of twin triodes in direct-coupled cathode-drive circuits makes it possible to obtain stable operation along with the low-noise characteristics of triodes.

Pentodes or tetrodes do not provide the useful sensitivity of triodes because of the "partition noise" introduced by the screen grid. The direct-coupled cathode-drive circuit provides both the gain and the stability capabilities of the pentode, as well as the advantages of a low-noise triode input stage. Because the cathode-drive stage provides a low-impedance load to the grounded-cathode stage, the gain of the latter stage is very low and there is no necessity for neutralizing the grid-plate capacitance. An *interstage impedance*, usually an inductance in series with the plate of the first stage and the cathode of the second stage, is often used at higher frequencies to provide a degree of impedance matching between the units. The cathode-drive portion of the circuit is matched to the input network and provides most of the stage gain. Because the feedback path of the cathode-drive circuit is the plate-cathode capacitance, which in most cases is very small, excellent isolation is provided between the antenna and the local oscillator.

Development of single triodes having low grid-plate capacitance, such as the 6BN4, has made possible the design of neutralized triode rf circuits. Tubes such as the 6GK5 and 6CW4 are specially designed to minimize grid-plate capacitance to permit easier neutralization of a grounded-cathode circuit over the wide frequency band. Bridge-neutralized rf-amplifier stages are widely used in television tuners; in this arrangement, a portion of the output signal is returned to the grid out of phase with the feedback signal from the grid-plate capacitance. This circuit provides excellent gain and noise performance with stable operation across the band.

The **mixer** stage of a vhf tuner usually employs a pentode tube, or the pentode unit of a triode-pentode tube. Although triodes such as the 6J6 were used as mixers in early receivers, they have been replaced by pentodes because the higher output impedance of a pentode provides a higher mixer gain than can be obtained with a triode.

The amplified signal from the rf stage in Fig. 72 is applied to the mixer grid along with a local-oscillator signal of much larger amplitude. The local-oscillator signal varies the mixer grid voltage from cutoff into the grid-current region. This signal develops a grid-resistor bias, called the **injection voltage,** which is a measure of the local-oscillator voltage. Because the transfer curve of the mixer tube is nonlinear, mixing action between the rf signal and the local-oscillator signal produces sum and difference frequencies. The output circuit of the mixer is tuned to the difference frequency (about 44 MHz) and rejects all other frequencies. This signal is then fed to the intermediate-frequency amplifier.

The mixer gain is a function of the amplitude of the local-oscillator signal. The gain has a broad maximum over a range of injection voltages from −2.5 to −5.0 volts for conventional-grid mixers and slightly lower for frame-grid mixers. Good impedance matching between the rf-amplifier plate and the mixer grid, consistent with bandpass requirements, is important to achieve maximum signal power transfer. A slight amount of regeneration is provided by a small screen-grid inductance. This regeneration effectively increases the mixer-grid input impedance and thus improves power gain.

The **local-oscillator** stage shown in Fig. 72 is a Colpitts type in which the tuned circuit is located between the grid and plate and the feedback path is through the tube interelectrode capacitances. A large signal is developed in the local oscillator and coupled loosely to the mixer grid to minimize the effects of changes in the mixer input on the frequency of oscillation. The circuit is designed to keep frequency shift within a very narrow range with supply-voltage and temperature changes. Fine tuning is provided by a variable inductance or capacitance across the tuned circuit. Tubes commonly used in local-oscillator and mixer circuits are the 6EA8, 6KZ8, and 6KE8.

Television IF Amplifiers

The intermediate-frequency (if) amplifier stages in a television receiver provide the additional gain required to bring the signal level to an amplitude suitable for final detection. A constant peak signal of about three to five volts is required at the input to the detector. The mixer output signal is passed through two or three stages of amplification to attain this level. High-transconductance pentodes having low grid-No.1-to-plate capacitances are normally used in if amplifiers. The coupling circuits are usually tuned transformers which may be single- or double-tuned. The transformers are either synchronously (same frequency) tuned or stagger-tuned, depending on circuit requirements. The over-all bandwidth varies from a maximum of 3.58 MHz at the 6-dB points for color receivers to values in the order of 2.0 to 2.5 MHz for the most inexpensive receivers. An expression for the figure of merit for a single tuned if-amplifier tube is the gain-bandwidth product G × B, which is given by

$$G \times B = \frac{g_m}{2 \pi C}$$

where C is the total tuning capacitance. This relationship again demonstrates the need for high transconductance and low interelectrode capacitance.

The first stage (or first two stages in the case of a three-stage if) is gain-controlled like the rf amplifier. However, the bias applied to the if-amplifier tube varies the input resistance and capacitance of the tube and thus detunes the circuit. It is important for proper reception to maintain the frequency response of the if stages constant, particularly in the case of the color receiver. Therefore, a small unbypassed cathode resistor is used which provides degenerative feedback to minimize the effect of bias changes. In addition, the effects on input impedance caused by the grid-plate capacitance are reduced by use of a partial bypass capacitor at the screen grid to provide neutralization of the grid-to-plate capacitance.

Tubes used in the gain-controlled stages of the if amplifier have remote- or semiremote-cutoff characteristics to reduce cross-modulation or intermodulation interference. Tube types commonly used in this application include the 6BZ6, 6GM6, 6JH6, 6JD6A, and 6KT6.

The last if-amplifier stage is a relatively-large-signal amplifier. For this reason, the tube must be biased so that it will operate over a region of linear operation for large voltage excursions. Because such a quiescent operating point provides a transconductance somewhat below the maximum value for the tube, the selection of the operating point involves a compromise between signal-handling capacity and gain. For purposes of linearity, the final if-amplifier stage is not gain-controlled, and operates with the cathode bypassed to ground. Because fixed bias is used, a sharp-cutoff tube is used to provide higher transconductance than could be obtained with an equivalent remote- or semiremote-cutoff tube. Examples of types used in this stage are the 6EW6 and 6JC6A.

Wideband (Video) Amplifiers

In some applications, it is necessary for a circuit to amplify signals ranging from very low frequencies (several hertz) to high frequencies (tens of megahertz) with a minimum of frequency and time-delay distortion. For example, very exacting requirements are demanded for such applications as television camera chains, ac voltmeters, and vertical amplifiers for oscilloscopes. In response to these demands, circuit compensation techniques have been developed to minimize the amplitude and time-delay variation as the upper or lower frequency limits of the amplifier are approached.

The need for such compensation is evident when many identical stages of amplification are employed. If ten cascaded stages are used, a variation of 0.3 dB per stage results in a total variation of 3 dB. In an uncompensated amplifier, this total variation occurs two octaves (a frequency ratio of four) prior to the half-power point. Because two octaves are lost from both the high and low frequencies, the bandwidth of ten cascaded uncompensated amplifies stages is only one-sixteenth that of a single amplifier stage. Fig. 73 shows the amplitude response characteristics of various numbers of identical uncompensated amplifiers.

In general, the output of an amplifier may be represented by a current generator i_{out} and a load resistance R_L, as shown in Fig. 74(a). Because the signal current is shunted by various capacitances at high frequencies, as shown in Fig. 74(b), there is a loss in gain at these frequencies. If an inductor L is placed in series with the load resistor R_L, as shown in Fig. 74(c), a low-Q circuit is formed which somewhat suppresses the capacitive loading. This method of gain compensation, called **shunt peaking,** can be effective for improving high-frequency response. Fig. 74 shows the frequency response for the circuits in Fig. 74(a), (b), and (c). If the inductor L in Fig. 74(c) is made **self-resonant** approximately one octave above the 3-dB frequency of the circuit of Fig. 74(b), the amplifier response is extended by about another 30 per cent.

If the stray capacitance C shown in Fig. 74(b) is broken into two parts C′ and C″ and an inductor L_1 is placed between them, a heavily damped form of series resonance may be employed for further improvement. This form of compensation, called **series peaking,** is shown in Fig. 75(a). If C′ and C″ are within a factor of two of each other, series peaking produces an appreciable improvement in frequency response as compared to shunt peaking. A more complex form of compensation embodying both self-resonant shunt peaking and series peaking is shown in Fig. 75(b).

The effects of various high-fre-

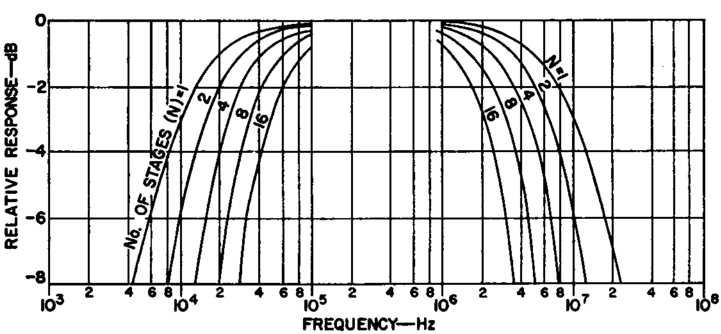


Fig. 73—Amplitude response characteristics of various numbers (N) of identical uncompensated amplifiers.

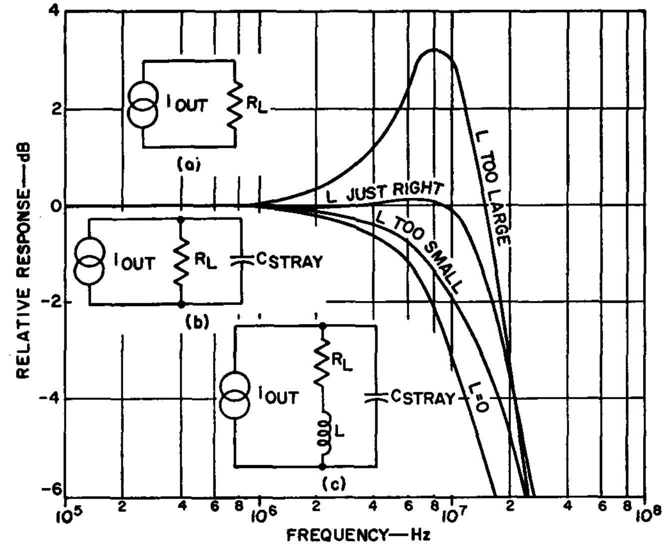


Fig. 74—Equivalent circuits and frequency response of uncompensated and shunt-peaked amplifiers.

quency compensation systems can be demonstrated by consideration of an amplifier consisting of three identical stages. If each of the three stages is down 3 dB at 1 MHz, and if a total gain variation of plus 1 dB and minus 3 dB is allowed, the bandwidth of the amplifier is 0.5 MHz without compensation. Shunt peaking raises the bandwidth to 1.3 MHz. Self-resonant shunt

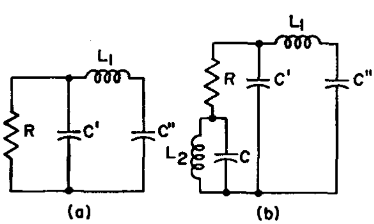


Fig. 75—Circuits using (a) series peaking, and (b) both self-resonant shunt peaking and series peaking.

peaking raises it to 1.5 MHz. An infinitely complicated network of shunt-peaking techniques could raise it to 2 MHz. If the distribution of capacitance permits it, series peaking alone can provide a bandwidth of about 2 MHz, while a combination of shunt and series peaking can provide a bandwidth of approximately 2.8 MHz. If the capacitance is perfectly distributed, and if an infinitely complex network of shunt and series peaking is employed, the ultimate capability is about 4 MHz.

The frequency response of a wideband amplifier is influenced greatly by variations in component values due to temperature effects, variation of tube parameters with voltage and current (normal large-signal excursions), changes of stray capacitance due to relocated lead wires, or other variations. A change of 20 per cent in any of the critical parameters can cause a change of 0.7 dB in gain per stage over the last half-octave of the response for the most simple case of shunt peaking. As the bandwidth is extended by more complex peaking, a circuit becomes substantially more critical. (Measurement probes generally alter circuit performance because of their capacitance; this effect should be considered during frequency-response measurements.)

In the design of wideband amplifiers using many stages of amplification, it is necessary to consider time-delay variations as well as amplitude variation. When feedback capacitance is a major contributor to response limitation, the more complex compensating networks may produce severe ringing or even sustained oscillation. If feedback capacitance is treated as input capacitance produced by the Miller effect, the added input capacitance C_f' caused by the feedback capacitor C_f is given by

$$C_f' = C_f\ (1 - VG)$$

where VG is the input-to-output voltage gain. The gain VG, however, has a phase angle that varies with frequency. The phase angle is 180 degrees at low frequencies, but may lead or lag this value at high frequencies; the magnitude of VG then also varies. In the design of very wideband amplifiers (20 MHz or more), the phase of the transconductance g_m must be considered.

The **video amplifier** stage in a television receiver usually employs a pentode-type tube specially designed to amplify the wide band of frequencies contained in the video signal and, at the same time, to provide high gain per stage. Pentodes are more useful than triodes in such stages because they have high transconductance (to provide high gain) together with low input and output interelectrode capacitances (to permit the broadband requirements to be satisfied). An approximate "figure of merit" for a particular tube for this application can be determined from the ratio of its transconductance, g_m, to the sum of its input and output capacitances, C_{in} and C_{out}, as follows:

$$\text{Figure of Merit} = \frac{g_m}{C_{in} + C_{out}}$$

Typical values for this figure are in the order of 500 x 10^6 or greater.

A typical video amplifier stage, such as that shown in Fig. 76, is connected between the second detector of the television receiver and the picture tube. The contrast control, R_1, in this circuit controls the gain of the video amplifier tube. The inductance, L_2, in series with the load resistor, R_L, maintains the plate load impedance at a relatively constant value with increasing frequency. The inductance L_1 isolates the output capacitance of the tube so that only stray capacitance is placed

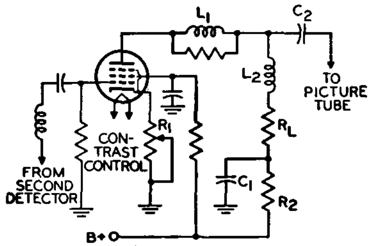


Fig. 76—Typical video amplifier stage.

across the load. As a result, a higher-value load resistor is used to provide higher gain without affecting frequency response or phase relations. The decoupling circuit, C_1R_2, is used to improve the low-frequency response. Tubes used as video amplifiers include types 6CL6 and 12BY7A, or the pentode sections of types 6AW8A and 6AN8A.

The **luminance** amplifier in a color-television receiver is a conventional video amplifier having a bandwidth of approximately 3.5 MHz. In a color receiver, the portion of the output of the second detector which lies within the frequency band from approximately 2.4 to 4.5 MHz is fed to a bandpass amplifier, as shown in the block diagram in Fig. 77. The color

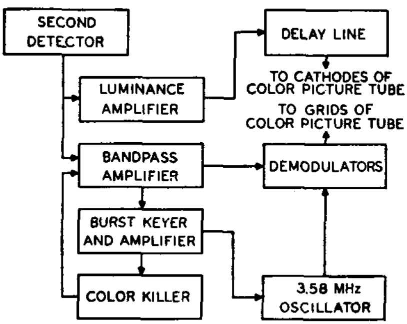


Fig. 77—Block diagram of video-amplifier section of color television receiver.

synchronizing signal, or "burst," contained in this signal may then be fed to a "burst-keyer" tube. At the same time, a delayed horizontal pulse may be applied to the keyer tube. The output of the keyer tube is applied to the burst amplifier tube and the signal is then fed to the 3.58-MHz oscillator and to the "color-killer" stage.

The color killer applies a bias voltage to the bandpass amplifier in the absence of burst so that the color section, or **chrominance** channel, of the receiver remains inoperative during black-and-white broadcasts. A threshold control varies the bias and controls the burst level at which the killer stage operates.

The output of the 3.58-MHz oscillator and the output of the bandpass amplifier are fed into phase and amplitude demodulator circuits. The output of each demodulator circuit is an electrical representation of a color-difference signal, *i.e.*, an actual color signal minus the black-and-white, or luminance, signal. The two color-difference signals are combined to produce the third color-difference signal; each of the three signals then represents one of the primary colors.

The three color-difference signals are usually applied to the grids of the three electron guns of the color picture tube, in which case the black-and-white signal from the luminance amplifier may be applied simultaneously to the cathodes. The chrominance and luminance signals then combine to produce the color picture. In the absence of transmitted color information, the chrominance channel is cut off by the color killer, as described above, and only the luminance signal is applied to the picture tube, producing a black-and-white picture.

TV Scanning, Sync, and Deflection

For reproduction of a transmitted picture in a television receiver, the face of a cathode-ray tube is scanned with an electron beam while the intensity of the beam is varied to control the emitted light at the phosphor screen. The scanning is synchronized with a scanned image at the TV transmitter, and the black-through-white picture areas of the scanned image are converted into an electrical signal that controls the intensity of the electron beam in the picture tube at the receiver.

Scanning Fundamentals

The scanning procedure used in the United States employs horizontal linear scanning in an odd-line interlaced pattern. The standard scanning pattern for television systems includes a total of 525 horizontal scanning lines in a rectangular frame having an aspect ratio of 4 to 3. The frames are repeated at a rate of 30 per second, with two fields interlaced in each frame. The first field in each frame consists of all odd-number scanning lines, and the second field in each frame consists of all even-number scanning lines. The field repetition rate is thus 60 per second, and the vertical scanning rate is 60 Hz.

The geometry of the standard odd-line interlaced scanning pattern is illustrated in Fig. 78. The scanning beam starts at the upper left corner of the frame at point A, and sweeps across the frame with uniform velocity to cover all the picture elements in one horizontal line. At the end of each trace, the beam is rapidly returned to the left side of the frame, as shown by the dashed line, to begin the next horizontal line. The horizontal lines slope downward in the direction of scanning because the vertical deflecting signal simultaneously produces a verti-

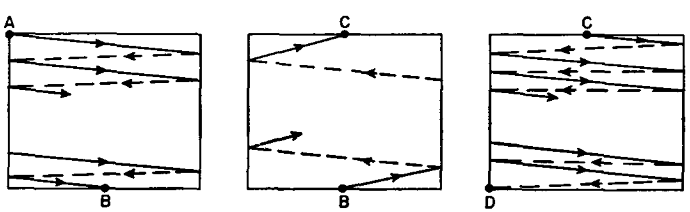


Fig. 78—The odd-line interlaced scanning procedure.

cal scanning motion, which is very slow compared with the horizontal scanning speed. The slope of the horizontal line trace from left to right is greater than the slope of the retrace from right to left because the shorter time of the retrace does not allow as much time for vertical deflection of the beam. Thus, the beam is continuously and slowly deflected downward as it scans the horizontal lines, and its position is successively lower as the horizontal scanning proceeds.

At the bottom of the field, the vertical retrace begins, and the beam is brought back to the top of the frame to begin the second or even-number field. The vertical "flyback" time is very fast compared to the trace, but is slow compared to the horizontal scanning speed; therefore, some horizontal lines are produced during the vertical flyback.

All odd-number fields begin at point A in Fig. 78 and are the same. All even-number fields begin at point C and are the same. Because the beginning of the even-field scanning at C is on the same horizontal level as A, with a separation of one-half line, and the slope of all lines is the same, the even-number lines in the even fields fall exactly between the odd-number lines in the odd field.

Sync

In addition to picture information, the composite video signal from the video detector of a television receiver contains timing pulses to assure that the picture is produced on the faceplate of the picture tube at the right instant and in the right location. These pulses, which are called sync pulses, control the horizontal and vertical scanning generators of the receiver.

Fig. 79 shows a portion of the detected video signal. When the picture is bright, the amplitude of the signal is low. Successively deeper grays are represented by higher amplitudes until, at the "blanking level" shown in the diagram, the amplitude represents a complete absence of light. This "black level" is held constant at a value equal to 75 per cent of the maximum amplitude of the signal during transmission. The remaining 25 per cent of the signal amplitude is used for synchronization information. Portions of the signal in this region (above the black level) cannot produce light.

In the transmission of a television picture, the camera becomes inactive at the conclusion of each horizontal line and no picture information is transmitted while the scanning beam is retracing to the beginning of the next line. The scanning beam of the receiver is maintained at the black level during this retrace interval by means of the blanking pulse shown in Fig. 79. Immediately after the beginning of the blanking period, the signal amplitude rises further above the black level to provide a horizontal-synchronization pulse that initiates the action of the horizontal scanning generator. When the bottom line of the picture is reached, a similar vertical-synchronization pulse initiates the action of the vertical scanning generator to move the scanning spot back to the top of the pattern.

The sync pulses in the composite video signal may be separated from the video information in the output of the second or video detector by means of

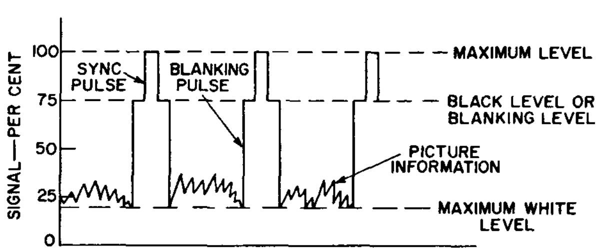


Fig. 79—Detected video signal.

the *triode circuit* shown in Fig. 80. In this circuit, the time constant of the network R_1C_1 is long with respect to the interval between pulses. During each pulse, the grid is driven positive and draws current, thereby charging capacitor C_1. Consequently, the grid develops a bias which is slightly greater

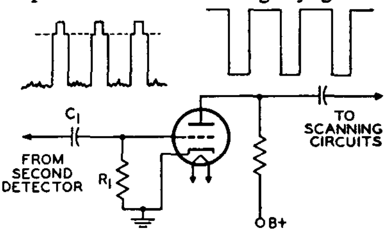


Fig. 80—Sync-separator circuit.

than the cutoff voltage of the tube. Because plate current flows only during the sync-pulse period, only the amplified pulse appears in the output. This **sync-separator** stage discriminates against the video information. Because the bias developed on the grid is proportional to the strength of the incoming signal, the circuit also has the advantage of being relatively independent of signal fluctuations.

After the synchronizing signals are separated from the composite video signal, it is necessary to filter out the horizontal and vertical sync signals so that each can be applied to its respective deflection generator. This filtering is accomplished by RC circuits designed to filter out all but the desired synchronizing signals. Although the horizontal, vertical, and equalizing pulses are all rectangular pulses of the same amplitude, they differ in frequency and pulse width, as shown in Fig. 81. The horizontal sync pulses have a repetition rate of 15,750 per second (one for each horizontal line) and a pulse width of 5.1 microseconds. The equalizing pulses have a width approximately half the horizontal pulse width, and a repetition rate of 31,500 per second; they occur at half-line intervals, with six pulses immediately preceding and six following the vertical synchronizing pulse. The vertical pulse is repeated at a rate of 60 per second (one for each field), and has a width of approximately 190 microseconds. The serrations in the vertical pulse occur at half-line intervals, dividing the complete pulse into six individual pulses that provide horizontal synchronization during the vertical retrace. (Although the picture is blanked out during the vertical retrace time, it is necessary to keep the horizontal scanning generator synchronized.)

All the pulses described above are produced at the transmitter by the synchronizing-pulse generator; their waveshapes and spacings are held within very close tolerances to provide the required synchronization of receiver and transmitter scanning.

The horizontal sync signals are separated from the total sync in a differentiating circuit that has a short time constant compared to the width of the horizontal pulses. When the total sync signal is applied to the *differentiating circuit* shown in Fig. 82, the capacitor charges completely very soon after the leading edge of each pulse, and remains charged for a period of time equal to practically the entire pulse width. When the applied voltage is removed at the time corresponding to the trailing edge of each pulse, the capacitor discharges completely within a very short time. As a result, a positive peak of voltage is obtained for

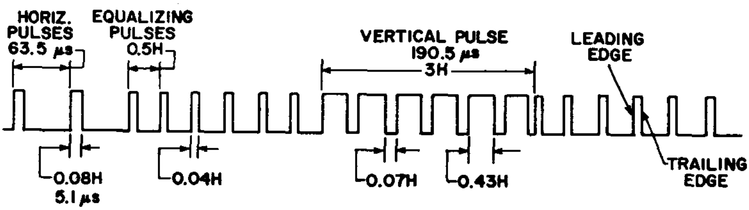


Fig. 81—Waveform of TV synchronizing pulses (H = horizontal line period of 1/15,750 seconds, or 63.5 μs).

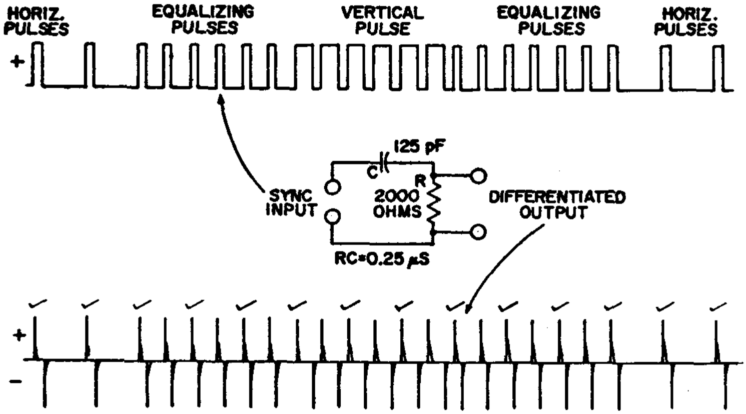


Fig. 82—Separation of the horizontal sync signals from the total sync by a differentiating circuit.

each leading edge and a negative peak for the trailing edge of every pulse. One polarity is produced by the charging current for the leading edge of the applied pulse, and the opposite polarity is obtained from the discharge current corresponding to the trailing edge of the pulse.

As mentioned above, the serrations in the vertical pulse are inserted to provide the differentiated output needed to synchronize the horizontal scanning generator during the time of vertical synchronization. During the vertical blanking period, many more voltage peaks are available than are necessary for horizontal synchronization (only one pulse is used for each horizontal line period). The check marks above the differentiated output in Fig. 82 indicate the voltage peaks used to synchronize the horizontal deflection generator for one field. Because the sync system is made sensitive only to positive pulses occurring at approximately the right horizontal timing, the negative sync pulses and alternate differentiated positive pulses produced by the equalizing pulses and the serrated vertical information have no effect on horizontal timing. It can be seen that although the total sync signal (including vertical synchronizing information) is applied to the circuit of Fig. 82, only horizontal synchronization information appears at the output.

The vertical sync signal is separated from the total sync in an integrating circuit which has a time constant that is long compared with the duration of the 5-microsecond horizontal pulses, but short compared with the 190-microsecond vertical pulse width. Fig. 83 shows the general circuit configuration

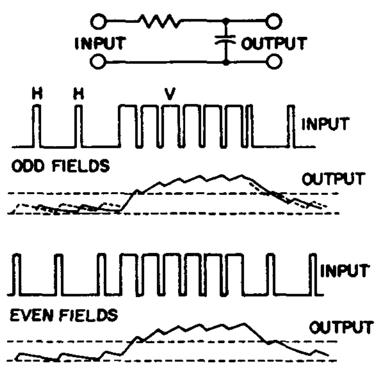


Fig. 83—Separation of vertical sync signals from the total sync for odd and even fields with no equalizing pulses. (Dashed line indicates triggering level for vertical scanning generator.)

used, together with the input and output signals for both odd and even fields. The period between horizontal pulses, when no voltage is applied to the RC circuit, is so much longer than the horizontal pulse width that the capacitor has time to discharge almost down to zero. When the vertical pulse is applied, however, the integrated voltage across the capacitor builds up to the value required for triggering the vertical scanning generator. This integrated voltage across the capacitor reaches its maximum amplitude at the end of the vertical pulse, and then declines practically to zero, producing a pulse of the triangular wave shape shown for the complete vertical synchronizing pulse. Although the total sync signal (including horizontal information) is applied to the circuit of Fig. 83, therefore, only vertical synchronization information appears at the output.

The vertical synchronizing pulses are repeated in the total sync signal at the field frequency of 60 per second. Therefore, the integrated output voltage across the capacitor of the RC circuit of Fig. 83 can be coupled to the vertical scanning generator to provide vertical synchronization. The six equalizing pulses immediately preceding and following the vertical pulse improve the accuracy of the vertical synchronization for better interlacing. The equalizing pulses that precede the vertical pulses make the average value of applied voltage more nearly the same for even and odd fields, so that the integrated voltage across the capacitor adjusts to practically equal values for the two fields before the vertical pulse begins. The equalizing pulses that follow the vertical pulse minimize any difference in the trailing edge of the vertical synchronizing signal for even and odd fields.

In fringe areas, two conditions complicate the process of sync separation. First, the incoming signal available at the antenna is weak and susceptible to fading and other variations; second, the receiver is operating at or near maximum gain, which makes it extremely susceptible to interference from pulse-type noise generated by certain types of electrical equipment, ignition systems, switches, or the like. Some type of **noise-immunity** provision is almost essential for acceptable performance. Noise may be reduced or eliminated from the sync and agc circuits by gating or by a combination of gating, inversion, and cancellation. An example, of the latter method is shown in Fig. 84. In this circuit the 6GY6, which has two independent control grids, serves the dual function of agc amplifier and noise inverter. Because the sync tips of the video signal at grid No. 1 of the 6GY6 drive the tube near its cutoff region, any noise signal extending above the tip level will appear inverted across the grid-No.2 load resistor R. This inverted noise signal is re-combined with the video signal and fed to the sync separator at point "A" in Fig. 84, where noise cancellation takes place. This process leaves the sync pulses relatively free of disturbing noise and results in a stable picture.

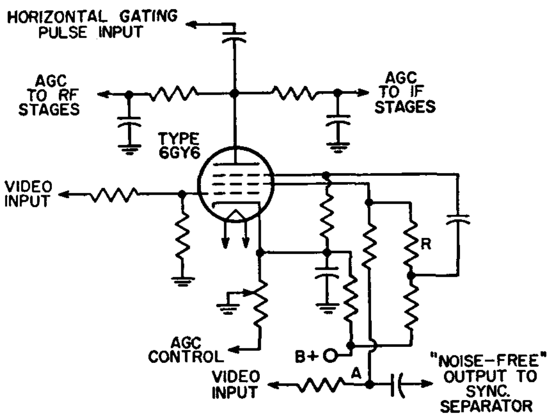


Fig. 84—Typical noise-cancellation circuit.

To prevent reduction of receiver gain due to the effect of noise on the agc amplifier, a portion of the inverted noise signal is fed to the second control grid, grid No.3, of the 6GY6 to cut off or gate the agc amplifier when a noise pulse occurs.

Horizontal Deflection

In the horizontal-deflection stages of a television receiver, a current that varies linearly with time and has a sufficient peak-to-peak amplitude must be passed through the horizontal-deflection-yoke winding to develop a magnetic field adequate to deflect the electron beam of the television picture tube. (This type of deflection is different from that used in a cathode-ray oscilloscope, where the beam is deflected electrostatically.) After the beam is deflected completely across the face of the picture tube, it must be returned very quickly to its starting point. (As explained previously, the beam is extinguished during this retrace by the blanking pulse incorporated in the composite video signal, or in some cases by additional external blanking derived from the horizontal-deflection system.)

The simplest form of a deflection circuit is shown in Fig. 85. In this circuit, the yoke impedance L is assumed to be a perfect inductor. When the

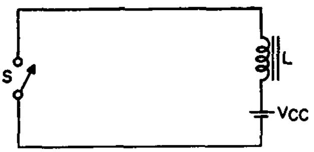


Fig. 85—Simplest form of deflection circuit.

switch is closed, the yoke current starts from zero and increases linearly. At any time t, the current i is equal to Et/L, where E is the applied voltage. When the switch is opened at a later time t_1, the current instantly drops from a value of Et_1/L to zero.

Although the basic circuit of Fig. 85 crudely approaches the requirements for deflection, it presents some obvious problems and limitations. The voltage across the switch becomes extremely high, theoretically approaching infinity. In addition, if very little of the total time is spent at zero current, the circuit would require a tremendous amount of dc power. Furthermore, the operation of the switch would be rather critical with regard to both its opening and its closing. Finally, because the deflection field would be phased in only one direction, the beam would have to be centered at the extreme left of the screen for zero yoke current.

If a capacitor is placed across the switch, as shown in Fig. 86, the yoke

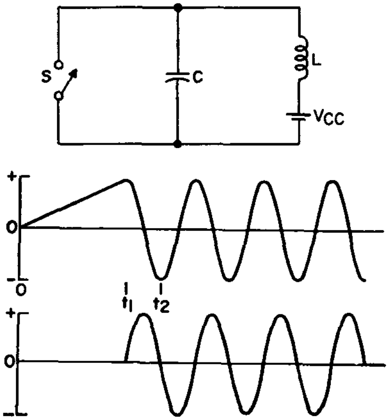


Fig. 86—Addition of capacitor to permit flyback ringing, and yoke-current (upper) and switch-voltage (lower) waveforms.

current still increases linearly when the switch is closed at time t = 0. However, when the switch is opened at time t = t_1, a tuned circuit is formed by the parallel combination of L and C. The resulting yoke currents and switch voltages are then as shown in Fig. 86. The current is at a maximum when the voltage equals zero, and the voltage is at a maximum when the current equals zero. If it is assumed that there are no losses, the ringing frequency f_{osc} is equal to $1/(2\pi\sqrt{LC})$.

If the switch is closed again at any time the capacitor voltage is not equal to zero, an infinite switch current flows as a result of the capacitive discharge. However, if the switch is closed at the precise moment t_2 that the capacitor voltage equals zero, the capacitor current effortlessly transfers to the switch, and a new transient condition results. Fig. 87 shows the yoke-current and switch-voltage waveforms for this new condition.

If the switch is again opened at t_4, closed at t_5, and so on, the desired

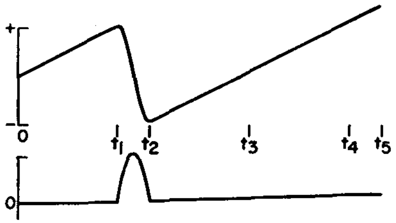


Fig. 87—Yoke-current (upper) and switch-voltage (lower) waveforms when switch is closed at t_2.

sweep results, the peak switch voltage is finite, and the average supply current is zero. The deflection system is then lossless and efficient and, because the average yoke current is zero, beam decentering is avoided. The only fault of the circuit of Fig. 86 is the critical timing of the switch, particularly at time $t = t_2$. However, if the switch is shunted by a damper diode, as shown in Fig. 88, the diode acts as a closed switch as soon as the capacitor voltage reverses slightly. The switch may then be closed at any time between t_2 and t_3.

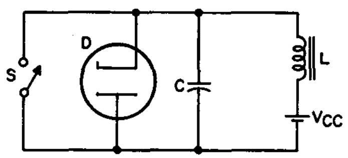


Fig. 88—Incorporation of damper diode.

Fig. 89 shows a typical horizontal-output-and-deflection circuit used in television receivers. In addition to supplying the deflection energy required for horizontal deflection of the picture-tube beam, this circuit provides the high dc voltage required for the ultor (anode) of the picture tube and the "boosted" B voltage for other portions of the receiver. The horizontal-output tube is usually a beam power tube such as the 6JB6A, 6JG6A, or 6JE6A.

In this circuit, a sawtooth voltage from the horizontal-oscillator tube is applied to the grid No. 1 of the horizontal-output tube. When this voltage rises above the cutoff point of the output tube, the tube conducts a sawtooth of plate current which is fed through the auto-transformer to the horizontal-deflecting yoke. At the end of the horizontal-scanning cycle, which lasts for 63.4 microseconds, the sawtooth voltage on the grid suddenly cuts off the output tube. This sudden change sets up an oscillation of about 50 to 70 kHz in the output circuit, which may be considered as an inductor shunted by the stray capacitance of the circuit. During the first half of this oscillation, a positive voltage appears across the transformer. In the

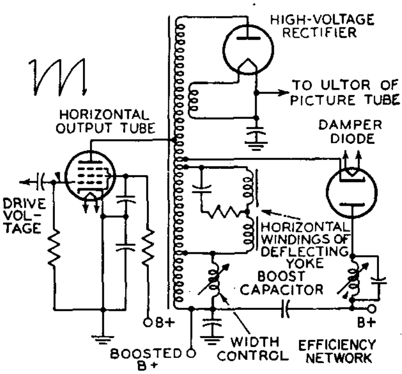


Fig. 89—Typical horizontal-deflection and high-voltage circuit.

second half of the cycle, the voltage swings below the plate supply voltage, and the damper diode conducts, damping out the oscillation. At the same time, the current through the deflecting yoke reverses and reaches its negative peak. As the damper-diode current decays to zero, the output tube begins to conduct again. The yoke current, therefore, is composed of current resulting from damper-diode conduction followed by output-tube conduction.

When the output tube is suddenly cut off, the high-voltage pulse produced is increased by means of an extra winding on the transformer. This high-voltage pulse charges a high-voltage capacitor through the high-voltage rectifier. The output of this circuit is the dc high-voltage supply for the picture tube. The high-voltage rectifier also obtains its filament power through a separate winding on the horizontal-output transformer.

Current flowing through the damper diode charges the "boost" capacitor through the damper portion of the transformer winding. The polarity of the charge on the capacitor is such that the voltage at the low end of the winding is increased above the plate supply voltage, or B+. This higher voltage or "boost" is used for the output-tube plate supply, and may also supply the deflection oscillators and other low-current-drain circuits in the receiver.

Vertical Deflection

The vertical-deflection circuit in a television receiver is essentially a class A audio amplifier with a complex load line, severe low-frequency requirements (much lower than 60 Hz), and a need for controlled linearity. The equivalent low-frequency response for a 10-percent deviation from linearity is 1 Hz.

The required performance can be obtained in a vertical-deflection circuit in any of three ways. The amplifier may be designed to provide a flat response down to 1 Hz. This design, however, requires an extremely large output transformer and immense capacitors. Another arrangement is to design the amplifier for fairly good low-frequency response and predistort the generated signal.

The third method is to provide extra gain so that feedback techniques can be used to provide linearity. If loop feedback of 20 or 30 dB is used, tube gain variations and nonlinearities become fairly insignificant. The feedback automatically provides the necessary "predistortion" to correct low-frequency limitations. In addition, the coupling of miscellaneous signals (such as power-supply hum or horizontal-deflection signals) in the amplifying loop is suppressed.

A modified multivibrator in which the vertical-output tube is part of the oscillator circuit is used in the vertical-deflection stage of many television receivers. This stage supplies the deflection energy required for vertical deflection of the picture-tube beam. A simplified combined vertical-oscillator-output stage is shown in Fig. 90. Waveshapes at critical points of the circuit are included to illustrate the development of the desired current through the vertical-output transformer and deflecting yoke.

The current waveform through the deflecting yoke and output transformer should be a sawtooth to provide the desired deflection. The grid and plate voltage waveforms of the output tube could also be sawtooth except for the effect of the inductive components in the yoke and transformer. The effect of these inductive components must be taken into consideration, however, particularly during retrace. The fast rate of current change during retrace time (which is approximately 1/15 as long as trace

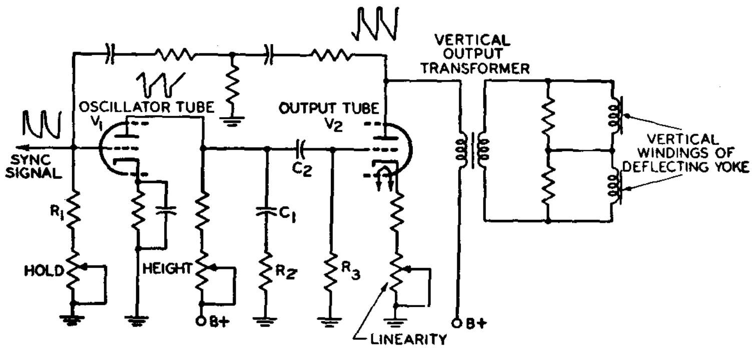


Fig. 90—Simplified combined vertical-oscillator-and-output stage.

time) causes a high-voltage pulse at the plate which could give a trapezoidal waveshape to the plate voltage and cause increased plate current, excess damping, and lengthened retrace time. However, the grid voltage is made sufficiently negative during retrace to keep the tube close to cutoff, as described below.

The frequency, and the relative deviation of the positive and negative portions of each cycle, are dependent on the values of resistors R_1 and R_3 and the RC combination R_3C_2, as explained in the section on multivibrators. The desired trapezoidal waveshape at the grid of V_2 is created by capacitor C_1 and resistor R_2. If R_2 were equal to zero, C_1 would cause the grid-voltage waveshape to take the form shown in Fig. 91(a). When R_2 is sufficiently large, C_1 does not discharge completely when V_1 conducts. When V_1 is cut off, therefore, the voltage on the grid of V_2 immediately rises to the voltage across C_1. The resulting waveshape is shown in Fig. 91(b). The negative-going pulse of the grid-voltage waveshape prevents the high plate pulse from causing excess conduction, and thereby prevents overdamping.

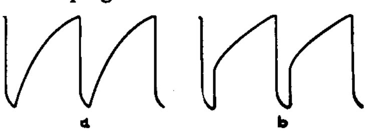


Fig. 91—Waveforms showing effect of R_2 in Fig. 90.

This vertical-deflection stage utilizes twin-triode tubes such as the 6DR7 and 6GF7. The 6GF7 is particularly suitable for this application because it incorporates dissimilar units to provide for the different operating requirements of the oscillator and output sections.

High-Voltage Regulation

In color television receivers, it is very important to regulate the high-voltage supply for the picture tube. Poor regulation of the high voltage can adversely affect the performance of the focusing and convergence circuits so that picture blooming results. In addition, excessive voltage or current may be applied to the high-voltage rectifier, horizontal-output tube, and horizontal-output (flyback) transformer so that the useful life of these components is substantially shortened. In modern color television receivers, regulation of the high voltage is accomplished by use of a shunt-type electronic voltage regulator connected across the output of the high-voltage power supply or by use of a pulse-type regulator connected in shunt with the flyback transformer.

Shunt Regulator Circuit—Fig. 92 shows the schematic diagram of a typical shunt regulator circuit. This circuit uses a 6BK4B sharp-cutoff beam triode for the regulator tube and is suitable for regulation of the output of a high-voltage, high-impedance supply. The cathode of the 6BK4B beam triode is held at a fixed positive potential with respect to ground. Because the grid potential is kept slightly less positive by the voltage drop across resistor R_2, the tube operates in the negative grid region and no grid current is drawn.

When the output voltage, e_o, rises as a result of a decrease in load current,

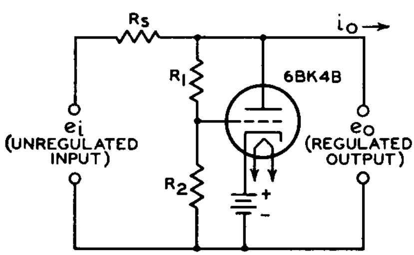


Fig. 92—High-voltage regulator circuit for color television.

a small fraction of the additional voltage is applied to the grid of the tube by the voltage-divider circuit consisting of R_1 and R_2. This increased grid voltage causes the tube to draw an increased current from the unregulated supply. The increased current, in turn, causes a voltage drop across the high internal impedance of the unregulated supply, R_s, which tends to counteract the original rise of the voltage. If desired, the grid may be connected to a variable point on the voltage divider to allow some adjustment of the output-voltage level.

The grid voltage for the 6BK4B can also be obtained from a tap on the B-boost voltage supply. The use of this lower voltage (about 375 volts) eliminates the need for costly and troublesome high-voltage resistors. In this arrangement, variations in high voltage also vary the tapped-down B-boost voltage at the regulator grid, and the resulting variations in conduction of the regulator increase or decrease the loading of the high-voltage supply so that the total load remains nearly constant.

The shunt regulator circuit, in effect, presents a variable load impedance to the output of the high-voltage rectifier. Because the regulator is connected directly across the output of the rectifier, the regulator tube is required to handle the full amount of the high voltage (approximately 25 kilovolts) applied to the picture tube. The tube area, therefore, must be well shielded to provide adequate X-ray protection, and a relatively large area is required for voltage insulation. In addition, the high-voltage rectifier is required to conduct full-load current continuously. The shunt regulator maintains a constant high voltage by sensing changes in the B Boost voltage, which are indicative of changes in beam current, and increasing or decreasing conduction accordingly.

Pulse Regulator Circuit—In a pulse-regulator system, the regulator circuit is effectively shunted across part of the horizontal winding of the horizontal-output transformer. During operation, the pulse-regulator circuit maintains a substantially constant pulse amplitude in the primary winding of the horizontal-output transformer with changing loads on the high-voltage power supply. A constant-amplitude, stepped-up pulse is then applied to the high-voltage rectifier tube, and the high voltage developed from this pulse is maintained at a constant value. In the pulse-regulator system, regulator control is achieved by sampling the picture-tube current by means of a special winding on the fly-back transformer and use of the resultant voltage drop (across a resistor) to control the grid circuit of the regulator tube.

Fig. 93 shows the schematic diagram and significant waveforms for a circuit that uses a 17KV6A beam-power pentode for the regulator tube. During trace and retrace, the cathode of the 17KV6A is held at B+. During the trace period, the screen grid of the 17KV6A is biased well below the cathode voltage and is unaffected by the beam current drawn by the picture tube. The control-grid bias is determined by the resistive voltage-divider network R_2, R_3, R_4, and R_5 and is directly dependent on the beam current of the picture tube. The damper tube conducts during the trace period and holds the plate potential of the 17KV6A at B+. With the plate-to-cathode potential at zero and the screen grid negative with respect to the cathode, the regulator tube is completely cut off during the trace period. At the start of the retrace period, however, the damper tube becomes reverse-biased, and the voltage on the plate of the regulator tube begins to rise. This increasing voltage is coupled to the screen grid through C_1 and R_1 and to the control grid through the interelectrode capacitance of the tube.

The waveforms in Fig. 93 show that at the start of retrace the plate and screen grid of the 17KV6A have both been driven positive with respect to the cathode and the control grid has become less negative with respect to the cathode. The regulator tube then begins to conduct. The pulses impressed on the screen and control grids are short in duration so that the screen grid remains positive with respect to the cathode and the control grid remains near cathode potential for only a short time. The regulator tube is driven into conduction for approximately 2 to 4 microseconds at the start of retrace and is then cut off. As the beam current increases or decreases, the voltage developed across the re-

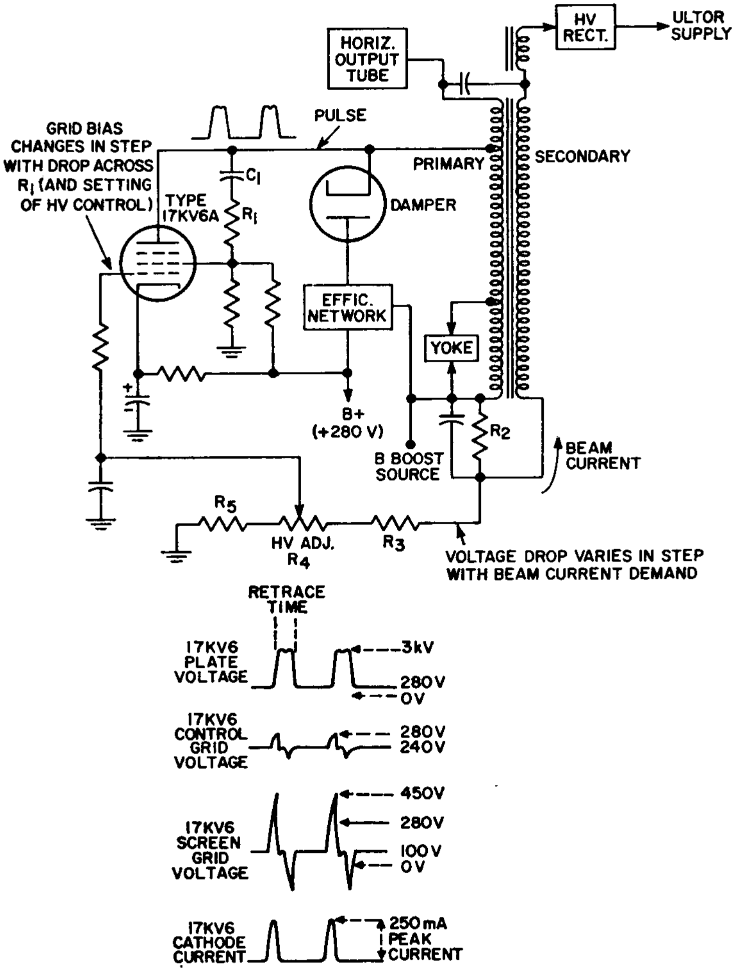


Fig. 93—Schematic diagram and significant waveforms for a typical pulse-regulator circuit.

sistive voltage-divider network R_2, R_3, R_4, and R_5 tracks these changes and is applied to the control grid of the regulator tube. In this way, the conduction of the regulator tube is increased or decreased as required to maintain a constant high-voltage output. By removal of the energy from the rising edge of the flyback pulse in this fashion, the height of the pulse used to develop the high voltage is controlled. At the same time interference with the shape of the deflection pulse is held to a minimum.

Color Demodulation

In the transmission of picture signals for color-television receivers, all the color information is contained in three signals, a luminance (black-and-white) or monochrome signal and two chrominance signals. The luminance signal, which is called the Y signal, contains brightness information only. The voltage response of the Y signal is made similar to the brightness response of the human eye by use of a composite signal that contains definite proportions of the red, green, and blue signals from the color-television camera (30 per cent red, 59 per cent green, and 11 per cent blue). This Y signal, which includes sync and blanking pulses, provides a correct monochrome picture in a conventional black-and-white television receiver.

For the generation of color-television signals, the Y signal is subtracted from the red, green, and blue signals to provide a new set of color-difference signals, which are designated as R-Y, B-Y, and G-Y. All of the original picture information is contained in the Y signal, the R-Y signal, and the B-Y signal. Therefore, the G-Y signal is not contained in the transmitted signal, but is synthesized in the receiver by proper combination of the R-Y and B-Y signals.

(Color signals transmitted under present color-television standards are not R-Y and B-Y, but a similar pair of signals designated as I and Q. In the color-television receiver, R-Y and B-Y signals are demodulated directly from the I and Q signals with negligible loss of color quality. For purposes of simplicity, only R-Y and B-Y signals are considered in this explanation. In addition, a 90-degree phase-shift network is shown; the phase-shift angle could be, and often is, some other value.)

Because the luminance signal and the two color-difference signals must be transmitted with a standard 6-MHz channel, the two color signals are combined into one signal at the transmitter and are independently recovered at the receiver by proper detection techniques. A color subcarrier of approximately 3.58 MHz is used for transmitting the color information within the 6-MHz spectrum of the television station. As shown in Fig. 94, the 3.58-MHz subcarrier and one of the color-difference signals are applied directly to a balanced AM modulator. The other color-difference signal is applied directly to a second balanced AM modulator, and the 3.58-MHz subcarrier is applied to this second modulator through a 90-degree phase-shifting network. The balanced modulators effectively cancel both the individual color-difference signals and the subcarrier signal, and the output contains only the sidebands of the combined chrominance signal.

Recovery of the color information at the receiver involves a process called **synchronous detection.** In this process, two separate detectors are used to recover the separate color information, just as two separate modulators were used to combine the information at the transmitter. The 3.58-MHz subcarrier, which was suppressed during transmission, must be reinserted at the receiver for recovery of the color information.

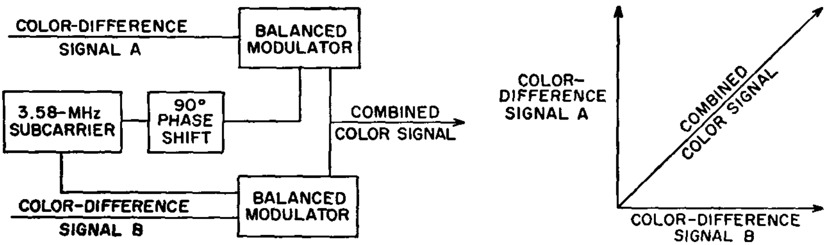


Fig. 94—Formation of combined color signal for transmission.

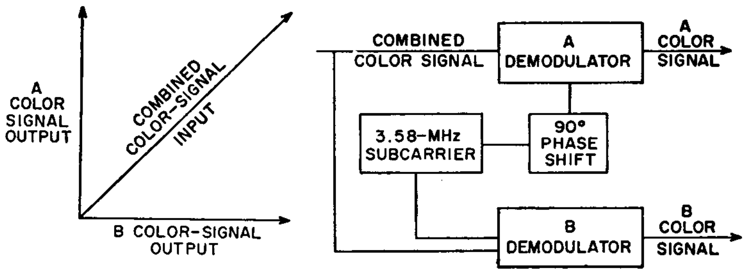


Fig. 95—Separation of combined color signal into two signals at the receiver.

The basis of synchronous detection is the phase relationship of this reinserted 3.58-MHz subcarrier.

For example, the original color information is represented in Fig. 94 by the color-difference signals A and B. At the receiver, the combined color signal is fed to two demodulators A and B, as shown in Fig. 95. At the same time, a 3.58-MHz subcarrier is also fed to the two demodulators, with the same phase relationship that was used in the modulators at the transmitter. This locally generated subcarrier essentially duplicates or replaces the original subcarrier, which was removed at the transmitter.

The local 3.58-MHz oscillator in the color-television receiver is made to function at the proper frequency and phase by means of a synchronizing signal sent out by the transmitter. This synchronizing signal consists of a short **burst** of 3.58-MHz signals transmitted during the horizontal blanking interval, immediately after the horizontal sync pulse, as shown in Fig. 96.

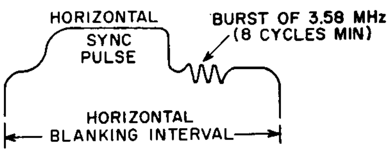


Fig. 96—Waveform for synchronizing signal.

Fig. 97 shows a simplified diagram of a low-level color demodulator frequently used in color-television receivers. The locally generated 3.58-MHz signal is applied to the grid No. 3 of the pentode. The transmitted color signal containing the 3.58-MHz sidebands is applied to grid No. 1. The phase of the 3.58-MHz color signal constantly changes in accordance with its color content. For example, the following table shows six variations in color (hue) as a function of subcarrier phase:

Subcarrier Phase-degrees (with respect to 3.58-MHz local signal in phase with burst)	Hue
13	Yellow
77	Red
119	Magenta
193	Blue
257	Cyan
299	Green

The basic operating principle of the color demodulator shown in Fig. 97 is that plate current from the pentode is zero (or quite low) unless both grid No. 1 and grid No. 3 are simultaneously positive. For example, when the signals applied to the two grids are in phase, plate current can be expected to flow for 180 degrees of each ac cycle. Conversely, when the signals are 180 degrees out of phase, plate current is cut off. The output signal from the detector, therefore, is a function of the phase relationship between the transmitted color signal and the locally generated subcarrier.

In a typical color-television receiver, two color demodulators of the type shown in Fig. 97 are required. In one demodulator, the 3.58-MHz subcarrier signal is applied directly to the pentode grid No. 3 from the local "burst" oscillator. In the other demodulator, the 3.58-MHz signal from the

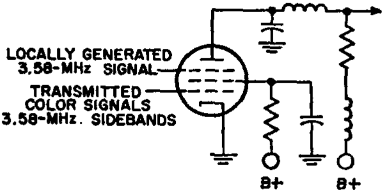


Fig. 97—Low-level color demodulator.

burst oscillator is shifted 90 degrees in phase before it is applied to the pentode grid No. 3. As shown previously in Fig. 95, the demodulator B produces R-Y signals. These B-Y and R-Y signals are then combined (matrixed) to produce the G-Y signal, as discussed earlier. The complete luminance signal is then amplified to the required level in a conventional video-amplifier circuit.

In some color-television receivers, the demodulators are designed so that the color output signals can be applied directly to the color picture tube. In the diagram shown in Fig. 98, for example, the 6JH8 sheet-beam demodula-

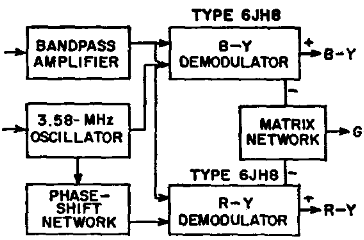


Fig. 98—Block diagram of demodulator circuit used to apply signals directly to color picture tube.

tors produce both positive and negative B-Y and R-Y signals. The positive signals are applied directly to the control grids (grid No. 1) of the blue and red guns of the color picture tube. At the same time, the negative color-difference signals are added (matrixed) in the correct proportions to produce the G-Y signal, which is applied to grid No. 1 of the green gun.

Oscillation

As an oscillator, an electron tube can be employed to generate a *continuously alternating* voltage. In present-day radio broadcast receivers, this application is limited practically to superheterodyne receivers for supplying the heterodyning frequency. Several circuits (represented in Fig. 99) may be utilized, but they all depend on feeding more energy from the plate circuit to the grid circuit than is required to equal the power loss in the grid circuit. Feedback may be produced by electrostatic or electromagnetic coupling between the grid and plate circuits. When sufficient energy is fed back to more than compensate for the loss in the grid circuit, the tube will oscillate.

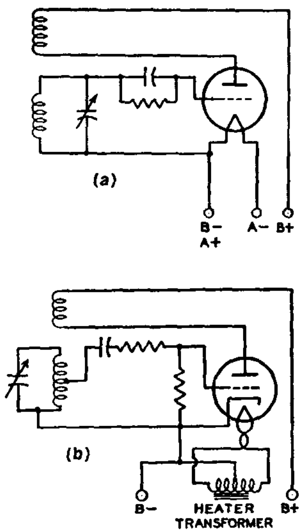


Fig. 99—Tuned-grid triode oscillator circuit: (a) using filament-type tube; (b) using heater-cathode-type tube.

The action consists of regular surges of power between the plate and the grid circuit at a frequency dependent on the circuit constants of inductance and capacitance. By proper choice of these values, the frequency may be adjusted over a very wide range.

Multivibrators

Relaxation oscillators, which are widely used in present-day electronic equipment, are used to produce non-sinusoidal waveshapes such as rectangular and sawtooth pulses. Probably the most common relaxation oscillator is the multivibrator, which may be considered as a two-stage resistance-coupled amplifier in which the output of each tube is coupled into the input of the other tube.

Fig. 100 is a basic multivibrator circuit of the free-running type. In this circuit, oscillations are maintained by the alternate shifting of conduction from one tube to the other. The cycle usually starts with one tube, V_1, at zero bias, and the other, V_2, at cutoff or beyond. At this point, the capacitor C_1 is charged sufficiently to cut off V_2. C_1 then begins to discharge through the resistor R_4, and the voltage on the grid of V_2 rises until V_2 begins to conduct. The voltage on the plate of V_2 then decreases, causing V_1 to conduct less and less. At the same time, the plate voltage of V_1 begins to rise, causing V_2 to conduct still more heavily. Because of the amplification, this cumulative effect builds up extremely fast, and conduction switches from V_1 to V_2

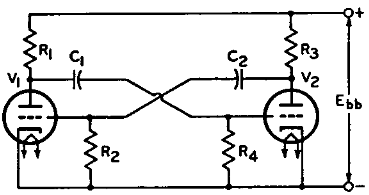


Fig. 100—Basic multivibrator circuit of the free-running type.

within a few microseconds, depending on the circuit components.

In this circuit, therefore, conduction switches from V_1 to V_2 over the interval during which C_1 discharges from the voltage across R_4 to the cutoff voltage for V_2. The actual transfer of conduction does not occur until cutoff is reached. Conduction switches back to V_1 through a similar process to complete the cycle. The plate waveform is essentially rectangular in shape, and may be adjusted as to symmetry, frequency, and amplitude by proper choice of circuit constants, tubes, and voltages.

Although this type of multivibrator is free-running, it may be triggered by pulses of a given amplitude and frequency to provide a frequency-stabilized output. Multivibrator circuits may also be designed so that they are not free-running, but must be triggered externally to shift conduction from one tube to the other. Depending on the type of circuit, conduction may shift back to the first tube after a given time interval, or the second tube may continue conducting until another trigger signal is applied.

Synchroguide Circuits

The "synchroguide" is a controlled type of oscillator used in television receivers to generate and control the synchronized sawtooth voltage necessary for adequate line- or horizontal-frequency scanning. A simplified synchroguide circuit is shown in Fig. 101. This circuit provides stable, noise-free control of a blocking oscillator which generates a horizontal-frequency signal. It permits comparison of the received sync pulses and the generated sawtooth voltages so that properly locked-in horizontal scanning results.

The triode V_2 in Fig. 101 is a conventional blocking oscillator which enables a sawtooth voltage to be developed across the capacitor C_2. A portion of this sawtooth is fed back to the grid of

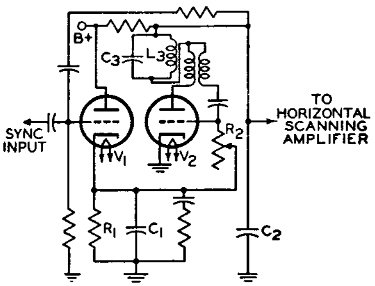


Fig. 101—Simplified synchroguide circuit.

the control tube, V_1. The positive sync pulses are also applied to the grid of

V_1. The waveforms shown in Fig. 102 illustrate the sawtooth and sync pulses (A and B) and their proper "in-sync"

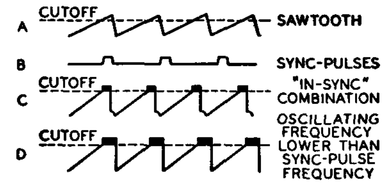


Fig. 102—Sawtooth and sync pulses in synchroguide circuit.

combination (C). The sync pulse occurs partly during the portion of the sawtooth voltage in which the triode V_1 draws current. Any shift in sync pulse as it is superimposed on the sawtooth, therefore, will affect the amount of conduction of the control tube. A change in control-tube conduction ultimately affects the bias on the oscillator-tube grid by changing the voltage to which the capacitor C_1 in the cathode circuit may charge. An increase in the positive bias increases the frequency of oscillation.

For example, waveform D in Fig. 102 illustrates a condition in which the sawtooth voltage is advanced in phase with respect to the sync pulses. The widening of the pulse which occurs at the corner of the sawtooth waveform allows the control tube to conduct more current and, consequently, allows the capacitor C_1 to charge to a higher voltage. This increased reference voltage also appears in the grid circuit of V_2 and makes the grid more positive. The increased grid voltage then speeds up the frequency of oscillations until proper synchronization results.

The blocking oscillator can be made more immune to changes in frequency and noise if V_2 is brought out of cutoff very sharply. This effect is obtained by sine-wave stabilization. The tuned circuit L_3C_3 in the plate circuit of Fig. 101 superimposes a shock-excited sine wave on the plate and grid waveforms, as shown in Fig. 103.

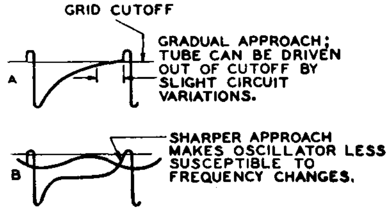


Fig. 103—Waveforms showing effect of tuned circuit L_3C_3 in Fig. 101.

Automatic Frequency Control

An automatic frequency control (afc) circuit provides a means of correcting automatically the intermediate frequency of a superheterodyne receiver when, for any reason, it drifts from the frequency to which the if stages are tuned. This correction is made by adjusting the frequency of the oscillator. Such a circuit will automatically compensate for slight changes in rf carrier or oscillator frequency as well as for inaccurate manual or push-button tuning.

An afc system requires two sections: a frequency detector and a variable reactance. The detector section may be essentially the same as the FM detector illustrated in Fig. 30 and discussed under **Detection.** In the afc system, however, the output is a dc control voltage, the magnitude of which is proportional to the amount of frequency shift. This dc control voltage is used to control the grid bias of an electron tube which comprises the variable reactance section (Fig. 104).

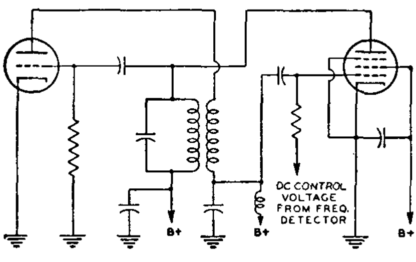


Fig. 104—Automatic-frequency-control (afc) circuit.

The plate current of the reactance tube is shunted across the oscillator tank circuit. Because the plate current and plate voltage of the reactance tube are almost 90 degrees out of phase, the control tube affects the tank circuit in the same manner as a reactance. The grid bias of the tube determines the magnitude of the effiective reactance and, consequently, a control of this grid bias can be used to control the oscillator frequency.

Automatic frequency control is also used in television receivers to keep the horizontal oscillator in step with the horizontal-scanning frequency (15,750 Hz) at the transmitter. A widely used horizontal afc circuit is shown in Fig. 105. This circuit, which is often referred to as a **balanced-phase-detector** or **phase-discriminator** circuit, is usually employed to control the frequency of a multivibrator-type horizontal-oscillator circuit. The 6AL5 detector supplies a dc control voltage to the grid of the horizontal-oscillator tube which counteracts changes in its operating frequency. The magnitude and polarity of the control voltages are determined by phase relationships in the afc circuit at a given moment.

The horizontal sync pulses obtained from the sync-separator circuit are fed through a single-triode phase-inverter or phase-splitter circuit to the two diode units of the 6AL5. Because of the action of the phase-inverter circuit, the signals applied to the two diode units are equal in amplitude but 180 degrees out of phase. A reference sawtooth voltage obtained from the horizontal output circuit is also applied simultaneously to both units. Any change in the oscillator frequency alters the phase relationship between the reference sawtooth and the incoming horizontal sync pulses, causing one diode unit of the 6AL5 to conduct more heavily than the other, and thus producing a correction signal. The system remains balanced at all times, therefore, because momentary changes in oscillator frequency are instantaneously corrected by the action of the control voltage.

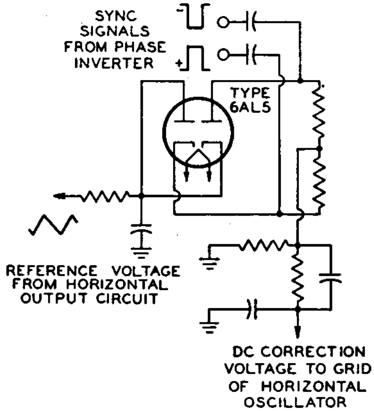


Fig. 105—Balanced phase-detector or phase-discriminator circuit for horizontal afc.

The diode units of the 6AL5 are biased so that conduction takes place only during the tips of the sync pulses. The relative position of the sync pulses on the retrace portion of the sawtooth waveform at any given instant determines which diode unit conducts more heavily, and thereby establishes the magnitude and polarity of the control voltage. The network between the diode units and the grid of the horizontal-oscillator tube is essentially a low-pass filter which prevents the horizontal sync pulses from affecting the horizontal-oscillator performance.

Frequency Conversion

Frequency conversion is used in superheterodyne receivers to change the frequency of the rf signal to an intermediate frequency. To perform this change in frequency, a frequency-converting device consisting of an oscillator and a frequency mixer is employed. In such a device, shown diagrammatically in Fig. 106, two voltages of different frequency, the rf signal voltage and the voltage generated by the oscillator, are applied to the input of the frequency mixer. These voltages beat, or heterodyne, within the mixer tube to produce a plate current having, in addition to

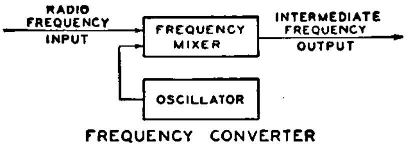


Fig. 106—Block diagram of simple frequency-converter circuit.

the frequencies of the input voltages, numerous sum and difference frequencies.

The output circuit of the mixer stage is provided with a tuned circuit which is adjusted to select only one beat frequency, *i.e.,* the frequency equal to the difference between the signal frequency and the oscillator frequency The selected output frequency is known as the intermediate frequency, or if. The output frequency of the mixer tube is kept constant for all values of signal frequency by tuning the oscillator to the proper frequency.

Important advantages gained in a receiver by the conversion of signal frequency to a fixed intermediate frequency are high selectivity with few tuning stages and a high, as well as stable, overall gain for the receiver.

Several methods of frequency conversion for superheterodyne receivers are of interest. These methods are alike in that they employ a frequency-mixer tube in which plate current is varied at a combination frequency of the signal frequency and the oscillator frequency. These variations in plate current produce across the tuned plate load a voltage of the desired intermediate frequency. The methods differ in the types of tubes employed and in the means of supply input voltages to the mixer tube.

A method widely used before the availability of tubes especially designed for frequency-conversion service, and currently used in many FM, television, and standard broadcast receivers, employs as mixer tube either a triode, a tetrode, or a pentode, in which oscillator voltage and signal voltage are applied to the same grid. In this method, coupling between the oscillator and mixer circuits is obtained by means of inductance or capacitance.

A second method employs a tube having an oscillator and frequency mixer combined in the same envelope. In one form of such a tube, coupling between the two units is obtained by means of the electron stream within the tube. Because five grids are used, the tube is called a pentagrid converter.

Grids No. 1 and No. 2 and the cathode are connected to an external circuit to act as a triode oscillator. Grid No. 1 is the grid of the oscillator and grid No. 2 is the anode. These grids and the cathode can be considered as a composite cathode which supplies to the rest of the tube an electron stream that varies at the oscillator frequency.

This varying electron stream is further controlled by the rf signal voltage on grid No. 4. Thus, the variations in plate current are due to the combination of the oscillator and the signal frequencies. The purpose of grids No. 3 and No. 5, which are connected together within the tube, is to accelerate the electron stream and to shield grid No. 4 electrostatically from the other electrodes.

Pentagrid-converter tubes of this design are good frequency-converting devices at medium frequencies. However, their performance is better at the lower frequencies because the output of the oscillator drops off as the frequency is raised and because certain undesirable effects produced by interaction between oscillator and signal sections of the tube increase with frequency.

To minimize these effects, several of the pentagrid-converter tubes are designed so that no electrode functions alone as the oscillator anode. In these tubes, grid No. 1 functions as the oscillator grid, and grid No. 2 is connected within the tube to the screen grid (grid No. 4). The combined two grids, Nos. 2 and 4, shield the signal grid (grid No. 3) and act as the composite anode of the oscillator triode. Grid No. 5 acts as the suppressor grid.

Converter tubes of this type are designed so that the space charge around the cathode is unaffected by electrons from the signal grid. Furthermore, the electrostatic field of the signal grid also has little effect on the space charge. The result is that rf voltage on the signal grid produces little effect on the cathode current. There is, therefore, little detuning of the oscillator by avc bias because changes in avc bias produce little change in oscillator transconductance or in the input capacitance of grid No. 1.

Examples of the pentagrid converters discussed in the preceding paragraph are the single-ended types 1R5 and 6BE6. A schematic diagram illustrating the use of the 6BE6 with self-excitation is given in Fig. 107. The 6BE6 may also be used with separate excitation. A complete circuit is shown in the **Circuits** section.

Another method of frequency conversion utilizes a separate oscillator having its grid connected to the No. 1 grid of a mixer hexode. The cathode, triode grid, and triode plate form the oscillator unit of the tube. The cathode, hexode mixer grid (grid No. 1), hexode screen grids (grids Nos. 2 and 4), hexode signal grid (grid No. 3), and hexode plate constitute the mixer unit. The internal shields are connected to the shell of the tube and act as a suppressor grid for the hexode unit.

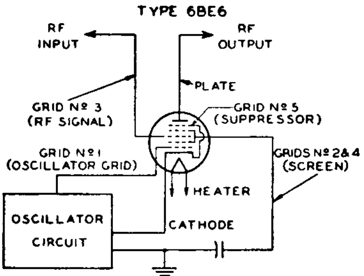


Fig. 107—Frequency-converter circuit using the 6BE6 pentagrid converter with self-excitation.

The action of this tube in converting a radio-frequency signal to an intermediate frequency depends on (1) the generation of a local frequency by the triode unit, (2) the transferring of this frequency to the hexode grid No. 1, and (3) the mixing in the hexode unit of this frequency with that of the rf signal applied to the hexode grid No. 3. The tube is not critical to changes in oscillator-plate voltage or signal-grid bias and, therefore, finds important use in all-wave receivers to minimize frequency-shift effects at the higher frequencies.

A further method of frequency conversion employs a tube called a pentagrid mixer. This type has two independent control grids and is used with a separate oscillator tube. RF signal voltage is applied to one of the control grids and oscillator voltage is applied to the other. It follows, therefore, that the variations in plate current are due to the combination of the oscillator and signal frequencies.

The tube contains a heater-cathode, five grids, and a plate. Grids Nos. 1 and 3 are control grids. The rf signal voltage is applied to grid No. 1. This grid has a remote-cutoff characteristic and is suited for control by avc bias voltage. The oscillator voltage is applied to grid No. 3. This grid has a sharp-cutoff characteristic and produces a comparatively large effect on plate current for a small amount of oscillator voltage. Grids Nos. 2 and 4 are connected together within the tube. They accelerate the electron stream and shield grid No. 3 electrostatically from the other electrodes. Grid No. 5, connected within the tube to the cathode, functions similarly to the suppressor grid in a pentode.

In the converter or mixer stage of a television receiver, stable oscillator operation is most readily obtained when separate tubes or tube sections are used for the oscillator and mixer functions. A typical television mixer-oscillator circuit is shown in Fig. 108. In such circuits, the oscillator voltage is applied to the mixer grid by inductive coupling, capacitive coupling, or a combination of the two. Tubes containing electrically independent oscillator and mixer units in the same envelope, such as the 6U8A and 6X8, are designed especially for this application.

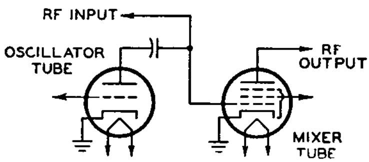


Fig. 108—Typical television mixer-oscillator circuit.

Tuning Indication With Electron-Ray Tubes

Electron-ray tubes are designed to indicate visually by means of a fluorescent target the effects of a change in controlling voltage. One application of them is as tuning indicators in radio receivers. Types such as the 6U5, 6E5, and the 6AB5/6N5 contain two main parts: (1) a triode which operates as a dc amplifier and (2) an electron-ray indicator which is located in the bulb as shown in Fig. 109. The target is operated at a positive voltage and, therefore, attracts electrons from the cathode. When the electrons strike the target they produce a glow on the fluorescent coating of the target. Under these conditions, the target appears as a ring of light.

A ray-control electrode is mounted between the cathode and target. When the potential of this electrode is less positive than the target, electrons flowing to the target are repelled by the electrostatic field of the electrode, and do not reach that portion of the target behind the electrode. Because the target does not glow where it is shielded from electrons, the control electrode casts a

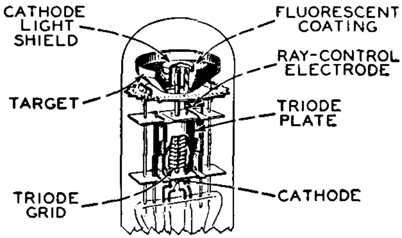


Fig. 109—Structure of electron-ray tube.

shadow on the glowing target. The extent of this shadow varies from approximately 100 degrees of the target when the control electrode is much more negative than the target to 0 degrees when the control electrode is at approximately the same potential as the target.

In the application of the electron-ray tube, the potential of the control electrode is determined by the voltage on the grid of the triode section, as can be seen in Fig. 110. The flow of the triode plate current through resistor R produces a voltage drop which determines the potential of the control electrode. When the voltage of the triode grid changes in the positive direction, plate current increases, the potential of the control electrode goes down because of the increased drop across R, and the shadow angle widens. When the potential of the triode grid changes in the

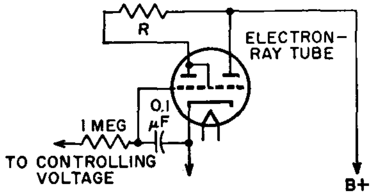


Fig. 110—Indicating circuit using an electron-ray tube.

negative direction, the shadow angle narrows.

Another type of indicator tube is the 6AF6G. This tube contains only an indicator unit but employs two ray-control electrodes mounted on opposite sides of the cathode and connected to individual base pins. It employs an external dc amplifier. (See Fig. 111.) Thus, two symmetrically opposite shadow angles may be obtained by connecting the two ray-control electrodes together; or, two unlike patterns may be obtained by individual connection of each ray-control electrode to its respective amplifier.

In radio receivers, avc voltage is applied to the grid of the dc amplifier.

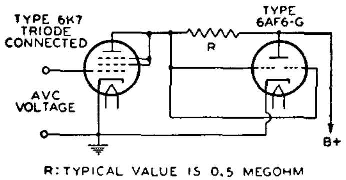


Fig. 111—Indicating circuit using 6AF6G electron-ray tube and external dc amplifier.

Because avc voltage is at maximum when the set is tuned to give maximum response to a station, the shadow angle is at minimum when the receiver is tuned to resonance with the desired station.

The choice between electron-ray tubes depends on the avc characteristic of the receiver. The 6E5 contains a sharp-cutoff triode which closes the shadow angle on a comparatively low value of avc voltage. The 6AB5/6N5 and 6U5 each have a remote-cutoff triode which closes the shadow on a larger value of avc voltage than the 6E5. The 6AF6G may be used in conjunction with dc amplifier tubes having either remote- or sharp-cutoff characteristics.

Electron Tube Installation

THE installation of electron tubes requires care if high-quality performance is to be obtained from the associated circuits. Installation suggestions and precautions which are generally common to all types of tubes are covered in this section. Careful observance of these suggestions will do much to help the experimenter and electronic technician obtain the full performance capabilities of radio tubes and circuits. Additional pertinent information is given under each tube type and in the **Circuits** section.

Filament and Heater Power Supply

The design of electron tubes allows for some variation in the voltage and current supplied to the filament or heater, but most satisfactory results are obtained from operation at the rated values. When the voltage is low, the temperature of the cathode is below normal, with the result that electron emission is limited. The limited emission may cause unsatisfactory operation and reduced tube life. On the other hand, high cathode voltage may cause rapid evaporation of cathode material and shorten tube life.

To insure proper tube operation, it is important that the filament or heater voltage be checked at the socket terminals by means of a high-resistance voltmeter while the equipment is in operation. In the case of series operation of heaters or filaments, correct adjustment can be checked by means of an ammeter in the heater or filament circuit.

The filament or heater voltage supply may be a direct-current source (a battery or a dc power line) or an alternating-current power line, depending on the type of service and type of tube. Frequently, a resistor (either variable or fixed) is used with a dc supply to permit compensation for battery voltage variations or to adjust the tube voltage at the socket terminals to the correct value. Ordinarily, a step-down transformer is used with an ac supply to provide the proper filament or heater voltage. Receivers intended for operation on both dc and ac power lines have the heaters connected in series with a suitable resistor and supplied directly from the power line.

DC filament or heater operation should be considered on the basis of the source of power. In the case of the battery supply for the 1.4-volt filament tubes, it is unnecessary to use a voltage-dropping resistor in series with the filament and a single dry-cell; the filaments of these tubes are designed to operate satisfactorily over the range of voltage variations that normally occur during the life of a dry-cell. Likewise no series resistor is required when the 1.25-volt filament subminiatures are operated from a single 1.5-volt flashlight-type dry-cell, when the 2-volt filament-type tubes are operated from a single storage cell, or when the 6.3-volt series are operated from a 6-volt storage battery.

In the case of dry-battery supply for 2-volt filament tubes, a variable resistor in series with the filament and the battery is required to compensate for battery variations. Turning the set on and off by means of the rheostat is advised to prevent over-voltage conditions after an off-period because the voltage of dry-cells rises during off periods.

In the case of storage-battery supply, air-cell-battery supply, or dc power supply, a non-adjustable resistor of suitable value may be used. It is well to check initial operating conditions, and thus the resistor value, by means of a voltmeter or ammeter.

AC filament or heater operation should be considered on the basis of either a parallel or a series arrangement of filaments and/or heaters. In the case of the parallel arrangements, a step-down transformer is employed. Precautions should be taken to see that the line voltage is the same as that for which the primary of the transformer is designed. The line voltage may be determined by measurement with an ac voltmeter (0-150 volts).

If the line voltage measures in excess of that for which the transformer is designed, a resistor should be placed in series with the primary to reduce the line voltage to the rated value of the transformer primary. Unless this is done, the excess input voltage will cause proportionally excessive voltage to be applied to the tubes. Any electron tube may be damaged or made inoperative by excessive operating voltages.

If the line voltage is consistently below that for which the primary of the transformer is designed, it may be necessary to install a booster transformer between the ac outlet and the transformer primary. Before such a transformer is installed, the ac line fluctuations should be very carefully noted. Some radio sets are equipped with a line-voltage switch which permits adjustment of the power transformer primary to the line voltage. When this switch is properly adjusted, the series-resistor or booster-transformer method of controlling line voltage is seldom required.

In the case of the series arrangements of filaments and/or heaters, a voltage-dropping resistance in series with the heaters and the supply line is usually required. This resistance should be of such value that, for normal line voltage tubes will operate at their rated heater or filament current. The method for calculating the resistor value is given below.

When the filaments of battery-type tubes are connected in series, the total filament current is the sum of the current due to the filament supply and the plate and grid-No. 2 currents (cathode current) returning to B(—) through the tube filaments. Consequently, in a series filament string it is necessary to add shunt resistors across each filament section to bypass this cathode current in order to maintain the filament voltage at its rated value.

The **filament** or **heater resistor** required when filaments and/or heaters are operated in parallel can be determined easily by a simple formula derived from Ohm's law.

$$\text{Required resistance (ohms)} = \frac{\text{supply volts} - \text{rated volts of tube type}}{\text{total rated filament current (amperes)}}$$

Thus, if a receiver using two IT4's, one IR5, one IU5, and one 3V4 is to be operated from a storage battery, the series resistor is equal to 2 volts (the voltage from a single storage cell) minus 1.4 volts (voltage rating for these tubes) divided by 0.3 ampere (the sum of 4×0.05 ampere $+ 1 \times 0.1$ ampere), *i.e.*, approximately 2 ohms. Because this resistor should be variable to allow adjustment for battery depreciation, it is advisable to obtain the next larger commercial size, although any value between 2 and 3 ohms will be quite satisfactory.

Where much power is dissipated in the resistor, the wattage rating should be sufficiently large to prevent overheating. The power dissipation in watts is equal to the voltage drop in the resistor multiplied by the total filament current in amperes. Thus, for the example above, $0.6 \times 0.3 = 0.18$ watt. In this case, the value is so small that any commercial rheostat with suitable resistance will be adequate.

For the case where the heaters and/or filaments of several tubes are operated in series, the resistor value is calculated by the following formula, also derived from Ohm's law.

$$\text{Required resistance (ohms)} = \frac{\text{supply volts} - \text{total rated volts of tubes}}{\text{rated amperes of tubes}}$$

Thus, if a receiver having one 6BE6, one 6BA6, one 6AT6, one 25L6GT, and one 25Z6GT is to be operated from a 117-volt power line, the series resistor is equal to 117 volts (the supply voltage) minus 68.9 volts (the sum of 3 $\times$ 6.3 volts + 2 $\times$ 25 volts) divided by 0.3 ampere (current rating of these tubes), *i.e.*, approximately 160 ohms. The wattage dissipation in the resistor will be 117 volts minus 68.9 volts times 0.3 ampere, or approximately 14.4 watts. A resistor having a wattage rating in excess of this value should be chosen.

When the series-heater connection is used in ac/dc receivers, it is usually advisable to arrange the heaters in the circuit so that the tubes most sensitive to hum disturbances are at or near the ground potential of the circuit. This arrangement reduces the amount of ac voltage between the heaters and cathodes of these tubes and minimizes the hum output of the receiver. The order of heater connection, by tube function, from chassis to the rectifier-cathode side of the ac line is shown in Fig. 112.

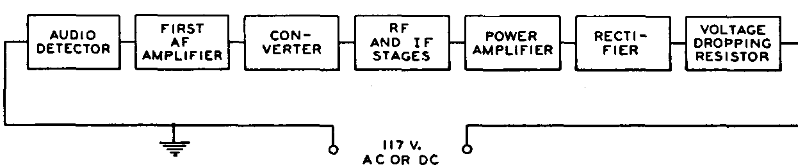


Fig. 112—Order of series heater-string connection, by tube function, to minimize hum.

Heater-to-Cathode Connection

When heater-type tubes are operated from ac, their cathodes may be returned (through resistors, capacitors, or other components) to the mid-tap on the heater supply winding, to the mid-tap of a small resistor (about 50 ohms) connected across the winding, or to one end of the heater supply winding, depending on circuit requirements. In all circuits, it is important to keep the heater-cathode voltage within the maximum ratings specified for the tube.

Heater-type tubes may produce hum as a result of conduction between heater and cathode or between heater and control grid, or by modulation of the electron stream by the alternating magnetic field surrounding the heater. When a large resistor is used between heater and cathode (as in series-connected heater strings), or when one side of the heater is grounded, even a minute pulsating leakage current between heater and cathode can develop a small voltage across the cathode-circuit impedance and cause objectionable hum. The use of a large cathode bypass capacitor is recommended to minimize this source of hum.

Much lower hum levels can be achieved when heaters are connected in parallel systems in which the center-tap of the heater supply is grounded or, preferably, connected to a positive bias source of 15 to 80 volts dc to reduce the flow of alternating current. The heater leads of the tubes should be twisted and kept awav from high-impedance circuits. The balanced ac supply provides almost complete cancellation of the alternating-current components.

The balanced arrangement described above also minimizes heater-grid hum. High grid-circuit impedances should be avoided, if possible. High heater voltages should also be avoided because heater-cathode hum rises sharply when the heater voltage is increased above the published value.

Certain tube types are designed especially to minimize hum in high-quality, high-fidelity audio equipment. Examples are the 5879, 7025, and 7199.

Plate Voltage Supply

The plate voltage for electron tubes is obtained from batteries, rectifiers, direct-current power lines, and small local generators. The maximum plate-voltage value for any tube type should not be exceeded if most satisfactory performance is to be obtained. Plate voltage should not be applied to a tube unless the corresponding recommended voltage is also supplied to the grid.

It is recommended that the primary circuit of the power transformer be fused to protect the rectifier tube(s), the power transformer, filter capacitor, and chokes in case a rectifier tube fails.

Grid Voltage Supply

The recommended grid voltages for different operating conditions have been carefully determined to give the most satisfactory performance. Grid voltage may be obtained from a fixed source such as a separate C-battery or a tap on the voltage divider of the high-voltage dc supply, from the voltage drop across a resistor in the cathode circuit, or from the voltage drop across a resistor in the grid circuit. The first method is called "fixed bias"; the second is called "cathode bias" or "self bias"; the third is called "grid-resistor bias" and is sometimes incorrectly referred to in receiving-tube practice as "zero-bias operation."

In any case, the object is to make the grid negative with respect to the cathode by the specified voltage. When a C-battery is used, the negative terminal is connected to the grid return and the positive terminal is connected to the negative filament socket terminal, or to the cathode terminal if the tube is of the heater-cathode type. If the filament is supplied with alternating current, this connection is usually made to the center-tap of a low resistance (20 to 50 ohms) shunted across the filament terminals. This method reduces hum disturbances caused by the ac supply. If bias voltages are obtained from the voltage divider of a high-voltage dc supply, the grid return is connected to a more negative tap than the cathode.

The **cathode-biasing** method utilizes the voltage drop produced by the cathode current flowing through a resistor connected between the cathode and the negative terminal of the B-supply. (See Fig. 113.) The cathode current is, of course, equal to the plate current in the case of a triode, or to the sum of the plate and grid-No. 2 currents in the case of a tetrode, pentode, or beam power tube. Because the voltage drop along the resistance is increasingly negative with respect to the cathode, the required negative grid-bias voltage can be obtained by connecting the grid return to the negative end of the resistance.

The value of the resistance for cathode-biasing a single tube can be determined from the following formula:

$$\text{Resistance (ohms)} = \frac{\text{desired grid-bias voltage} \times 1000}{\text{rated cathode current in milliamperes}}$$

Thus, the resistance required to produce 9 volts bias for a triode which operates at 3 milliamperes plate current is $9 \times 1000/3 = 3000$ ohms. If the cathode current of more than one tube passes through the resistor, or if the tube or tubes employ more than three electrodes, the total current determines the size of the resistor.

Bypassing of the cathode-bias resistor depends on circuit-design requirements. In rf circuits the cathode resistor usually is bypassed. In af circuits the use of an unbypassed resistor will re-

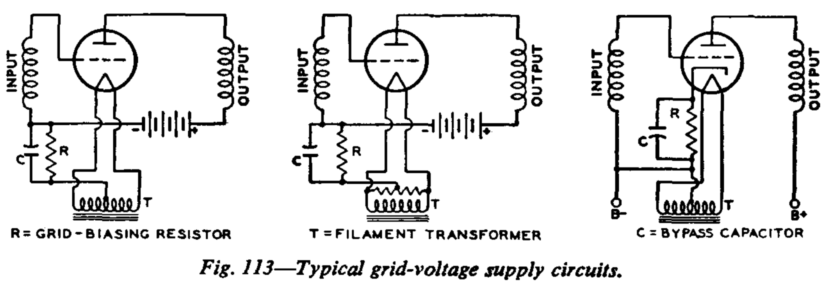


Fig. 113—Typical grid-voltage supply circuits.

duce distortion by introducing degeneration into the circuit. However, the use of an unbypassed resistor decreases gain and power sensitivity. When bypassing is used, it is important that the bypass capacitor be sufficiently large to have negligible reactance at the lowest frequency to be amplified.

In the case of power-output tubes having high transconductance, such as beam power tubes, it may be necessary to shunt the bias resistor with a small mica capacitor (approximately 0.001μF) in order to prevent oscillations. The usual af bypass may or may not be used, depending on whether or not degeneration is desired. In tubes having high values of transconductance, such as the 6BA6, 6CB6, and 6AC7, input capacitance and input conductance change appreciably with plate current. When such a tube having a separate suppressor-grid connection is used as an rf amplifier, these changes may be minimized by leaving a certain portion of the cathode-bias resistor unbypassed. In order to minimize feedback when this method is used, the external grid-No. 1-to-plate (wiring) capacitances should be kept to a minimum, the grid No. 2 should be bypassed to ac ground, and the grid No. 3 should be connected to ac ground.

The use of a cathode resistor to obtain bias voltage is not recommended for amplifiers in which there is appreciable shift of electrode currents with the application of a signal. In such amplifiers, a separate fixed supply is recommended.

The **grid-resistor biasing** method is also a self-bias method because it utilizes the voltage drop across the grid resistor produced by small amounts of grid current flowing in the grid-cathode circuit. This current is due to (1) an electromotive potential difference between the materials comprising the grid and cathode and (2) grid rectification when the grid is driven positive. A large value of resistance is required in order to limit this current to a very small value and to avoid undesirable loading effects on the preceding stage.

Examples of this method of bias are given in the **Circuits** section. In these circuits, the audio amplifier type 1U5 or 12AV6 has a 10-megohm resistor between the grid and the negative filament or cathode to furnish the required bias, which is usually less than 1 volt. This method of biasing is used principally in the early voltage-amplifier stages (usually employing high-mu triodes) of audio amplifier circuits, where the tube dissipation will not be excessive under zero-signal conditions.

A grid resistor is also used in many oscillator circuits for obtaining the required bias. In these circuits, the grid voltage is relatively constant and its magnitude is usually in the order of 5 volts or more. Consequently, the bias voltage is obtained only through grid rectification. A relatively low value of resistor, 0.1 megohm or less, is used. Oscillator circuits employing this method of bias are given in the **Circuits** section.

Grid-bias variation for the rf and if amplifier stages is a convenient and frequently used method for controlling receiver volume. The variable voltage supplied to the grid may be obtained: (1) from a variable cathode resistor as shown in Figs. 114 and 115; (2) from a bleeder circuit by means of a potentiometer as shown in Fig. 116; or (3) from a bleeder circuit in which the bleeder current is varied by a tube

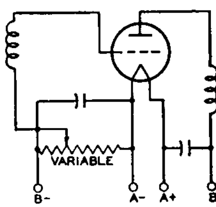


Fig. 114—Amplifier stage using a variable cathode-bias resistor for volume control.

used for automatic volume control. The latter circuit is shown in Fig 61.

In all cases it is important that the control be arranged so that at no time will the bias be less than the recommended minimum grid-bias voltage for the particular tubes used. This requirement can be met by providing a fixed stop on the potentiometer, by connecting a fixed resistance in series with the variable resistance, or by connecting a fixed cathode resistance in series with the variable resistance used for regulation. Where receiver gain is controlled by grid-bias variation, it is advisable to have the control voltages extend over a wide range in order to minimize cross-modulation and modulation-distortion. A remote-cutoff type of tube should, therefore, be used in the controlled stages.

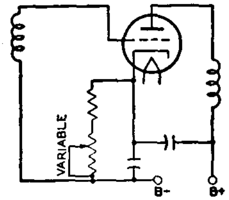


Fig. 115—Amplifier stage similar to Fig. 114 but using heater-cathode-type tube.

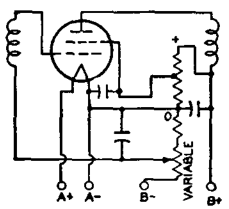


Fig. 116—Amplifier stage using a bleeder circuit and potentiometer for volume control.

In most tubes employing a unipotential cathode, a **positive grid current** begins to flow when the grid is slightly negative and increases rapidly as the grid is made more positive, as shown in Fig. 117. The value of grid voltage at which the grid-current curve intercepts the horizontal axis is determined by several different physical processes, including an electrothermal effect due to the differences in temperature and in material composition of the grid and the cathode, and by the positive grid current. For values of grid potentials which are larger than this intercept, the direction of the grid current is positive (*i.e.,* from the cathode to the grid). At smaller values of grid potential, the direction of the grid current is negative (*i.e.,* from the grid to the cathode).

Positive grid current consists of electrons emitted from the cathode which are intercepted by the control grid. Negative grid current, which becomes appreciable only when the grid potential is more negative than the value of the intercept, is a result of the emission of electrons from the heated control grid to the cathode, the effect of gas molecules in the tube, and the influence of leakage currents between the grid and cathode and the grid and the plate.

The value of grid potential at the intercept of the grid-current curve on the horizontal axis (often mistakenly called **contact potential**) may be as high as 1½ volts. If the operating bias of the tube is less than this intercept, it is found that two effects are present. Direct current flows in the grid circuit, and the dynamic input resistance of the tube may be relatively low. It is generally desirable to supply the tube with a value of bias sufficiently high so that the operating point of the tube is not near the value of this intercept. If the value of the operating bias is near the value of the intercept, care should be taken to avoid undesirable effects in the grid circuit due to grid current or low input resistance.

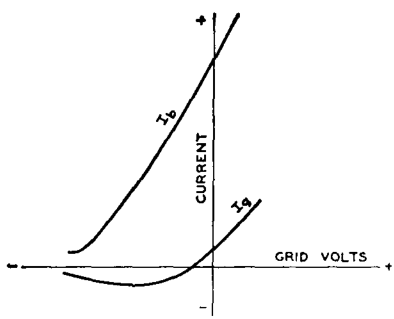


Fig. 117—Curves showing flow of positive grid current in tubes employing unipotential cathodes.

Screen-Grid Voltage Supply

The positive voltage for the screen grid (grid No. 2) of screen-grid tubes may be obtained from a tap on a voltage divider, from a potentiometer, or from a series resistor connected to a high-voltage source, depending on the particular tube type and its application. The screen-grid voltage for tetrodes should be obtained from a voltage divider or a potentiometer rather than through a series resistor from a high-voltage source because of the characteristic screen-grid current variations of tetrodes. Fig. 118 shows a tetrode with its screen-grid voltage obtained from a potentiometer.

When pentodes or beam power tubes are operated under conditions where a large shift of plate and screen-grid currents does not take place with the application of the signal, the screen-grid voltage may be obtained through a series resistor from a high-voltage source. This method of supply is possible because of the high uniformity of the screen-grid current characteristic in pentodes and beam power tubes. Because the screen-grid voltage rises with increase in bias and resulting decrease in screen-grid current, the cutoff characteristic of a pentode is extended by this method of supply.

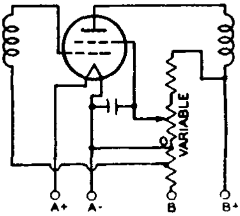


Fig. 118—Tetrode circuit in which screen-grid voltage is obtained from a potentiometer.

This method is sometimes used to increase the range of signals which can be handled by a pentode. When used in resistance-coupled amplifier circuits employing pentodes in combination with the cathode-biasing method, it minimizes the need for circuit adjustments. Fig. 119 shows a pentode with its screen-grid voltage supplied through a series resistor.

When power pentodes and beam power tubes are operated under conditions such that there is a large change in plate and screen-grid currents with the application of signal, the series-resistor method of obtaining screen-grid voltage should not be used. A change in screen-grid current appears as a change in the voltage drop across the series resistor in the screen-grid circuit; the result is a change in the power output and an increase in distortion. The screen-grid voltage should be obtained from a point in the plate-voltage-supply filter system having the correct voltage, or from a separate source.

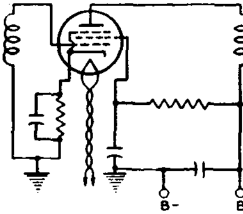


Fig. 119—Pentode circuit in which screen-grid voltage is supplied through a series resistor.

It is important to note that the plate voltage of tetrodes, pentodes, and beam power tubes should be applied before or simultaneously with the screen-grid voltage. Otherwise, with voltage on the screen grid only, the screen-grid current may rise high enough to cause excessive screen-grid dissipation.

Screen-grid voltage variation for the rf amplifier stages has sometimes been used for volume control in older-type receivers. Reduced screen-grid voltage decreases the transconductance of the tube and results in reduced gain per stage. The voltage variation is obtained by means of a potentiometer shunted across the screen-grid voltage supply. (See Fig. 118.) When the screen-grid voltage is varied, it must never exceed the rating of the tube. This requirement can be met by providing a fixed stop on the potentiometer.

Shielding

In high-frequency stages having high gain, the output circuit of each stage must be shielded from the input circuit of that stage. Each high-frequency stage also must be shielded from the other high-frequency stages. Unless shielding is employed, undesired feedback may occur and may produce many harmful effects on receiver performance.

To prevent this feedback, it is a desirable practice to shield separately each unit of the high-frequency stages. For instance, in a superheterodyne receiver, each if and rf coil may be mounted in a separate shield can. Baffle plates may be mounted on the ganged tuning capacitor to shield each section of the capacitor from the other section. The oscillator coil may be especially well shielded by being mounted under the chassis.

The shielding precautions required in a receiver depend on the design of the receiver and the layout of the parts. In all receivers having high-gain high-frequency stages, it is necessary to shield separately each tube in high-frequency stages. When metal tubes, and in particular the single-ended types, are used, complete shielding of each tube is provided by the metal shell which is grounded through its grounding pin at the socket terminal. The grounding connection should be short and sturdy. Many modern tubes of glass construction have internal shields, usually connected to the cathode; where present, these shields are indicated in the socket diagram.

Dress of Circuit Leads

At high frequencies such as are encountered in FM and television receivers, lead dress, that is, the location and arrangement of the leads used for connections in the receiver, is very important. Because even a short lead provides a large impedance at high frequencies, it is necessary to keep all high-frequency leads as short as possible. This precaution is especially important for ground connections and for all connections to bypass capacitors and high-frequency filter capacitors. The ground connections of plate and screen-grid bypass capacitors of each tube should be kept short and made directly to cathode ground.

Particular care should be taken with the lead dress of the input and output circuits of high-frequency stages so that the possibility of stray coupling is minimized. Unshielded leads connected to shielded components should be dressed close to the chassis. As the frequency increases, the need for careful lead dress becomes increasingly important.

In high-gain audio amplifiers, these same precautions should be taken to minimize the possibility of self-oscillation.

Filters

Feedback effects also are caused in radio or television receivers by coupling between stages through common voltage-supply circuits. Filters find an important use in minimizing such effects. They should be placed in voltage-supply leads to each tube in order to return the signal current through a low-impedance path direct to the tube cathode rather than by way of the voltage-supply circuit. Fig. 120 illustrates several forms of filter circuits. Capacitor C

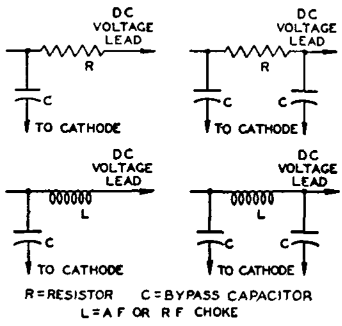


Fig. 120—Typical filter circuits.

forms the low-impedance path, while the choke or resistor assists in diverting the signal through the capacitor by offering a high impedance to the power-supply circuit.

The choice between a resistor and a choke depends chiefly upon the permissible dc voltage drop through the filter. In circuits where the current is small (a few milliamperes), resistors are practical; where the current is large or regulation important, chokes are more suitable.

The minimum practical size of the capacitors may be estimated in most cases by the following rule: The impedance of the capacitor at the lowest frequency amplified should not be more *than one-fifth of the impedance of the* filter choke or resistor at that frequency. Better results will be obtained in special cases if the ratio is not more than one-tenth.

Radio-frequency circuits, particularly at high frequencies, require high-quality capacitors. Mica or ceramic capacitors are preferable. Where stage shields are employed, filters should be placed within the shield.

Another important application of filters is to smooth the output of a rectifier tube. (See **Rectification.**) A smoothing filter usually consists of capacitors and iron-core chokes. In any filter-design problem, the load impedance must be considered as an integral part of the filter because the load is an important factor in filter performance. Smoothing effect is obtained from the chokes because they are in series with the load and offer a high impedance to the ripple voltage. Smoothing effect is obtained from the capacitors because they are in parallel with the load and store energy on the voltage peaks; this energy is released on the voltage dips and serves to maintain the voltage at the load substantially constant. Smoothing filters are classified as choke-input or capacitor-input according to whether a choke or capacitor is placed next to the rectifier tube. (See Fig. 121.)

The **Circuits** section gives a number of examples of rectifier circuits with recommended filter constants.

If an input capacitor is used, consideration must be given to the instantaneous peak value of the ac input voltage. This peak value is about 1.4 times the rms value as measured by an ac voltmeter. Filter capacitors, therefore, especially the input capacitor, should have a rating high enough to withstand the instantaneous peak value if breakdown is to be avoided. When the input-choke method is used, the available dc output voltage will be somewhat lower than with the input-capacitor method for a given ac plate voltage. However, improved regulation together with lower peak current will be obtained.

Mercury-vapor and gas-filled rectifier tubes occasionally produce a form of local interference in radio receivers through direct radiation or through the power line. This interference is generally identified in the receiver as a broadly tunable 120-Hz buzz (100 Hz for 50-Hz supply line, etc.). It is usually caused by the formation of a steep wave front when plate current within the tube begins to flow on the positive half of each cycle of the ac supply voltage.

There are several ways of eliminating this type of interference. One is to shield the tube. Another is to insert an rf choke having an inductance of one millihenry or more between each plate and transformer winding and to connect high-voltage, rf bypass capacitors between the outside ends of the transformer winding and the center tap. (See Fig. 122.) The rf chokes should be placed within the shielding of the tube. The rf bypass capacitors should have a voltage rating high enough to withstand the peak voltage of each half of the secondary, which is approximately 1.4 times the rms value.

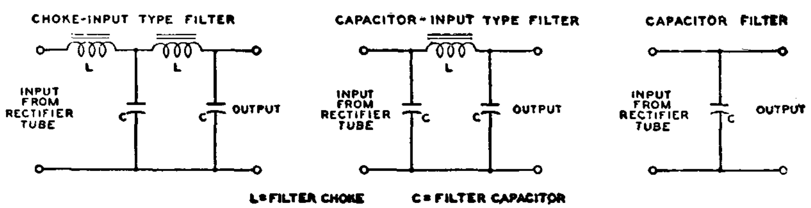


Fig. 121—Typical smoothing filters for rectifier tubes.

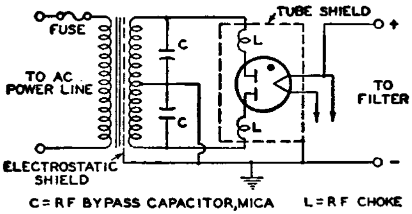


Fig. 122—Filter circuit used to eliminate interference produced by mercury-vapor or gas-filled rectifier tubes.

Transformers having electrostatic shielding between primary and secondary are not likely to transmit rf disturbances to the line. Often the interference may be eliminated simply by making the plate leads of the rectifier extremely short. In general, the particular method of interference elimination must be selected by experiment for each installation.

Output Coupling Devices

An output-coupling device is used in the plate circuit of a power output tube to keep the comparatively high dc plate current from the winding of an electromagnetic speaker and, also, to transfer power efficiently from the output stage to a loudspeaker of either the electromagnetic or dynamic type.

Output-coupling devices are of two types, (1) choke-capacitor and (2) transformer. The choke-capacitor type includes an iron-core choke having an inductance of not less than 10 henries which is placed in series with the plate and B-supply. The choke offers a very low resistance to the dc plate current component of the signal voltage but opposes the flow of the fluctuating component. A bypass capacitor of 2 to 6 microfarads supplies a path to the speaker winding for the signal voltage. The choke-coil output coupling device, however, is now only of historical interest.

The transformer type is constructed with two separate windings, a primary and a secondary wound on an iron core. This construction permits designing each winding to meet the requirements of its position in the circuit. Typical arrangements of each type of coupling device are shown in Fig. 123. Examples of transformers for push-pull stages are shown in several of the circuits given in the **Circuits** section.

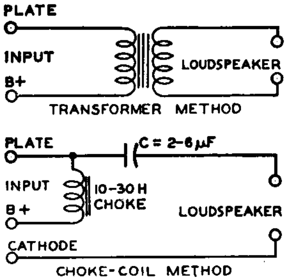


Fig. 123—Typical output-coupling devices.

High-Fidelity Systems

The results achieved from any high-fidelity amplifier system depend to a large degree upon the skill and care with which the system is constructed. Improper placement of transformers, other components, and wiring, and attempts to achieve excessive compactness, can only result in instability, oscillation, hum, and other operating difficulties, as well as in damage to components by overheating. It is important, therefore, that construction of high-fidelity amplifier systems be undertaken only by persons who have had some experience in the layout, mechanical construction, and wiring of audio equipment.

It is impractical to give specific construction data for various amplifiers and supplementary units because the best arrangement for each unit or combination of units will depend on the requirements of the user. It is possible, however, to list some general considerations which should be observed in the construction of any high-fidelity amplifier system.

Any amplifier having two or more stages should be constructed with a straight-line layout so that maximum separation is provided between the signal input and output circuits and terminals. Power-supply connections, particularly those carrying ac, should be isolated as far as possible from signal connections, especially from the input connection. Signal-carrying conductors, even when shielded, should not be cabled together with power-supply conductors. Internal wiring for ac-operated tube heaters, switches, pilot-light sockets, and other devices, should be twisted and placed flat against the chassis. All connections to the ground side of the circuit in each unit should be made to a common bus of heavy wire. This bus should be connected to the chassis only at the point of minimum signal voltage, *i.e.*, at the signal-input terminal of the unit.

All internal wiring that carries signal voltages should be as short as possible, and as far as possible above the chassis, to minimize losses at the higher audio frequencies due to stray shunt capacitance. All connections between units should be made with shielded cable having a capacitance of not more than 30 picofarads per foot, such as Alpha Type 1249 or 1704, Belden Type 8401 or 8410, or equivalent cable.

Because power amplifiers and power-supply units of high-fidelity systems normally dissipate large amounts of heat, they should be constructed and installed in such a manner as to assure adequate ventilation for the tubes and other components. A beam power tube or rectifier tube should be separated from any other tube or component on the same side of the chassis by at least 1½ tube diameters.

Power amplifiers and power-supply units which are to be installed horizontally (*i.e.*, with the tubes vertical) in cabinets or on shelves should be provided with mounting feet, perforated bottom covers, and a number of small holes around each tube socket to permit relatively cool air to enter from below and provide ventilation for the under side of the chassis and tubes.

If a power amplifier, tone-control amplifier, and one or more preamplifiers are to be constructed on the same chassis, the mechanical layout should be planned so that the circuits operating at the lowest signal levels are farthest from the output stage and power supply. Amplifier units which normally operate at comparable signal levels but are not used simultaneously (such as preamplifiers for tape pickup heads and magnetic phonograph pickups) may be installed side by side on the same chassis without danger of interaction. Units which operate simultaneously, however (such as the channels of a stereophonic system), should not be installed side by side on the same chassis without careful consideration to placement of components and wiring, and the possible use of shielding to prevent interaction.

When an amplifier, preamplifier, mixer, or other unit requiring heater power is located more than five or six feet from its power-supply unit, the heater-current conductors in the power-supply cable must be large enough to assure that each tube receives its rated heater voltage. In cases where very large heater currents or very long power-supply cables are involved, it may be desirable to install a heater-supply transformer on or near the amplifier unit. If such a transformer is installed on or near a preamplifier for a magnetic-tape pickup head, a magnetic phonograph pickup, or a dynamic microphone, the transformer should be completely shielded and positioned to prevent its field from inducing hum in the pickup device.

High-Voltage Considerations for Television Picture Tubes

Like other high-voltage devices, television picture tubes require that certain precautions be observed to minimize the possibility of failure caused by humidity, dust, and corona.

Humidity Considerations. When humidity is high, a continuous film of moisture may form on the glass bulb immediately surrounding the anode cavity cap of all-glass picture tubes or on the glass part of the envelope of metal picture tubes. This film may permit sparking to take place over the glass surface to the external conductive coating or to the metal shell. Such sparking may introduce noise into the receiver. To prevent such a possibility, the uncoated bulb surface around the cap and the glass part of the envelope of metal picture tubes should be kept clean and dry.

Dust Considerations. The accumulation of dust on the uncoated area of the bulb around the anode cap of all-glass picture tubes or on the glass part of the envelope or insulating supports for metal picture tubes will decrease the insulating qualities of these parts. The dust usually consists of fibrous materials and may contain soluble salts. The fibers absorb and retain moisture; the soluble salts provide electrical leakage paths that increase in conductivity as the humidity increases. The resulting high leakage currents may overload the high-voltage power supply.

It is recommended, therefore, that the uncoated bulb surface of all-glass picture tubes and the coated glass surface and insulating supports for metal picture tubes be kept clean and free from dust or other contamination such as finger-prints. The frosted Filterglass faceplate of the metal picture tubes may be cleaned with a soapless detergent, such as Dreft, then rinsed with clean water, and immediately dried.

Corona Considerations. A high-voltage system may be subject to corona, especially when the humidity is high, unless suitable precautions are taken. Corona, which is an electrical discharge appearing on the surface of a conductor when the voltage gradient exceeds the breakdown value of air, causes deterioration of organic insulating materials through formation of ozone, and induces arc-over at points and sharp edges. Sharp points or other irregularities on any part of the high-voltage system may increase the possibility of corona and should be avoided.

In the metal-shell picture tubes, the metal lip at the maximum diameter has rounded edges to prevent corona. Adequate spacing between the lip and any grounded element in the receiver, or between the small end of the metal shell and any grounded element, should be provided to preclude the possibility of corona. Such spacing should not be less than 1 inch of air. Similarly, an air space of 1 inch, or equivalent, should be provided around the body of the metal shell. As a further precaution to prevent corona, the deflecting-yoke surface on the end adjacent to the shell should present a smooth electrical surface with respect to the small end of the metal shell or the anode terminal of all-glass tubes.

Picture-Tube Safety Considerations

Tube Handling. Breakage of picture tubes, which contain a high vacuum, may result in injury from flying glass. Do not strike or scratch the tube or subject it to more than moderate pressure when installing it in or removing it from electronic equipment.

High-Voltage Precautions. In picture-tube circuits, high voltages may appear at normally low-potential points in the circuit because of capacitor breakdown or incorrect circuit connections. Therefore, before any part of the circuit is touched the power-supply switch should be turned off, the power plug disconnected, and both terminals of any capacitors grounded.

X-Ray Radiation Precautions. All types of picture tubes may be operated at voltages (if ratings permit) up to 16 kilovolts without producing harmful x-ray radiation or danger of personal injury on prolonged exposure at close range. Above 16 kilovolts, special x-ray shielding precautions may be necessary

Interpretation of Tube Data

THE tube data given in the following **Technical Data** section include ratings, typical operation values, characteristics, and characteristic curves.

The values for grid-bias voltages, other electrode voltages, and electrode supply voltages are given with reference to a specified **datum point** as follows: For types having filaments heated with dc, the negative filament terminal is taken as the datum point to which other electrode voltages are referred. For types having filaments heated with ac, the mid-point (*i.e.*, the center tap on the filament-transformer secondary, or the mid-point on a resistor shunting the filament) is taken as the datum point. For types having unipotential cathodes indirectly heated, the cathode is taken as the datum point.

Ratings are established on electron tube types to help equipment designers utilize the performance and service capabilities of each tube type to best advantage. Ratings are given for those characteristics which careful study and experience indicate must be kept within certain limits to insure satisfactory performance.

Three rating systems are in use by the electron-tube industry. The oldest is known as the Absolute Maximum system, the next as the Design Center system, and the latest and newest as the Design Maximum system. Definitions of these systems have been formulated by the Joint Electron Device Engineering Council (JEDEC) and standardized by the National Electrical Manufacturers Association (NEMA) and the Electronic Industries Association (EIA) as follows:

Absolute Maximum ratings are limiting values which should not be exceeded with any tube of the specified type under any condition of operation. These ratings are used only in rare instances for receiving types, but are generally used for transmitting and industrial types.

Design Center ratings are limiting values which should not be exceeded with a tube of the specified type having characteristics equal to the published values under normal operating conditions. These ratings, which include allowances for normal variations in both tube characteristics and operating conditions, were used for most receiving tubes prior to 1957.

Design Maximum ratings are limiting values which should not be exceeded with a tube of the specified type having characteristics equal to the published values under any conditions of operation. These ratings include allowances for normal variations in tube characteristics, but do not provide for variations in operating conditions. Design Maximum ratings were adopted for receiving tubes in 1957.

Electrode voltage and current ratings are in general self-explanatory, but a brief explanation of other ratings will aid in the understanding and interpretation of tube data.

Heater warm-up time is defined as the time required for the voltage across the heater to reach 80 per cent of the rated value in the circuit shown in Fig. 124. The heater is placed in series with a resistance having a value 3 times the nominal heater operating resistance

($R = 3\ E_f/I_f$), and a voltage having a value 4 times the rated heater voltage ($V = 4\ E_f$) is then applied. The warm-up time is determined when $E = 0.8\ E_f$.

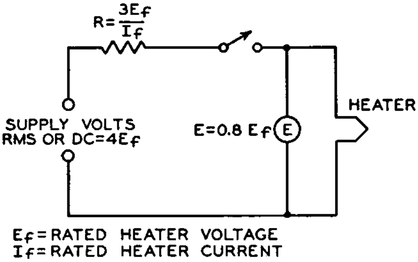


Fig. 124—Test circuit for measuring heater warm-up time.

Plate dissipation is the power dissipated in the form of heat by the plate as a result of electron bombardment It is the difference between the power supplied to the plate of the tube and the power delivered by the tube to the load.

Peak heater-cathode voltage is the highest instantaneous value of voltage that a tube can safely stand between its heater and cathode. This rating is applied to tubes having a separate cathode terminal and used in applications where excessive voltage may be introduced between heater and cathode.

Maximum dc output current is the highest average plate current which can be handled continuously by a rectifier tube. Its value for any rectifier tube type is based on the permissible plate dissipation of that type. Under operating conditions involving a rapidly repeating duty cycle (steady load), the average plate current may be measured with a dc meter.

The nomograph shown in Fig. 125 can be used to determine tube voltage drop or plate current for any diode unit when values for a single plate-voltage, plate-current condition are available from the data. It can also be used to compare the relative perveance ($G = I_b/E_b\,^{3/2}$) of several diodes. **Perveance** can be considered a figure of merit for diodes; high-perveance units have lower voltage drop at a fixed current level.

Tube voltage drop or plate current for a specific diode unit can be determined as follows: First, convenient values are selected for the plate-voltage and plate-current scales of the nomograph. The published plate-current and plate-voltage values are then located on the scales and connected with a straight edge. The intersection of the connecting line with the perveance scale is then used as a pivot point to determine the value of tube voltage drop corresponding to a desired current value, or the value of plate current corresponding to a desired tube voltage drop. Because the pivot point for a specific diode unit represents its perveance, the pivot points for several units (plotted to the same scales) can be used to compare their relative perveance.

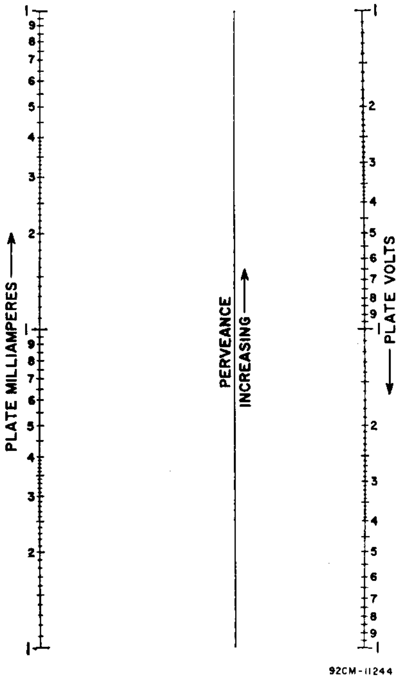


Fig. 125—Diode perveance nomograph.

For example, type 5U4GB has a tube voltage drop (per plate) of 44 volts at a plate current of 225 milliamperes. Convenient scales for this type are from 1 to 100 volts for plate voltage and from 10 to 1000 milliamperes for plate current. The points 44 volts and 225 milliamperes are then connected with a straight line to determine the pivot point. Using this pivot point, it is easy to determine such values as a plate current of 150 milliamperes at a tube voltage drop of 33 volts, or a voltage drop of 25 for a current of 100 milliamperes.

For readings in the order of one volt and/or one milliampere, the nomograph is not accurate because of the effects of contact potential and initial electron velocity.

Maximum peak plate current is the highest instantaneous plate current that a tube can safely carry recurrently in the direction of normal current flow. The safe value of this peak current in hot-cathode types of rectifier tubes is a function of the electron emission available and the duration of the pulsating current flow from the rectifier tube in each half-cycle.

The value of peak plate current in a given rectifier circuit is largely determined by filter constants. If a large choke is used at the filter input, the peak plate current is not much greater than the load current; but if a large capacitor is used as the filter input, the peak current may be many times the load current. In order to determine accurately the peak plate current in any rectifier circuit, measure it with a peak-indicating meter or use an oscillograph.

Maximum peak inverse plate voltage is the highest instantaneous plate voltage which the tube can withstand recurrently in the direction opposite to that in which it is designed to pass current. For mercury-vapor tubes and gas-filled tubes, it is the safe top value to prevent arc-back in the tube operating within the specified temperature range.

Referring to Fig. 126, when plate A of a full-wave rectifier tube is positive, current flows from A to C, but not from B to C, because B is negative. At the instant plate A is positive, the filament is positive (at high voltage) with respect to plate B. The voltage between the positive filament and the negative plate B is in inverse relation to that causing current flow. The peak value of this voltage is limited by the resistance and nature of the path between plate B and filament. The maximum value of this voltage at which there is no danger of breakdown of the tube is known as maximum peak inverse voltage.

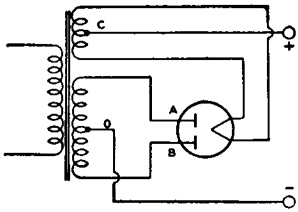


Fig. 126—Schematic diagram of full-wave rectifier tube and circuit connections.

The relations between peak inverse voltage, rms value of ac input voltage, and dc output voltage depend largely on the individual characteristics of the rectifier circuit and the power supply. The presence of line surges or any other transient, or wave-form distortion, may raise the actual peak voltage to a value higher than that calculated for sine-wave voltages. Therefore, the **actual** inverse voltage, and not the calculated value, should be such as not to exceed the rated maximum peak inverse voltage for the rectifier tube. A calibrated cathode-ray oscillograph or a peak-indicating electronic voltmeter is useful in determining the actual peak inverse voltage.

In single-phase, full-wave circuits with sine-wave input and with no capacitor across the output, the peak inverse voltage on a rectifier tube is approximately 1.4 times the rms value of the plate voltage applied to the tube. In single-phase, half-wave circuits with sine-wave input and with capacitor input to the filter, the peak inverse voltage may be as high as 2.8 times the rms value of the applied plate voltage.

In polyphase circuits, mathematical determination of peak inverse voltage requires the use of vectors.

The **Rating Chart** for full-wave rectifiers presents graphically the relationships between maximum ac voltage input and maximum dc output current derived from the fundamental ratings for conditions of capacitor-input and choke-input filters. This graphical presentation provides for considerable latitude in choice of operating conditions.

The **Operation Characteristics** for a full-wave rectifier with capacitor-input filter show by means of boundary line the limiting current and voltage relationships presented in the Rating Chart.

The **Operation Characteristics** for a full-wave rectifier with choke-input filter not only show by means of boundary line the limiting current and voltage relationships presented in the Rating Chart, but also give some information as to the effect on regulation of various sizes of chokes. The solid-line curves show the dc voltage outputs which would be obtained if the filter chokes had infinite inductance. The long-dash lines radiating from the zero position are boundary lines for various sizes of chokes as indicated. The intersection of one of these lines with a solid-line curve indicates the point on the curve at which the choke no longer behaves as though it had infinite inductance. To the left of the choke boundary line, the regulation curves depart from the solid-line curves as shown by the representative short-dash regulation curves.

Typical Operation Values. Values for typical operation are given for many types in the **Technical Data** section. These typical operating values are given to show concisely some guiding information for the use of each type. These values should not be confused with ratings, because a tube can be used under any suitable conditions within its maximum ratings, according to the application.

The power output value for any operating condition is an approximate tube output—that is, plate input minus plate loss. Circuit losses must be subtracted from tube output in order to determine the useful output.

Characteristics are covered in the **Electron Tube Characteristics** section and such data should be interpreted in accordance with the definitions given in that section. **Characteristic curves** represent the characteristics of an average tube. Individual tubes, like any manufactured product, may have characteristics that range above or below the values given in the characteristic curves.

Although some curves are extended well beyond the maximum ratings of the tube, this extension has been made only for convenience in calculations. Do NOT operate a tube outside of its *maximum ratings.*

Interelectrode capacitances are direct capacitances measured between specified elements or groups of elements in electron tubes. Unless otherwise indicated in the data, all capacitances are measured with filament or heater cold, with no direct voltages present, and with no external shields. All electrodes other than those between which capacitance is being measured are grounded. In twin or multi-unit types, inactive units are also grounded.

The capacitance between the input electrode and all other electrodes, except the output electrode, connected together is commonly known as the input capacitance. The capacitance between the output electrode and all other electrodes, except the input electrode, connected together is known as the output capacitance.

Hum and noise characteristics of high-fidelity audio amplifier tube types such as the 7025 and the 7199 are tested in an amplifier circuit such as that shown in Fig. 127. The output of the test circuit is fed into a low-noise amplifier. The bandwidth of this amplifier depends on the characteristic being measured. If hum alone is being tested, a relatively narrow bandwidth is used to include both the line frequency and the major harmonics generated by the tube under test. In noise or combination hum-and-noise measurements, the bandwidth is defined in the registration of the tube type.

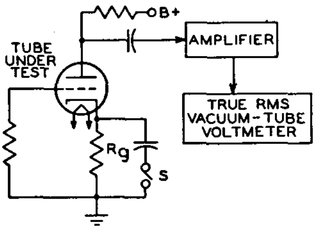


Fig. 127—Test circuit for measuring hum and noise characteristics of high-fidelity audio-amplifier tubes.

The amplifier gain is calibrated so that the vacuum-tube voltmeter measures hum and noise in microvolts referenced to the grid of the tube under test. A pentode can also be evaluated in this manner by the addition of a screen-grid supply adequately bypassed at the tube screen-grid pin connection. Power-supply ripple at the plate of the tube under test must be negligible compared to its hum and noise output. Extraordinary shielding of both the test socket and the associated operating circuit is required to minimize capacitances between heater leads and high-impedance connections.

The test-circuit components are determined by the tube type being tested and the type of hum to be controlled. Heater-cathode hum can be eliminated from the measurement by closing the switch S. The circuit can also be made more or less sensitive to heater-grid hum by increasing or decreasing the grid resistance R_g. No circuit changes affect the component of magnetic hum generated by the tube.

Grid-No. 2 (Screen-grid) Input is the power applied to the grid-No. 2 electrode and consists essentially of the power dissipated in the form of heat by grid No. 2 as a result of electron bombardment. With tetrodes and pentodes, the power dissipated in the screen-grid circuit is added to the power in the plate circuit to obtain the total B-supply input power.

When the screen-grid voltage is supplied through a series voltage-dropping resistor, the maximum screen-grid voltage rating may be exceeded, provided the maximum screen-grid dissipation rating is not exceeded at any signal condition, and the maximum screen-grid voltage rating is not exceeded at the maximum-signal condition. Provided these conditions are fulfilled, the screen-grid supply voltage may be as high as, but not above, the maximum plate voltage rating.

For certain voltage amplifier types, as listed in the data section, the maximum permissible screen-grid (grid-No. 2) input varies with the screen-grid voltage, as shown in Fig. 128. (This curve cannot be assumed to apply to types other than those for which it is specified in the data section.) Full rated screen-grid input is permissible at screen-grid voltages up to 50 per cent of the maximum rated screen-grid supply voltage. From the 50-per-cent point to the full rated value of supply voltage, the screen-grid input must be decreased. The decrease in allowable screen-grid input follows a curve of the parabolic form. This rating chart is useful for applications utilizing either a fixed screen-grid voltage or a series screen-grid voltage-dropping resistor. When a fixed voltage is used, it is necessary only to determine that the screen-grid input is within the boundary of the operating area on the chart at the selected value of screen-grid voltage to be used. When a voltage-dropping resistor is used, the minimum value of resistor that will assure tube operation within the boundary of the curve can be determined from the following relation:

$$R_{g2} \geqq \frac{E_{c2}\ (E_{cc2} - E_{c2})}{P_{c2}}$$

where R_{g2} is the minimum value for the voltage-dropping resistor in ohms, E_{c2} is the selected screen-grid voltage in volts, E_{cc2} is the screen-grid supply voltage in volts, and P_{c2} is the screen-grid input in watts corresponding to E_{c2}.

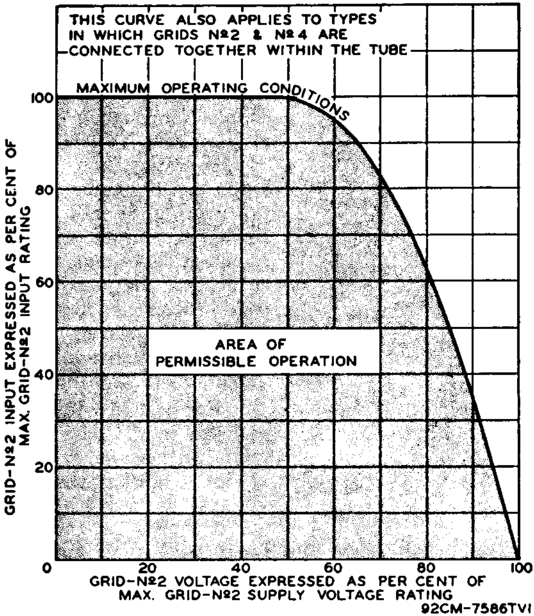


Fig. 128—Grid-No. 2 input rating curve.

Application Guide for RCA Receiving Tubes

In the Application Guide on the following pages, RCA receiving tubes are classified in two ways: (a) by function, and (b) by structure (diode, triode, etc.). The functional classification covers 42 principal types of application, as listed below.

Tube types are grouped by structure under each classification; they are also keyed to indicate miniature, octal, nuvistor, duodecar, and novar types.

Triodes are designated as *low*, *medium-*, or *high-mu* types on the following basis: *low*, less than 10; *medium*, 10 or more, but less than 50; *high*, 50 or more. Where applicable, tubes are designated as *sharp-*, *semiremote*, or *remote-cutoff* on the basis of the ratio, in per cent, of the negative control-grid voltage to the screen-grid voltage (or, for triodes, the plate voltage) for cutoff, as given in the characteristics or typical operation values. These terms are defined as follows: *sharp*, less than 10 per cent; *semiremote*, 10 or more, but less than 20 per cent, *remote*, 20 per cent or more.

For more complete data on these types, refer to the **Technical Data For RCA Receiving Tubes.**

APPLICATIONS

1. **Audio-Frequency Amplifiers**
2. **Automatic Gain Control Circuits (AGC and AVC)**
3. **Bandpass Amplifiers (Color TV)**
4. **Blankers**
5. **Burst Amplifiers**
6. **Cathode-Drive RF Amplifiers (Grounded-Grid)**
7. **Chroma Amplifiers**
8. **Color Killers**
9. **Color Matrixing Circuits**
10. **Complex-Wave Generators**
11. **Converters**
12. **Dampers**
13. **Demodulators (Color TV)**
14. **Detectors**
15. **DC Restorers**
16. **Discriminators**
17. **Frequency Dividers**
18. **FM Detectors**
19. **Gated Noise, AGC, and Sync Amplifiers**
20. **Grounded-Grid RF Amplifiers**
21. **Harmonic Generators**
22. **Horizontal-Deflection Circuits**
23. **Intermediate-Frequency Amplifiers**
24. **Keyed AGC Amplifiers**
25. **Limiters**
26. **Mixers—RF**
27. **Mixer-Oscillators—RF**
28. **Multivibrators**
29. **Noise Inverters (Noise Immune Circuits)**
30. **Oscillators**
31. **Phase Inverters**
32. **Phase Splitters**
33. **Radio-Frequency Amplifiers**
34. **Reactance Circuits**
35. **Rectifiers**
36. **Regulators (High Voltage)**
37. **Sync Amplifiers**
38. **Sync Clippers**
39. **Sync Separators**
40. **Tuning Indicators**
41. **Vertical-Deflection Circuits (Oscillator and Amplifier)**
42. **Video Amplifiers**

1. AUDIO-FREQUENCY AMPLIFIERS

Voltage Amplifiers

Medium-Mu Triode with Twin Diode
- **6BF6**

Medium-Mu Triode—Sharp-Cutoff Pentode
- **6LQ8** • **11LQ8** • **7199†**

Medium-Mu Twin Triode
- **5J6** • **7AU7** ⊙ **12SN7GTA**
- **6J6A** • **9AU7** • **19J6**
- ⊙ **6SN7GTB** • **12AU7A**

• **Miniature** ⊙ **Octal** † **For high-fidelity equipment**

Twin Diode—High-Mu Triode

• 3AV6 • 4AV6 • 6AT6 • 6AV6 • 6BN8 • 6CN7 • 8BN8 • 12AT6 • 12AV6 • 14GT8 • 18FY6A

Triple Diode—High-Mu Triode

• 5T8 • 6T8A

High-Mu Twin Triode

• 6EU7† ⊙ 6SL7GT • 12AX7A† • 12AZ7A • 12BZ7 ⊙ 12SL7GT • 20EZ7 • 7025†

High-Mu Triode—Sharp-Cutoff Pentode

• 6KT8

Sharp-Cutoff Pentode

• 3DT6A* • 4DT6A* • 5HZ6* • 6DT6A* • 6GX6* • 6HZ6* • 5879† • 7543†

Power Amplifiers

Beam Power Tube

• 5AQ5 • 5CZ5 ⊙ 5V6GT • 6AQ5A • 6AS5 • 6CM6 • 6CU5 • 6CZ5 ⊙ 6DG6GT • 6DS5 § 6GC5 ⊙ 6HG5 ⊙ 6L6 ⊙ 6L6GC† ⊙ 6V6 ⊙ 6V6GTA ⊙ 6W6GT ⊙ 6Y6GA/6Y6G • 12AB5 • 12AQ5 • 12CA5 • 12CU5/12C5 ⊙ 12V6GT ⊙ 12W6GT • 17CU5/17C5 • 25C5 • 25F5A • 34GD5A • 35C5 ⊙ 35L6GT • 50B5 • 50C5 ⊙ 50L6GT • 6973† ⊙ 7408†

Beam Power Tube—Sharp-Cutoff Pentode

‡ 6AD10 ‡ 6BF11* ‡ 6AL11 ‡ 12AL11 ‡ 12BF11* ‡ 17BF11*

Power Pentode

• 6BQ5 • 6BQ5/EL84 • 6EH5 ⊙ 6F6 • 6GK6 ⊙ 6K6GT • 8BQ5 • 10BQ5 • 12EH5 • 12FX5 • 25EH5 • 35EH5 • 50EH5 • 60FX5 • 7189† ▴ 7868†

Pentode—Beam Power Tube

‡ 6J10 ‡ 6Z10 ‡ 6Z10/6J10 ‡ 13Z10 ‡ 13Z10/13J10

2. AUTOMATIC GAIN CONTROL CIRCUITS (AGC & AVC)

Diode—Remote-Cutoff Pentode

• 6EQ7 • 12EQ7

Twin Diode—High-Mu Triode

• 3AV6 • 4AV6 • 6AT6 • 6AV6 • 12AT6 • 12AV6 • 18FY6A

Medium-Mu Triode—Sharp-Cutoff Pentode

• 5AN8 • 5GH8A • 6AN8A • 6AZ8 • 6BA8A • 6BH8 • 6CU8 • 6GH8A • 8BA8A • 8BH8

High-Mu Triode—Sharp-Cutoff Pentode

• 6AW8A • 6HF8 • 6JV8 • 8AW8A • 8JV8 • 10HF8

Sharp-Cutoff Twin Pentode

• 3BU8 • 4HS8 • 6BU8 • 6HS8

3. BANDPASS AMPLIFIER (COLOR TV)

Medium-Mu Triode—Sharp-Cutoff Pentode

• 5GH8A • 6GH8A • 6HL8

High-Mu Triode—Sharp-Cutoff Pentode

• 6AW8A • 6KT8 • 6KV8 • 6LF8 • 8AW8A • 11KV8

4. BLANKERS

Medium-Mu Triode—Sharp-Cutoff Pentode

• 5GH8A • 6GH8A

Medium-Mu Twin Triode

• 6FQ7/6CG7 • 6GU7 • 8FQ7/8CG7 • 8GU7 • 12BH7A

Medium-Mu Triode—Semiremote-Cutoff Pentode

• 6LM8

High-Mu Triode—Sharp-Cutoff Pentode

• 6KT8

5. BURST AMPLIFIERS

Beam-Deflection Tube

• 6JH8

Medium-Mu Triode—Sharp-Cutoff Pentode

• 5EA8 • 5GH8A • 6EA8 • 6GH8A • 19EA8

Medium-Mu Triode—Semiremote-Cutoff Pentode

• 6LM8

Twin Diode–High-Mu Triode

• 6BN8 • 8BN8

Sharp-Cutoff Pentode

• 3JC6A • 4EW6 • 4JC6A • 5EW6 • 6EW6 • 6JC6A

6. CATHODE-DRIVE RF AMPLIFIERS (GROUNDED-GRID)

Medium-Mu Triode

• 6BC4

• Miniature ‡ Duodecar ⊙ Octal ▴ Novar * Dual-control grids † For high-fidelity equipment § Neonoval

Medium-Mu Twin Triode

- • 4BC8
- • 4BQ7A
- • 4BS8
- • 4BZ7
- • 5BK7A
- • 5BQ7A
- • 6BC8
- • 6BK7B
- • 6BQ7A
- • 6BS8
- • 6BZ7

High-Mu Triode

- △ 2CW4
- △ 2DS4
- • 2HQ5
- • 3HQ5
- • 4HQ5
- • 6AB4
- △ 6CW4
- △ 6DS4
- • 6HQ5
- △ 13CW4

High-Mu Twin Triode

- • 6DT8
- • 12AT7
- • 12AZ7A
- • 12DT8

7. CHROMA AMPLIFIERS

Medium-Mu Triode—Sharp-Cutoff Pentode

- • 5GH8A
- • 6GH8A

Medium-Mu Triple Triode

- ▲ 6MD8
- ▲ 12MD8

Medium-Mu Twin Triode

- • 6FQ7/6CQ7
- • 6GU7
- • 8FQ7/8CG7
- • 8GU7
- • 12BH7A

8. COLOR KILLERS

Quadruple Diode

- • 6JU8A
- • 8JU8A

Medium-Mu Triode—Sharp-Cutoff Pentode

- • 5GH8A
- • 6GH8A

High-Mu Triode—Sharp-Cutoff Pentode

- • 6KT8

9. COLOR MATRIXING CIRCUITS

Medium-Mu Twin Triode

- • 6FQ7/6CG7
- • 6GU7
- • 8FQ7/8CG7
- • 8GU7
- • 12BH7A

Medium-Mu Triode—Sharp Cutoff Pentode

- • 5GH8A
- • 6GH8A

Medium-Mu Triple Triode

- ▲ 6MD8
- ▲ 12MD8

Twin Pentode

- • 6LE8
- • 10LE8
- • 15LE8

Quadruple Diode

- • 6JU8A
- • 8JU8A

10. COMPLEX-WAVE GENERATORS

High-Mu Twin Double-Plate Triode

- • 12FQ8

Diode—Sharp-Cutoff, Twin-Plate Tetrode

- • 6FA7

Diode—Sharp-Cutoff, Three-Plate Tetrode

- • 6KM8

Medium-Mu Triode—Three-Plate Tetrode

- • 6FH8

11. CONVERTERS

Medium-Mu Triode—Sharp-Cutoff Pentode

- • 4KE8
- • 5EA8
- • 5GH8
- • 5KE8
- • 5U8
- • 5X8
- • 6EA8
- • 6GH8A
- • 6KE8
- • 6KZ8
- • 6U8A
- • 6U8A/6KD8
- • 9KZ8
- • 19X8
- • 19EA8

High-Mu Twin Triode

- • 6DT8
- • 12AT7
- • 12AZ7A
- • 12DT8

Sharp-Cutoff Pentode

- • 3AU6
- • 4AU6
- • 6AU6A
- • 12AU6
- • 18GD6A

Pentagrid

- • 6BA7
- • 6BE6
- • 12BE6
- • 18FX6A

12. DAMPERS

Half-Wave (Diode)

- ⊙ 6AU4GTA
- ⊙ 6AX4GTB
- ▲ 6AY3B
- ▲ 6BA3
- ‡ 6BE3
- ‡ 6BE3/6BZ3
- ▲ 6BH3A
- ▲ 6BS3A
- ▲ 6CJ3
- ▲ 6CK3
- ▲ 6CL3
- ▲ 6CM3
- ⊙ 6CQ4
- • 6CT3
- ⊙ 6DA4
- ⊙ 6DE4
- ⊙ 6DM4A
- ▲ 6DW4A
- ▲ 6DW4B
- ⊙ 6W4GT
- ⊙ 12AX4GTB
- ▲ 12AY3A
- ‡ 12BE3
- ▲ 12BS3A
- ▲ 12CK3
- ▲ 12CL3
- ⊙ 12D4
- ⊙ 17AX4GTA
- ▲ 17AY3A
- ‡ 17BE3/17BZ3
- ▲ 17BH3A
- ▲ 17BS3A
- ▲ 17CK3
- • 17CT3
- ⊙ 17D4
- ⊙ 17DE4
- ⊙ 17DM4A
- ▲ 22BH3A
- ⊙ 22DE4
- ⊙ 25AX4GTA
- ▲ 25CM3
- ▲ 34CM3

13. DEMODULATORS (COLOR TV)

Medium-Mu Twin Triode

- • 12BH7A

Medium-Mu Triode—Sharp-Cutoff Pentode

- • 5GH8A
- • 6GH8A

High-Mu Twin Triode

- • 12AZ7A

Sharp-Cutoff Pentode

- • 5HZ6
- • 6GY6
- • 6HZ6

Pentagrid Amplifier

- • 3BY6
- • 6BY6

• Miniature ⊙ Octal △ Nuvistor ▲ Novar

Twin Pentode
• 6LE8 • 10LE8 • 15LE8

Beam Deflection Tube
• 6JH8 • 6ME8

Sharp-Cutoff Twin Pentode
• 6MK8

14. DETECTORS

Diode—Sharp-Cutoff Pentode
• 5AM8 • 6AM8A
• 5AS8 • 6AS8

Diode—Remote-Cutoff Pentode
• 6CR6 • 12CR6 • 12EQ7
• 6EQ7

Twin Diode
• 3AL5 • 6AL5 • 12AL5

Twin Diode—High-Mu Triode
• 3AV6 • 6BN8 • 12AV6
• 4AV6 • 6CN7 • 14GT8
• 6AT6 • 8BN8 • 18FY6A
• 6AV6 • 12AT6

Triple Diode
• 6BJ7

Triple Diode—High-Mu Triode
•5T8 • 6T8A

Quadruple Diode
• 6JU8A • 8JU8A

Sharp-Cutoff Pentode
• 3DT6A* • 5HZ6* • 6GX6*
• 4DT6A* • 6DT6A* • 6HZ6*
• 5GX6*

15. DC RESTORERS

Diode—Sharp-Cutoff Pentode
• 5AM8 • 6AM8A • 6AS8
• 5AS8

Triple Diode
• 6BJ7

16. DISCRIMINATORS

FM

Twin Diode
• 3AL5 • 6AL5 • 12AL5

Twin Diode—High-Mu Triode
• 6BN8 • 14GT8

Triple Diode—High-Mu Triode
• 5T8 • 6T8A

Beam Tube
• 3BN6 • 4BN6 • 6BN6

Beam Power Tube—Sharp-Cutoff Pentode
‡ 6AL11 ‡ 12AL11 ‡ 17BF11
‡ 6BF11 ‡ 12BF11

Pentode—Beam Power Tube
‡ 6J10 ‡ 13Z10 ‡ 17AB10/17AZ10
‡ 6Z10 ‡ 13Z10/13J10
‡ 6Z10/6J10 ‡ 17AB10

FM Quadrature-Grid

Sharp-Cutoff Pentode
• 3DT6A* • 6DT6A* • 6GY6*
• 4DT6A* • 6GX6* • 6HZ6*
• 5HZ6*

Beam Tube
• 3BN6 • 4BN6 • 6BN6

Horizontal AFC

Twin Diode—High-Mu Triode
• 6BN8 • 8BN8 • 8CN7
• 6CN7

17. FREQUENCY DIVIDERS

High-Mu Twin Double-Plate Triode
• 12FQ8

18. FM DETECTORS
(See 16. Discriminators)

19. GATED NOISE, AGC, AND SYNC AMPLIFIERS

High-Mu Triode—Sharp-Cutoff Pentode
• 6KA8 • 8KA8 • 8LC8
• 6LC8

Sharp-Cutoff Pentode
• 6GY6*

Sharp-Cutoff Twin Pentode
• 3BU8 • 6BU8 • 6HS8
• 4HS8

Pentagrid Amplifier
• 3BY6 • 4CS6 • 6CS6
• 3CS6 • 6BY6

20. GROUNDED-GRID RF AMPLIFIERS
(See 6. Cathode-Drive RF Amplifiers)

21. HARMONIC GENERATORS
(See 10. Complete-Wave Generators)

22. HORIZONTAL-DEFLECTION CIRCUITS

Oscillators

Medium-Mu Triode—Sharp-Cutoff Pentode
• 5GH8A • 6GH8A

• Miniature ⊚ Octal * Dual-control grids ‡ Duodecar

Medium-Mu Twin Triode
- • 6FQ7/6CG7 • 8FQ7/8CG7 • 12BH7A
- ⊙ 6SN7GTB • 9AU7 ⊙ 12SN7GTA
- • 7AU7 • 12AU7A

Amplifiers

Beam Power Tube

⊙ 6AU5GT ▴ 6JT6A ‡ 17JM6A
⊙ 6AV5GA ▴ 6JU6 ▴ 17JR6
⊙ 6BQ6GTB/6CU6 ▴ 6KM6 ▴ 17JT6A
⊙ 6CB5A ▴ 6LQ6 ▴ 22JF6
⊙ 6CD6GA ▴ 6LQ6/6JE6C ▴ 22JG6
⊙ 6DQ5 ⊙ 12AV5GA ▴ 22JR6
▴ 6GJ5A ⊙ 12BQ6GTB/12CU6 ▴ 22KM6
▴ 6GT5A ▴ 12JB6A ▴ 24JE6A
⊙ 6GW6/6DQ6B ▴ 12JT6A ▴ 24LQ6
▴ 6JB6A ⊙ 17BQ6GTB ▴ 24LQ6/24JE6B
▴ 6JE6A ▴ 17GJ5A ⊙ 25AV5GA
▴ 6JF6 ▴ 17GT5A ⊙ 25BQ6GTB/25CU6
▴ 6JG6A ⊙ 17GW6/17GW6B ⊙ 25CD6GB
‡ 6JM6A ▴ 17JB6A ⊙ 25DN6
▴ 6JR6 ▴ 17JG6A ‡ 31JS6A
‡ 6JS6A

23. INTERMEDIATE-FREQUENCY AMPLIFIERS

Medium-Mu Triode—Sharp-Cutoff Tetrode
- • 5CQ8 • 6CQ8

Medium-Mu Triode—Sharp-Cutoff Pentode
- • 5AN8 • 6AZ8 • 6GH8A
- • 5GH8A • 6BH8 • 11LQ8
- • 6AN8A • 6CU8

High-Mu Triode—Sharp-Cutoff Pentode
- • 6AW8A • 6KV8 • 10GN8
- • 6GN8 • 8AW8A • 10HF8
- • 6HF8 • 8GN8 • 10JA8
- • 6JV8 • 8JV8 • 11KV8
- • 6KT8

Sharp-Cutoff Pentode
- • 3AU6 • 4JC6A • 6DK6
- • 3BC5/3CE5 • 4JD6▪ • 6EJ7/EF184
- • 3CB6 • 5EW6 • 6EW6
- • 3CB6/3CF6 • 6AG5 • 6HS6
- • 3DK6 • 6AK5 • 6JC6A
- • 3JC6A • 6AU6A • 6JD6▪
- • 3JD6▪ • 6BC5 • 12AU6
- • 4AU6 • 6CB6/6CF6 • 12AW6
- • 4CB6 • 6CB6A • 12DK6
- • 4DE6 • 6DC6 • 18GD6A
- • 4DK6 • 6DE6
- • 4EW6

Diode—Sharp-Cutoff Pentode
- • 5AM8 • 6AM8A • 6AS8
- • 5AS8

Semiremote-Cutoff Pentode
- • 3BZ6 • 4KT6 • 6HR6
- • 3KT6 • 5GM6 • 6JH6
- • 4BZ6 • 6BZ6 • 6KT6
- • 4EH7 • 6EH7/EF183 • 12BZ6
- • 4GM6 • 6GM6 • 19HR6
- • 4JH6

Remote-Cutoff Pentode
- • 6BA6 • 12BA6 • 18FW6A

Remote-Cutoff Pentode with Diode
- • 6EQ7 • 12EQ7

24. KEYED AGC AMPLIFIERS

(See 19. Gated Noise, AGC, and Sync Amplifiers)

25. LIMITERS

Beam Tube
- • 3BN6 • 4BN6 • 6BN6

Sharp-Cutoff Pentode
- • 3AU6 • 6GX6 • 6HZ6
- • 4AU6 • 6HS6 • 12AU6
- • 6AU6A

Power Pentode—Beam Power Tube
- ‡ 6J10 ‡ 13Z10 ‡ 17AB10
- ‡ 6Z10 ‡ 13Z10/13J10 ‡ 17AB10/17AX10
- ‡ 6Z10/6J10

26. MIXERS—RF

Medium-Mu Twin Triode
- • 5J6 • 6J6A

High-Mu Triode
- △ 2CW4 △ 6CW4 △ 13CW4
- • 6AB4

27. MIXER-OSCILLATORS—RF

Medium-Mu Triode—Sharp-Cutoff Tetrode
- • 5CL8A • 6CL8A • 19CL8A
- • 5CQ8 • 6CQ8

Medium-Mu Triode—Sharp-Cutoff Pentode
- • 4KE8 • 5U8 • 6KE8
- • 5AT8 • 5X8 • 6KZ8
- • 5B8 • 6AT8A • 6U8A
- • 5BR8 • 6BR8A • 6X8A
- • 5CG8 • 6CG8A • 9EA8
- • 5EA8 • 6EA8 • 9KZ8
- • 5FG7 • 6FG7 • 9U8A
- • 5KE8 • 6HB7 • 19EA8
- • 19X8

High-Mu Twin Triode
- • 6DT8 • 12AT7 • 12DT8

• Miniature ⊙ Octal ▴ Novar
* Dual-control grids △ Nuvistor
▪ Approaches semiremote-cutoff characteristics; used in first-if amplifier applications

28. MULTIVIBRATORS

Medium-Mu Triode—Sharp-Cutoff Pentode

• 5GH8A • 6GH8A

Medium-Mu Twin Triode

• 6FQ7/6CG7 • 6GU7 ⊙ 6SN7GTB • 7AU7 • 8FQ7/8CG7 • 8GU7 • 9AU7 • 12AU7A • 12BH7A ⊙ 12SN7-GTA

High-Mu Twin Triode

• 12AX7A

29. NOISE INVERTERS (NOISE IMMUNE CIRCUITS)

High-Mu Triode—Sharp-Cutoff Pentode

• 6KA8 • 6LC8 • 8KA8 • 8LC8

Sharp-Cutoff Pentode

• 6GY6*

Quadruple Diode

• 6JU8A • 8JU8A

30. OSCILLATORS

Radio Frequency—UHF

Medium-Mu Triode

• 2AF4B/2DZ4 △ 2DV4 • 3AF4A/3DZ4 • 6AF4A △ 6DV4 • 6DZ4

Radio Frequency—VHF

Medium-Mu Twin Triode

• 5J6 • 6J6A

High-Mu Triode

• 6AB4

Power Triode

• 6C4 (Class C)

3.58-MHz (Color TV)

Medium-Mu Triode—Sharp-Cutoff Pentode

• 5GH8A • 6GH8A

High-Mu Triode—Sharp-Cutoff Pentode

• 6KT8

Low Frequency, Sweep Type

Medium-Mu Triode—Sharp-Cutoff Pentode

• 5AN8 • 6AN8A • 6AU8A • 6AZ8 • 6BA8A • 6BH8 • 6CH8 • 8AU8 • 8BA8B • 8BH8

Twin Diode—High-Mu Triode

• 6BN8 • 6CN7 • 8BN8 • 8CN7

High-Mu Twin Triode

• 12AX7A

31. PHASE INVERTERS

Medium-Mu Twin Triode

• 6FQ7/6CG7 • 6GU7 ⊙ 6SN7GTB • 7AU7 • 8GU7 • 8FQ7/8CG7 • 9AU7 • 12AU7A • 12BH7A ⊙ 12SN7-GTA

High-Mu Triode—Sharp-Cutoff Pentode

• 6AW8A • 6EB8 • 6GN8 • 6HF8 • 8AW8A • 8EB8 • 8GN8 • 10GN8 • 10HF8 • 10JA8

High-Mu Twin Triode

⊙ 6SL7GT • 12AX7A ⊙ 12SL7GT • 7025

Medium-Mu Triple Triode

‡ 6AV11

32. PHASE SPLITTERS

Medium-Mu Triode—Sharp-Cutoff Tetrode

• 5CQ8 • 6CQ8

Medium-Mu Triode—Sharp-Cutoff Pentode

• 5AN8 • 6AN8A • 6AZ8 • 6BA8A • 6CU8 • 8BA8A • 7199

High-Mu Triode—Sharp-Cutoff Pentode

• 6AW8A • 8AW8A

33. RADIO-FREQUENCY AMPLIFIERS

Medium-Mu Triode

• 2BN4A • 3BN4A • 6BC4 • 6BN4A

Medium-Mu Triode—Sharp-Cutoff Tetrode

• 5CQ8 • 6CQ8

Medium-Mu Twin Triode

• 4BC8 • 4BQ7A • 4BS8 • 4BZ7 • 5BK7A • 5BQ7A • 5J6 • 6BC8 • 6BK7B • 6BQ7A • 6BS8 • 6BZ7 • 6J6A

High-Mu Triode

△ 2CW4 △ 2DS4 △ 2EG4 • 2ER5 • 2FH5 • 2FQ5A • 2GK5 • 3ER5 • 3FH5 • 3GK5 • 3HM5/3HA5 • 4GK5 • 6AB4 △ 6CW4 △ 6DS4 • 6ER5 • 6FH5 • 6FQ5A • 6GK5 • 6HM5/6HA5 △ 13CW4

High-Mu Twin Triode

• 6DT8 • 12AZ7A • 12DT8

• Miniature ⊙ Octal △ Nuvistor * Dual-control grids

Power Triode
- • 6C4 (Class C)

Sharp-Cutoff Tetrode
- • 2CY5 • 6CY5 • 6FV6
- • 3CY5

Sharp-Cutoff Pentode
- • 3AU6 ⊙ 4DE6 • 6CB6A
- • 3BC5/3CE5 • 6AG5 • 6DC6
- • 3CB6 • 6AK5 • 6DE6
- • 3CF6 • 6AU6A • 12AU6
- • 4AU6 • 6BC5 • 12AW6
- • 4CB6 • 6BH6 • 18GD6A

Remote-Cutoff Pentode
- • 6BA6 • 12BA6 • 18FW6A
- • 6BJ6A

Remote-Cutoff Pentode with Diode
- • 6EQ7 • 12EQ7

34. REACTANCE CIRCUITS

Medium-Mu Triode—Sharp-Cutoff Pentode
- • 5AN8 • 6AZ8 • 6CU8
- • 6AN8A • 6BA8A • 8BA8A

Twin Diodes—High-Mu Triode
- • 6CN7 • 8CN7

High-Mu Triode—Sharp-Cutoff Pentode
- • 6AW8A • 8AW8A

35. RECTIFIERS

Power-Supply Types—Vacuum

Half-Wave (Diode)
- • 35W4 • 36AM3B • 50DC4
- ⊙ 35Z5GT

Full-Wave (Twin Diode)
- ⊙ 3DG4 ⊙ 5V3A • 6CA4
- ⊙ 5AS4A ⊙ 5V4GA • 6X4
- ▲ 5BC3A ⊙ 5Y3GT ⊙ 6X5GT
- ⊙ 5DJ4 ⊙ 5Z4 • 12X4
- ⊙ 5U4GB

High-Voltage Types (For rf-rectifier or pulsed low-current applications)—Vacuum

Half-Wave (Diode)
- • 1BC2 • 1V2 • 2BJ2
- ⊙ 1G3GT/1B3GT • 1X2B ⊙ 3A3A
- • 1X2B/1X2A ⊙ 3CA3
- ⊙ 1K3/1J3 • 2AV2

36. REGULATORS (HIGH VOLTAGE)

Beam Triode
- ⊙ 6BK4B ⊙ 6LJ6

Beam Power Tube
- ▲ 6KV6 ▲ 17KV6

37. SYNC AMPLIFIERS

Medium-Mu Triode—Sharp-Cutoff Pentode
- • 6AU8A • 6CX8 • 8CX8
- • 6AZ8 • 8AU8

Medium-Mu Twin Triode
- • 6FQ7/6CG7 • 8FQ7/8CG7 • 12AU7A
- • 7AU7 • 9AU7

High-Mu Triode with Twin Diode
- • 6CN7 • 8CN7

High-Mu Triode—Sharp-Cutoff Pentode
- • 6AW8A • 6JV8 • 8JV8
- • 6HF8 • 8AW8A • 10HF8

High-Mu Twin Triode
- • 12BZ7

38. SYNC CLIPPERS

Medium-Mu Triode—Sharp-Cutoff Tetrode
- • 5CQ8 • 6CQ8

Medium-Mu Triode—Sharp-Cutoff Pentode
- • 5AN8 • 6AZ8 • 8AU8
- • 6AN8A • 6CU8 • 8CX8
- • 6AU8A • 6CX8

High-Mu Triode—Sharp-Cutoff Pentode
- • 6AW8A • 6HF8 • 8JV8
- • 6EB8 • 6JV8 • 10GN8
- • 6GN8 • 8AW8A • 10HF8
- • 6GW8/ECL86 • 8EB8 • 10JA8
- • 8GN8

High-Mu Twin Triode
- • 12BZ7

Sharp-Cutoff Twin Pentode
- • 3BU8 • 6BU8 • 6HS8
- • 4HS8

Pentagrid Amplifier
- • 3BY6 • 4CS6 • 6CS6
- • 3CS6 • 6BY6

39. SYNC SEPARATORS

Medium-Mu Triode—Sharp-Cutoff Tetrode
- • 5CQ8 • 6CQ8

• Miniature ⊙ Octal ▲ Novar

Medium-Mu Triode—Sharp-Cutoff Pentode

• 5AN8 • 6CU8 • 6LQ8
• 5GH8A • 6CX8 • 8AU8
• 6AN8A • 6GH8 • 8CX8
• 6AU8A • 6GH8A • 11LQ8
• 6AZ8 • 6HL8

Medium-Mu Twin Triode

• 6FQ7/6CG7 • 8FQ7/8CG7 • 12AU7A
• 7AU7 • 9AU7

Twin Diode—High-Mu Triode

• 6CN7 • 8CN7

High-Mu Triode—Sharp-Cutoff Pentode

• 6AW8A • 6KV8 • 8KA8
• 6EB8 • 6LC8 • 8LC8
• 6GN8 • 8AW8A • 10GN8
• 6HF8 • 8EB8 • 10HF8
• 6JV8 • 8GN8 • 10JA8
• 6KA8 • 8JV8 • 11KV8
• 6KT8

High-Mu Twin Triode

• 12BZ7

Sharp-Cutoff Twin Pentode

• 3BU8 • 6BU8 • 6MK8
• 4HS8 • 6HS8

Pentagrid Amplifier

• 3BY6 • 4CS6 • 6CS6
• 3CS6 • 6BY6

40. TUNING INDICATORS

Indicator with Triode Unit

6E5

Twin Indicator Units

⊙ 6AF6G

41. VERTICAL-DEFLECTION CIRCUITS

Oscillators and Amplifiers (Combined)

Medium-Mu Triode—Low-Mu Triode

• 6DE7 • 10DE7 • 13DE7
§ 6EW7 § 10EW7

Medium-Mu Dual Triode

• 6CM7 • 8CM7 • 8CS7
• 6CS7

Medium-Mu Twin Triode

• 6FQ7/6CG7 • 8FQ7/8CG7

High-Mu Triode—Low-Mu Triode

• 6CY7 ▴ 6GF7A • 11CY7
• 6DR7 ⊙ 6GL7 • 13DR7
• 6EA7 • 10DR7 ⊙ 13EM7
⊙ 6EM7 ⊙ 10EM7 ▴ 13FD7
▴ 6FD7 ▴ 10GF7A ▴ 13GF7A

High-Mu Triode—Beam Power Tube

▴ 6KY8A ▴ 15KY8A

Dual Triode

⊙ 6EM7 ▴ 6GF7A ⊙ 13EM7/15EA7

Amplifiers

Low-Mu Triode

• 12B4A

Medium-Mu Triode

• 6S4A

Beam Power Tube

• 5AQ5 • 6EM5 • 12AQ5
• 5CZ5 • 6HR5 • 12JQ6#
⊙ 5V6GT • 6JQ6# ⊙ 12V6GT
• 6AQ5A ⊙ 6V6 • 17JQ6#
• 6CM6 ⊙ 6V6GT • 25JQ6#
• 6CZ5 • 8EM5

Power Pentode

⊙ 6K6GT

42. VIDEO AMPLIFIERS

Medium-Mu Triode—Sharp-Cutoff Pentode

• 5AN8 • 6BH8 • 8AU8
• 5GH8A • 6CU8 • 8BA8A
• 6AN8A • 6CX8 • 8BH8
• 6AU8A • 6GH8A • 8CX8
• 6AZ8 • 6HL8 • 11LQ8
• 6BA8A • 6LQ8

High-Mu Triode—Sharp-Cutoff Pentode

• 6AW8A • 6KV8 • 8JV8
• 6EB8 • 6LF8 • 10GN8
• 6GN8 • 8AW8A • 10HF8
• 6HF8 • 8EB8 • 10JA8
• 6JV8 • 8GN8 • 11KV8
• 6KT8

Sharp-Cutoff Pentode

• 3JC6A • 6KY6 • 11HM7
• 4JC6A • 7KY6 • 12BY7A
• 6JC6A • 7KZ6 § 12HG7

Sharp-Cutoff Pentode

• 5AM8 • 6AM8A • 6AS8
• 5AS8

Power Pentode

⊙ 6AG7 • 6CL6 • 6GK6

• Miniature ⊙ Octal ▴ Novar # With an integral diode
§ Neonoval

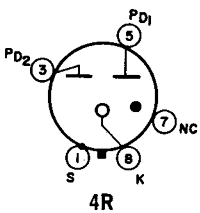
Technical Data for RCA Tube Types

THIS section contains technical descriptions of RCA tubes used in standard broadcast, FM, and television receivers, in audio amplifiers, and in many other diverse applications. It includes detailed data on current types, including characteristics curves in many cases. Essential information on types intended primarily for replacement use and on discontinued types in which there may still be some interest is given in chart form at the end of the section. Characteristics charts for RCA television picture tubes for replacement use and for RCA voltage-regulator and voltage-reference tubes are given in the following section.

In choosing tube types for the design of new electronic equipment, the designer should refer to the **Application Guide for RCA Receiving Tubes** in the pages immediately preceding this section.

Tube types are listed in this section according to the numerical-alphabetical-numerical sequence of their type designations. For **Key: Basing Diagrams**, see inside back cover.

Refer to chart at end of data section. **0Z4**



FULL-WAVE GAS RECTIFIER **0Z4A**

Metal type used as a power rectifier in equipment with vibrator-type power supplies. **Outlines section, 2A**; requires octal socket. This tube, like other power-handling tubes, should be adequately ventilated.

Full-Wave Rectifier

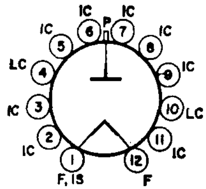
MAXIMUM AND MINIMUM RATINGS (Design-Center Values)		
Peak Inverse Plate Voltage (Per Plate)	880 max	volts
Peak Starting-Supply Voltage (Per Plate)	300▲ min	volts
Peak Plate Current (Per Plate)	330 max	mA
DC Output Current	110 max	mA
	30▲ min	mA
TYPICAL OPERATION WITH VIBRATOR-TYPE POWER SUPPLY AND CAPACITOR INPUT TO FILTER		
Peak Plate Supply Voltage (Per Plate)‡	440	volts
Filter-Input Capacitor	8	μF
Total Effective Plate Supply Impedance (Per Plate)	600	ohms
DC Output at Input to Filter	310	volts
DC Output Current	100	mA
CHARACTERISTICS		
Tube Voltage Drop for current of 110 mA (Per Plate)	24	volts
MINIMUM CIRCUIT VALUE		
Total Effective Plate-Supply Impedance (Per Plate)	300	ohms

▲ Absolute value. Under no circumstances should the tube be operated below the value shown.
‡ Open-circuit voltage (flat portion of transformer voltage wave).

Refer to chart at end of section. **0Z4G**

1A3	Refer to chart at end of section.
1A4P	Refer to chart at end of section.
1A5GT	Refer to chart at end of section.
1A6	Refer to chart at end of section.
1A7GT	Refer to chart at end of section.
1AC5	Refer to chart at end of section.

1AD2 HALF-WAVE VACUUM RECTIFIER



12DQ

Duodecar type used as a rectifier in high-voltage pulse circuits of color and black-and-white television receivers. **Outlines section,** 9A; requires duodecar 12-contact socket. Socket terminals 4 and 10 may be used as tie points for components at or near filament potential. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT.

Filament Voltage (ac/dc)	1.25	volts
Filament Current	0.2	ampere
Direct Interelectrode Capacitance (Approx.):		
Plate to Filament	1.6	pF

Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	26000▪	volts
Peak Plate Current	50	mA
Average Plate Current	0.5	mA

CHARACTERISTICS, Instantaneous Value

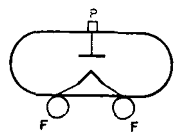
Tube Voltage Drop for plate current of 7 mA	225	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds)

▪ The dc component must not exceed 22000 volts.

1AD5	Refer to chart at end of section.
1AX2	Refer to chart at end of section.

1AY2 HALF-WAVE VACUUM RECTIFIER



Miniature type used to supply high voltage to the anode of the picture tube in television receivers. **Outlines section,** 33A; requires 2-contact socket. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT.

Filament Voltage (ac/dc)	1.25	volts
Filament Current	0.2	ampere
Direct Interelectrode Capacitances:		
Plate to Filament	1.4	pF

Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	26000*	volts
Peak Plate Current	50	mA
Average Plate Current	0.5	mA

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 7 mA	75	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

* The dc component must not exceed 22000 volts.

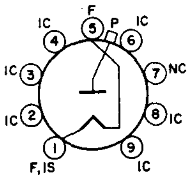
Refer to chart at end of section. **1B3GT**

Refer to chart at end of section. **1B4P**

Refer to chart at end of section. **1B5/25S**

Refer to chart at end of section. **1B7GT**

HALF-WAVE VACUUM RECTIFIER 1BC2



9RG

Miniature type used as a rectifier in high-voltage pulse circuits of black-and-white television receivers. **Outlines section,** 7E; requires miniature 9-contact socket. Socket terminal 7 may be used as a tie point for components at or near filament potential. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT.

Filament Voltage (ac/dc)	1.25	volts
Filament Current	0.2	ampere
Direct Interelectrode Capacitance:		
Plate to Filament	1	pF

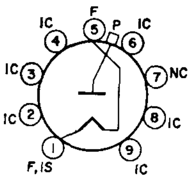
Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
Peak Inverse Plate Voltage#	18000•	volts
Peak Plate Current	45	mA
Average Plate Current	0.5	mA
CHARACTERISTICS, Instantaneous Value		
Tube Voltage Drop for plate current of 7 mA	80	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
• The dc component must not exceed 15000 volts.

HALF-WAVE VACUUM RECTIFIER 1BH2



9RG

Miniature type used as a high-voltage rectifier to supply power to the anode of the television picture tube. **Outlines section,** 7G. Filament pins 1 and 5 have solder lugs to eliminate the need for a tube socket. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT. Heater: volts (ac/dc), 1.25; amperes, 0.2.

Flyback Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
Peak Inverse Plate Voltage#	18000▲	volts
Peak Plate Current	45	mA
Average Plate Current	0.5	mA
CHARACTERISTICS, Instantaneous Value		
Tube Voltage Drop for plate current of 7 mA	80	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
▲ The dc component must not exceed 15000 volts.

Refer to chart at end of section. **1C5GT**

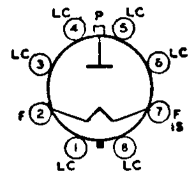
Refer to chart at end of section. **1C6**

Refer to chart at end of section. **1C7G**

1D5GP **1D5GT**	Refer to chart at end of section.
1D7G	Refer to chart at end of section.
1D8GT	Refer to chart at end of section.
1DN5	Refer to chart at end of section.
1E5GP	Refer to chart at end of section.
1E7GT	Refer to chart at end of section.
1E8	Refer to chart at end of section.
1F4	Refer to chart at end of section.
1F5G	Refer to chart at end of section.
1F6	Refer to chart at end of section.
1F7G	Refer to chart at end of section.

1G3GT/ 1B3GT

HALF-WAVE VACUUM RECTIFIER



3C

Glass octal type used as a rectifier in high-voltage pulse circuits of color and black-and-white television receivers or as a rectifier in a high-voltage rf-operated power supply.

Filament Voltage (ac/dc)	1.25*	volts
Filament Current	0.2	ampere
Direct Interelectrode Capacitance (Approx.):		
Plate to Filament and Internal Shield	1.3	pF

* Under no circumstances should the filament voltage be less than 1.05 volts or greater than 1.45 volts.

Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	26000*	volts
Peak Plate Current	50	mA
Average Plate Current	0.5	mA

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 7 mA	100	volts

Radio-Frequency Rectifier

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage	33000	volts
Peak Plate Current	35	mA
Average Plate Current	1.1	mA
Frequency Range of Supply Voltage	1.5 to 100	kHz

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
* The dc component must not exceed 22000 volts.

Installation and Application

Type 1G3GT/1B3GT requires an octal socket. Plate connection is cap at top of bulb. Socket terminals 1, 3, 4, 5, 6, and 8 may be connected to socket terminal 7 or to a corona shield which is connected to socket terminal 7. Socket terminals 4 and 6 may be used as tie points for components at or near filament potential. This type may be supplied with pins 1, 4, and/or 6 omitted. Outlines section, 14B.

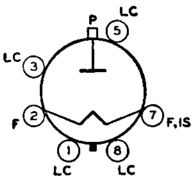
The high voltages at which the 1G3GT/1B3GT is operated are very dangerous. Great care should be taken during the adjustment of circuits. The tube and its associated apparatus, especially all parts which may be at high potential with respect to ground, should be housed in a protective enclosure. The protective housing should be designed with interlocks so that personnel cannot possibly come in contact with any high-potential point in the electrical system. The interlock devices should function to break the primary circuit of the high-voltage supply when any gate or door on the protective housing is opened, and should prevent the closing of this primary circuit until the door is locked again.

It should be noted that **high voltages** may appear at normally low-potential points in the circuit as a result of capacitor breakdown or incorrect circuit connections. Therefore, before any part of the circuit is touched, the power-supply switch should be turned off and both terminals of any capacitor should be grounded.

Operation of the 1G3GT/1B3GT with a plate voltage above approximately 16000 volts (absolute value) results in the production of X-rays which can constitute a health hazard on prolonged exposure at close range unless the tube is adequately shielded. Relatively simple shielding should prove adequate, but the need for this precaution should be considered.

Refer to chart at end of section.	**1G4GT**
Refer to chart at end of section.	**1G5G**
Refer to chart at end of section.	**1G6GT**
Refer to chart at end of section.	**1H4G**
Refer to chart at end of section.	**1H5GT**
Refer to chart at end of section.	**1H6G**
Refer to chart at end of section.	**1J3**
Refer to chart at end of section.	**1J5G**
Refer to chart at end of section.	**1J6G** **1J6GT**
Refer to chart at end of section.	**1K3**

HALF-WAVE VACUUM RECTIFIER **1K3/1J3**



3C

Glass octal type used as a rectifier in high-voltage pulse circuits of black-and-white television receivers. Plate connection is cap at top of bulb. Socket terminals 1, 3, 4, 5, 6, and 8 may be connected to socket terminal 7 or to a corona shield which is connected to socket terminal 7. Socket terminals 4 and 6 may be used as tie points for components at or near filament potential. Outlines section, 14B; requires octal socket. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT.

Filament Voltage (ac/dc)	1.25	volts
Filament Current	0.2	ampere
Direct Interelectrode Capacitance (Approx.):		
Plate to Filament and Internal Shield	1.6	pF

* Under no circumstances should the filament voltage be less than 1.05 volts or greater than 1.45 volts.

Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	26000•	volts
Peak Plate Current	50	mA
Average Plate Current	0.5	mA

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 7 mA	225	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
• The dc component must not exceed 22000 volts.

1L6 Refer to chart at end of section.

1LA4 Refer to chart at end of section.

1LA6 Refer to chart at end of section.

1LB4 Refer to chart at end of section.

1LC5 Refer to chart at end of section.

1LC6 Refer to chart at end of section.

1LD5 Refer to chart at end of section.

1LE3 Refer to chart at end of section.

1LG5 Refer to chart at end of section.

1LH4 Refer to chart at end of section.

1LN5 Refer to chart at end of section.

1N2A Refer to chart at end of section.

1N5GT Refer to chart at end of section.

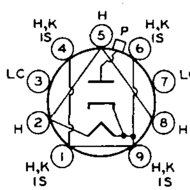
1N6G Refer to chart at end of section.

1P5GT Refer to chart at end of section.

1Q5GT Refer to chart at end of section.

1R5 Refer to chart at end of section.

1S2A/DY87 HALF-WAVE VACUUM RECTIFIER



9DT

Miniature type used in high-voltage, low-current applications in television scanning circuits. Outlines section, 7F; requires miniature 9-contact socket. Socket terminals 3 and 7 may be used as tie points for components at or near heater potential. For high-voltage considerations, refer to type 1G3GT/1B3GT.

Heater Voltage (ac/dc)	1.4	volts
Heater Current	0.55	ampere
Direct Interelectrode Capacitance:		
Plate to cathode and heater	1.8	pF

Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)

Peak Inverse Plate Voltage#	22000	volts
Peak Plate Current	40	mA
Average Plate Current	0.8	mA
CHARACTERISTICS, Instantaneous Value		
Tube Voltage Drop for plate current of 7 mA	100	volts

Pulse duration must not exceed 10% of a horizontal scanning cycle (10 microseconds).

Refer to chart at end of section. **1S4**

Refer to chart at end of section. **1S5**

Refer to chart at end of section. **1T4**

Refer to chart at end of section. **1T5GT**

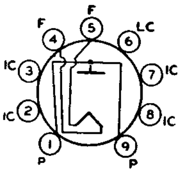
Refer to chart at end of section. **1T6**

Refer to chart at end of section. **1U4**

Refer to chart at end of section. **1U5**

Refer to chart at end of section. **1V**

HALF-WAVE VACUUM RECTIFIER **1V2**



9U

Miniature type used as a doubler in high-voltage pulse rectifier circuits of black-and-white television receivers and as a focus rectifier in color television receivers. The very low power required by the filament permits the use of a rectifier transformer having small size and light weight. **Outlines section,** 6B; requires miniature 9-contact socket.

Filament Voltage (ac)	0.625▪	volt
Filament Current	0.3	ampere
Direct Interelectrode Capacitance:		
Plate to Filament (Approx.)	0.8	pF

▪ Under no circumstances should the filament voltage be less than 0.525 volt or greater than 0.725 volt.

Pulsed Rectifier

For operation in a 525-line, 30-frame system

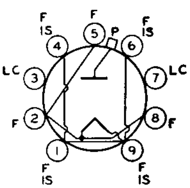
MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	8250•	volts
Peak Plate Current	11	mA
Average Plate Current	0.6	mA

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
• The dc component must not exceed 7000 volts.

Refer to chart at end of section. **1X2A**

HALF-WAVE VACUUM RECTIFIER **1X2B** **1X2B/1X2A**



9Y

Miniature type used as a rectifier in high-voltage pulse circuits of black-and-white television receivers and as a focus rectifier in color television receivers. **Outlines** section, 7A; requires miniature 9-contact socket. Socket terminals 3 and 7 may be used as tie points for components at or near filament potential. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT.

Filament Voltage (ac)	1.25*	volts
Filament Current	0.2	ampere
Direct Interelectrode Capacitance:		
Plate to Filament and Internal Shield (Approx.)	1	pF

* Under no circumstances should the filament voltage be less than 1.05 volts or greater than 1.45 volts.

Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
Peak Inverse Plate Voltage#	22000•	volts
Peak Plate Current	45	mA
Average Plate Current	0.5	mA
CHARACTERISTICS, Instantaneous Value		
Tube Voltage Drop for plate current of 7 mA	100	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
• The dc component must not exceed 18000 volts.

2A3 Refer to chart at end of section.

2A5 Refer to chart at end of section.

2A6 Refer to chart at end of section.

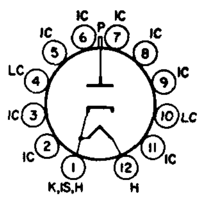
2A7 Refer to chart at end of section.

2AF4A Refer to chart at end of section.

2AF4B Refer to chart at end of section.

2AF4B/2DZ4 Refer to type 6AF4A.

2AH2 HALF-WAVE VACUUM RECTIFIER



12DG

Duodecar type used as a rectifier in high-voltage pulse circuits of black-and-white television receivers. Outlines section, 9A; requires duodecar 12-contact socket. Socket terminals 4 and 10 may be used as tie points for components at or near heater potential. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT. Heater: volts (ac/dc), 2.5; amperes, 0.3.

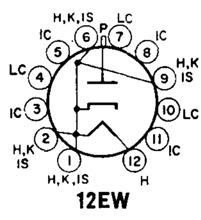
Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
Peak Inverse Plate Voltage#	30000•	volts
Peak Plate Current	80	mA
Average Plate Current	1.5	mA
CHARACTERISTICS, Instantaneous Value		
Tube Voltage Drop for plate current of 7 mA	100	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
• The dc component must not exceed 24000 volts.

2AS2 HALF-WAVE VACUUM RECTIFIER



12EW

Duodecar type used as a rectifier in high-voltage pulse circuits of black-and-white television receivers. Outlines section, 9B; requires duodecar 12-contact socket. Socket terminals 4, 7, and 10 may be used as tie points for components at or near heater potential. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT. Heater: volts (ac/dc), 2.5; amperes, 0.33.

Pulsed Rectifier

For operation in a 525-line, 30-frame system

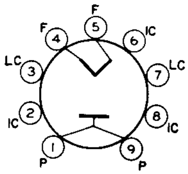
MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	30000•	volts
Peak Plate Current	80	mA
Average Plate Current	1.5	mA

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 7 mA	100	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
• The dc component must not exceed 24000 volts.



9U

HALF-WAVE VACUUM RECTIFIER — 2AV2

Miniature type used as a high-voltage, low-current pulse-operated focus rectifier in color television receivers. The filament of the tube can be operated directly across the filament winding of the horizontal-output transformer without a series voltage-dropping resistor. Outlines section, 6B; requires miniature 9-contact socket.

Filament Voltage (ac)	1.8	volts
Filament Current	0.225	ampere
Direct Interelectrode Capacitance (Approx.):		
Plate to Filament	0.8	pF

Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

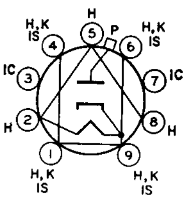
Peak Inverse Plate Voltage#	8250**	volts
Peak Plate Current	50	mA
Average Plate Current	0.6	mA

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 1 mA	20	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
** Under no circumstances should this absolute value be exceeded; the dc component must not exceed 7000 volts.

Refer to chart at end of section. **2B7**



9RT

HALF-WAVE VACUUM RECTIFIER — 2BJ2

Miniature type used as a rectifier in high-voltage pulse circuits of transistorized black-and-white television receivers. Outlines section, 7A; requires miniature 9-contact socket. Socket terminals 3 and 7 should not be used as tie points for external-circuit connections. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT.

Heater Voltage (ac/dc)	2.3	volts
Heater Current	0.3	ampere
Direct Interelectrode Capacitance:		
Plate to Cathode, Heater, and Internal Shield	1	pF

Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	20000	volts
Peak Plate Current	80	mA
Average Plate Current	1	mA

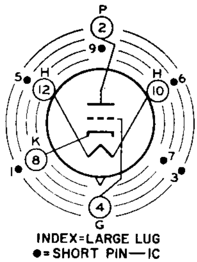
CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 7 mA	80	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

2BN4	Refer to chart at end of section.
2BN4A	Refer to type 6BN4A.
2CW4	Refer to type 6CW4.
2CY5	Refer to type 6CY5.
2DS4	Refer to type 6DS4.
2DV4	Refer to type 6DV4.
2DZ4	Refer to chart at end of section.
2E5	Refer to chart at end of section.

2EG4 HIGH-MU TRIODE



12AQ

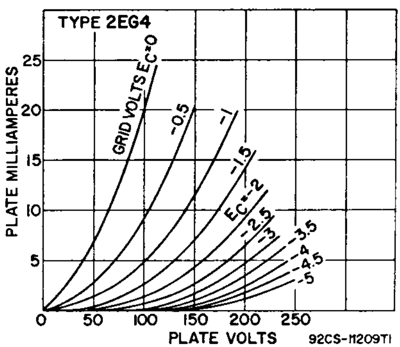
Nuvistor type used as a grounded-cathode, neutralized rf amplifier in vhf tuners of television and FM receivers. **Outlines section,** 1; requires nuvistor socket.

Heater Voltage (ac/dc)	1.7	volts
Heater Current	0.6	ampere
Heater Warm-up Time (Average)	8	seconds
Peak Heater-Cathode Voltage	±100	volts
Direct Interelectrode Capacitances (Approx.):		
Grid to Plate	0.92	pF
Grid to Cathode, Heater, and Shell	4.3	pF
Plate to Cathode, Heater, and Shell	1.8	pF
Plate to Cathode	0.18	pF
Heater to Cathode	1.6	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	300°	volts
Plate Voltage	135	volts
Grid Voltage:		
Negative-bias value	55	volts
Peak or dc positive value	0	volts
Plate Dissipation	1.5	watts
Cathode Current	15	mA



CHARACTERISTICS AND TYPICAL OPERATION	Characteristics	Typical Operation	
Plate Supply Voltage	110	70	volts
Grid Supply Voltage	—	0	volts
Cathode-Bias Resistor	130	—	ohms
Grid Resistor	—	47000	ohms
Amplification Factor	63	68	
Plate Resistance (Approx.)	7000	5440	ohms
Transconductance	9000	12500	μmhos
Grid Voltage (Approx.) for plate current of 100 μA	—5	—	volts
Grid Voltage (Approx.) for plate current of 10 μA	—6.8	—	volts
Plate Current	6.5	7	mA

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance:▪		
For fixed-bias operation	2.2	megohms
For cathode-bias operation	0.5	megohm

° A plate supply voltage of 300 volts may be used provided that a sufficiently large resistor is used in the plate circuit to limit the plate dissipation to 1.5 watts under any condition of operation.

▪ *For operation at metal-shell temperatures up to 135° C.*

Refer to chart at end of section. **2EN5**

Refer to type 6ER5. **2ER5**

Refer to type 6FH5. **2FH5**

Refer to type 6FQ5A. **2FQ5A**

Refer to type 6FS5. **2FS5**

Refer to type 6GK5. **2GK5**

Refer to type 6GU5. **2GU5**

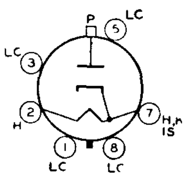
Refer to type 6HA5. **2HA5**

Refer to type 6HQ5. **2HQ5**

Refer to chart at end of section. **3A2**

Refer to chart at end of section. **3A3**

Refer to chart at end of section. **3A3/3B2**



8EZ

HALF-WAVE VACUUM RECTIFIER

3A3A
3A3A/3B2

Glass octal type used as a rectifier in high-voltage pulse circuits of color television receivers. Outlines section, 14E; require octal socket. Socket terminals 1, 3, 4, 5, 6, and 8 may be connected to socket terminal 7. Socket terminals 4 and 6 may be used as tie points for components at or near heater potential. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT.

Heater Voltage (ac)	3.15*	volts
Heater Current	0.22	ampere
Direct Interelectrode Capacitance (Approx.):		
Plate to Heater, Cathode, and Internal Shield	1.5	pF

* Under no circumstances should the heater voltage be less than 2.65 volts or greater than 3.65 volts.

Pulsed Rectifier

For operation in a 525-line, 30-frame system

***MAXIMUM RATINGS** (Design-Maximum Values)*

Peak Inverse Plate Voltage#	30000	volts

Peak Plate Current	100	mA
Average Plate Current	2	mA

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

3A8GT Refer to chart at end of section.

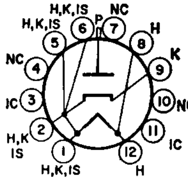
3AF4A Refer to chart at end of section.

3AF4A/3DZ4 Refer to type 6AF4A.

3AL5 Refer to type 6AL5.

3AT2

HALF-WAVE VACUUM RECTIFIER



12FV

Duodecar type used as a rectifier in high-voltage pulse circuits of color television receivers. **Outlines section,** 9B; requires duodecar 12-contact socket. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT.

Heater Voltage (ac/dc)	3.15	volts
Heater Current	0.22	ampere
Direct Interelectrode Capacitance:		
Plate to Cathode, Heater, and Internal Shield	1.5	pF

Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	30000	volts
Peak Plate Current	88	mA
Average Plate Current	1.7	mA

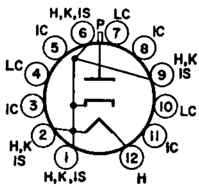
Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

3AU6 Refer to type 6AU6A.

3AV6 Refer to type 6AV6.

3AW2

HALF-WAVE VACUUM RECTIFIER



12EW

Duodecar type used as a high-voltage rectifier to supply power to the anode of the picture tube in color and black-and-white television receivers. **Outlines section,** 9B; requires duodecar 12-contact socket. Socket terminals 3, 5, 8 and 11 should not be used as tie points. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT. Heater: volts (ac/dc), 3.15; amperes, 0.35.

Flyback Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	38000•	volts
Peak Plate Current	110	mA
Average Plate Current	2.2	mA
CHARACTERISTICS, Instantaneous Value		
Tube Voltage Drop for plate current of 7 mA	60	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
• The dc component must not exceed 30000 volts.

3AW3 Refer to chart at end of section.

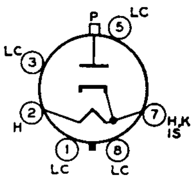
3B2 Refer to chart at end of section.

3BA6 Refer to chart at end of section.

3BC5 Refer to chart at end of section.

Refer to type 6BC5.	**3BC5/3CE5**
Refer to chart at end of section.	**3BE6**
Refer to chart at end of section.	**3BN4**
Refer to type 6BN4A.	**3BN4A**
Refer to type 6BN6.	**3BN6**
Refer to chart at end of section.	**3BU8**
Refer to type 6BU8.	**3BU8/3GS8**
Refer to type 6BY6.	**3BY6**
Refer to type 6BZ6.	**3BZ6**

HALF-WAVE VACUUM RECTIFIER — 3CA3



8MH

Glass octal type used as a rectifier in high-voltage pulse circuits of color television receivers. **Outlines section,** 14E; requires octal socket. Socket terminals 1, 3, 4, 5, 6, and 8 may be connected to terminal 7 or to a corona shield which connects to terminal 7. Socket terminals 4 and 6 may be used as tie points at or near cathode potential. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT.

Heater Voltage (ac)	3.6	volts
Heater Current	0.225	ampere
Direct Interelectrode Capacitance (Approx.):		
Plate to Heater, Cathode, and Internal Shield	1.6	pF

Pulsed Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	30000	volts
Peak Plate Current	100	mA
Average Plate Current	2	mA

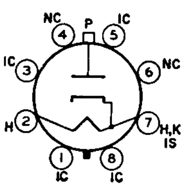
CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 11 mA	100	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

Refer to type 6CB6A.	**3CB6**
Refer to type 6CB6A.	**3CB6/3CF6**
Refer to chart at end of section.	**3CE5**
Refer to chart at end of section.	**3CF6**

HALF-WAVE VACUUM RECTIFIER — 3CN3A



8MU

Glass octal type used as a high-voltage rectifier to supply power to the anode of the picture tube in color and black-and-white television receivers. **Outlines section,** 14E; requires octal socket. Socket terminals 1, 3, 4, 5, 6, and 8 should not be used as tie points although they may be connected to terminal 7. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT. **Heater:** volts (ac/dc), 3.15; amperes, 0.48; operational warm-up time, 4 seconds.

Flyback Rectifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	38000•	volts

Peak Plate Current	110	mA
Average Plate Current	2.2	mA

CHARACTERISTICS, Instantaneous Value

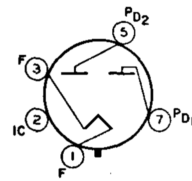
Tube Voltage Drop for plate current of 7 mA	60	volts

\# Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
• The dc component must not exceed 30000 volts.

3CS6 Refer to type 6CS6.

3CY5 Refer to type 6CY5.

3DG4 FULL-WAVE VACUUM RECTIFIER



5DE

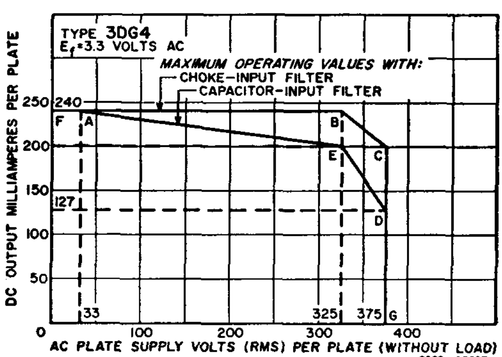
Glass octal type used in power supplies of color and black-and-white television receivers and other equipment having high dc requirements. **Outlines section,** 19E; requires octal socket. It is especially important that this tube, like other power-handling tubes, be adequately ventilated. For discussion of Rating Chart, refer to **Interpretation of Tube Data. Filament:** volts (ac/dc), 3.3; amperes, 3.8.

Full-Wave Rectifier

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage	1050	volts
Peak Plate Current (Per Plate)	1.2	amperes
Hot-Switching Transient Plate Current (Per Plate)	6.5	amperes
AC Plate Supply Voltage (Per Plate, rms)	See Rating Chart	
DC Output Current (Per Plate)	See Rating Chart	
Bulb Temperature (At hottest point on bulb surface)	200	°C

RATING CHART



TYPICAL OPERATION WITH CAPACITOR INPUT TO FILTER

AC Plate-to-Plate Supply Voltage (rms)	550	volts
Filter-Input Capacitor•	40	μF
Effective Plate-Supply Impedance per Plate	32	ohms
DC Output Voltage at Input to Filter (Approx.):		
At full-load current of 350 mA	300	volts

CHARACTERISTICS

Tube Voltage Drop for plate current of 350 mA (per plate)	25	volts

• Higher values of capacitance than indicated may be used, but the effective plate-supply impedance may have to be increased to prevent exceeding the maximum peak-plate-current rating.

3DK6 Refer to type 6DK6.

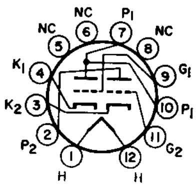
3DT6 Refer to chart at end of section.

3DT6A Refer to type 6DT6A.

Refer to chart at end of section.	**3DZ4**
Refer to chart at end of section.	**3EA5**
Refer to chart at end of section.	**3EH7**
Refer to type 6EH7/EF183.	**3EH7/XF183**
Refer to chart at end of section.	**3EJ7**
Refer to type 6EJ7/EF184.	**3EJ7/XF184**
Refer to type 6ER5.	**3ER5**
Refer to type 6FH5.	**3FH5**
Refer to type 6FS5.	**3FS5**
Refer to type 6GK5.	**3GK5**
Refer to chart at end of section.	**3GS8**
Refer to chart at end of section.	**3GS8/3BU8**
Refer to chart at end of section.	**3HA5**
Refer to type 6HM5/6HA5.	**3HM5/3HA5**
Refer to type 6HQ5.	**3HQ5**
Refer to chart at end of section.	**3HS8**
Refer to chart at end of section.	**3JC6**
Refer to type 6JC6A.	**3JC6A**
Refer to type 6JD6.	**3JD6**
Refer to type 6KT6.	**3KT6**
Refer to chart at end of section.	**3LF4**
Refer to chart at end of section.	**3Q4**
Refer to chart at end of section.	**3Q5GT**
Refer to chart at end of section.	**3S4**
Refer to chart at end of section.	**3V4**
Refer to type 6AU6A.	**4AU6**
Refer to type 6AV6.	**4AV6**
Refer to chart at end of section.	**4BC5**
Refer to type 6BC8.	**4BC8**
Refer to chart at end of section.	**4BL8**
Refer to type 6BL8/ECF80.	**4BL8/XCF80**
Refer to type 6BN6.	**4BN6**
Refer to type 6BQ7A.	**4BQ7A**
Refer to type 6BS8.	**4BS8**
Refer to chart at end of section.	**4BU8**
Refer to type 6BU8.	**4BU8/4GS8**
Refer to type 6BZ6.	**4BZ6**
Refer to type 6BZ7.	**4BZ7**
Refer to type 6CB6A.	**4CB6**

4CS6	Refer to type 6CS6.
4CY5	Refer to chart at end of section.
4DE6	Refer to type 6DE6.
4DK6	Refer to type 6DK6.
4DT6	Refer to chart at end of section.
4DT6A	Refer to type 6DT6A.
4EH7	Refer to type 6EH7/EF183.
4EJ7	Refer to type 6EJ7/EF184.
4ES8	Refer to chart at end of section.
4ES8/XCC189	Refer to type 6ES8/ECC189.
4EW6	Refer to type 6EW6.
4GK5	Refer to type 6GK5.
4GM6	Refer to type 6GM6.
4GS8 **4GS8/4BU8**	Refer to chart at end of section.
4GX7	Refer to chart at end of section.
4GZ5	Refer to chart at end of section.
4HA5	Refer to chart at end of section.
4HA5/PC900	Refer to type 6HA5.
4HA7	Refer to type 5HA7.

4HC7 DUAL TRIODE



12FR

Duodecar type used for sync clipper and agc-amplifier service in television receivers. Outlines section, 30E; requires duodecar 12-contact socket. Heater: volts (ac/dc), 4.2; amperes, 0.6; warm-up time, 11 seconds; maximum heater-cathode volts, ± 200 peak, 100 average.

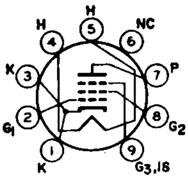
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1	Unit No.2	
Plate Voltage	330	330	volts
Grid Voltage:			
Positive-bias value	0	0	volts
Negative-bias value	100	100	volts
Peak Positive-Pulse Grid Voltage	60	—	volts
Plate Dissipation#	3	1.2	watts
CHARACTERISTICS			
Plate Voltage	150	150	volts
Grid Voltage	—1	—1	volt
Amplification Factor	23	100	
Plate Resistance (Approx.)	5200	53000	ohms
Transconductance	4400	1900	μmhos
Plate Current	18	1	mA
Grid Voltage (Approx.) for plate current of 10 μA	—13	—2.2	volts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance	5	5	megohms

A bias resistor or other means is required to protect the tube in absence of excitation.

4HM6	Refer to type 6HM6.
4HQ5	Refer to type 6HQ5.

Refer to type 6HS8. **4HS8**



9PM

SEMIREMOTE-CUTOFF PENTODE 4HT6

Miniature type with frame grid used in the if-amplifier stage of television receivers. **Outlines section, 6B**; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	4.2		volts
Heater Current	0.45		ampere
Heater Warm-up Time	11		seconds
Heater-Cathode Voltage			
Peak value	±200 max		volts
Average value	100 max		volts
Direct Interelectrode Capacitances:	**Unshielded**	**Shielded**	
Grid No.1 to Plate	0.031	0.024	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	8.7	8.7	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.15	3	pF

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	250	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Negative-bias value	50	volts
Cathode Current	25	mA
Plate Dissipation	2.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 125 volts	0.6	watt
For grid-No.2 voltages between 125 and 250 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	125	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.143	megohm
Transconductance	14000	μmhos
Plate Current	15	mA
Grid-No.2 Current	4	mA
Grid-No.1 Voltage (Approx.) for transconductance of 100 μmhos	—4.5	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

Refer to type 6JC6A. **4JC6 4JC6A**

Refer to type 6JD6. **4JD6**

Refer to type 6JH6. **4JH6**

Refer to type 6KE8. **4KE8**

Refer to type 6KT6. **4KT6**

Refer to type 6LJ8. **4LJ8**

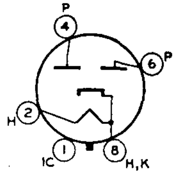
Refer to type 6AM8A. **5AM8**

Refer to type 6AN8A. **5AN8**

Refer to type 6AQ5A. **5AQ5**

5AR4/ GZ34

FULL-WAVE VACUUM RECTIFIER



5DA

Glass octal type used in power supply of television receivers and other equipment having high dc requirements. **Outlines section, 13F;** requires octal socket. It is especially important that this tube, like other power-handling tubes, be adequately ventilated. **Heater:** volts (ac/dc), 5; amperes, 1.9.

Full-Wave Rectifier

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage	1700	volts
Peak Plate Current (Per Plate)	825	mA
Hot-Switching Transient Plate Current (Per Plate)	3.7	amperes
AC Plate-Supply Voltage (Per Plate, rms, without load)	See Rating Chart	
Average Output Current (Per Plate)	See Rating Chart	

TYPICAL OPERATION WITH CAPACITOR INPUT TO FILTER

AC Plate-to-Plate Supply Voltage (rms)	450	550	volts
Effective Plate-Supply Impedance per Plate	160	200	ohms
Average Output Current	225	160	mA
DC Output Voltage at Input to Filter	475	620	volts

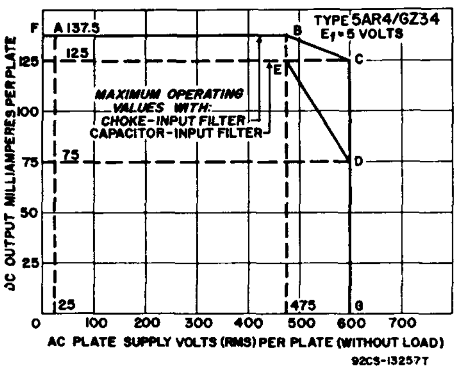
TYPICAL OPERATION WITH CHOKE INPUT TO FILTER

AC Plate-to-Plate Supply Voltage (rms)	450	550	volts
Filter Input Choke	10	10	henries
Average Output Current	250	225	mA
DC Output Voltage at Input to Filter	375	465	volts

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 225 mA (Per Plate)	—	17	volts

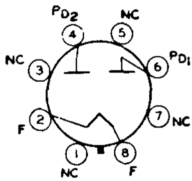
RATING CHART



5AS4 Refer to chart at end of section.

5AS4A

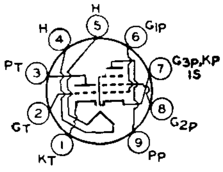
FULL-WAVE VACUUM RECTIFIER



5T

Glass octal type used in power supplies of television receivers having high dc requirements. **Outlines section,** 19D; requires octal socket. This type may be supplied with pins 3, 5, and 7 omitted. Vertical mounting is preferred, but horizontal mounting is permissible if pins 1 and 4 are in vertical plane. It is especially important that this tube, like other power-handling tubes, be adequately ventilated. **Heater:** volts (ac), 5; amperes, 3. For maximum ratings, typical operation, and curves, refer to type 5U4GB.

Refer to type 6AS8.	**5AS8**
Refer to type 6AT8A.	**5AT8**
Refer to chart at end of section.	**5AU4**



9DZ

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE **5AV8**

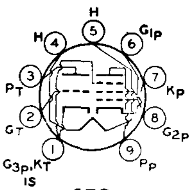
Miniature type used in television receiver applications, **Outlines section, 6B**; requires miniature 9-contact socket. **Heater**: volts (ac/dc), 4.7; amperes, 0.6; warm-up time (average), 11 seconds; maximum heater-cathode volts, ± 200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	300	volts
Grid No.2 Supply Voltage	—	300	volts
Grid-No.2 (Screen-Grid) Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	2	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 150 volts	—	0.5	watt
For grid-No.2 voltages between 150 and 300 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	200	200	volts
Grid-No.2 Supply Voltage	—	150	volts
Grid-No.1 Voltage	—6	—	volts
Cathode-Bias Resistor	—	180	ohms
Amplification Factor	19	—	
Plate Resistance (Approx.)	5750	300000	ohms
Transconductance	3300	6200	µmhos
Plate Current	13	9.5	mA
Grid-No.2 Current	—	2.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 µA	—19	—8	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:*			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

* If either unit is operating at maximum rated conditions, grid-No.1-circuit resistance for both units should not exceed the stated values.

Refer to chart at end of section.	**5AW4**
Refer to chart at end of section.	**5AZ4**



9EC

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE **5B8**

Miniature type used as combined vhf oscillator and mixer in television receivers. **Outlines section, 6B**; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	4.7	volts
Heater Current	0.6	ampere
Heater Warm-up Time	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	1.7	pF
Grid to Cathode, Heater, Pentode Grid No.3, and Internal Shield	1.9	pF
Plate to Cathode, Heater, Pentode Grid No.3, and Internal Shield	1.4	pF

Pentode Unit:		
Grid No.1 to Plate	0.05 max	pF
Grid No.1 to Cathode, Heater, and Grid No.2	6	pF
Plate to Cathode, Heater, Grid No.2, and Internal Shield	2.6	pF
Plate to Cathode, Heater, and Grid No.2	0.15	pF
Triode Grid to Pentode Plate	0.0078	pF
Pentode Grid No.1 to Triode Plate	0.0033	pF
Pentode Plate to Triode Plate	0.06	pF

Class A_1 Amplifier

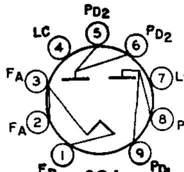
MAXIMUM RATINGS (Design-Center Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	300	volts
Grid No.2 (Screen-Grid) Supply Voltage	—	300	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	2	watts
Grid No.2 Input:			
For grid-No.2 voltages up to 150 volts	—	0.5	watt
For grid-No.2 voltages between 150 and 300 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	200	200	volts
Grid-No.2 Supply Voltage	—	150	volts
Grid Voltage	—6	—	volts
Cathode-Bias Resistor	—	180	ohms
Amplification Factor	19	—	
Plate Resistance (Approx.)	5750	300000	ohms
Transconductance	3300	6200	μmhos
Plate Current	13	9.5	mA
Grid-No.2 Current	—	2.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—19	—8	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance*:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

* If either unit is operated at maximum rated conditions, grid-No.1-circuit resistance for both units should not exceed the stated values.

5BC3 Refer to chart at end of section.

5BC3A

FULL-WAVE VACUUM RECTIFIER



9QJ

Novar types used in power supplies of radio equipment and television receivers having high dc requirements. **Outlines section,** 31C; requires novar 9-contact socket. Vertical operation is preferred, but tubes may be operated in horizontal position if pins 2 and 7 are in vertical plane. It is especially important that these tubes, like other power-handling tubes, be adequately ventilated. **Filament:** volts (ac), 5; amperes, 3.

Full-Wave Rectifier

MAXIMUM RATINGS (Design-Maximum Values)		
Peak Inverse Plate Voltage	1700	volts
Peak Plate Current (Per Plate)	1	ampere
Hot-Switching Transient Plate Current (Per Plate)°	5	amperes
AC Plate-Supply Voltage (Per Plate, rms)	See Rating Chart	
Average Output Current (Per Plate)	See Rating Chart	

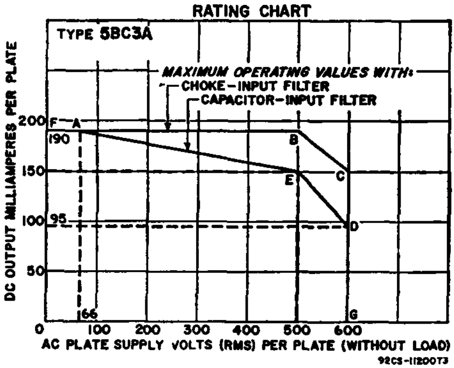
TYPICAL OPERATION WITH CAPACITOR INPUT TO FILTER				
AC Plate-to-Plate Supply Voltage (rms)	600	900	1100	volts
Filter-Input Capacitor▪	40	40	40	μF
Total Effective Plate-Supply Impedance per Plate				
DC Output Voltage at Input to Filter (Approx.):	21	67	97	ohms
At load current of: 300 mA	290	—	—	volts
275 mA	—	460	—	volts
162 mA	—	—	630	volts
150 mA	335	—	—	volts
137.5 mA	—	520	—	volts
81 mA	—	—	680	volts

TYPICAL OPERATION WITH CHOKE INPUT TO FILTER

AC Plate-to-Plate Supply Voltage (rms)	900	1100	volts
Filter-Input Choke	10	10	henries
DC Output Voltage at Input to Filter (Approx.):			
At load current of: 348 mA	340	—	volts
275 mA	—	440	volts
174 mA	355	—	volts
137.5 mA	—	445	volts

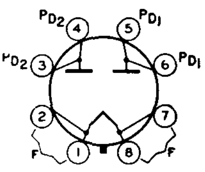
° If hot switching is regularly required in operation, the use of choke-input circuits is recommended. Such circuits limit the hot-switching current to a value no higher than that of the peak plate current. When capacitor-input circuits are used, a maximum peak current value per plate of 5 amperes during the initial cycles of the hot-switching transient should not be exceeded.

▪ Higher values of capacitance than indicated may be used, but the effective plate-supply impedance may have to be increased to prevent exceeding the maximum rating for peak plate current.



Refer to chart at end of section.	**5BE8**
Refer to type 6BK7B.	**5BK7A**
Refer to type 6BQ7A.	**5BQ7A**
Refer to type 6BR8A.	**5BR8**
Refer to chart at end of section.	**5BT8**
Refer to type 6BW8.	**5BW8**
Refer to type 6CG8A.	**5CG8**
Refer to chart at end of section.	**5CL8**
Refer to type 6CL8A.	**5CL8A**
Refer to chart at end of section.	**5CM8**
Refer to type 6CQ8.	**5CQ8**
Refer to type 6CZ5.	**5CZ5**
Refer to chart at end of section.	**5DH8**

FULL-WAVE VACUUM RECTIFIER — 5DJ4



Glass octal type used in power supplies of radio and television receivers having high dc requirements. Outlines section, 19E; requires octal socket. Operation in vertical position is preferred, but horizontal operation is permissible if pins 2 and 4 are in vertical plane. It is especially important that this tube, like other power-handling tubes, be adequately ventilated. Filament: volts (ac/dc), 5; amperes, 3.

Full-Wave Rectifier

MAXIMUM RATINGS (Design-Maximum Values)

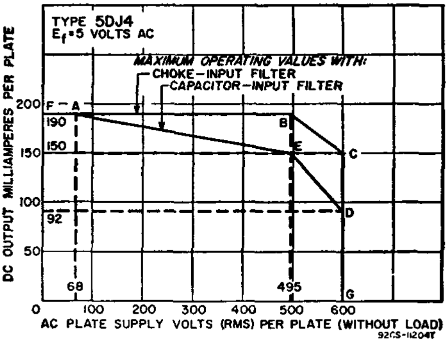
Peak Inverse Plate Voltage	1700	volts
Peak Plate Current (Per Plate)	1	ampere
Hot-Switching Transient Plate Current (Per Plate)	5	amperes
AC Plate-Supply Voltage (Per Plate, rms, without load)	See Rating Chart	
Average Output Current (Per Plate)	See Rating Chart	

TYPICAL OPERATION

Filter Input	Capacitor		Choke	
AC Plate-to-Plate Supply Voltage (rms, without load)	600	900	1100	volts
Filter-Input Capacitor°	40	40	—	μF
Filter-Input Choke	—	—	10	henries
Effective Plate-Supply Impedance per Plate	21	67	—	ohms
DC Output Voltage at Input to Filter (Approx.)	290	460	420	volts
Average Output Current	300	275	275	mA

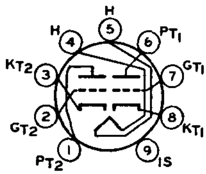
° When capacitor values greater than 40 μF are used, the effective plate-supply impedance should be increased so that the maximum rating for peak plate current is not exceeded.

RATING CHART



5EA8 Refer to type 6EA8.

5ES8
5ES8/
YCC189 VARIABLE-MU TWIN TRIODE



9AJ

Miniature type used as cascode-type amplifier in tuners of television receivers. **Outlines section,** 6B; requires miniature 9-contact socket. **Heater:** volts (ac/dc), 5.6; amperes, 0.45.

Class A_1 Amplifier (Each Unit)

CHARACTERISTICS

Plate Voltage	90	90	90	volts
Grid Voltage	−1.4	−5	−9	volts
Plate Resistance	2500	—	—	ohms
Transconductance	12500	625	125	μmhos
Plate Current	15	—	—	mA

Cascode-Type Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage with plate current of 0 mA	550	volts
Plate Voltage	130	volts
Grid Voltage, Negative-bias value	50	volts
Cathode Current	22	mA
Plate Dissipation	1.8	watts
Heater-Cathode Voltage:		
Grounded-cathode section▲	80	volts
Grounded-grid section (cathode positive)•	180*	volts

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance:		
Grounded-cathode section	1	megohm
Grounded-grid section	0.5	megohm

* The dc component must not exceed 130 volts.
▲ Pins 6, 7, and 8.
• Pins 1, 2, and 3.

Refer to type 6EU8. **5EU8**

Refer to type 6EW6. **5EW6**

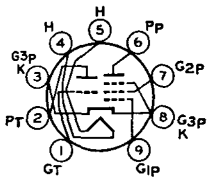
Refer to type 6FG7. **5FG7**

Refer to type 6FV8A. **5FV8**

Refer to type 6GH8A. **5GH8A**

Refer to type 6GJ7/ECF801. **5GJ7**

Refer to type 6GM6. **5GM6**



9GF

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE 5GS7

Miniature type used as frequency changer in vhf television tuners. **Outlines section,** 6B; requires miniature 9-contact socket. **Heater:** volts, 5.4; amperes 0.45; maximum heater-cathode volts, ±100.

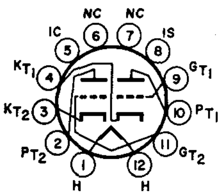
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)	**Triode Unit**	**Pentode Unit**	
Plate Voltage	125	250	volts
Grid-No.2 Input	—	150	volts
Cathode Current	15	18	mA
Plate Dissipation	1.5	2	watts
Grid-No. 2 Input	—	0.5	watt
CHARACTERISTICS			
Plate Voltage	100	170	volts
Grid-No.2 Voltage	—	150	volts
Grid-No.1 (Control-Grid) Voltage	—3	—1.2	volts
Plate Resistance	—	0.35 min	megohm
Transconductance	5500	12000	megohm
Plate Current	14	10	mA
Grid-No.2 Current	—	3.3	mA
Amplification Factor	17	—	

MAXIMUM CIRCUIT VALUES	**Triode Unit**	**Pentode Unit**	
Grid-No.1-Circuit Resistance	0.5	—	megohm
For fixed-bias operation	—	0.25	megohm
For automatic-bias operation	—	0.5	megohm

Refer to chart at end of section. **5GX6**

Refer to chart at end of section. **5GX7**



12FQ

DUAL TRIODE 5HA7

4HA7

Duodecar type used as a sync clipper and agc amplifier in television receivers. **Outlines section,** 8A; requires duodecar 12-contact socket. Type 4HA7 is identical with type 5HA7 except for the heater ratings.

	4HA7	**5HA7**	
Heater Voltage (ac/dc)	4.2	5.6	volts
Heater Current	0.6	0.45	ampere

Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1	Unit No.2	
Plate Voltage	330	330	volts
Grid Voltage:			
Positive-bias value	0	0	volts
Negative-bias value	50	50	volts
Cathode Current	20	—	mA
Plate Dissipation	2.75	0.8	watts
CHARACTERISTICS			
Plate Voltage	250	250	volts
Grid Voltage	—8.5	—2	volts
Amplification Factor	17	100	
Plate Resistance (Approx.)	7700	62500	ohms
Transconductance	2200	1600	µmhos
Plate Current	10.5	1.2	mA
Grid Voltage (Approx.) *for plate current of 10 µA*	—24	—	volts

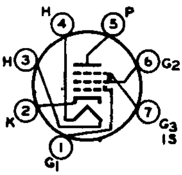
5HB7 Refer to type 6HB7.

5HG8 Refer to chart at end of section.

5HZ6 Refer to type 6HZ6.

5J6 Refer to type 6J6A.

5JK6 SHARP-CUTOFF PENTODE



7CM

Miniature type used for if-amplifier applications in color and black-and-white television receivers. Outlines section, 5C; requires miniature 7-contact socket.

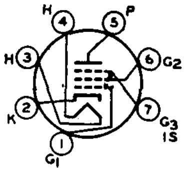
Heater Voltage (ac/dc)	4.9	volts
Heater Current	0.45	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.02 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	9.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.7	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	275	volts
Grid-No.2 (Screen-Grid) Supply Voltage	275	volts
Grid-No.2 Voltage	See curve page 98	
Cathode Current	22	mA
Plate Dissipation	2.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 137.5 volts	0.6	watt
For grid-No.2 voltages between 137.5 and 275 volts	See curve page 98	
CHARACTERISTICS		
Plate Voltage	125	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Voltage	125	volts
Cathode-Bias Resistor	68	ohms
Plate Resistance (Approx.)	18000	µmhos
Transconductance	0.15	megohm
Plate Current	11.5	mA
Grid-No.2 Current	3.9	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—3.5	volts
Input Resistance at 44 MHz	4000	ohms

MAXIMUM CIRCUIT VALUE

Grid-No.1 Circuit Resistance, for cathode-bias operation	0.5	megohm



7CM

SEMIREMOTE-CUTOFF PENTODE 5JL6

Miniature type with frame grid used for agc-controlled if-amplifier applications in television receivers. Outlines section, 5C; requires miniature 7-contact socket. Heater: volts (ac/dc), 4.9; amperes, 0.45; warm-up time (average), 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

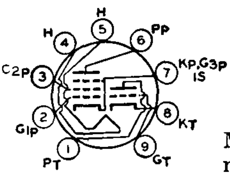
Plate Voltage	275	volts
Grid-No.2 (Screen-Grid) Supply Voltage	275	volts
Grid-No.2 Voltage	See curve page 98	
Plate Dissipation	2.5	watts
Cathode Current	22	mA
Grid-No.2 Input	0.6	watt

CHARACTERISTICS

Plate Voltage	125	volts
Grid-No.3 (Suppressor Grid)	Connected to cathode at socket	
Grid-No.2 Voltage	60	volts
Cathode-Bias Resistor	68	ohms
Plate Resistance (Approx.)	120000	ohms
Transconductance	15500	μmhos
Plate Current	12.5	mA
Grid-No.2 Current	4	mA
Grid-No.1 Voltage (Approx.) for transconductance of 1500 μmhos	—2.7	volts
Grid-No.1 Voltage (Approx.) for transconductance of 150 μmhos	—5.5	volts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, for fixed-bias operation	0.5	megohm



9AE

MEDIUM-MU TRIODE—SHARP-CUTOFF PENTODE 5KD8

Miniature type used as combined vhf oscillator and mixer tube in television receivers. Outlines section, 6B; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	5.6	volts
Heater Current	0.45	ampere
Heater Warm-up Time	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage		See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watt
For grid-No.2 voltages between 165 and 330 volts		See curve page 98	
CHARACTERISTICS			
Plate Voltage	125	125	volts
Grid-No.2 Voltage	—	110	volts
Grid-No.1 Voltage	—1	—1	volt

	Triode Unit	Pentode Unit	
Amplification Factor	40	—	
Plate Resistance (Approx.)	—	0.2	megohm
Transconductance	7500	5000	µmhos
Plate Current	13.5	9.5	mA
Grid-No.2 Current	—	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—9	—8	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	1	megohm

5KE8 Refer to type 6KE8.

5KZ8 Refer to type 6KZ8.

5LJ8 Refer to type 6LJ8.

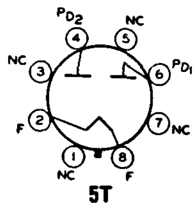
5T4 Refer to chart at end of section.

5T8 Refer to type 6T8A.

5U4G Refer to chart at end of section.

5U4GB

FULL-WAVE VACUUM RECTIFIER



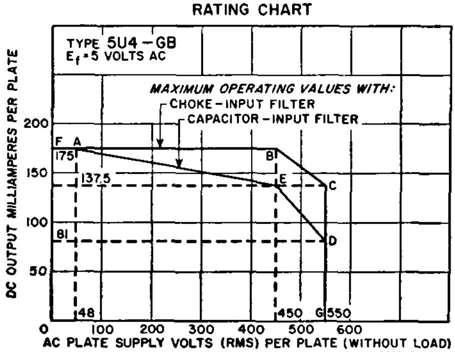
Glass octal type used in power supplies of radio and color and black-and-white television receivers having high dc requirements. **Outlines section,** 19E; requires octal socket. This type may be supplied with pins 3, 5, and 7 omitted. Vertical mounting is preferred, but horizontal mounting is permissible if pins 1 and 4 are in vertical plane. The coated filament is designed to operate from the ac line through a step-down transformer. The voltage at the filament terminals should be 5 volts at an average line voltage of 117 volts. It is especially important that this tube, like other power-handling tubes, be adequately ventilated. For discussion of Rating Chart and Operation Characteristics, refer to **Interpretation of Tube Data. Filament:** volts (ac), 5; amperes, 3.

Full-Wave Rectifier

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage	1550	volts
Peak Plate Current (Per Plate)	1	ampere
Hot-Switching Transient Plate Current (Per Plate)	#	
AC Plate Supply Voltage (Per Plate, rms)	See Rating Chart	
Average Output Current (Per Plate)	See Rating Chart	

RATING CHART



TYPICAL OPERATION WITH CAPACITOR INPUT TO FILTER

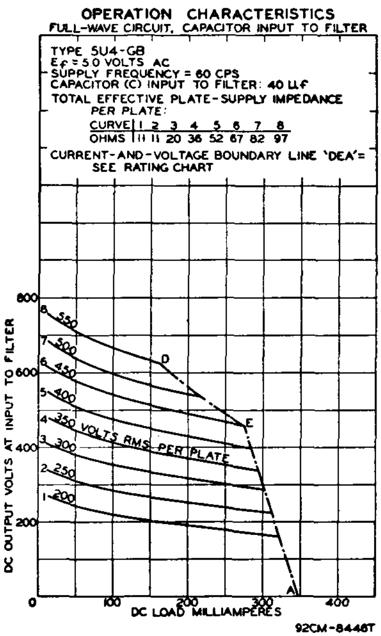
AC Plate-to-Plate Supply Voltage (rms)	600	900	1100	volts
Filter-Input Capacitor*	40	40	40	µF
Total Effective Plate-Supply Impedance per Plate	21	67	97	ohms
DC Output Voltage at Input to Filter (Approx.):				
At half-load current of 150 mA	335	—	—	volts
At half-load current of 137.5 mA	—	520	—	volts
At half-load current of 81 mA	—	—	680	volts
At full-load current of 300 mA	290	—	—	volts
At full-load current of 275 mA	—	460	—	volts
At full-load current of 162 mA	—	—	630	volts
Voltage Regulation (Approx.): Half-load to full-load current	45	60	50	volts

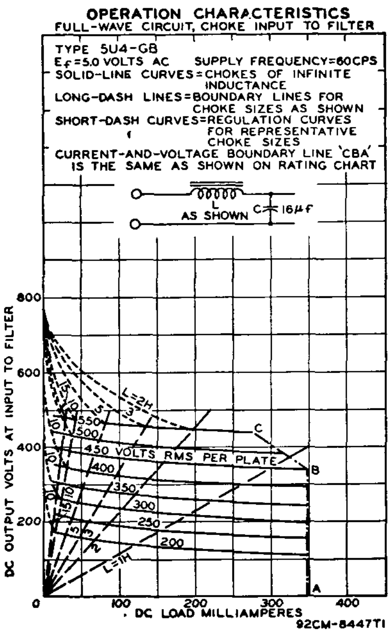
TYPICAL OPERATION WITH CHOKE INPUT TO FILTER

AC Plate-to-Plate Supply Voltage (rms)	900	1100	volts
Filter-Input Choke	10	10	henries
DC Output Voltage at Input to Filter (Approx.):			
At half-load current of 174 mA	355	—	volts
At half-load current of 137.5 mA	—	455	volts
At full-load current of 348 mA	340	—	volts
At full-load current of 275 mA	—	440	volts
Voltage Regulaton (Approx.): Half-load to full-load current	15	15	volts

If hot switching is regularly required in operation, the use of choke-input circuits is recommended. Such circuits limit the hot-switching current to a value no higher than that of the peak plate current. When capacitor-input circuits are used, a maximum peak current value per plate of 4.6 amperes during the initial cycles of the hot-switching transient should not be exceeded.

* Higher values of capacitance than indicated may be used, but the effective plate-supply impedance may have to be increased to prevent exceeding the maximum rating for peak plate current.



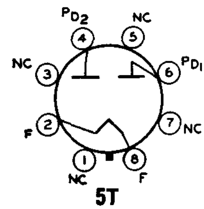


Refer to type 6U8A. **5U8**

5U9/LCF201 Refer to type 6U9/ECF201.

5V3 Refer to chart at end of section.

5V3A FULL-WAVE VACUUM RECTIFIER



Glass octal type used in power supplies of color and black-and-white television receivers and other equipment having high dc requirements. **Outlines section, 19E**; requires octal socket. Vertical mounting is preferred, but horizontal mounting is permissible if pins 2 and 4 are in vertical plane. It is especially important that this tube, like other power-handling tubes, be adequately ventilated. For discussion of Rating Chart, refer to **Interpretation of Tube Data. Filament:** volts (ac/dc), 5; amperes, 3.

Full-Wave Rectifier

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage	1550	volts
Peak Plate Current (Per Plate)	1.4	amperes
Hot-Switching Transient Plate Current (Per Plate)	6.6	amperes
AC Plate-Supply Voltage (Per Plate, rms, without load)	550	volts
Average Output Current (Per Plate)	415°	mA

° With capacitor-input filter for ac plate-supply volts (rms, per plate, without load) = 470.

TYPICAL OPERATION

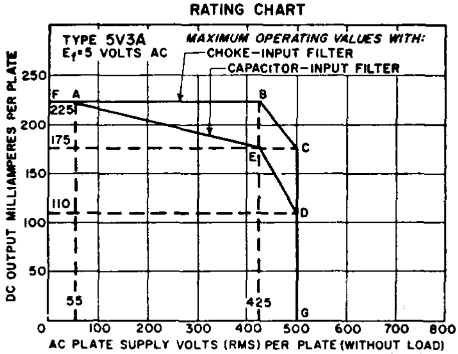
Filter Input	Capacitor	Choke	
AC Plate-to-Plate Supply Voltage (rms)	850	1000	volts
Filter-Input Capacitor■	40	—	µF
Effective Plate-Supply Impedance per Plate	50	—	ohms
Minimum Filter-Input Choke	—	10	henries
Average Output Current	350	350	mA
DC Output at Input to Filter (Approx.)	440	390	volts

CHARACTERISTICS

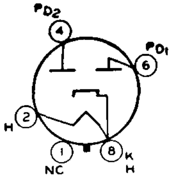
Tube Voltage Drop for plate current of 350 mA (per plate)	42	volts

■ When capacitor values greater than 40 µF are used, the effective plate-supply impedance should be increased so that the maximum rating for peak plate current is not exceeded.

RATING CHART



5V4G Refer to chart at end of section.



5L

FULL-WAVE VACUUM RECTIFIER

5V4GA

Glass octal types used in full-wave power supplies having high dc requirements. **Outlines section,** 19B; requires octal socket. The heater is designed to operate from the ac line through a step-down transformer. The voltage at the heater terminals should be 5 volts under operating conditions at an average line voltage of 117 volts. It is especially important that these tubes, like other power-handling tubes, be adequately ventilated. **Heater:** volts (ac/dc) 5; amperes, 2.

Full-Wave Rectifier

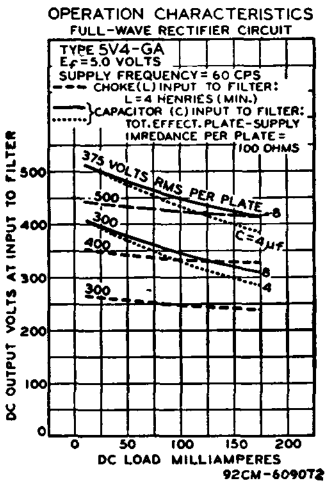
MAXIMUM RATINGS (Design-Center Values)

Peak Inverse Plate Voltage	1400	volts
AC Plate-Supply Voltage (Per Plate, rms):		
With capacitor-input filter	375	volts
With choke-input filter	500	volts
Peak Plate Current (Per Plate)	525	mA
Average Output Current	175	mA

TYPICAL OPERATION

Filter Input	Capacitor	Choke	
AC Plate-to-Plate Supply Voltage (rms)	750	1000	volts
Filter-Input Capacitor*	10	—	µF
Total Effective Plate-Supply Impedance per Plate	100	—	ohms
Filter-Input Choke	—	4	henries
DC Output Voltage at Input to Filter (Approx.):			
At output current of 175 mA	410	410	volts

* Higher values of capacitance than indicated may be used, but the effective plate-supply impedance may have to be increased to prevent exceeding the maximum rating for peak plate current.



Refer to type 6V6. **5V6GT**

Refer to chart at end of section. **5W4** **5W4GT**

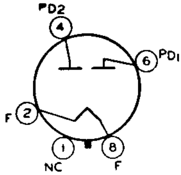
Refer to chart at end of section. **5X4G**

5X8 Refer to type 6X8A.

5Y3G Refer to chart at end of section.

5Y3GT

FULL-WAVE VACUUM RECTIFIER



5T

Glass octal type used in power supplies of radio and television equipment having moderate dc requirements. **Outlines section,** 13E; requires octal socket. Vertical mounting is preferred, but horizontal mounting is permissible if pins 2 and 8 are in horizontal plane. It is especially important that this tube, like other power-handling tubes, be adequately ventilated. For discussion of Rating Chart and Operating Characteristics, refer to **Interpretation of Tube Data. Filament:** volts (ac), 5; amperes, 2.

Full-Wave Rectifier

MAXIMUM RATINGS (Design-Center Values)

Peak Inverse Plate Voltage	1400	volts
Peak Plate Current (Per Plate)	440	mA
Hot-Switching Transient Plate Current (Per Plate)	2.5	amperes
AC Plate Supply Voltage (Per Plate, rms)	See Rating Chart	
DC Output Current (Per Plate)	See Rating Chart	

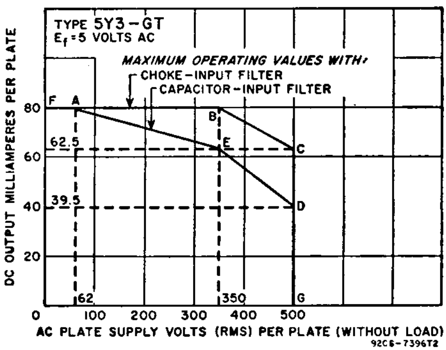
TYPICAL OPERATION WITH CAPACITOR INPUT TO FILTER

AC Plate-to-Plate Supply Voltage (rms)	700	1000	volts
Filter Input Capacitor*	20	10	µF
Effective Plate-Supply Impedance per Plate	50	140	ohms
DC Output Voltage at Input to Filter (Approx.):			
At half-load current of 62.5 mA	390	—	volts
At half-load current of 42 mA	—	610	volts
At full-load current of 125 mA	360	—	volts
At full-load current of 84 mA	—	560	volts
Voltage Regulation (Approx.):			
Half-load to full-load current	40	50	volts

TYPICAL OPERATION WITH CHOKE INPUT TO FILTER

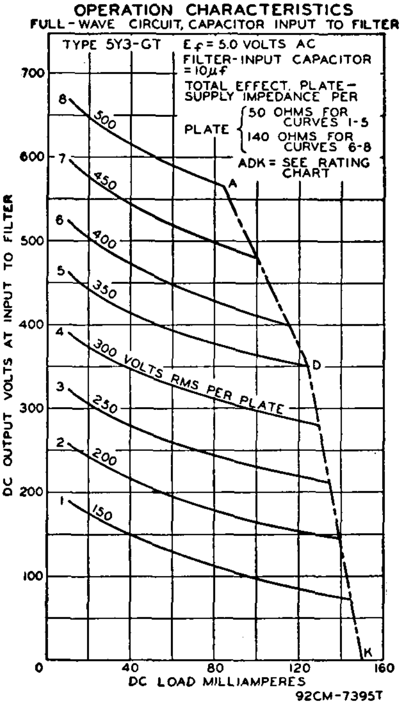
AC Plate-to-Plate Supply Voltage (rms)	700	1000	volts
Filter Input Choke#	10	10	henries
DC Output Voltage at Input to Filter (Approx.):			
At half-load current of 75 mA	270	—	volts
At half-load current of 62.5 mA	—	405	volts
At full-load current of 150 mA	245	—	volts
At full-load current of 125 mA	—	380	volts
Voltage Regulation (Approx.):			
Half-load to full-load current	25	15	volts

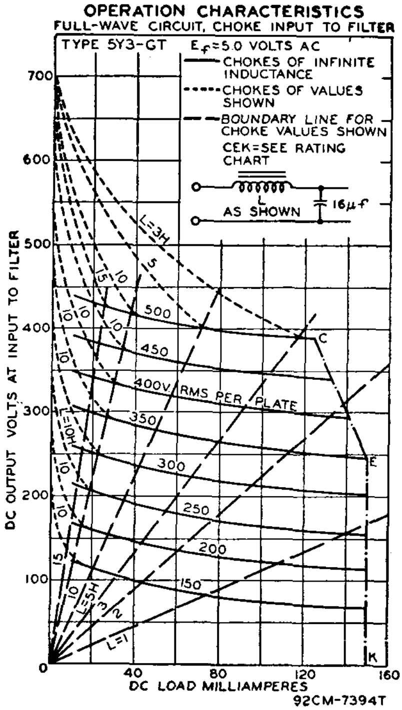
RATING CHART



* *Higher values of capacitance than indicated may be used but the effective plate supply impedance may have to be increased to prevent exceeding the maximum rating for hot-switching transient plate current.*

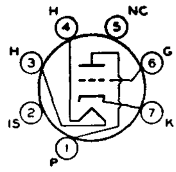
This value is adequate to maintain optimum regulation in the region to the right of line L = 10H on curve OPERATION CHARACTERISTICS with Choke Input to Filter, provided the load currents are not less than 35 mA and 50 mA, respectively, for plate-to-plate supply voltages of 700 and 1000 volts (rms).





Refer to chart at end of section.	**5Y4G**
Refer to chart at end of section.	**5Y4GA** **5Y4GT**
Refer to chart at end of section.	**5Z3**
Refer to chart at end of section.	**5Z4**
Refer to chart at end of section.	**6A3**
Refer to chart at end of section.	**6A6**
Refer to chart at end of section.	**6A7**
Refer to chart at end of section.	**6A7S**
Refer to chart at end of section.	**6A8**
Refer to chart at end of section.	**6A8G** **6A8GT**

6AB4 HIGH-MU TRIODE

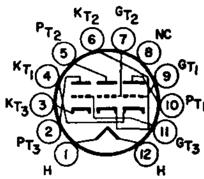


5CE

Miniature type used as cathode-drive amplifier, frequency converter, or oscillator at frequencies up to 300 MHz in television and FM receivers. Outlines section, 5C; requires miniature 7-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.15. For maximum ratings, characteristics, and curves, refer to type 12AT7.

6AB5/6N5	Refer to chart at end of section.
6AB7	Refer to chart at end of section.
6AC5GT	Refer to chart at end of section.
6AC7	Refer to chart at end of section.

6AC10 HIGH-MU TRIPLE TRIODE


12FE

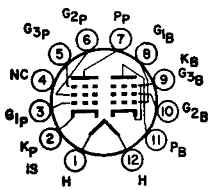
Duodecar type used in matrixing (color-difference) circuits of color television receivers. Outlines section, 8C; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.6; average warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	330	volts
Plate Dissipation	2	watts
CHARACTERISTICS		
Plate Voltage	200	volts
Cathode-Bias Resistor	150	ohms
Amplification Factor	62	
Plate Resistance (Approx.)	10700	ohms
Transconductance	5800	μmhos
Plate Current	9	mA
Grid Voltage (approx.) for plate current of 100 μA	—5	volts
MAXIMUM CIRCUIT VALUE		
Grid-Circuit Resistance	0.5	megohm

6AD6G	Refer to chart at end of section.
6AD7G	Refer to chart at end of section.

6AD10 BEAM POWER TUBE— SHARP-CUTOFF PENTODE


12EZ

Duodecar type used as FM detector and audio-frequency output amplifier in color and black-and-white television receivers. Outlines section, 8B; requires duodecar 12-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	1.05	amperes
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Direct Interelectrode Capacitances:		
Beam Power Unit:		
Grid No.1 to Plate	0.26	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	11	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	11	pF
Pentode Unit:		
Grid No.1 to Plate	0.024	pF
Grid No.3 to Plate	3.4	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	8	pF
Grid No.3 to Cathode, Heater, Grid No.1, Grid No.2, Plate, and Internal Shield	9.5	pF
Grid No.1 to Grid No.3	0.12	pF
Plate of Beam Power Unit to Plate of Pentode Unit	0.34	pF

Beam Power Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	275	volts
Grid-No.2 (Screen-Grid) Voltage	275	volts
Plate Dissipation	10	watts
Grid-No.2 Input	2	watts

TYPICAL OPERATION

Plate Voltage	250	volts
Grid-No.2 Voltage	250	volts
Grid-No.1 (Control-Grid) Voltage	—8	volts
Peak AF Grid-No.1 Voltage	8	volts
Zero-Signal Plate Current	35	mA
Maximum-Signal Plate Current	39	mA
Zero-Signal Grid-No.2 Current	2.5	mA
Maximum-Signal Grid-No.2 Current	7	mA
Plate Resistance (Approx.)	0.1	megohm
Transconductance	6500	μmhos
Load Resistance	5000	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	4.2	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	0.5	megohm

Pentode Unit as Class A_1 Amplifier

CHARACTERISTICS

Plate Supply Voltage	150	volts
Grid No.3 (Control Grid)	Connected to negative end of cathode resistor	
Grid-No.2 (Screen-Grid) Voltage	100	volts
Grid No.1 (Control Grid)	Connected to negative end of cathode resistor	
Cathode-Bias Resistor	180	ohms
Plate Resistance (Approx.)	0.11	megohm
Transconductance, Grid No.1 to Plate	3400	μmhos
Transconductance, Grid No.3 to Plate	600	μmhos
Plate Current	3.2	mA
Grid-No.2 Current	3.2	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—4.5	volts
Grid-No.3 Voltage (Approx.) for plate current of 20 μA	—7	volts

Pentode Unit as FM Sound Detector

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	300	volts
Grid-No.3 Voltage:		
Negative-bias value	100	volts
Positive-bias value	25	volts
Grid-No.2 Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 Voltage:		
Negative-bias value	—50	volts
Positive-bias value	0	volts
Plate Dissipation	1.7	watts
Grid-No.3 Input	0.1	watt
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	1	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

MAXIMUM CIRCUIT VALUES

Grid-No.3-Circuit Resistance	0.68	megohm
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.22	megohm
For cathode-bias operation	0.47	megohm

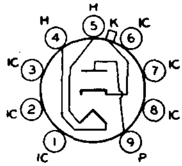
6AE5GT Refer to chart at end of section.

6AE6G Refer to chart at end of section.

6AE7GT Refer to chart at end of section.

6AF3
12AF3

HALF-WAVE VACUUM RECTIFIER



9CB

Miniature type used as a damper tube in horizontal-deflection circuits of television receivers. **Outlines section**, 7C; requires miniature 9-contact socket. Socket terminals 1, 2, 3, 6, 7, and 8 should not be used as tie points. It is especially important that this tube, like other power-handling tubes, be adequately ventilated. Type 12AF3 is identical with type 6AF3 except for heater ratings.

	6AF3	12AF3	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	1.2	0.6	amperes
Heater Warm-up Time (Average)	—	11	seconds

Damper Service

For operation in a 525-line, 30-frame system

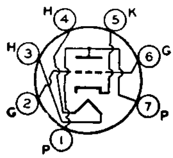
MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		4500	volts
Peak Plate Current		750	mA
Average Plate Current		185	mA
Bulb Temperature (At hottest point)		210	°C
Heater-Cathode Voltage:			
Peak value	+300	—4500	volts
Average value	+100	—1000	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6AF4
6AF4A
2AF4B/2DZ4, 3AF4A/3DZ4

MEDIUM-MU TRIODE



7DK

Miniature types used as local oscillators in uhf television receivers covering the frequency range of 470 to 890 MHz. **Outlines section**, 5C and 5B, respectively; requires miniature 7-contact socket. Types 2AF4B/2DZ4 and 3AF4A/3DZ4 are identical with type 6AF4A except for heater and heater-cathode ratings.

	2AF4B/ 2DZ4	3AF4A/ 3DZ4	6AF4 6AF4A	
Heater Voltage (ac/dc)	2.35	3.15	6.3	volts
Heater Current	0.6	0.45	0.225	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±180 max	±50 max	±50 max	volts
Average value	100 max	25 max	25 max	volts

Direct Interelectrode Capacitances:•		
Grid to Plate	1.9	pF
Grid to Cathode and Heater	2.2	pF
Plate to Cathode and Heater	1.4	pF
Heater to Cathode (External Shield connected to plate)	2.2	pF

• With external shield connected to cathode, except as noted.

Class A_1 Amplifier

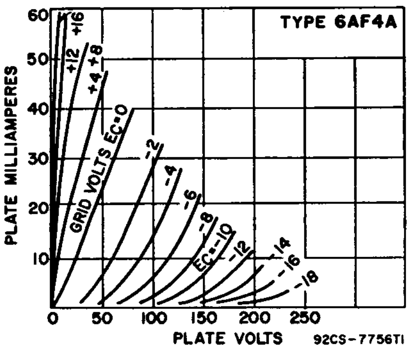
CHARACTERISTICS

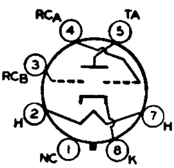
Plate Supply Voltage	80	volts
Cathode-Bias Resistor	150	ohms
Amplification Factor	13.5	
Plate Resistance (Approx.)	2100	ohms
Transconductance	6500	µmhos
Plate Current	17.5	mA

UHF Oscillator

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	150	volts
Grid Voltage, Negative-bias value	50	volts
Grid Current	2	mA
Plate Dissipation	2.5	watts
Average Cathode Current	24	mA
TYPICAL OPERATION AS OSCILLATOR AT 1000 MHz		
Plate Supply Voltage	100	volts
Plate Resistor	220	ohms
Grid Resistor	10000	ohms
Plate Current	17	mA
Grid Current (Approx.)	750	μA
MAXIMUM CIRCUIT VALUES		
Grid-Circuit Resistance:		
For fixed-bias operation	Not recommended	
For cathode-bias operation	0.5	megohm

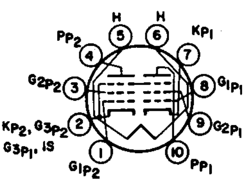




7AG

ELECTRON-RAY TUBE 6AF6G

Glass octal type used to indicate visually, by means of two shadows on the fluorescent target, the effects of changes in the controlling voltages. It is a twin-indicator type and is used as a convenient means of indicating accurate radio-receiver turning. This type may be supplied with pin No. 1 omitted. Tube requires octal socket. Heater: volts (ac/dc), 6.3; amperes, 0.15. **Maximum ratings in indicator service:** fluorescent-target volts, 250 max, 125 min; ray-control-electrode supply volts, 250 max; peak heater-cathode volts, 90 max. **Typical operation:** fluorescent-target volts, 250; fluorescent-target mA, 3.75; ray-contact-electrode volts (approx. for 0° shadow angle), 155; ray-control-electrode volts (approx. for 100° shadow angle), 0.



10L

DUAL PENTODE 6AF9

Miniature type used in television receiver applications. Unit No.1 is used as a video output pentode, and unit No.2 as a sound if amplifier, agc amplifier, or sync separator. **Outlines section, 6L,** except has 10-pin base; requires miniature 10-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.85	ampere
Peak Heater-Cathode Voltage	±200 max	volts

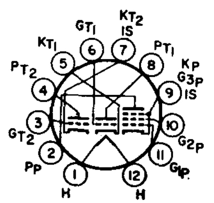
Direct Interelectrode Capacitances:	Unit No.1	Unit No.2	
Plate to All Other Electrodes (except grid No.1)	7	11	pF
Grid No.1 to All Other Electrodes (except plate)	12	10	pF
Plate to Grid No.1	0.105	0.140	pF
Grid No.1 to Heater	—	0.140	pF
Plate of Unit No.1 to Plate of Unit No. 2	0.150 max		pF
Grid No.1 of Unit No.1 to Grid No.1 of Unit No. 2	0.010 max		pF
Plate of Unit No.1 to Grid No.1 of Unit No.2	0.100 max		pF
Plate of Unit No.2 to Grid No.1 of Unit No.1	0.005 max		pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1	Unit No.2	
Plate Supply Voltage	550	550	volts
Plate Voltage	250	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	550	550	volts
Grid-No.2 Voltage	250	250	volts
Cathode Current	60	15	mA
Plate Dissipation	5.1	1.5	watts
Grid-No.2 Input	2.5	0.5	watts
CHARACTERISTICS			
Plate Voltage	170	150	volts
Grid-No.2 Voltage	170	150	volts
Grid-No.1 (Control-Grid) Voltage	—2.6	—2.1	volts
Mu Factor, Grid No.1 to Grid No.2	38	38	
Internal Resistance	0.032	0.16	megohm
Transconductance	22000	8500	μmhos
Plate Current	30	10	mA
Grid-No.2 Current	7.2	3	mA
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance	1	1	megohm

6AF11
15AF11

DUAL TRIODE— SHARP-CUTOFF PENTODE



12DP

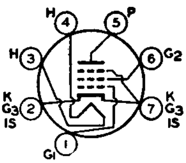
Duodecar type used in television receiver applications. The high-mu triode unit is used for agc keyer service, the medium-mu triode unit for sync separator service, and the pentode unit for video amplifier service. Outlines section, 8C; requires duodecar 12-contact socket. Type 15AF11 is identical with type 6AF11 except for heater ratings.

	6AF11	15AF11	
Heater Voltage (ac/dc)	6.3	14.7	volts
Heater Current	1.05	0.45	amperes
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit No.1	Triode Unit No.2	Pentode Unit	
Plate Voltage	330	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	—	330	volts
Grid-No.2 Voltage	—	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	0	volts
Plate Dissipation	1.1	2	5	watts
Grid-No.2 Input:				
For grid-No.2 voltages up to 165 volts	—	—	1.25	watts
For grid-No.2 voltages between 165 and 330 volts	—	—	See curve page 98	
CHARACTERISTICS				
Plate Supply Voltage	200	200	250	volts
Grid-No.2 Supply Voltage	—	—	150	volts
Grid-No.1 Voltage	—2	—	—	volts
Cathode-Bias Resistor	—	220	100	ohms
Amplification Factor	68	41	—	
Plate Resistance (Approx.)	12400	9400	68000	ohms
Transconductance	5500	4400	11000	μmhos
Plate Current	7	9.2	24	mA

	Triode Unit No.1	Triode Unit No.2	Pentode Unit	
Grid-No.2 Current	—	—	4.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—6.5	—10	volts
MAXIMUM CIRCUIT VALUES				
Grid-No.1-Circuit Resistance:				
For fixed-bias operation	0.5	0.5	0.25	megohm
For cathode-bias operation	1	1	1	megohm



7BD

SHARP-CUTOFF PENTODE 6AG5

Miniature type used in compact radio equipment as an rf or if amplifier up to 400 MHz. **Outlines section, 5C;** requires miniature 7-contact socket. For typical operation as a resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section.

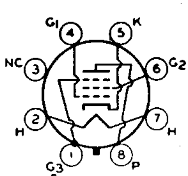
Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.3	ampere
Direct Interelectrode Capacitances:		
Pentode Unit:		
Grid No.1 to Plate	0.030 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	6.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	1.8	pF
Triode Unit:		
Grid No.1 to Plate and Grid No.2	2.5	pF
Grid No.1 to Cathode, Heater, Grid No.3, and Internal Shield	3.6	pF
Grid No.2 to Cathode, Heater, Grid No.3, and Internal Shield	3	pF
Plate to Cathode, Heater, Grid No.3, and Internal Shield	3	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)	Triode Connection*	Pentode Connection	
Plate Voltage	300	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	300	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	2	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 150 volts	—	0.5	watt
For grid No.2 voltages between 150 and 300 volts	—	See curve page 98	

CHARACTERISTICS						
Plate Supply Voltage	180	250	100	125	250	volts
Grid-No.2 Supply Voltage	—	—	100	125	150	volts
Cathode-Bias Resistor	330	820	180	100	180	ohms
Amplification Factor	45	42	—	—	—	
Plate Resistance (Approx.)	0.008	0.01	0.6	0.5	0.8	megohm
Transconductance	5700	3800	4500	5100	5000	μmhos
Plate Current	7	5.5	4.5	7.2	6.5	mA
Grid-No.2 Current	—	—	1.4	2.1	2	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—	—	—5	—6	—8	volts

* Grid No.2 connected to plate.



8Y

POWER PENTODE 6AG7

Metal type used in output stage of video amplifier of color and black-and-white television receivers. **Outlines section, 2B;** requires octal socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.65	ampere

Peak Heater-Cathode Voltage	±90 max	volts
Direct Interelectrode Capacitances:•		
Grid No.1 to Plate	0.06 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, Shell, and Internal Shield	13	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, Shell, and Internal Shield	7.5	pF

• Pins 1 and 3 connected to Pin No.5.

Class A_1 Amplifier

MAXIMUM RATINGS *(Design-Center Values)*

Plate Voltage	300	volts
Grid-No.2 Voltage	300	volts
Grid-No.1 Voltage, Positive-bias value	0	volts
Plate Dissipation	9	watts
Grid-No.2 Input	1.5	watts

CHARACTERISTICS

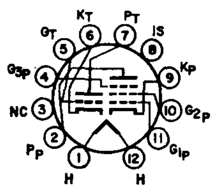
Plate Voltage	300	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket	
Grid-No.2 (Screen-Grid) Voltage	150	volts
Grid-No.1 (Control-Grid) Voltage	—3	volts
Peak AF Grid-No.1 Voltage	3	volts
Zero-Signal Grid-No.2 Current	30	mA
Maximum-Signal Grid-No.2 Current	30.5	mA
Zero-Signal Grid-No.2 Current	7	mA
Maximum-Signal Grid-No.2 Current	9	mA
Plate Resistance	0.13	megohm
Transconductance	11000	μmhos
Load Resistance	10000	ohms
Total Harmonic Distortion	7	per cent
Maximum-Signal Power Output	3	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

6AG9

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE



12HE

Duodecar type with frame grid pentode unit used in color and black-and-white television receiver applications. The pentode unit is used as a video amplifier; the triode unit is used as an agc amplifier. **Outlines section,** 8C; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.82; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

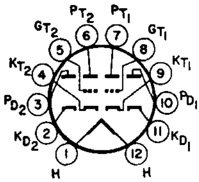
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Voltage	—	200	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	1.1	10	watts
Grid-No.2 Input	—	1.5	watts

CHARACTERISTICS

	Triode Unit	Pentode Unit		
Plate Voltage	150	55	250	volts
Grid-No.2 Voltage	—	125	150	volts
Grid-No.1 Voltage	—	0	—	volts
Cathode-Bias Resistor	350	—	56	ohms
Amplification Factor	39	—	—	
Plate Resistance (Approx.)	8500		40000	ohms
Transconductance	4600	—	30000	μmhos
Plate Current	6.2		28	mA
Grid-No.2 Current	—	—	5.6	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—7	— —	—	volts
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	56 21	—5.4	volts

MAXIMUM CIRCUIT VALUES

	Triode Unit	Pentode Unit	
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.1	megohm
For cathode-bias operation	1	0.25	megohm



12DA

TWIN DIODE—TWIN TRIODE 6AG11

30AG11

Duodecar type containing two diodes and two high-mu triodes, used primarily in FM stereo multiplex service. **Outlines section,** 8A; requires duodecar 12-contact socket. Type 30AG11 is identical with type 6AG11 except for heater ratings.

	6AG11	30AG11	
Heater Voltage	6.3	30	volts
Heater Current	0.75	0.15	ampere
Heater Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average	100 max	100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Plate Dissipation	2	watts

CHARACTERISTICS

Plate Voltage	125	volts
Grid Voltage	—1	volt
Amplification Factor	66	
Plate Resistance (Approx.)	8500	ohms
Transconductance	7800	µmhos
Plate Current	7.5	mA
Grid Voltage (Approx.) for plate current of 30 µA	—5	volts

Diode Units (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

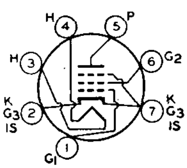
Plate Current	5	mA

CHARACTERISTICS

Tube Voltage Drop for plate current of 18 mA	5	volts

Refer to chart at end of section. **6AH4GT**

Refer to chart at end of section. **6AH6**



7BD

SHARP-CUTOFF PENTODE 6AK5 6AK5/EF95

Miniature types used as rf or if amplifiers especially in high-frequency wide-band applications at frequencies up to 400 MHz. **Outlines section,** 5B; require miniature 7-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.175	ampere
Peak Heater-Cathode Voltage	±90 max	volts
Direct Interelectrode Capacitances (Approx.):▪		
Grid No.1 to Plate	0.02 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.8	pF

▪ With external shield connected to pin 2 or 7.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	180	volts
Grid-No.2 (Screen-Grid) Voltage	See curve page 98	
Grid-No.2 Supply Voltage	180	volts
Grid-No.1 Voltage, Positive-bias value	0	volts
Plate Dissipation	1.7	watts

Grid-No.2 Input:		
For grid-No.2 voltages up to 90 volts	0.5	watt
For grid-No.2 voltages between 90 and 180 volts	See curve page 98	
Cathode Current	18	mA

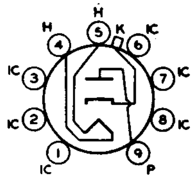
CHARACTERISTICS

Plate Supply Voltage	120	180	volts
Grid-No.2 Supply Voltage	120	120	volts
Cathode-Bias Resistor	180	180	ohms
Plate Resistance (Approx.)	0.3	0.5	megohm
Transconductance	5000	5100	μmhos
Plate Current	7.5	7.7	mA
Grid-No.2 Current	2.5	2.4	mA
Grid-No.1 Voltage for plate current of 10 μA	—8.5	—8.5	volts

6AL3 Refer to chart at end of section.

6AL3/EY88

HALF-WAVE VACUUM RECTIFIER



9CB

Miniature type used as damper tube in horizontal-deflection circuits of black-and-white television receivers. **Outlines section,** 7D; requires miniature 9-contact socket. Socket terminals 1, 2, 3, 6, 7, and 8 should not be used as tie points. It is especially important that this tube, like other power-handling tubes, be adequately ventilated. **Heater:** volts (ac/dc), 6.3; amperes, 1.55.

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)

Peak Inverse Plate Voltage# (Absolute maximum)	7500°	volts
Peak Plate Current	550	mA
Average Plate Current	220	mA
Plate Dissipation	5	watts
Peak Heater-Cathode Voltage	6600	volts

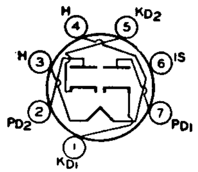
° Under no circumstances should this absolute value be exceeded.

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6AL5

3AL5, 12AL5

TWIN DIODE



6BT

Miniature, high-perveance type used as detector in FM and television circuits, especially as a ratio detector in ac-operated FM receivers. Each diode section can be used independently of the other, or the two sections can be combined in parallel or full-wave arrangement. Resonant frequency of each unit is approximately 700 MHz. **Outlines section,** 5B; requires miniature 7-contact socket. Types 3AL5 and 12AL5 are identical with type 6AL5 except for heater ratings.

	3AL5	6AL5	12AL5	
Heater Voltage (ac/dc)	3.15	6.3	12.6	volts
Heater Current	0.6	0.3	0.15	ampere
Heater Warm-up Time (Average)	11	—	—	seconds
Peak Heater-Cathode Voltage	±330 max	±330 max	±330 max	volts

Direct Interelectrode Capacitances:		
Plate No.1 to Cathode No.1, Heater, and Internal Shield	2.5	pF
Plate No.2 to Cathode No.2, Heater, and Internal Shield	2.5	pF
Cathode No.1 to Plate No.1, Heater, and Internal Shield	3.4	pF
Cathode No.2 to Plate No.2, Heater, and Internal Shield	3.4	pF
Plate No.1 to Plate No.2	0.068 max	pF

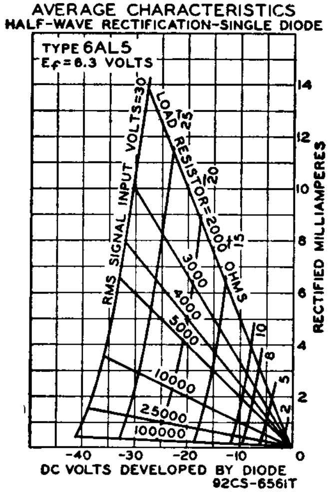
Half-Wave Rectifier

MAXIMUM RATINGS (Design-Center Values)

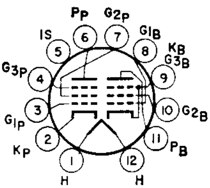
Peak Inverse Plate Voltage	330	volts
Peak Plate Current (Per Plate)	54	mA
Average Output Current (Per Plate)	9	mA

TYPICAL OPERATION

AC Plate Voltage per Plate (rms)	117	volts
Min. Total Effective Plate-Supply Impedance per Plate	300	ohms
Average Output Current per Plate	9	mA



Refer to chart at end of section. **6AL7GT**



12BU

BEAM POWER TUBE— SHARP-CUTOFF PENTODE

6AL11

10AL11, 12AL11

Duodecar type used as FM detector and audio-frequency output amplifier in television receivers. **Outlines section, 8C; requires duodecar 12-contact socket. Types** 10AL11 and 12AL11 are identical with type 6AL11 except for heater ratings.

	6AL11	10AL11	12AL11	
Heater Voltage (ac/dc)	6.3	9.8	12.6	volts
Heater Current	0.9	0.6	0.45	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitance:		
Beam Power Unit:		
Grid No.1 to Plate	0.26	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	11	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	12	pF
Pentode Unit:		
Grid No.1 to Plate	0.034	pF
Grid No.3 to Plate	3.2	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	6.5	pF
Grid No.3 to Cathode, Heater, Grid No.1, Grid No.2, Plate, and Internal Shield	7.5	pF
Grid No.1 to Grid No.3	0.24	pF
Pentode Plate to Beam Power Plate	0.12	pF

Beam Power Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	275	volts
Grid-No.2 (Screen-Grid) Voltage	275	volts
Plate Dissipation	10	watts
Grid-No.2 Input	2	watts

TYPICAL OPERATION

Plate Voltage	250	volts
Grid-No.2 Voltage	250	volts
Grid-No.1 (Control-Grid) Voltage	—8	volts
Peak AF Grid-No.1 Voltage	8	volts
Zero-Signal Plate Current	35	mA
Maximum-Signal Plate Current	39	mA
Zero-Signal Grid-No.2 Current	2.5	mA
Maximum-Signal Grid-No.2 Current	7	mA
Plate Resistance (Approx.)	0.1	megohm
Transconductance	6500	μmhos
Load Resistance	5000	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	4.2	watts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	0.5	megohm

Pentode Unit as Class A_1 Amplifier

CHARACTERISTICS

Plate Supply Voltage	150	volts
Grid-No.3 (Suppressor-Grid) Voltage	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	100	volts
Cathode-Bias Resistor	560	ohms
Plate Resistance (Approx.)	0.15	megohm
Transconductance, Grid No.1 to Plate	1000	μmhos
Transconductance, Grid No.3 to Plate	400	μmhos
Plate Current	1.3	mA
Grid-No.2 Current	2.1	mA
Grid-No.1 Voltage (Approx.) for plate current of 30 μA	—4.5	volts
Grid-No.3 Voltage (Approx.) for plate current of 50 μA	—4.5	volts

Pentode Unit as FM Detector

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.3 Voltage	28	volts
Grid-No.2 Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	1.7	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

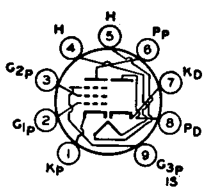
6AM4 Refer to chart at end of section.

6AM8 Refer to chart at end of section.

6AM8A

5AM8

DIODE— SHARP-CUTOFF PENTODE

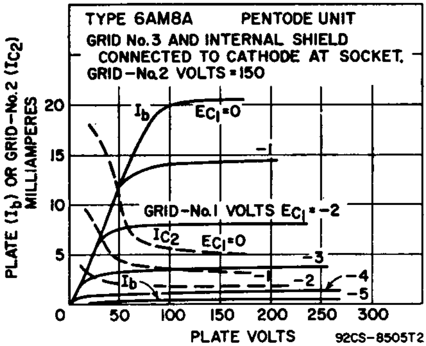


9CY

Miniature type used in television receiver applications. The pentode unit is used as an if amplifier, video amplifier, or agc amplifier. The high-perveance diode is used as an audio detector, video detector, or dc restorer. Outlines section, 6B; requires miniature 9-contact socket. Type 5AM8 is identical with type 6AM8A except for heater ratings.

	5AM8	6AM8A	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	100 max	100 max	volts

Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:			
Diode Unit:			
Plate to Cathode and Heater		1.8	pF
Cathode to Plate and Heater		3	pF
Pentode Unit:			
Grid No.1 to Plate		0.015	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3 and Internal Shield		6.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield		2.6	pF
Pentode Grid No.1 to Diode Plate		0.006	pF
Pentode Plate to Diode Cathode		0.15	pF
Pentode Plate to Diode Plate		0.1	pF



Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	3.2	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	125	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.3	megohm
Transconductance	7800	μmhos
Plate Current	12.5	mA
Grid-No.2 Current	3.2	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—6	volts
Grid-No.1 Voltage (Approx.) for plate current of 2 mA	—3	volts

MAXIMUM CIRCUIT VALUES

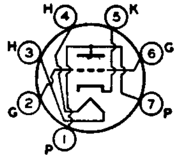
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

Diode Unit

MAXIMUM RATINGS (Design-Maximum Values)

Average Plate Current	5	mA

6AN4 HIGH-MU TRIODE



7DK

Miniature type used as mixer or rf amplifier in cathode-drive circuits of uhf television tuners covering the frequency range of 470 to 890 MHz. Outlines section, 5B; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.225	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid to Plate	1.7°	pF
Grid to Cathode and Heater	3.3°	pF
Plate to Cathode and Heater	1.8°	pF
Heater to Cathode	2.9▲	pF
Grid to Cathode	2.6▲	pF
Plate to Cathode	0.18▲	pF
Cathode to Grid and Heater	5.7*	pF
Plate to Grid and Heater	3.4*	pF

° With external shield connected to cathode.
▲ With external shield connected to ground.
* With external shield connected to grid.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Cathode Current	30	mA
Plate Dissipation	4	watts

CHARACTERISTICS

Plate-Supply Voltage	200	volts
Cathode-Bias Resistor	100	ohms
Amplification Factor	70	
Transconductance	10000	μmhos
Plate Current	13	mA
Grid Voltage (Approx.) for plate current of 20 μA	—7	volts

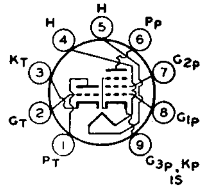
MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

6AN8 Refer to chart at end of section.

6AN8A MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

5AN8



Miniature type used in color television receiver applications. The pentode unit is used as an intermediate-frequency amplifier, a video amplifier, an agc amplifier, or a reactance tube. The triode unit is used in low-frequency oscillator, sync-separator, sync-clipper, and phase-splitter circuits. Outlines section, 6B; requires miniature 9-contact socket. Type 5AN8 is identical with 6AN8A except for heater ratings.

	5AN8	6AN8A	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

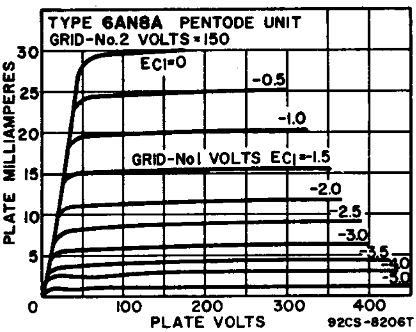
Direct Interelectrode Capacitances:

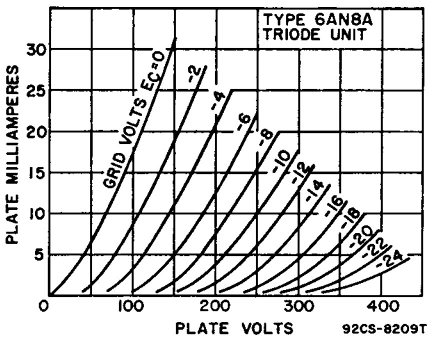
Triode Unit:		
Grid to Plate	1.5	pF
Grid to Cathode and Heater	2	pF
Plate to Cathode and Heater	0.26	pF
Pentode Unit:		
Grid No.1 to Plate	0.04 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	7	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.4	pF
Triode Grid to Pentode Plate	0.02	pF
Pentode Grid No.1 to Triode Plate	0.02	pF
Pentode Plate to Triode Plate	0.15	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 Supply Voltage	—	330	volts
Grid-No.2 (Screen-Grid) Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.8	2.3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	150	125	volts
Grid-No.2 Supply Voltage	—	125	volts
Grid-No.1 Voltage	—3	—	volts
Cathode-Bias Resistor	—	56	ohms
Amplification Factor	21	—	
Plate Resistance (Approx.)	4700	170000	ohms
Transconductance	4500	7800	μmhos
Plate Current	15	12	mA
Grid-No.2 Current	—	3.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—17	—6	volts
Grid-No.1 Voltage (Approx.) for plate current of 1.6 mA	—	—3	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:*			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

* If either unit is operating at maximum rated conditions, grid-No.1-circuit resistance for both units should not exceed the stated values.



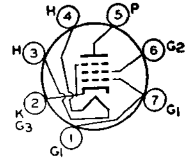


Refer to chart at end of section.

6AQ5

6AQ5A BEAM POWER TUBE

5AQ5, 12AQ5



7BZ

Miniature type used as output amplifier primarily in automobile receivers and in ac-operated receivers and, triode-connected, as a vertical-deflection amplifier in television receivers. **Outlines section, 5D;** requires miniature 7-*contact* socket. Within its maximum ratings, the performance of this type is equivalent to that of larger types 6V6 and 6V6GTA. Types 5AQ5 and 12AQ5 are identical with type 6AQ5A except for heater ratings.

	5AQ5	6AQ5A	12AQ5	
Heater Voltage (ac/dc)	4.7	6.3	12.6	volts
Heater Current	0.6	0.45	0.225	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.4	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	8	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	8.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	275	volts
Grid-No.2 (Screen-Grid) Voltage	275	volts
Plate Dissipation	12	watts
Grid-No.2 Input	2	watts
Bulb *Temperature* (At hottest point)	250	°C

CHARACTERISTICS (Triode Connection)

Plate Voltage	250	volts
Grid-No.1 Voltage	—12.5	volts
Amplification Factor	9.5	
Plate Resistance (Approx.)	1970	ohms
Transconductance	4800	µmhos
Plate Current	49.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 0.5 mA	—37	volts

TYPICAL OPERATION

Same as for type 6V6GTA within the limitations of the maximum ratings.

MAXIMUM CIRCUIT VALUES

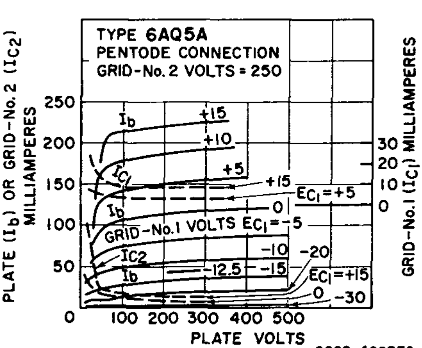
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

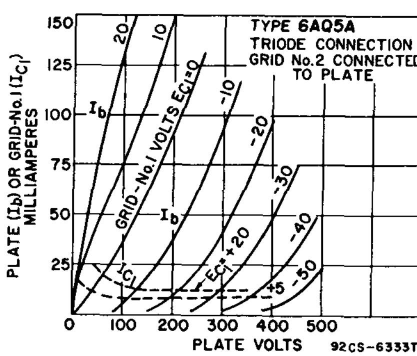
Vertical Deflection Amplifier (Triode Connection)°

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Voltage	275	volts
Peak Positive-Pulse Plate Voltage#	1100	volts
Peak Negative-Pulse Grid-No.1 (Control-Grid) Voltage	275	volts





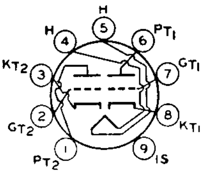
Peak Cathode Current	115	mA
Average Cathode Current	40	mA
Plate Dissipation	10	watts
Bulb Temperature (At hottest point)	250	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance, for cathode-bias operation	2.2	megohms

° Grid No.2 connected to plate.
Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

Refer to chart at end of section. **6AQ6**

Refer to chart at end of section. **6AQ7GT**

Refer to chart at end of section. **6AQ8**



9AJ

HIGH-MU TWIN TRIODE 6AQ8/ECC85

Miniature types used as rf amplifier and self-oscillating mixer in FM/AM radio receivers. **Outlines section,** 6B; requires 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.435	ampere
Peak Heater-Cathode Voltage	±90 max	volts

Direct Interelectrode Capacitances:	Unit No.1	Unit No.2	
Grid to Plate	1.5	1.5	pF
Cathode to Plate	0.18	0.18	pF
Grid to Cathode, Heater, and Internal Shield	3	3	pF
Plate to Cathode, Heater, and Internal Shield	1.2	1.2	pF
Plate to Grid of Other Unit	0.008 max	0.008 max	pF
Plate to Cathode of Other Unit	0.008 max	0.008 max	pF
Grid to Cathode of Other Unit	0.003 max	0.003 max	pF
Plate of Unit No.1 to Plate of Unit No.2		0.04 max	pF
Grid of Unit No.1 to Grid of Unit No.2		0.003 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values, Each Unit)		
Plate Supply Voltage	550	volts
Plate Voltage	300	volts
Grid Voltage, Negative-bias value	100	volts
Cathode Current	15	mA
Plate Dissipation:		
For either plate	2.5	watts
For both plates with both units operating	4.5	watts
CHARACTERISTICS		
Plate Voltage	250	volts
Grid Voltage, Negative-bias value	2.3	volts
Plate Current	10	mA
Transconductance	5900	μmhos
Amplification Factor	57	

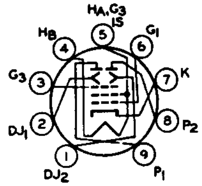
TYPICAL OPERATION (Each Unit)	RF Amplifier	Converter	
Plate Supply Voltage	250	250	volts
Plate Voltage	230	—	volts
Plate Resistor	1800	12000	ohms
Grid Resistor	—	1	megohm
Grid Voltage	—2	—	volts
RMS Oscillator Voltage	—	3	volts
Cathode-Bias Resistor	200	—	ohms
Plate Resistance (Approx.)	9700	22000	ohms
Transconductance	6000	—	μmhos
Conversion Transconductance	—	2300	μmhos
Input Resistance at frequency of 100 MHz	6000	15000	ohms
Plate Current	10	5.2	mA
Equivalent Noise Resistance	500	—	ohms

MAXIMUM CIRCUIT VALUES (Each Unit)		
Grid-Circuit Resistance	1	megohm
Resistance between Cathode and Heater	20000	ohms

6AR5 Refer to chart at end of section.

6AR8

BEAM-DEFLECTION TUBE



9DP

Miniature type used in color-demodulator and burst-gate circuits in color television receivers. This type has two plates and two deflecting electrodes; the control grid varies beam deflection. **Outlines section,** 6E; requires miniature nine-contact socket. Pin 5 should be conected to cathode at socket. The 6AR8 should be so located in the equipment that it is not subjected to stray magnetic fields. **Heater:** volts (ac/dc), 6.3; amperes, 0.3.

Color TV Demodulator

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage (Each Plate)	300	volts
Peak Deflecting-Electrode Voltage (Each Electrode):		
Negative value	150	volts
Positive value	150	volts
Grid-No.3 (Accelerating-Grid) Voltage	300	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Cathode Current	30	mA
Plate Dissipation (Each Plate)	2	watts
Grid-No.3 Input	1	watt
MAXIMUM CIRCUIT VALUES		
Grid-No.1 Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.25	megohm

Class A_1 Amplifier

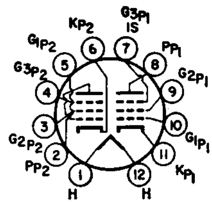
With both plates connected together and with both deflecting electrodes connected to cathode at socket

CHARACTERISTICS		
Plate-No.1 Supply Voltage	250	volts
Plate-No.2 Supply Voltage	250	volts
Grid-No.3 Voltage	250	volts
Cathode-Bias Resistor	300	ohms
Transconductance	4000	µmhos
Total Plate Current	10	mA
Grid-No.3 Current	0.4	mA
Grid-No.1 Voltage (Approx.) for total plate current of 10 µA	—14	volts

6AR11

8AR11, 11AR11

SEMIREMOTE-CUTOFF TWIN PENTODE



12DM

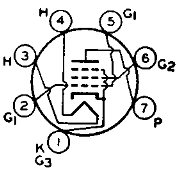
Duodecar type used as if-amplifier tube in television receivers. **Outlines section,** 8A; requires duodecar 12-contact-socket. Types 8AR11 and 11AR11 are identical with type 6AR11 except for heater ratings.

	6AR11	8AR11	11AR11	
Heater Voltage (ac/dc)	6.3	8.4	11.2	volts
Heater Current	0.8	0.6	0.45	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

	Unit No.1	Unit No.2	
Direct Interelectrode Capacitances:			
Grid No.1 to Plate	0.026	0.026	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No. 3, and Internal Shield	10	10	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.8	3	pF
Grid No.1 to Plate of Other Unit	0.002	0.002	pF
Plate of Unit No.1 to Plate of Unit No.2		0.02	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values, Each Unit)		
Plate Voltage	330	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	3.1	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.65	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	
CHARACTERISTICS (Each Unit)		
Plate Supply Voltage	125	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.2	megohm
Transconductance	10500	μmhos
Plate Current	11	mA
Grid-No.2 Current	3.5	mA
Grid-No.1 Voltage (Approx.) for transconductance of 50 μmhos	—15	volts



7CV

BEAM POWER TUBE 6AS5

Miniature type used as output amplifier primarily in automobile and in ac-operated receivers. **Outlines section, 5D;** requires miniature 7-contact socket. For curves of average plate characteristics, refer to type 35C5.

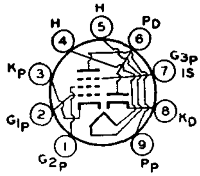
Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.8	ampere
Peak Heater-Cathode Voltage	±100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.6	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	12	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	9	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)		
Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	117	volts
Plate Dissipation	5.5	watts
Grid-No.2 Input	1.0	watt
Bulb Temperature (At hottest point)	250	°C
TYPICAL OPERATION		
Plate Voltage	150	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 (Control-Grid) Voltage	—8.5	volts
Peak AF Grid-No.1 Voltage	8.5	volts
Zero-Signal Plate Current	35	mA
Maximum-Signal Plate Current	36	mA
Zero-Signal Grid-No.2 Current (Approx.)	2	mA
Maximum-Signal Grid-No.2 Current (Approx.)	6.5	mA
Transconductance	5600	μmhos
Load Resistance	4500	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	2.2	watts
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

6AS8
5AS8

DIODE— SHARP-CUTOFF PENTODE



9DS

Miniature type used in television and radio receiver applications. The pentode unit is used as an if amplifier, video amplifier, or agc amplifier. The high-perveance diode is used as an audio detector, video detector, or dc restorer. Outlines section, 6B; requires miniature 9-contact socket. For curve of average plate characteristics of pentode unit, see type 6AN8A. Type 5AS8 is identical with type 6AS8 except for heater ratings.

	5AS8	6AS8	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Diode Unit:		
Plate to Cathoe, Heater, Pentode Grid No.3, and Internal Shield	3	pF
Pentode Unit:		
Grid No.1 to Plate	0.03	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	7	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.4	pF
Pentode Grid No.1 to Diode Plate	0.005 max	pF
Pentode Plate to Diode Cathode	0.15 max	pF
Pentode Plate to Diode Plate	0.10 max	pF

Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 Supply Voltage	300	volts
Grid-No.2 (Screen-Grid) Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	2.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	0.5	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	200	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	150	volts
Cathode-Bias Resistor	180	ohms
Plate Resistance (Approx.)	300000	ohms
Transconductance	6200	μmhos
Plate Current	9.5	mA
Grid-No.2 Current	3	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—8	volts

MAXIMUM CIRCUIT VALUES

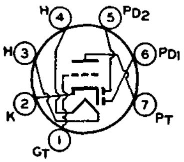
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

Diode Unit

MAXIMUM RATINGS (Design-Center Values)

Peak Inverse Plate Voltage	330	volts
Peak Plate Current	50	mA
Average Plate Current	5	mA

6AS11 Refer to chart at end of section.



7BT

TWIN DIODE— HIGH-MU TRIODE

6AT6
12AT6

Miniature type used as a combined detector, amplifier, and avc tube in automobile and ac-operated radio receivers. **Outlines section, 5C**; requires miniature 7-contact socket. For typical operation as resistance-coupled amplifier refer to **Resistance-Coupled Amplifier** section. Type 12AT6 is identical with type 6AT6 except for heater ratings.

	6AT6	12AT6	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	0.3	0.15	ampere
Peak Heater-Cathode Voltage	±90 max	±90 max	volts
Direct Interelectrode Capacitances:			
Triode Grid to Triode Plate		2	pF
Triode Grid to Cathode and Heater		2.2	pF
Triode Plate to Cathode and Heater		0.8	pF
Plate of Diode Unit No.2 to Triode Grid		0.04 max	pF

Triode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)			
Plate Voltage		300	volts
Plate Dissipation		0.5	watts
Grid Voltage, Positive-bias value		0	volts
CHARACTERISTICS			
Plate Voltage	100	250	volts
Grid Voltage	—1	—3	volts
Amplification Factor	70	70	
Plate Resistance	54000	58000	ohms
Transconductance	1300	1200	μmhos
Plate Current	0.8	1	mA

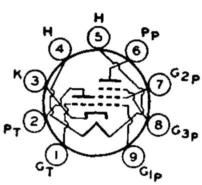
Diode Units

MAXIMUM RATING (Design-Center Value)		
Plate Current (Each Unit)	1	mA

The two diode plates are placed around a cathode whose sleeve is common to the triode unit. Each diode plate has its own base pin. For diode operation curves, refer to type 6AV6.

6AT8

Refer to chart at end of section.



9DW

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

6AT8A
5AT8

Miniature types used as combined oscillator and mixer tubes in television receivers utilizing an intermediate frequency in the order of 40 MHz. **Outlines section, 6B**; requires miniature 9-contact socket. Except for interlectrode capacitances and basing arrangement, this type is identical with miniature type 6X8. The basing arrangement is particularly suitable for connection to the coils of certain designs of turret tuners. Type 5AT8 is identical with type 6AT8A except for heater ratings.

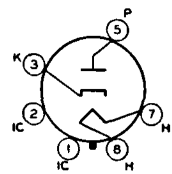
	5AT8	6AT8A	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Direct Interelectrode Capacitances:			
Triode Unit:	Unshielded	Shielded•	
Grid to Plate	1.5	1.5	pF
Grid to Cathode and Heater	2	2.4	pF
Plate to Cathode and Heater	0.5	1	pF
Pentode Unit:			
Grid No.1 to Plate	0.06 max	0.03 max	pF
Grid No.1 to Cathode, Heater, Grid No.2 and Grid No.3	4.6	4.8	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	0.9	1.6	pF

Pentode Grid No.1 to Triode Plate	0.05 max	0.04 max	pF
Pentode Plate to Triode Plate	0.05 max	0.008 max	pF
Heater to Cathode	6	6†	pF

• With external shield connected to cathode except as noted.
† With external shield connected to plate.

6AU4GT Refer to chart at end of section.

6AU4GTA HALF-WAVE VACUUM RECTIFIER



4CG

Glass octal type used as damper tube in horizontal-deflection circuits of color and wide-angle picture-tube television receivers. Outlines section, 13G; requires octal socket. Type may be supplied with pin No. 1 omitted. Socket terminals 1, 2, 4, and 6 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	1.8	amperes
Direct Interelectrode Capacitances (Approx.):		
Plate to Heater and Cathode	8.5	pF
Cathode to Heater and Plate	11.5	pF
Heater to Cathode	4	pF

Damper Service

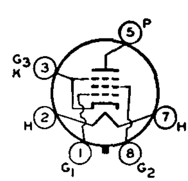
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		4500	volts
Peak Plate Current		1300	mA
Average Plate Current		210	mA
Plate Dissipation		6.5	watts
Heater-Cathode-Voltage:			
Peak value	+300	—4500	volts
Average value	+100	—900	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6AU5GT BEAM POWER TUBE



6CK

Glass octal type used as horizontal-deflection amplifier in low-cost, high-efficiency deflection circuits of television receivers. Outlines section, 13D; requires octal socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	1.25	amperes
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.5	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	11.3	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	7	pF

Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection	Triode† Connection	
Plate Voltage	115	110	volts
Grid-No.2 (Screen-Grid) Voltage	175	100	volts
Grid-No.1 (Control-Grid) Voltage	—20	—4.5	volts
Plate Resistance	6000	—	ohms
Transconductance	5600	—	μmhos
Plate Current	60	—	mA
Grid No.2 Current	6.8	—	mA

† Grid No.2 connected to plate.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)

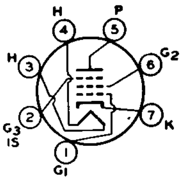
DC Plate Voltage	550	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	5500°	volts
Peak Negative-Pulse Plate Voltage	1250	volts
DC Grid-No.2 (Screen-Grid) Voltage▪	200	volts
Peak Negative-Pulse Grid-No.1 (Control-Grid) Voltage	300	volts
Peak Cathode Current	400	mA
Average Cathode Current	110	mA
Grid-No.2 Input	2.5	watts
Plate Dissipation††	10	watts
Bulb Temperature (At hottest point)	210	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	0.47	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
° Under no circumstances should this absolute value be exceeded.

▪ Obtained through a series dropping resistor of sufficient magnitude to limit the grid-No.2 input to the rated maximum value.

†† A bias resistor or other means is required to protect the tube in absence of excitation.

Refer to chart at end of section. **6AU6**



7BK

6AU6A

SHARP-CUTOFF PENTODE

3AU6, 4AU6, 12AU6

Miniature type used in compact radio equipment as rf amplifier especially in high-frequency, wide-band applications; also used as limiter tube in FM equipment. Outlines section, 5C; requires miniature 7-contact socket. For a discussion of limiters, refer to Electron Tube Applications section. For typical operation as resistance-coupled amplifier, refer to Resistance-Coupled Amplifier section. Types 3AU6, 4AU6, and 12AU6 are identical with type 6AU6A except for heater ratings.

	3AU6	4AU6	6AU6	12AU6	
Heater Voltage (ac/dc)	3.15	4.2	6.3	12.6	volts
Heater Current	0.6	0.45	0.3	0.15	ampere
Heater Warm-up Time (Average)	11	11	11	—	seconds
Heater-Cathode Voltage:					
Peak value	±200 max	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Pentode Connection:		
Grid No.1 to Plate	0.0035 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5	pF
Triode Connection:†		
Grid No.1 to Plate, Grid No.2, Grid No.3, and Internal Shield	2.6	pF
Grid No.1 to Cathode and Heater	3.2	pF
Plate, Grid No.2, Grid No.3, and Internal Shield to Cathode and Heater	1.2▪	pF

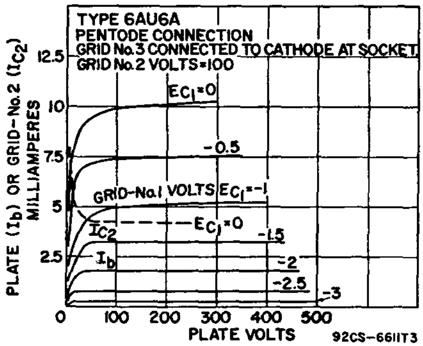
† Grid No.2, grid No.3, and internal shield connected to plate.
▪ Value is 8.5 pF with external shield connected to cathode.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode† Connection	Pentode Connection	
Plate Voltage	275	330	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	—	0	volts
Grid-No.2 (Screen-Grid) Voltage		See curve page 98	
Grid-No.2 Supply Voltage	—	330	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	vols
Plate Dissipation	3.5	3.5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.75	watt
For grid-No.2 voltages between 165 and 330 volts		See curve page 98	

CHARACTERISTICS	Triode† Connection	Pentode Connection			
Plate Supply Voltage	250	100	250	150	volts
Grid No.3	—	Connected to cathode at socket			
Grid-No.2 Supply Voltage	—	100	125	150	volts
Cathode-Bias Resistor	330	150	100	68	ohms
Amplification Factor	—	0.5	1.5	1	megohms
Plate Resistance (Approx.)	36	—	—	—	
Transconductance	4800	3900	4500	5200	μmhos
Plate Current	12.2	5	7.6	10.6	mA
Grid-No.2 Current	—	2.1	3	4.3	mA
Grid-No.1 Voltage for plate current of 10 μA	—	—4.2	—5.5	—6.5	volts

† Grid No.2, grid No.3, and internal shield connected to plate.



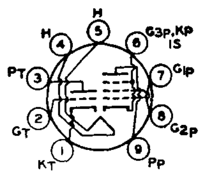
6AU7 Refer to chart at end of section.

6AU8 Refer to chart at end of section.

6AU8A MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

8AU8

Miniature type used in television receiver applications. Pentode unit is used as video amplifier, if amplifier, and agc amplifier. Triode unit is used in sync-amplifier, sync-separator, sync-clipper, and phase-inverter circuits. Outlines section, 6E; requires 9-contact socket. Type 8AU8 is identical with type 6AU8A except for heater ratings.



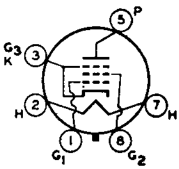
9DX

	6AU8A	8AU8	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	2.2	pF
Grid to Cathode and Heater	2.6	pF
Plate to Cathode and Heater	0.34	pF
Pentode Unit:		
Grid No.1 to Plate	0.06	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	7.5	pF
Plate to Cathode,, Heater, Grid No.2, Grid No.3, and Internal Shield	3.4	pF
Triode Grid to Pentode Plate	0.022 max	pF
Pentode Grid No.1 to Triode Plate	0.006 max	pF
Pentode Plate to Triode Plate	0.12 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage		See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.8	3.3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	1	watt
For grid-No.2 voltages between 165 and 330 volts		See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	150	200	volts
Grid-No.2 Supply Voltage	—	125	volts
Cathode-Bias Resistor	150	82	ohms
Amplification Factor	43	—	
Plate Resistance (Approx.)	8100	100000	ohms
Transconductance	5300	8000	μmhos
Plate Current	9.5	17	mA
Grid-No.2 Current	—	3.4	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—6.5	—7.5	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm



6CK

6AV5GA

BEAM POWER TUBE

12AV5GA, 25AV5GA

Glass octal type used as horizontal-deflection amplifier in television receivers. Outlines section, 19C; requires octal socket. Types 12AV5GA and 25AV5GA are identical with type 6AV5GA except for heater ratings.

	6AV5GA	12AV5GA	25AV5GA	
Heater Voltage (ac/dc)	6.3	12.6	25	volts
Heater Current	1.2	0.6	0.3	amperes
Heater Warm-up Time (Average)	—	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.)		
Grid No.1 to Plate	0.5	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	14	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	7	pF

Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection		Triode• Connection	
Plate Voltage	60	250	150	volts
Grid-No.2 (Screen-Grid) Voltage	150	150	150	volts
Grid-No.1 (Control-Grid) Voltage	0	—22.5	—22.5	volts
Plate Resistance	—	14500	—	ohms
Transconductance	—	5900	—	μmhos
Plate Current	260	57	—	mA
Screen Current	26	2.1	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—43	—	volts
Amplification Factor	—	—	4.3	

• Grid No.2 connected to plate.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)		
DC Plate Voltage	550	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	5500°	volts
Peak Negative-Pulse Plate Voltage	1250	volts
DC Grid-No.2 Voltage	175	volts
Peak Negative-Pulse Grid-No.1 Voltage	300	volts
Peak Cathode Current	400	mA
Average Cathode Current	110	mA
Grid-No.2 Input	2.5	watts
Plate Dissipation††	11	watts
Bulb Temperature (At hottest point)	210	°C

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	0.47	megohm

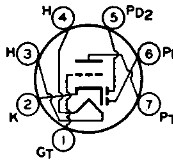
Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
° Under no circumstances should this absolute value be exceeded.
†† A bias resistor or other means is required to protect the tube in absence of excitation.

6AV5GT Refer to chart at end of section.

6AV6

3AV6, 4AV6, 12AV6

TWIN DIODE— HIGH-MU TRIODE



7BT

Miniature type used as combined detector, amplifier, and avc tube in automobile and ac-operated radio receivers. The 6AV6 may be substituted directly for the 6AT6 in applications where the higher amplification of the 6AV6 is advantageous. Outlines section, 5C; requires miniature 7-contact socket. Types 3AV6, 4AV6, and 12AV6 are identical with type 6AV6 except for heater ratings.

	3AV6	4AV6	6AV6	12AV6	
Heater Voltage (ac/dc)	3.15	4.2	6.3	12.6	volts
Heater Current	0.6	0.45	0.3	0.15	ampere
Heater Warm-up Time (Average)	11	11	—	—	seconds
Heater-Cathode Voltage:					
Peak value	±200 max	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	100 max	volts

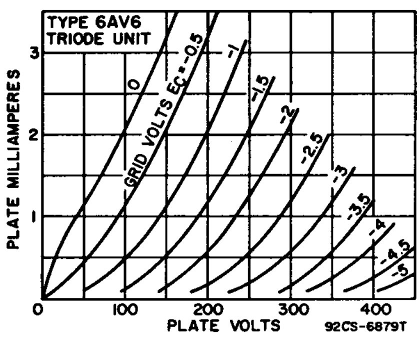
Direct Interelectrode Capacitances:		
Triode Grid to Triode Plate	2	pF
Triode Grid to Cathode and Heater	2.2	pF
Triode Plate to Cathode and Heater	0.8■	pF
Plate of Diode Unit No.2 to Triode Grid	0.04 max	pF

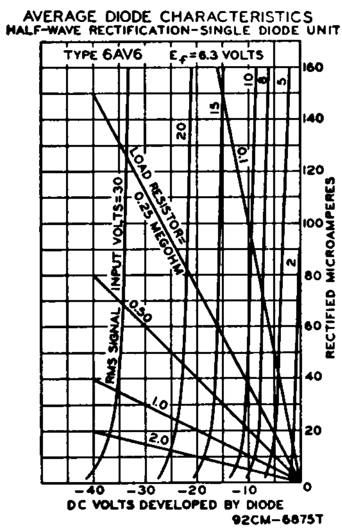
■ This value is 1.2 pF with external shield connected to cathode.

Triode Unit as Class A_1 Amplifier

MAXIMUM RATING (Design-Maximum Value)

Plate Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation	0.55	watt





CHARACTERISTICS

Plate Voltage	100	250	volts
Grid Voltage	—1	—2	volts
Amplification Factor	100	100	
Plate Resistance	80000	62500	ohms
Transconductance	1250	1600	μmhos
Plate Current	0.50	1.2	mA

Diode Units

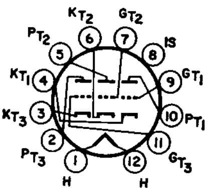
MAXIMUM RATING (Design-Maximum Value)

Plate Current (Each Unit)	1	mA

The two diode plates are placed around a cathode, the sleeve of which is common to the triode unit. Each diode plate has its own base pin. Diode biasing of the triode unit is not recommended.

Installation and Application

The triode unit of the 6AV6 is recommended for use only in resistance-coupled circuits. Refer to the **Resistance-Coupled Amplifier** section for typical operating conditions. **Grid bias for the triode unit of the 6AV6** may be obtained from a fixed source, such as a fixed-voltage tap on the dc power supply, or from a cathode-bias resistor. It should not be obtained by the diode-biasing method because of the probability of plate-current cutoff, even with relatively small signal voltages applied to the diode circuit.



12BY

MEDIUM-MU TRIPLE TRIODE 6AV11

Duodecar type used for general-purpose amplifier, phase inverter, or oscillator applications. **Outlines section, 8A**; requires duodecar 12-contact socket.

Heater Voltage (ac/dc)	6.3		volts
Heater Current	0.6		ampere
Heater-Cathode Voltage:			
Peak value	±200 max		volts
Average value	100 max		volts

Direct Interelectrode Capacitances (Approx.):	Unit No. 1	Unit No. 2	Unit No. 3	
Grid to Plate	1.2	1.2	1.2	pF
Grid to Cathode, Heater, and Internal Shield	1.9	1.9	1.9	pF
Plate to Cathode, Heater, and Internal Shield	1.8	0.7	2	pF

Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Plate Dissipation	2.75	watts
Total Plate Dissipation (All Plates)	6	watts
Average Cathode Current	20	mA

CHARACTERISTICS

Plate Voltage	100	250	volts
Grid Voltage	0	—8.5	volts
Amplification Factor	20	17	
Plate Resistance (Approx.)	6500	7700	ohms
Transconductance	3100	2200	μmhos
Plate Current	11.8	10.5	mA
Grid Voltage (Approx.) for plate current of 10 μA	—	—24	volts

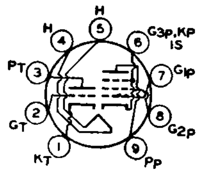
MAXIMUM CIRCUIT VALUES

Grid-No. 1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

Refer to chart at end of section. **6AW8**

6AW8A
8AW8A

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE



9DX

Miniature type used in television receiver applications. The pentode unit is used as an if amplifier, video amplifier, agc amplifier, or reactance tube. The triode unit is used in low-frequency oscillator, sync-separator, sync-clipper, and phase-splitter circuits. Outlines section, 6E; requires miniature 9-contact socket. Type 8AW8A is identical with type 6AW8A except for heater ratings.

	6AW8A	8AW8A	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

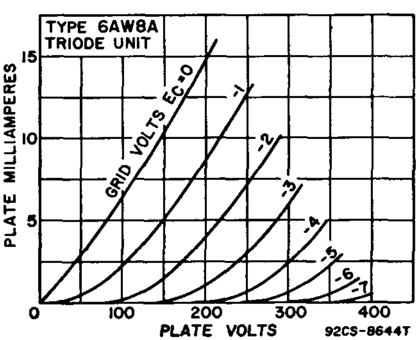
Direct Interelectrode Capacitances:	Unshielded	Shielded	
Triode Unit:			
Grid to Plate	2.2	2.2	pF
Grid to Cathode, Pentode Cathode, Pentode Grid No.3, Internal Shield, and Heater	3.2	3.4	pF
Plate to Cathode, Pentode Cathode, Pentode Grid No.3, Internal Shield, and Heater	1.8	3	pF
Pentode Unit:			
Grid No.1 to Plate	0.06 max	0.05 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	10	10	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3.6	4.5	pF
Pentode Grid No.1 to Triode Plate	0.008 max	0.005 max	pF
Pentode Plate to Triode Plate	0.15 max	0.025 max	pF

■ With external shield connected to pins 4 and 5.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage		See curve page 98	
Grid-No.1 (Control-Grid) Voltage, positive-bias value	0	0	volts
Plate Dissipation	1.1	3.75	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	1.1	watts
For grid-No.2 voltages between 165 and 330 volts		See curve page 98	

CHARACTERISTICS			
Plate Supply Voltage	200	150	volts
Grid-No.2 Supply Voltage	—	150	volts
Grid-No.1 Voltage	—2	—	volts
Cathode-Bias Resistor	—	150	ohms
Amplification Factor	70	—	



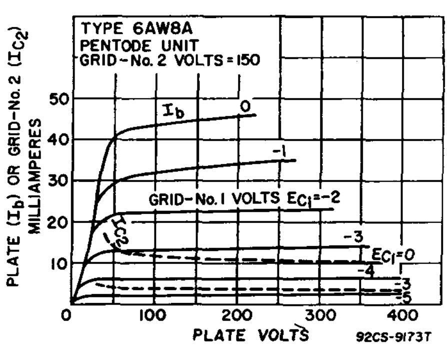
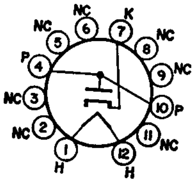


Plate Resistance (Approx.)	—	0.2	megohm
Transconductance	4000	9500	μmhos
Plate Current	4	15	mA
Grid-No.2 Current	—	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—5	—8	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm



12BL

HALF-WAVE VACUUM RECTIFIER

6AX3

12AX3, 17AX3

Duodecar type used as damper tube in horizontal-deflection circuits of television receivers. **Outlines section,** 8C; requires 12-contact socket. Socket terminals 5, 6, 8, and 9 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated. Types 12AX3 and 17AX3 are identical with type 6AX3 except for heater ratings.

	6AX3	12AX3	17AX3	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Direct Interelectrode Capacitances:				
Plate to Cathode and Heater			5.5	pF
Cathode to Plate and Heater			7.5	pF
Heater to Cathode			2.8	pF

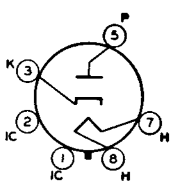
Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)			
Peak Inverse Plate Voltage#		5000	volts
Peak Plate Current		1000	mA
Average Plate Current		165	mA
Plate Dissipation		5.3	watts
Heater-Cathode Voltage:			
Peak value	+300	—5000	volts
Average value	+100	—900	volts
CHARACTERISTICS			
Tube Voltage Drop for plate current of 250 mA		32	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

Refer to chart at end of section. **6AX4GT**



4CG

HALF-WAVE VACUUM RECTIFIER

6AX4GTB

12AX4GTB, 17AX4GTA, 25AX4GT

Glass octal type used as damper tube in horizontal-deflection circuits of color and black-and-white television receivers. **Outlines section,** 13D; requires octal socket. May be supplied with pin No. 1 omitted. Socket terminals 1, 2, 4, and 6 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated. Types 12AX4/GTB, 17AX4GTA, and 25AX4GT are identical with type 6AX4GTB except for heater ratings.

	6AX4-GTB	12AX4-GTB	17AX4-GTA	25AX4GT	
Heater Voltage (ac/dc)	6.3	12.6	16.8	25	volts
Heater Current	1.2	0.6	0.45	0.3	amperes
Heater Warm-up Time (Average)	—	11	11	11	seconds
Direct Interelectrode Capacitances (Approx.):					
Cathode to Plate and Heater				8.5	pF
Plate to Cathode and Heater				5	pF
Heater to Cathode				4	pF

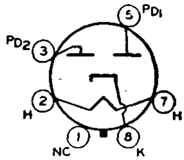
Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)			
Peak Inverse Plate Voltage#		5000	volts
Peak Plate Current		1000	mA
Average Plate Current		165	mA
Plate Dissipation		5.3	watts
Heater-Cathode Voltage:			
Peak value	+300	—5000	volts
Average value	+100	—900	volts
CHARACTERISTICS, Instantaneous Value			
Tube Voltage Drop for plate current of 250 mA		32	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6AX5GT FULL-WAVE VACUUM RECTIFIER



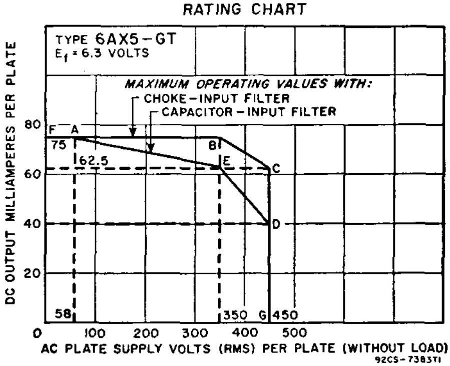
6S

Glass octal type used in power supplies of radio equipment having moderate dc requirements. Outlines section, 13D; requires octal socket. This type may be supplied with pin No. 1 omitted. This tube, like other power-handling tubes, should be adequately ventilated. Heater: volts (ac), 6.3; amperes, 1.2.

Full-Wave Rectifier

MAXIMUM RATINGS (Design-Center Values)			
Peak Inverse Plate Voltage		1250	volts
Peak Plate Current (Per Plate)		375	mA
Hot-Switching Transient Plate Current:			
For duration of 0.2 second maximum		2.6	amperes
AC Plate Supply Voltage (Per Plate, rms)		See Rating Chart	
Average Output Current (Per Plate, rms)		See Rating Chart	
Peak Heater-Cathode Voltage		±450	volts
TYPICAL OPERATION WITH CAPACITOR INPUT TO FILTER			
AC Plate-to-Plate Supply Voltage (rms)	700	900	volts
Filter Input Capacitor*	10	10	µF
Effective Plate-Supply Impedance Per Plate	50	105	ohms
DC Output Voltage at Input to Filter (Approx.):			
At half-load current of 62.5 mA	395	—	volts
At half-load current of 40 mA	—	540	volts
At full-load current of 125 mA	350	—	volts
At full-load current of 80 mA	—	490	volts
Voltage Regulation (Approx.):			
Half-load to full-load current	45	50	volts
TYPICAL OPERATION WITH CHOKE INPUT TO FILTER			
AC Plate-to-Plate Supply Voltage (rms)	700	900	volts
Filter Input Choke	10#	10##	henries
DC Output Voltage at Input to Filter (Approx.):			
At half-load current of 75 mA	270		volts
At half-load current of 62.5 mA	—	365	volts
At full-load current of 150 mA	250		volts
At full-load current of 125 mA	—	350	volts

RATING CHART



Voltage Regulation (Approx.):			
Half-load to full-load current	20	15	volts

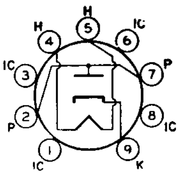
* Higher values of capacitance than indicated may be used but the effective plate-supply impedance may have to be increased to prevent exceeding the maximum rating for hot-switching transient plate current.

\# This value is adequate to maintain optimum regulation provided the load current is not less than 30 mA. For load currents less than 30 mA, a larger value of inductance is required for optimum regulation.

\#\# This value is adequate to maintain optimum regulation provided the load current is not less than 35 mA. For load currents less than 35 mA, a larger value of inductance is required for optimum regulation.

Refer to chart at end of section. **6AX8**

Refer to chart at end of section. **6AY3**



9HP

HALF-WAVE VACUUM RECTIFIER

6AY3B

12AY3A, 17AY3A

Novar type used as damper tubes in horizontal-deflection circuits of black-and-white television receivers. **Outlines section, 30B;** requires novar 9-contact socket. Socket terminals 1, 3, 6, and 8 should not be used as tie points. It is especially important that these tubes, like other power-handling tubes, be adequately ventilated. Types 12AY3A and 17AY3A are identical with type 6AY3B except for heater ratings.

	6AY3B	12AY3A	17AY3A	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Direct Interelectrode Capacitances (Approx.):				
Plate to Cathode and Heater			6.5	pF
Cathode to Plate and Heater			9	pF
Heater to Cathode			2.8	pF

Damper Service

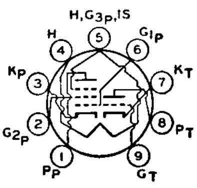
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		5000	volts
Peak Plate Current		1100	mA
Average Plate Current		175	mA
Plate Dissipation		6.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—5000	volts
Average value	+100	—900	volts

\# Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

Refer to chart at end of section. **6AY11**



9ED

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

6AZ8

Miniature type used in color and black-and-white television receiver applications. The pentode unit is used as an if amplifier, video amplifier, agc amplifier, or reactance tube. The triode unit is used in low-frequency oscillator, sync-separator, sync-clipper, and phase-splitter circuits. **Outlines section, 6B;** requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater-Cathode Voltage:▲		
Peak value	±200 max	volts
Average value	100 max	volts

Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	1.7	pF
Grid to Cathode, Heater, Pentode Grid No.3, and Internal Shield	2	pF
Plate to Cathode, Heater, Pentode Grid No.3, and Internal Shield	1.7	pF
Pentode Unit:		
Grid No.1 to Plate	0.02 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	6.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.2	pF
Triode Grid to Pentode Plate	0.027 max	pF
Pentode Grid No.1 to Triode Plate	0.020 max	pF
Pentode Plate to Triode Plate	0.045 max	pF

▲ The heater-cathode voltage of the pentode unit should not exceed the value of the operating cathode bias. Grid No.3 will be made negative with respect to cathode if this value is exceeded, and thus possibly cause a change in tube characteristics.

Class A₁ Amplifier

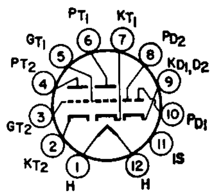
MAXIMUM RATINGS (Design-Center Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	300	volts
Grid-No.2 Voltage		See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.6	2	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 150 volts	—	0.5	watt
For grid-No.2 voltages between 150 and 300 volts		See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	200	200	volts
Grid-No.2 Voltage	—	150	volts
Grid-No.1 Voltage	−6	—	volts
Cathode-Bias Resistor	—	180	ohms
Amplification Factor	19	—	
Plate Resistance (Approx.)	5750	300000	ohms
Transconductance	3300	6000	μmhos
Plate Current	13	9.5	mA
Grid-No.2 Current	—	3	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	−19	—	volts
Grid-No.1 Voltage (Approx.) for transconductance of 100 μmhos	—	−12.5	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:*			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

* If either unit is operating at maximum rated conditions, grid-No.1-circuit resistance for both units should not exceed the stated values.

6B4G	Refer to chart at end of section.
6B5	Refer to chart at end of section.
6B6G	Refer to chart at end of section.
6B7 **6B7S**	Refer to chart at end of section.
6B8	Refer to chart at end of section.
6B8G	Refer to chart at end of section.

6B10
8B10

TWIN DIODE— MEDIUM-MU TWIN TRIODE



12BF

Duodecar type used in television receiver applications; diode units are used in horizontal-phase-detector circuits, and triode units are used in horizontal-oscillator circuits. Outlines section, 8A; requires duodecar 12-contact socket. Type 8B10 is identical with type 6B10 except for heater ratings.

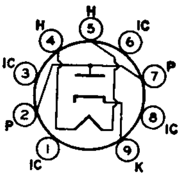
	6B10	8B10	
Heater Voltage (ac/dc)	6.3	8.5	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A₁ Amplifier (Each Triode Unit)

MAXIMUM RATING (Design-Maximum Value)		
Plate Voltage	330	volts
Average Cathode Current	20	mA
Plate Dissipation	3	watts
CHARACTERISTICS		
Plate Voltage	250	volts
Grid Voltage	—8	volts
Amplification Factor	18	
Plate Resistance (Approx.)	7200	ohms
Transconductance	2500	μmhos
Plate Current	10	mA
Grid Voltage (Approx.) for plate current of 50 μA	—20	volts
MAXIMUM CIRCUIT VALUES		
Grid-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

Diode Units (Each Unit)

MAXIMUM RATING (Design-Maximum Value)		
Plate Current	5	mA
CHARACTERISTICS, Instantaneous Value		
Tube Voltage Drop for plate current of 20 mA	5	volts



9HP

HALF-WAVE VACUUM RECTIFIER — 6BA3

Novar type used as damper tube in horizontal-deflection circuits of black-and-white television receivers. Outlines section, 11B or 30C; requires novar 9-contact socket. Socket terminals 1, 3, 6, and 8 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated.

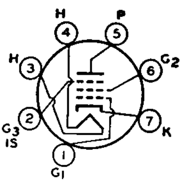
Heater Voltage (ac/dc)	6.3	volts
Heater Current		
Direct Interelectrode Capacitances (Approx.):	4.4	pF
Plate to Cathode and Heater	6	pF
Cathode to Plate and Heater	1.8	pF
Heater to Cathode	1.2	ampere

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)			
Peak Inverse Plate Voltage#		5000	volts
Peak Plate Current		1000	mA
Average Plate Current		165	mA
Plate Dissipation		5.3	watts
Heater-Cathode Voltage:			
Peak value	+300	—5000	volts
Average value	+100	—900	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).



7BK

REMOTE-CUTOFF PENTODE — 6BA6, 6BA6/EF93

12BA6

Miniature types used as rf amplifiers in standard broadcast and FM receivers, as well as in wide-band, high-frequency applications. The low value of grid-No.1-to-plate capacitance minimizes regenerative effects, while the high transconductance makes possible high signal-to-noise ratio. Outlines section, 5C; require miniature 7-contact socket. Type 12BA6 is identical with type 6BA6 except for heater ratings.

	6BA6 6BA6/EF93	12BA6	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	0.3	0.15	ampere
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:			
Grid No.1 to Plate		0.0035 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield		5.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield		5•	pF

• This value is 5.5 pF with external shield connected to cathode.

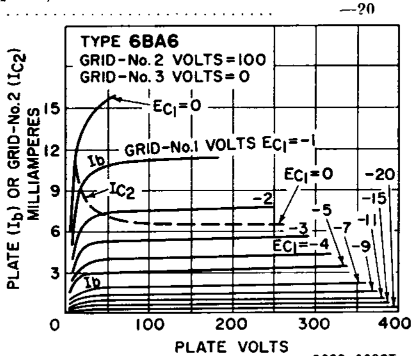
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Voltage	See curve page 98	
Grid-No.2 Supply Voltage	330	volts
Plate Dissipation	3.4	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.7	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	55	volts
Positive-bias value	0	volts

CHARACTERISTICS

Plate Supply Voltage	100	250	volts
Grid No.3 and Internal Shield	Connected to cathode at socket		
Grid-No.2 Supply Voltage	100	100	volts
Cathode-Bias Resistor	68	68	ohms
Plate Resistance (Approx.)	0.25	1	megohm
Transconductance	4300	4400	μmhos
Plate Current	10.8	11	mA
Grid-No.2 Current	4.4	4.2	mA
Grid-No.1 Voltage (Approx.) for transconductance of 40 μmhos	—20	—20	volts

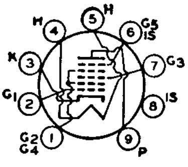


92CS-6609T

Installation and Application

Control-grid bias variation is effective in changing the volume of the receiver. To obtain adequate volume control, an available grid-No.1-bias voltage of approximately 50 volts is required. The exact value depends upon the circuit design and operating conditions. This voltage may be obtained, depending on the receiver requirements, from a potentiometer across a fixed supply voltage, from a variable cathode-bias resistor, from the avc system, or from a combination of these methods.

The **grid-No.2 (screen-grid)** voltage may be obtained from a potentiometer or bleeder circuit across the B-supply source, or through a dropping resistor from the plate supply. The use of series resistors for obtaining satisfactory control of grid-No.2 voltage in the case of four-electrode tubes is usually impossible because of secondary-emission phenomena. In the 6BA6, however, because grid No.3 practically removes these effects, it is practical to obtain grid-No.2 voltage through a series-dropping resistor from the plate supply or from some high intermediate voltage, provided the source does not exceed the plate-supply voltage. With this method, the grid-No.2-to-cathode voltage will fall off very little from minimum to maximum value of the resistor controlling cathode bias. In some cases, it may actually rise. This rise of grid-No.2-to-cathode voltage above the normal maximum value is allowable because both the grid-No.2 current and the plate current are reduced simultaneously by a sufficient amount to prevent damage to the tube. It should be recognized that, in general, the series-resistor method of obtaining grid-No.2 voltage from a higher voltage supply necessitates the use of the variable cathode-resistor method of controlling volume in order to prevent too high a voltage on grid No.2. When grid-No.2 and control-grid voltage are obtained in this manner, the remote "cutoff" advantage of the 6BA6 can be fully realized. However, it should be noted that the use of a resistor in the grid-No.2 circuit has an effect on the change in plate resistance with variation in grid-No.3 (suppressor-grid) voltage in case grid No.3 is utilized for control purposes.



8CT

PENTAGRID CONVERTER 6BA7

Miniature type used as converter in AM and FM receivers. **Outlines section, 6E; requires miniature 9-contact socket.**

Heater Voltage	6.3	volts
Heater Current	0.3	ampere
Peak Heater-Cathode Voltage	±90	volts
Direct Interelectrode Capacitances:		
Grid No. 3 to All Other Electrodes	9.5	pF
Plate to All Other Electrodes	8.3	pF
Grid No. 1 to All Other Electrodes	6.7	pF
Grid No. 3 to Plate	0.19 max	pF
Grid No. 3 to Grid No. 1	0.1 max	pF
Grid No. 1 to Plate	0.05 max	pF
Grid No. 1 to All Other Electrodes, except Cathode	3.4	pF
Grid No. 1 to Cathode	3.3	pF
Cathode to All Other Electrodes except Grid No. 1	4	pF

Converter Service

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Grid-No.5-and-Internal-Shield Voltage△	0	volts
Grids-No.-2-and-No.4 (Screen-Grid) Voltage	100	volts
Grids-No.2-and-No.4 Supply Voltage	300	volts
Plate Dissipation	2	watts
Grids-No.2-and-No.4 Input	1.5	watts
Total Cathode Current	22	mA
Grid-No.3 Voltage:		
Negative-bias value	100	volts
Positive-bias value	0	volts

CHARACTERISTICS (Separate Excitation)*

Plate Voltage	100	250	volts
Grid No.5 and Internal Shield△	Connected directly to ground		
Grids-No.2-and-No.4 (Screen-Grid) Voltage	100	100	volts

Grid-No.3 (Control-Grid) Voltage	—1	—1	volt
Grid-No.1 (Oscillator-Grid) Resistor	20000	20000	ohms
Plate Resistance (Approx.)	0.5	1	megohm
Conversion Transconductance	900	950	µmhos
Conversion Transconductance (Approx.)**	3.5	3.5	µmhos
Plate Current	3.6	3.8	mA
Grids-No.2-and-No.4 Current	10.2	10	mA
Grid-No.1 Current	0.35	0.35	mA
Total Cathode Current	14.2	14.2	mA

NOTE: The transconductance between grid No.1 and grids No.2 and No.4 connected to plate (not oscillating) is approximately 8000 µmhos under the following conditions: signal applied to grid No.1 at zero bias; grids No.2 and No.4 and plate at 100 volts; grid No.3 grounded. Under the same conditions, the plate current is 32 milliamperes, and the amplification factor is 16.5.

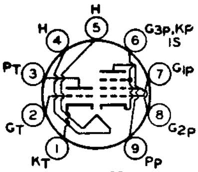
*** The characteristics shown with separate excitation correspond very closely with those obtained in a self-excited oscillator circuit operating with zero bias.**

** With grid-No.3 bias of —20 volts.

△ Internal Shield (pins No.6 and No.8) connected directly to ground.

6BA8A
8BA8A

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE



9DX

Miniature type used in color and black-and-white television receivers. The pentode unit is used as a video amplifier, an agc amplifier, or a reactance tube. The triode unit is used in low-frequency oscillator and phase-splitter circuits. **Outlines section,** 6E; requires miniature 9-contact socket. Type 8BA8A is identical with type 6BA8A except for the heater ratings.

	6BA8A	8BA8A	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.3	0.45	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances (Approx.):			
Triode Unit:			
Grid to Plate	2.2	2.2	pF
Grid to Cathode and Heater	2.5	2.7	pF
Plate to Cathode and Heater	0.4	1.9	pF
Pentode Unit:			
Grid No.1 to Plate	0.06	0.05	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	10	10	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3.6	4.5	pF
Triode Grid to Pentode Plate	0.016	0.006	pF
Pentode Grid No.1 to Triode Plate	0.006	0.003	pF
Pentode Plate to Triode Plate	0.15	0.023	pF

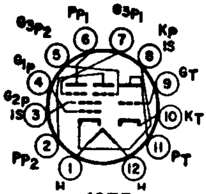
Class A_1 Amplifier

▪ With external shield connected to cathode of unit under test.

MAXIMUM RATINGS (Design-Center Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	300	volts
Grid-No.2 Voltage		See curve page 98	
Grid-No.1 (Control-Grid) Voltage:			
Negative-bias value	—	—50	volts
Positive-bias value	—	0	volts
Plate Dissipation	2	3.25	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 150 volts	—	1	watt
For grid-No.2 voltages between 150 and 300 volts		See curve page 98	
CHARACTERISTICS			
Plate-Supply Voltage	200	200	volts
Grid-No.2 Supply Voltage	—	150	volts
Grid-No.1 Voltage	—8	—	volts
Cathode-Bias Resistor	—	180	ohms
Amplification Factor	18	—	
Plate Resistance (Approx.)	6700	400000	ohms
Transconductance	2700	9000	µmhos
Plate Current	8	13	mA
Grid-No.2 Current	—	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 µA	—16	—10	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm



12ER

6BA11

TRIODE—TWIN PENTODE

8BA11

Duodecar type used as vertical-deflection oscillator and for combined sync-agc applications in color and black-and-white television receivers. **Outlines section, 8B**; requires duodecar 12-contact socket. Type 8BA11 is identical with type 6BA11 except for heater ratings.

	6BA11	8BA11	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	amperes
Heater Warm-up Time	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	2	pF
Grid to Cathode and Heater	2	pF
Plate to Cathode, Heater, and Internal Shield	1.9	pF
Pentode Unit (Each Unit):		
Grid No.3 to Plate	2	pF
Grid No.3 to all Other Electrodes	6	pF
Grid No.1 to all Other Electrodes	3	pF
Plate to all Other Electrodes	3	pF
Grid No.3 of Pentode 1 to Grid No.3 of Pentode 2	0.026 max	pF

Triode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

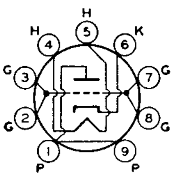
Plate Voltage	300	volts
Average Cathode Current	20	mA
Plate Dissipation	1.5	watts

CHARACTERISTICS

Plate Voltage	250	volts
Grid Voltage	—11	volts
Amplification Factor	18	
Transconductance	1800	µmhos
Plate Current	5	mA
Grid Voltage (Approx.) for plate current of 100 µA	—18	volts

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm



9DR

6BC4

MEDIUM-MU TRIODE

Miniature type used as an rf amplifier in the cathode-drive circuits of uhf television tuners covering the frequency range of 470 to 890 MHz. **Outlines section, 6A**; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.225	ampere
Peak Heater-Cathode Voltage	±75 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid to Plate	1.6	pF
Grid to Heater and Cathode	2.9	pF
Plate to Heater and Cathode	0.26	pF
Heater to Cathode	2.7	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	250	volts
Cathode Current	25	mA
Plate Dissipation	2.5	watts

CHARACTERISTICS

Plate Supply Voltage	150	volts
Cathode-Bias Resistor	100	ohms
Amplification Factor	48	
Plate Resistance (Approx.)	4800	ohms
Transconductance	10000	µmhos
Plate Current	14.5	mA
Grid Voltage (Approx.) for plate current of 10 µA	—10	volts

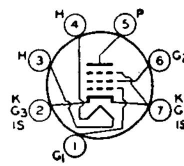
MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance:		
For fixed-bias operation	Not recommended	
For cathode-bias operation	0.5	megohm

6BC5

3BC5/3CE5, 6BC5/6CE5

SHARP-CUTOFF PENTODE



7BD

Miniature type used in compact radio equipment as an rf or if amplifier at frequencies up to 400 MHz. Outlines section, 5C; requires miniature 7-contact socket. For typical operation as resistance-coupled amplifier, refer to Resistance-Coupled Amplifier section. Types 3BC5/3CE5 and 6BC5/6CE5 are identical with type 6BC5 except for heater ratings.

	3BC5/3CE5	6BC5 6BC5/6CE5	
Heater Voltage (ac/dc)	3.15	6.3	volts
Heater Current	0.6	0.3	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±90 max	volts
Average value	100 max	—	volts

Direct Interelectrode Capacitances:		
Pentode Connection:		
Grid No.1 to Plate	0.030 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	6.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	1.8	pF
Triode Connection:*		
Grid No.1 to Plate and Grid No.2	2.5	pF
Grid No.1 to Cathode, Heater, Grid No.3, and Internal Shield	3.9	pF
Plate and Grid No.2 to Cathode, Heater, Grid No.3, and Internal Shield	3	pF

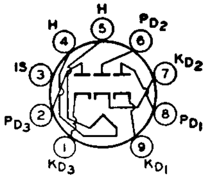
* Grid No.2 connected to plate.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)	Triode Connection*	Pentode Connection	
Plate Voltage	300	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	300	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	2	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 150 volts	—	0.5	watt
For grid-No.2 voltages between 150 and 300 volts	—	See curve page 98	

CHARACTERISTICS	Triode Connection*		Pentode Connection			
Plate Supply Voltage	180	250	100	125	250	volts
Grid-No.2 Supply Voltage	—	—	100	125	150	volts
Cathode-Bias Resistor	330	820	180	100	180	ohms
Amplification Factor	42	40	—	—	—	
Plate Resistance (Approx.)	0.006	0.009	0.6	0.5	0.8	megohm
Transconductance	6000	4400	4900	6100	5700	µmhos
Plate Current	8	6	4.7	8	7.5	mA
Grid-No.2 Current	—	—	1.4	2.4	2.1	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 µA	—	—	—5	—6	—8	volts

* Grid No.2 connected to plate.



9AX

TRIPLE DIODE 6BC7

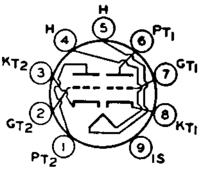
Miniature type containing three high-perveance diode units in one envelope; used in dc restorer circuits of color television receivers and in AM/FM radio receivers as a combination FM discriminator and AM detector tube. **Outlines section,** 6B; requires 9-contact miniature socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.450	ampere
Peak Heater-Cathode Voltage	±200 max	volts
Direct Interelectrode Capacitances (Approx.):		
Diode-No.1 Plate to Diode-No.1 Cathode, Heater, and Internal Shield	3.5	pF
Diode-No.2 Plate to Diode-No.2 Cathode, Heater, and Internal Shield	5.5	pF
Diode-No.3 Plate to Diode-No.3 Cathode, Heater, and Internal Shield	3.5	pF

MAXIMUM RATINGS (Design-Center Values, Each Diode Unit)

Peak Inverse Plate Votlage	330	volts
Peak Plate Current*	54	mA
Average Output Current	12	mA

* In rectifier service, the minimum total effective plate-supply impedance per plate is 560 ohms.



9AJ

MEDIUM-MU TWIN TRIODE 6BC8 4BC8

Miniature type used as a cascode amplifier in vhf television tuners and in push-pull cathode-drive rf amplifiers. **Outlines section,** 6B; requires miniature 9-contact socket. Type 4BC8 is identical with type 6BC8 except for heater ratings.

	4BC8	6BC8	
Heater Voltage (ac/dc)	4.2	6.3	volts
Heater Current	0.6	0.4	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200*max	±200*max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances*:	Unit No.1	Unit No.2	
Grid to Plase	1.2	1.2	pF
Grid to Catnode, Heater, and Internal Shield	2.6	—	pF
Cathode to Grid, Heater, and Internal Shield	—	5.5	pF
Plate to Cathode, Heater, and Internal Shield	1.3	—	pF
Plate to Grid, Heater, and Internal Shield	—	2.4	pF
Plate to Cathode	—	0.12	pF
Heater to Cathode	2.8	2.8	pF
Plate of Unit No.1 to Plate of Unit No.2	0.02 max		pF
Plate of Unit No.2 to Plate and Grid of Unit No.1	0.04 max		pF

* Rating may be as high as 300 volts under cutoff conditions, when tube is used as a cascode amplifier, the two units are connected in series, and heater is negative with respect to cathode.
* With external shield connected to internal shield.

Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	250*	volts
Cathode Current	22	mA
Plate Dissipation	2.2	watts

CHARACTERISTICS

Plate Supply Voltage	150	volts
Cathode-Bias Resistor	220	ohms
Plate Resistance (Approx.)	5300	ohms
Amplification Factor	35	
Transconductance	6200	μmhos
Plate Current	10	mA
Grid Voltage (Approx.) for transconductance of 50 μmhos	—13	volts

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance	0.5	megohm

* Rating may be as high as 300 volts under cutoff conditions, when tube is used as a cascode amplifier, the two units are connected in series, and heater is negative with respect to cathode.

6BD4
6BD4A

Refer to chart at end of section.

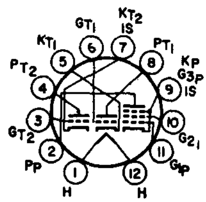
6BD6

Refer to chart at end of section.

6BD11

15BD11

DUAL TRIODE— SHARP-CUTOFF PENTODE



12DP

Duodecar type used in television receiver applications. The high-mu triode unit No.1 is used in general-purpose applications, the medium-mu triode unit No.2 in sync-separator circuits, and the pentode unit as a video amplifier. Outlines section, 8B; requires duodecar 12-contact socket. Type 15BD11 is identical with type 6BD11 except for heater ratings.

	6BD11	15BD11	
Heater Voltage (ac/dc)	6.3	14.7	volts
Heater Current	1.05	0.45	amperes
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit No.1	Triode Unit No.2	Pentode Unit	
Plate Voltage	330	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	—	330	volts
Grid-No.2 Voltage	—	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	0	volts
Plate Dissipation	1.5	2	4	watts
Grid-No.2 Input:				
For grid-No.2 voltages up to 165 volts	—	—	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	—	—	See curve page 98	

CHARACTERISTICS	Triode Unit No.1	Triode Unit No.2	Pentode Unit		
Plate Supply Voltage	200	200	35	135	volts
Grid-No.2 Supply Voltage	—	—	135	135	volts
Grid-No.1 Voltage	—2	—	0	0	volts
Cathode-Bias Resistor	—	220	—	100	ohms
Amplification Factor	68	41	—	—	
Plate Resistance (Approx.)	12400	9400	—	45000	ohms
Transconductance	5500	4400	—	10400	μmhos
Plate Current	7	9.2	34▪	17	mA
Grid-No.2 Current	—	—	13▪	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—5.5	—6.5	—	—6	volts

MAXIMUM CIRCUIT VALUES

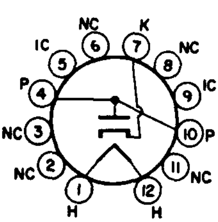
Grid-No.1-Circuit Resistance:				
For fixed-bias operation	0.5	0.5	1	megohm
For cathode-bias operation	1	1	1	megohm

▪ This value may be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

6BE3
6BE3/6BZ3

12BE3, 17BE3/17BZ3

HALF-WAVE VACUUM RECTIFIER



12GA

Duodecar type used as damper tube in horizontal-deflection circuits of color and black-and-white television receivers. *Outlines section*, 8D; requires duodecar 12-contact socket. Types 12BE3 and 17BE3/17BZ3 are identical with type 6BE3 except for heater ratings.

	6BE3 6BE3/6BZ3	12BE3	17BE3/ 17BZ3	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.46	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Direct Interelectrode Capacitances (Approx.):				
Plate to Cathode, and Heater			10	pF
Cathode to Heater, and Plate			8	pF
Heater to Cathode			3.4	pF

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		5000	volts
Peak Plate Current		1200	mA
Average Plate Current		200	mA
Plate Dissipation		6.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—5000	volts
Average value	+100	—900	volts

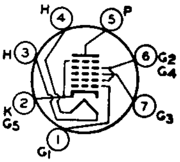
CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for dc plate current of 350 mA	25	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

PENTAGRID CONVERTER 6BE6

12BE6



7CH

Miniature type used as converter in AM and FM receivers. Outlines section, 5C; requires miniature 7-contact socket. The 6BE6 is similar in performance to metal type 6SA7. For general discussion of pentagrid types, see Frequency Conversion in Electron Tube Applications section. Type 12BE6 is identical with type 6BE6 except for heater ratings.

	6BE6	12BE6	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	0.3	0.15	ampere
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:	Unshielded	Shielded▪	
Grid No.3 to Plate	0.30 max	0.25 max	pF
Grid No.3 to Grid No.1	0.15 max	0.15 max	pF
Grid No.1 to Plate	0.10 max	0.05 max	pF
Grid No.3 to All Other Electrodes	7	7	pF
Grid No.1 to All Other Electrodes	5.5	5.5	pF
Plate to All Other Electrodes	8.0	13.0	pF
Grid No.1 to Cathode and Grid No.5	3	3	pF
Cathode and Grid No.5 to All Other Electrodes except Grid No.1	15	20	pF

▪ With external shield connected to cathode and grid No.5.

Converter

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grids-No.2-and-No.4 (Screen-Grid) Voltage	110	volts
Grids-No.2-and-No.4 Supply Voltage	330	volts
Cathode Current	15.5	mA
Plate Dissipation	1.1	watts
Grids-No.2-and-No.4 Input	1.1	watts
Grid-No.3 Voltage:		
Negative-bias value	55	volts
Positive-bias value	0	volts
Heater-Cathode Voltage:		
Peak value	200	volts
Average value	100	volts

TYPICAL OPERATION (Separate Excitation)*

Plate Voltage	100	250	volts
Grids-No.2-and-No.4 (Screen-Grid) Voltage	100	100	volts
Grid-No.1 (Oscillator-Grid) Voltage (rms)	10	10	volts
Grid-No.3 (Control-Grid) Voltage	—1.5	—1.5	volts
Grid-No.1 (Oscillator-Grid) Resistor	20000	20000	ohms
Plate Resistance (Approx.)	0.4	1	megohm

Conversion Transconductance	455	475	μmhos
Plate Current	2.6	2.9	mA
Grids-No.2-and-No.4 Current	7.0	6.8	mA
Grid-No.1 Current	0.5	0.5	mA
Cathode Current	10.1	10.2	mA
Grid-No.3 Voltage for conversion transconductance of 10 μmhos	—30	—30	volts

NOTE: The transconductance between grid No.1 and grids No.2 and No.4 connected to plate (not oscillating) is approximately 7250 μmhos under the following conditions: grids No.1 and No.3 at 0 volts; grids No.2 and No.4 and plate at 100 volts. Under the same conditions, the cathode current is 25 mA, and the amplification factor is 20. Grid-No.1 voltage (Approx.) for plate current of 10 μA is —11 volts.

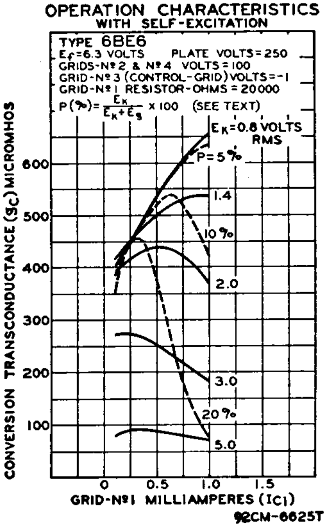
* The characteristics shown with separate excitation correspond very closely with those obtained in a self-excited circuit operating with zero bias.

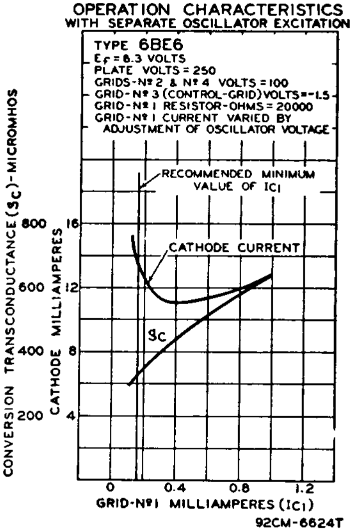
Installation and Application

Because of the special structural arrangement of the 6BE6, a change in signal-grid voltage produces little change in cathode current. Consequently, an rf voltage on the signal grid produces little modulation of the electron current flowing in the cathode circuit. This feature is important because it is desirable that the impedance in the cathode circuit should produce little degeneration or regeneration of the signal-frequency input and intermediate-frequency output. Another important feature is that, because signal-grid voltage has very little effect on the space charge near the cathode, changes in avc bias produce little change in oscillator transconductance and in the input capacitance of grid No.1. There is, therefore, little detuning of the oscillator by avc bias.

A typical self-excited oscillator circuit employing the 6BE6 is given in the **Circuits** section.

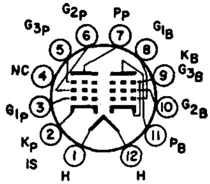
In the 6BE6 operation characteristics curves with self-excitation, E_k is the voltage across the oscillator-coil section between cathode and ground; E_g is the oscillator voltage between cathode and grid.





Refer to chart at end of section. **6BF5**

Refer to chart at end of section. **6BF6**



12EZ

BEAM POWER TUBE— SHARP-CUTOFF PENTODE

6BF11

12BF11, 17BF11

Duodecar type used as combined detector and amplifier tube in color and black-and-white television receivers. The dual-control, sharp-cutoff pentode unit is used as an FM detector and the beam power unit as an af output amplifier. Outlines section, 8C; requires duodecar 12-contact socket. Types 12BF11 and 17BF11 are identical with type 6BF11 except for heater ratings.

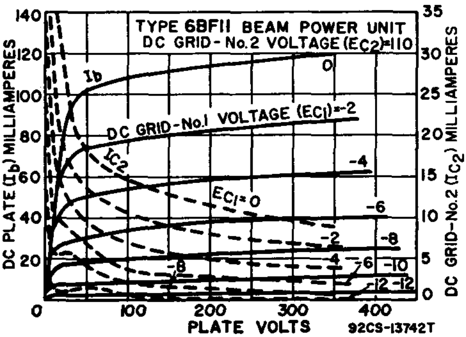
	6BF11	12BF11	17BF11	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Pentode Unit:		
Grid No.1 to Plate	0.36	pF
Grid No.3 to Plate	3.2	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	6.5	pF
Grid No.3 to Cathode, Heater, Grid No.1, Grid No.2, Plate, and Internal Shield	8	pF
Grid No.1 to Grid No.3	0.11	pF
Beam Power Unit:		
Grid No.1 to Plate	0.24	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	13	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	10	pF
Pentode Plate to Beam Power Plate	0.13	pF

Beam Power Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	165	volts
Grid-No.2 (Screen-Grid) Voltage	150	volts
Average Cathode Current	65	mA
Plate Dissipation	6.5	watts
Grid-No.2 Input	1.8	watts

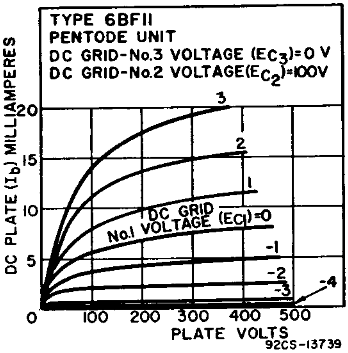


TYPICAL OPERATION

Plate Voltage	145	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 (Control-Grid) Voltage	—6	volts
Peak AF Grid-No.1 Voltage	6	volts
Zero-Signal Plate Current	36	mA
Maximum-Signal Plate Current	40	mA
Zero-Signal Grid No.2 Current	3	mA
Maximum-Signal Grid-No.2 Current	9	mA
Plate Resistance (Approx.)	0.03	megohm
Transconductance	8600	μmhos
Load Resistance	30000	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	2.4	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	0.5	megohm



Pentode Unit as Class A_1 Amplifier

CHARACTERISTICS

Plate Supply Voltage	150	volts
Grid No.3 (Control-Grid)	Connected to negative end of cathode resistor	
Grid-No.2 (Screen-Grid) Supply Voltage	100	volts
Grid No.1 (Control Grid)	Connected to negative end of cathode resistor	
Cathode-Bias Resistor	560	ohms
Plate Resistance (Approx.)	0.15	megohm
Transconductance, Grid No.1 to Plate	1000	μmhos
Transconductance, Grid No.3 to Plate	400	μmhos
Plate Current	1.3	mA
Grid-No.2 Current	2	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—4.5	volts
Grid-No.3 Voltage (Approx.) for plate current of 10 μA	—4.5	volts

Pentode Unit as FM Sound Detector

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.3 Voltage	28	volts
Grid No.2 Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	1.7	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

MAXIMUM CIRCUIT VALUES

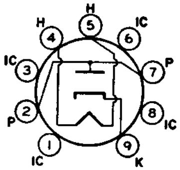
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	0.5	megohm

6BG6G

6BG6GA

Refer to chart at end of section.

Refer to chart at end of section. **6BH3**



9HP

HALF-WAVE VACUUM RECTIFIER

6BH3A

17BH3A, 22BH3A

Novar types used as damper tubes in horizontal-deflection circuits of black-and-white television receivers. Outlines section, 30B; requires novar 9-contact socket. Socket terminals 1, 3, 6, and 8 should not be used as tie points. These tubes, like other power-handling tubes, should be adequately ventilated. Types 17BH3A and 22BH3A are identical with type 6BH3A except for heater ratings.

	6BH3A	17BH3A	22BH3A	
Heater Voltage (ac/dc)	6.3	17	22.4	volts
Heater Current	1.6	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Direct Interelectrode Capacitances (Approx.):				
Plate to Cathode and Heater			6.5	pF
Cathode to Plate and Heater			9	pF
Heater to Cathode			2.8	pF

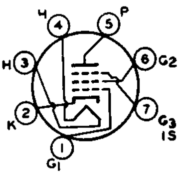
Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		5500	volts
Peak Plate Current		1100	mA
Average Plate Current		130	mA
Plate Dissipation		6.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—5500	volts
Average value	+100	—900	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).



7CM

SHARP-CUTOFF PENTODE

6BH6

Miniature type used as rf amplifier particularly in ac/dc receivers and in mobile equipment where low heater-current drain is important. It is particularly useful in high-frequency, wide-band applications. Outlines section, 5C; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.15	ampere
Peak Heater-Cathode Voltage	±90 max	volts
Direct Interelectrode Capacitances:•		
Grid No.1 to Plate	0.0035 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.4	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4.4	pF

• Without external shield, or with external shield connected to cathode.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Grid-No.2 (Screen-Grid) Voltage	See curve page 98	
Grid-No.2 Supply Voltage	300	volts
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts

Plate Dissipation		3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 150 volts		0.5	watt
For grid-No.2 voltages between 150 and 300 volts		See curve page 98	

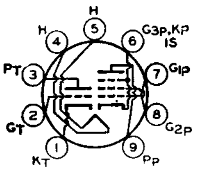
CHARACTERISTICS

Plate Voltage	100	250	volts
Grid No.3	Connected to cathode at socket		
Grid-No.2 Voltage	100	150	volts
Grid-No.1 Voltage	—1	—1	volt
Plate Resistance (Approx.)	0.7	1.4	megohms
Transconductance	3400	4600	μmhos
Plate Current	3.6	7.4	mA
Grid-No.2 Current	1.4	2.9	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—5	—7.7	volts

6BH8

8BH8

MEDIUM-MU TRIODE—SHARP-CUTOFF PENTODE



9DX

Miniature type used in color and black-and-white television receiver applications. The pentode unit is used as an if amplifier, video amplifier, or agc amplifier. The triode unit is used in low-frequency oscillator circuits. **Outlines section, 6E**; requires miniature 9-contact socket. Type 8BH8 is identical with type 6BH8 except for heater ratings.

	6BH8	8BH8	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Triode Unit:		
Grid to Plate	2.4	pF
Grid to Cathode and Heater	2.6	pF
Plate to Cathode and Heater	0.38	pF
Pentode Unit:		
Grid No.1 to Plate	0.046	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	7	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.4	pF
Triode Grid to Pentode Plate	0.016	pF
Pentode Grid No.1 to Triode Plate	0.004	pF
Pentode Plate to Triode Plate	0.095	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	300	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 150 volts	—	1	watt
For grid-No.2 voltages between 150 and 300 volts	—	See curve page 98	

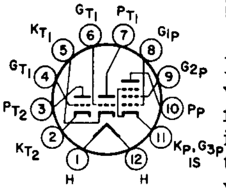
CHARACTERISTICS

Plate Supply Voltage	150	200	volts
Grid-No.2 Supply Voltage	—	125	volts
Grid-No.1 Voltage	—5	—	volts
Cathode-Bias Resistor	—	82	ohms
Amplification Factor	17	—	
Plate Resistance (Approx.)	5150	150000	ohms
Transconductance	3300	7000	μmhos
Plate Current	9.5	15	mA
Grid-No.2 Current	—	3.4	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—14	—8	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

MEDIUM-MU TWIN TRIODE— SHARP-CUTOFF PENTODE 6BH11



12FP

Duodecar type used in color and black-and-white television receiver applications. The triode units are used for general-purpose applications, and the pentode unit is used for horizontal-deflection service. **Outlines section,** 8B; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.8; maximum heater-cathode volts, ±200 peak, 100 average.

Pentode Unit as Horizontal-Deflection Oscillator

MAXIMUM RATINGS (Design-Maximum Values)

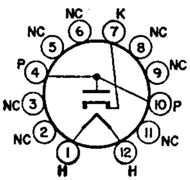
Plate Voltage	350	volts
Grid-No.2 (Screen-Grid) Voltage	330	volts
Grid-No.1 (Control-Grid) Voltage:		
Positive-bias value	0	volts
Peak negative value	175	volts
Peak Cathode Current	300	mA
Average Cathode Current	20	mA
Plate Dissipation	2.5	watts
Grid-No.2 Input	0.55	watt

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Each Triode Unit	
Plate Voltage	330	volts
Grid Voltage, Positive-bias Value	0	volts
Plate Dissipation	2.5	watts

CHARACTERISTICS	Pentode Unit	Each Triode Unit	
Plate Voltage	125	125	volts
Grid-No.2 Voltage	125	—	volts
Grid-No.1 Voltage	—1	—1	volt
Amplification Factor	—	46	
Plate Resistance (Approx.)	200000	5400	ohms
Transconductance	7500	8500	μmhos
Plate Current	12	13.5	mA
Grid-No.2 Current	4	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—8	—8	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	2.2	2.2	megohms
For cathode-bias operation	2.2	2.2	megohms

HALF-WAVE VACUUM RECTIFIER 6BJ3



12BL

Duodecar type used as damper tube in horizontal-deflection circuits of black-and-white television receivers. **Outlines section,** 8C; requires duodecar 12-contact socket. Socket terminals 5, 6, 8, and 9 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated. **Heater:** volts (ac/dc), 6.3; amperes, 1.2.

Damper Service

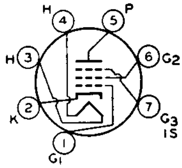
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		3300	volts
Peak Plate Current		840	mA
Average Plate Current		140	mA
Plate Dissipation		4	watts
Heater-Cathode Voltage:			
Peak value	+300	—3300	volts
Average value	+100	—600	volts
CHARACTERISTICS, Instantaneous Value			
Tube Voltage Drop for plate current of 250 mA		21	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6BJ6
6BJ6A REMOTE-CUTOFF PENTODE



7CM

Miniature type used as rf amplifier in high-frequency and wide-band applications. Features high transconductance and low grid-to-plate capacitance. **Outlines section,** 5C; requires miniature 7-contact socket. Type 6BJ6A is identical with type 6BJ6 except for test controlled interference impedance.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.15	ampere
Peak Heater-Cathode Voltage	±90 max	volts
Direct Interelectrode Capacitances:▪		
Grid No.1 to Plate	0.0035 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.5	pF

▪ Without external shield, or with external shield connected to cathode.

Class A_1 Amplifier

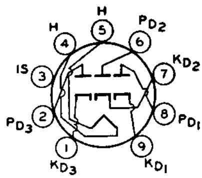
MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Grid-No.2 (Screen-Grid) Voltage	See curve page 98	
Grid-No.2 Supply Voltage	300	volts
Plate Dissipation	3	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	0.6	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts

CHARACTERISTICS

Plate Voltage	100	250	volts
Grid No.3	Connected to cathode at socket		
Grid-No.2 Voltage	100	100	volts
Grid-No.1 Voltage	—1	—1	volt
Plate Resistance (Approx.)	0.25	1.3	megohms
Transconductance	3650	3600	µmhos
Plate Current	9	9.2	mA
Grid-No.2 Current	3.5	3.3	mA
Grid-No.1 Voltage (Approx.) for transconductance of 10 µmhos	—20	—20	volts

6BJ7 TRIPLE DIODE



9AX

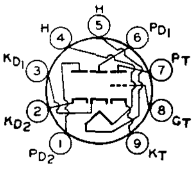
Miniature type used as a dc-restorer tube in each of the three signal channels of color-television receivers. Each diode has a separate cathode. **Outlines section,** 6B; requires miniature 9-contact socket.

Heater Voltage	6.3	volts
Heater Current	0.45	ampere
Direct Interelectrode Capacitances:		
Plate of Unit No.1 to Cathode of Unit No.1, Heater, and Internal Shield	3	pF
Plate of Unit No.2 to Cathode of Unit No.2, Heater, and Internal Shield	2.6	pF
Plate of Unit No.3 to Cathode of Unit No.3, Heater ,and Internal Shield	2.6	pF
Cathode of Unit No.1 to Plate of Unit No.1, Heater, and Internal Shield	4	pF
Cathode of Unit No.2 to Plate of Unit No.2, Heater, and Internal Shield	3.8	pF
Cathode of Unit No.3 to Plate of Unit No.3, Heater, and Internal Shield	4	pF
Plate of Unit No.1 to Plate of Unit No.2	0.055	pF
Plate of Unit No.2 to Plate of Unit No.3	0.036	pF
Plate of Unit No.3 to Plate of Unit No.1	0.036	pF

DC Restorer Service

MAXIMUM RATINGS (Design-Center Values, Each Unit)

Peak Inverse Plate Voltage		330	volts
Peak Plate Current		10	mA
Average Output Current		1	mA
Peak Heater-Cathode Voltage	+100	—330	volts



9ER

TWIN DIODE— MEDIUM-MU TRIODE 6BJ8

Miniature type used in black-and-white and color television receiver applications. The diode units are used in phase-detector, phase-comparator, ratio-detector or discriminator, and horizontal afc discriminator circuits. The triode unit is used in phase-splitter, audio-frequency amplifier, vertical-deflection amplifier, and low-frequency oscillator applications. **Outlines section,** 6E; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.6	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	2.6	pF
Grid to Cathode and Heater	2.8	pF
Plate to Cathode and Heater	0.31	pF
Diode Units:		
Plate to Cathode and Heater (Each Unit)	1.9	pF
Cathode to Plate and Heater (Each Unit)	4.6	pF
Plate of Unit No.1 to Plate of Unit No.2	0.06 max	pF
Plate of Diode Unit No.1 to Triode Grid	0.07 max	pF
Plate of Diode Unit No.2 to Triode Grid	0.11 max	pF
Plate of Either Diode Unit to All Other Electrodes	3	pF
Cathode of Either Diode Unit to All Other Electrodes	4.8	pF

Triode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Average Cathode Current	22	mA
Plate Dissipation	4	watts

CHARACTERISTICS

Plate Voltage	90	250	volts
Grid Voltage	0	—9	volts
Amplification Factor	22	20	
Plate Resistance (Approx.)	4700	7150	ohms
Transconductance	4700	2800	μmhos
Plate Current	13.5	8	mA
Plate Current for grid voltage of —12.5 volts	—	1.7	mA
Grid Voltage (Approx.) for plate current of 10 μA	—7	—18	volts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance	1	megohm

Triode Unit as Vertical-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Voltage	330	volts
Peak Positive-Pulse Plate Voltage#	1200	volts
Peak Negative-Pulse Grid Voltage	275	volts
Peak Cathode Current	77	mA
Average Cathode Current	22	mA
Plate Dissipation	4	watts

MAXIMUM CIRCUIT VALUE

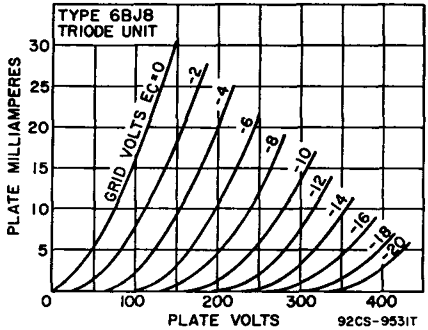
Grid-Circuit Resistance, for cathode-bias operation	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

Diode Units

MAXIMUM RATINGS (Design-Maximum Values)

Plate Current (Each Unit):		
Peak	54	mA
Average	9	mA

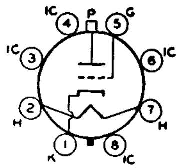


6BK4 Refer to chart at end of section.

6BK4A Refer to chart at end of section.

6BK4B

BEAM TRIODE



Glass octal type used for the voltage regulation of high-voltage, low-current dc power supplies in color and black-and-white television receivers. **Outlines section**, 21B; requires octal socket. Socket terminals 3, 4, 6, and 8 should not be used for tie points. For high voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.2	ampere
Peak Heater-Cathode Voltage	—450* max	volts
Direct Interelectrode Capacitances (Approx.):‡		
Grid to Plate	0.03	pF
Grid to Cathode and Heater	2.6	pF
Plate to Cathode and Heater	1	pF

* Series impedence should be used with the cathode to limit the cathode current under prolonged short-circuit conditions to 450 mA.
‡ Without external shield.

Shunt Voltage-Regulator Service

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Voltage	27000	volts
Unregulated DC Supply Voltage	60000	volts
DC Grid Voltage	—135	volts
Peak Grid Voltage▪	—440	volts
Average Plate Current	1.6	mA
Plate Dissipation	40	watts

TYPICAL OPERATION

Unregulated DC Supply Voltage	36000	volts
Equivalent Resistance of Unregulated Supply	11	megohms
Voltage Divider Values:		
R_1 (5 watts)	220	megohms
R_2 (2 watts)	1	megohm
R_3 (0.5 watt)	0.82	megohm
DC Reference Voltage Supply	200	volts
Equivalent Resistance of Reference Voltage	1000	ohms
Effective Grid-Plate Transconductance	200	µmhos
DC Plate Current for Load Current of 0 mA	1000	µA
DC Plate Current for Load Current of 1 mA	45	µA
Regulated DC Output Voltage for Load Current of 0 mA	25000	volts
Regulated DC Output Voltage for Load Current of 1 mA	24500	volts

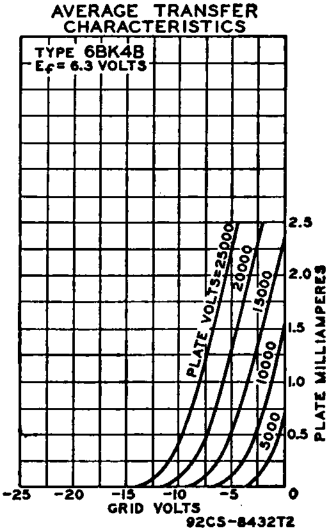
MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance	3	megohms

▪ For interval of 20 seconds maximum duration during equipment warm-up period.

CHARACTERISTICS RANGE VALUES	Note	Min	Max	
Grid Voltage (1)	1	—7	—	volts
Grid Voltage (2)	2	—	—40	volts
Grid-Voltage Change	3	—	9	volts

Note 1: With dc plate voltage of 30000 volts and dc plate current of 1 mA.
Note 2: With dc plate voltage of 30000 volts and dc plate current of 0.1 mA.
Note 3: Difference between grid voltage (1) and grid voltage (2).

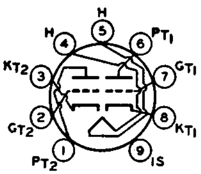


Refer to chart at end of section. **6BK5**

Refer to chart at end of section. **6BK7A**

MEDIUM-MU TWIN TRIODE 6BK7B

5BK7A



9AJ

Miniature type used as a cascode amplifier in vhf color and black-and-white television tuners and in push-pull cathode-drive rf amplifiers. **Outlines section,** 6B; requires miniature 9-contact socket. For typical operation as a resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section. Type 5BK7A is identical with type 6BK7B except for heater ratings.

	5BK7A	6BK7B	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200*max	±200*max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:	Unit No.1	Unit No.2	
Grid to Plate	1.8	1.8	pF
Grid to Cathode, Heater, and Internal Shield	3	3	pF
Plate to Cathode, Heater, and Internal Shield	1	0.9	pF
Cathode to Grid, Heater, and Internal Shield	6	6	pF
Plate to Grid, Heater, and Internal Shield	2.4	2.4	pF

Plate to Cathode	0.22	0.22	pF
Heater to Cathode	2.8	3	pF
Grid of Unit No.1 to Grid of Unit No.2		0.004 max	pF
Plate of Unit No.1 to Plate of Unit No.2		0.075 max	pF

* Rating may be as high as 300 volts under cutoff conditions when tube is used as a cascode amplifier, the units are connected in series, and heater is negative with respect to cathode.

Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Center Value)

Plate Voltage	300	volts
Grid Voltage, Negative-bias value	50	volts
Plate Dissipation	2.7	watts

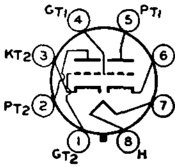
CHARACTERISTICS

Plate Supply Voltage	150	volts
Cathode-Bias Resistor	56	ohms
Amplification Factor	43	
Plate Resistance (Approx.)	4600	ohms
Transconductance	9300	μmhos
Plate Current	18	mA
Grid Voltage (Approx.) for plate current of 10 μA	—11	volts

6BL4 Refer to chart at end of section.

6BL7GT Refer to chart at end of section.

6BL7GTA MEDIUM-MU TWIN TRIODE



8BD

Glass octal type used as combined vertical-deflection amplifier and vertical-deflection oscillator in color and black-and-white television receivers. When so operated, it is recommended that unit No.1 (pins 4, 5, and 6) be used as the oscillator. **Outlines section, 13D;** requires octal socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	1.5	amperes
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Direct Interelectrode Capacitances (Approx.):	Unit No. 1	Unit No. 2	
Grid to Plate	6	6	pF
Grid to Cathode and Heater	4.2	4.6	pF
Plate to Cathode and Heater	0.9	0.9	pF

Class A_1 Amplifier

CHARACTERISTICS (Each Unit)

Plate Voltage	150	250	250	volts
Grid Voltage	0	—17	—9	volts
Amplification Factor	—	—	15	
Plate Resistance (Approx.)	—	—	2150	ohms
Transconductance	—	—	7000	μmhos
Plate Current	65■	4	40	mA
Grid Voltage (Approx.) for plate current of 50 μA	—	—	—23	volts

■ This value can be measured by a method involving a recurrent waveform such that *the maximum ratings of the tube will not be exceeded.*

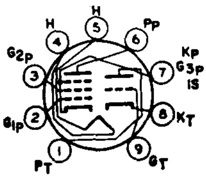
Vertical-Deflection Oscillator or Amplifier•

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)	Oscillator	Amplifier	
DC Plate Voltage	500	500	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	—	2000△	volts
Peak Negative-Pulse Grid Voltage	400	250	volts
Peak Cathode Current	210	210	mA
Average Cathode Current	60	60	mA
Plate Dissipation:			
For either plate	10	10	watts
For both plates with both units operating	12	12	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance	4.7	4.7†	megohms

• Unless otherwise specified, values are for each unit.
Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).
▲ Under no circumstances should this absolute value be exceeded.
† *For cathode-bias operation.*

Refer to chart at end of section. **6BL8**



9DC

MEDIUM-MU TRIODE—SHARP-CUTOFF PENTODE

6BL8/ECF80

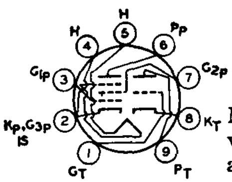
4BL8/XCF80

Miniature type used in frequency-changer service in color and black-and-white television receivers. **Outlines section, 6B**; require miniature 9-contact socket. Type 4BL8/XCF80 is identical with type 6BL8/ECF80 except for heater ratings.

	4BL8/ XCF80	6BL8/ ECF80	
Heater Voltage (ac/dc)	4.6	6.3	volts
Heater Current	0.6	0.45	ampere
Peak Heater-Cathode Voltage	±100 max	±100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	550	volts
Plate Voltage	250	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	550	volts
Grid-No.2 Voltage:			
With cathode current of 14 mA	—	175	volts
With cathode current less than 10 mA	—	200	volts
Cathode Current	14	14	mA
Plate Dissipation	1.5	1.7	watts
Grid-No.2 Input:			
With plate dissipation greater than 1.2 watts	—	0.5	watt
With plate dissipation less than 1.2 watts	—	0.75	watt
CHARACTERISTICS			
Plate Voltage	100	170	volts
Grid-No.2 Voltage	—	170	volts
Grid-No.1 Voltage	—2	—2	volts
Amplification Factor	20	—	
Mu-Factor, Grid No.2 to Grid No.1	—	47	
Plate Resistance (Approx.)	—	0.4	megohm
Transconductance	5000	6200	μmhos
Plate Current	14	10	mA
Grid-No.2 Current	—	2.8	mA
Input Resistance at frequency of 50 MHz	—	0.01	megohm
Equivalent Noise Resistance	—	1500	ohms
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.5	megohm
For cathode-bias operation	0.5	1	megohm



9EX

HIGH-MU TRIODE—POWER PENTODE

6BM8/ECL82

Miniature type used in color and black-and-white television receiver applications. The pentode unit is used as an audio output tube, and the triode unit as an oscillator and af voltage amplifier. **Outlines section, 6G**; requires miniature 9-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.78; *maximum* heater-cathode volts, 100 peak.

Class A₁ Amplifier

MAXIMUM RATINGS (Design-Center Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	900	volts
Plate Voltage	300	600	volts
Grid-No.2 Supply Voltage	—	550	volts
Grid-No.2 Voltage	—	300	volts
Cathode Current	15	50	mA
Plate Dissipation	1	7	watts
Grid-No.2 Input	—	1.8	watts
CHARACTERISTICS			
Plate Voltage	100	200	volts
Grid-No.2 Voltage	—	200	volts
Grid-No.1 Voltage	0	—16	volts
Amplification Factor	70	9.5*	
Plate Resistance (Approx.)	—	0.02	megohm
Transconductance	2500	6400	μmhos
Plate Current	3.5	35	mA
Grid-No.2 Current	—	7	mA
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	1	1	megohm
For cathode-bias operation	2	2	megohms

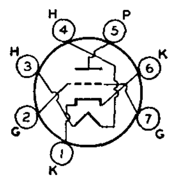
* Grid No.2 to Grid No.1

6BN4 Refer to chart at end of section.

6BN4A

2BN4A, 3BN4A **MEDIUM-MU TRIODE**

Miniature type used as rf amplifier tube in grid-drive circuits of vhf color and black-and-white television tuners. Outlines section, 5C; requires miniature 7-contact socket. Types 2BN4A and 3BN4A are identical with type 6BN4A except for heater ratings.



7EG

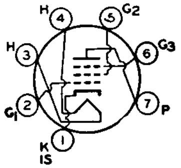
	2BN4A	3BN4A	6BN4A	
Heat Voltage (ac/dc)	2.35	3	6.3	volts
Heater Current	0.6	0.45	0.2	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	±100 max	volts

Direct Interelectrode Capacitances (Approx.):*		
Grid to Plate	1.2	pF
Grid to Cathode and Heater	3.2	pF
Plate to Cathode and Heater	1.4	pF

* With external shield connected to cathode.

Class A₁ Amplifier

MAXIMUM RATINGS (Design-Center Values)		
Plate Voltage	275	volts
Grid Voltage, Positive-bias value	0	volts
Cathode Current	22	mA
Plate Dissipaation	2.2	watts
CHARACTERISTICS		
Plate-Supply Voltage	150	volts
Cathode-Bias Resistor	220	ohms
Amplification Factor	43	
Plate Resistance (Approx.)	5400	ohms
Transconductance	7700	μmhos
Plate Current	9	mA
Grid Voltage (Approx.) for plate current of 100 μA	—6	volts
MAXIMUM CIRCUIT VALUE		
Grid-Circuit Resistance	0.5	megohm


7DF

6BN6

3BN6, 4BN6
12BN6

BEAM TUBE

Miniature type used as combined limiter, discriminator, and audio-voltage amplifier in intercarrier television and FM receivers. **Outlines section,** 5D; requires miniature 7-contact socket. Types 3BN6, 4BN6, and 12BN6 are identical with type 6BN6 except for heater ratings.

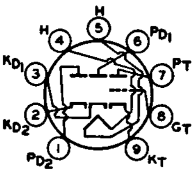
	3BN6	4BN6	6BN6	12BN6	
Heater Voltage (ac/dc)	3.15	4.2	6.3	12.6	volts
Heater Current	0.6	0.45	0.3	0.15	ampere
Heater Warm-up Time (Average)	11	11	—	—	seconds
Heater-Cathode Voltage:					
Peak value	±200 max	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4.2	pF
Grid No.3 to Cathode, Heater, Grid No.1, Grid No.2, and Internal Shield	3.3	pF
Grid No.1 to Grid No.3	0.004 max	pF

Limiter and Discriminator Service

MAXIMUM RATINGS (Design-Maximum Values)

Plate-Supply Voltage	330	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 Voltage, Positive peak value	60	volts
Cathode Current	13	mA


9ER

6BN8

8BN8

TWIN DIODE—HIGH-MU TWIN TRIODE

Miniature type used in color and black-and-white television receiver applications. The triode unit is used in burst-amplifier, af amplifier, and low-frequency oscillator applications. The diode units are used in phase-detector, ratio-detector or discriminator, and horizontal afc discriminator circuits. **Outlines section,** 6E; requires miniature 9-contact socket. Type 8BN8 is identical with type 6BN8 except for heater ratings.

	6BN8	8BN8	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Triode Grid to Triode Plate	2.5	pF
Triode Grid to Cathode and Heater	3.6	pF
Triode Plate to Cathode and Heater	0.25	pF
Plate of Diode Unit No.1 to Triode Grid	0.06 max	pF
Plate of Diode Unit No.2 to Triode Grid	0.1 max	pF
Plate of Diode Unit No.1 to Plate of Diode Unit No.2	0.07 max	pF
Diode Cathode to All Other Electrodes (Each Diode Unit)	5	pF
Diode Plate to Diode Cathode and Heater (Each Diode Unit)	1.9	pF
Diode Cathode to Diode Plate and Heater (Each Diode Unit)	4.8	pF
Diode Plate to All Other Electrodes (Each Diode Unit)	3	pF

Triode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation	1.7	watts

CHARACTERISTICS

Plate Voltage	100	250	volts
Grid Voltage	—1	—3	volts

Amplification Factor	75	70	
Plate Resistance (Approx.)	21000	28000	ohms
Transconductance	3500	2500	µmhos
Plate Current	1.5	1.6	mA
Grid Voltage (Approx.) for plate current of 10 µA	—2.5	—5.5	volts

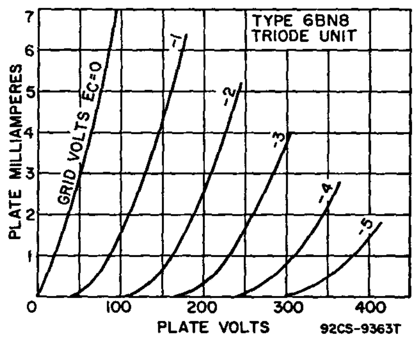
MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance	1	megohm

Diode Units

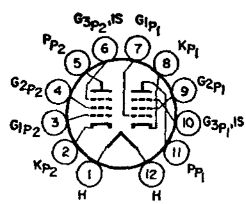
MAXIMUM RATINGS (Design-Maximum Values)

Plate Current (Each Unit):		
Peak	54	mA
Average	9	mA



6BN11

SHARP-CUTOFF TWIN PENTODE



Duodecar type used as if-amplifier tube in television receivers. **Outlines** section, 8B; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.8; maximum heater-cathode volts, ±200 peak, 100 average.

Class A₁ Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

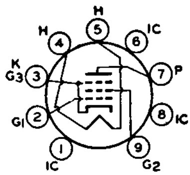
Plate Voltage	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	*0*	volts
Plate Dissipation	3.1	watts
Grid-No.2 Input	0.65	watt

CHARACTERISTICS

Plate Voltage	125	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket	
Grid-No.2 Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.2	megohm
Transconductance	13000	µmhos
Plate Current	11	mA
Grid-No.2 Current	3.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—3	volts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, *for cathode-bias operation*	0.25	megohm



9CV

6BQ5
6BQ5/EL84
8BQ5, 10BQ5

POWER PENTODE

Miniature type used in the output stage of audio-frequency amplifiers. Outlines section, 6G; requires miniature 9-contact socket. Types 8BQ5 and 10BQ5 are identical with type 6BQ5 except for heater ratings.

	6BQ5 6BQ5/EL84	8BQ5	10BQ5	
Heater Voltage (ac/dc)	6.3	8	10.6	volts
Heater Current	0.76	0.6	0.45	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±100 max	±100 max	±100 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.5 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	10.8	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	6.5	pF
Grid No.1 to Heater	0.25 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Grid-No.2 (Screen-Grid) Voltage	300	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Cathode Current	65	mA
Plate Dissipation	12	watts
Grid No.2 Input	2	watts

TYPICAL OPERATION

Plate Voltage	250	volts
Grid-No.2 Voltage	250	volts
Grid-No.1 (Control-Grid) Voltage	—7.3	volts
Peak AF Grid No.1 Voltage	6.2	volts
Zero-Signal Plate Current	48	mA
Maximum-Signal Plate Current	50.6	mA
Zero-Signal Grid-No.2 Current	5.5	mA
Maximum-Signal Grid-No.2 Current	10	mA
Plate Resistance (Approx.)	38000	ohms
Transconductance	11300	μmhos
Load Resistance	4500	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	5.7	watts

MAXIMUM CIRCUIT VALUES

Grid-No.-Circuit Resistance:		
For fixed-bias operation	0.3	megohm
For cathode-bias operation	1	megohm

Push-Pull Class AB_1 Amplifier

MAXIMUM RATINGS (Same as for Single-Tube Class A_1 Amplifier)

TYPICAL OPERATION (Values are for two tubes)

Plate Supply Voltage	250	300	volts
Grid-No.2 Supply Voltage	250	300	volts
Cathode-Bias Resistor	130	130	ohms
Peak AF Grid-No.1-to-Grid-No.1 Voltage	22.6	28.3	volts
Zero-Signal Plate Current	62	72	mA
Maximum-Signal Plate Current	75	92	mA
Zero-Signal Grid-No.2 Current	7	8	mA
Maximum-Signal Grid-No.2 Current	15	22	mA
Effective Load Resistance (Plate-to-plate)	8000	8000	ohms
Total Harmonic Distortion	3	4	per cent
Maximum-Signal Power Output	11	17	watts

MAXIMUM CIRCUIT VALUES

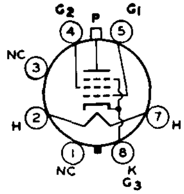
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.3	megohm
For cathode-bias operation	1	megohm

Refer to chart at end of section.

6BQ6GTB /6CU6

BEAM POWER TUBE

12BQ6GTB/12CU6, 17BQ6GTB, 25BQ6GTB/25CU6



6AM

Glass octal type used as horizontal-deflection amplifier in color and black-and-white television receivers. Outlines section, 14D; requires octal socket. This type may be supplied with pin No.1 omitted. Types 12BQ6GTB/12CU6, 17BQ6GTB, and 25BQ6GTB/25CU6 are identical with type 6BQ6GTB/6CU6 except for heater ratings.

	6BQ6GTB/ 6CU6	12BQ6G-TB/12CU6	17BQ6-GTB	25BQ6GTB/ 25CU6	
Heater Voltage (ac/dc)	6.3	12.6	16.8	25	volts
Heater Current	1.2	0.6	0.45	0.3	ampere
Heater Warm-up Time (Average)	—	11	11	—	seconds
Heater-Cathode Voltage:					
Peak value	±200 max	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.6	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	15	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	7	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	60	150	250	volts
Grid-No.2 Voltage	150	150	150	volts
Grid-No.1 Voltage	0	—22.5	—22.5	volts
Mu-Factor, Grid No.2 to Grid No.1	—	4.3	—	
Plate Resistance (Approx.)	—	—	14500	ohms
Transconductance	—	—	5900	µmhos
Plate Current	260•	—	57	mA
Grid-No.2 Current	26•	—	2.1	mA
Grid-No.1 Voltage (Approx.) for plate mA = 1	—	—	—43	volts

• These values can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)

DC Plate Voltage	600	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	6000†	volts
Peak Negative-Pulse Plate Voltage	1250	volts
DC Grid-No.2 (Screen-Grid) Voltage	200	volts
Peak Negative-Pulse Grid-No.1 (Control-Grid) Voltage	300	volts
Peak Cathode Current	400	mA
Average Cathode Current	110	mA
Plate Dissipation▪	11	watts
Grid-No.2 Input	2.5	watts
Bulb Temperature (At hottest point)	220	°C

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	0.47	megohm

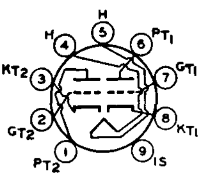
Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
† Under no circumstances should this absolute value be exceeded.
▪ A bias resistor or other means is required to protect the tube in absence of excitation.

6BQ7 Refer to chart at end of section.

6BQ7A

4BQ7A, 5BQ7A

MEDIUM-MU TWIN TRIODE



9AJ

Miniature type used as a cascode amplifier in vhf color and black-and-white television tuners in push-pull cathode-drive rf amplifiers. Outlines section, 6B; requires miniature 9-contact socket. For typical operation as a resistance-coupled amplifier, refer to Resistance-Coupled Amplifier section. Types 4BQ7A and 5BQ7A are identical with type 6BQ7A except for heater ratings.

	4BQ7A	5BQ7A	6BQ7A	
Heater Voltage (ac/dc)	4.2	5.6	6.3	volts
Heater Current	0.6	0.45	0.4	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200*max	±200*max	±200*max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:°	Unit No.1	Unit No.2	
Grid to Plate	1.2	1.2	pF
Grid to Cathode, Heater, and Internal Shield	2.6	—	pF
Cathode to Grid, Heater, and Internal Shield	—	5	pF
Plate to Cathode, Heater, and Internal Shield	1.2	—	pF
Plate to Grid, Heater, and Internal Shield	—	2.2	pF
Plate to Cathode	0.12	0.12	pF
Heater to Cathode	2.6	2.6	pF
Plate of Unit No.1 to Plate of Unit No.2	0.010 max		pF
Plate of Unit No.2 to Plate and Grid of Unit No.1	0.024 max		pF

* Rating may be high as 300 volts under cutoff conditions, when tube is used as a cascode amplifier, the two units are connected in series, and heater is negative with respect to cathode.

° With external shield connected to internal shield.

Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Center Values)

Plate Supply Voltage	250*	volts
Cathode Current	20	mA
Plate Dissipation	2	watts

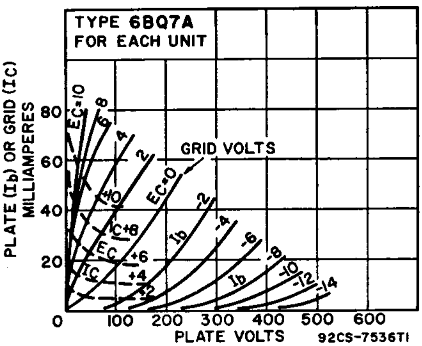
CHARACTERISTICS

Plate Supply Voltage	150	volts
Cathode-Bias Resistor	220	ohms
Amplification Factor	38	
Plate Resistance (Approx.)	5900	ohms
Transconductance	6400	μmhos
Plate Current	9	mA
Grid Voltage (Approx.):		
For plate current of 100 μA	—6.5	volts
For plate current of 10 μA	—	volts

MAXIMUM CIRCUIT VALUE

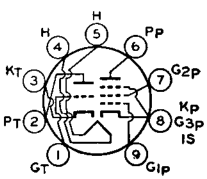
Grid-Circuit Resistance	0.5	megohm

* Rating may be high as 300 volts under cutoff conditions, when tube is used as a cascode amplifier, the two units are connected in series, and heater is negative with respect to cathode.



Refer to chart at end of section. **6BR8**

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE **6BR8A** 5BR8



9FA

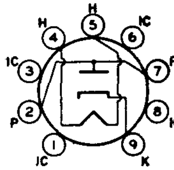
Miniature type used in color and black-and-white television receiver applications. Especially useful as combined triode oscillator and pentode mixer in vhf television tuners. Outlines section, 6B; requires miniature 9-contact socket. Except for basing arrangement and grid-No.1-to-plate capacitance of pentode unit, types 5BR8 and 6BR8A are identical with types 5U8 and 6U8A, respectively.

6BS3 Refer to chart at end of section.

6BS3A

12BS3A, 17BS3A

HALF-WAVE VACUUM RECTIFIER



9HP

Novar type used as damper tube in horizontal-deflection circuits of black-and-white television receivers. Outlines section, 30B; requires novar 9-contact socket. Socket terminals 1, 3, 6, and 8 should not be used as tie points; it is recommended that socket clips for these pins be removed to reduce the possibility of arc-over and to minimize leakage. These tubes, like other power-handling tubes, should be adequately ventilated. Types 12BS3A and 17BS3A are identical with type 6BS3A except for heater ratings.

	6BS3A	12BS3A	17BS3A	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds

Direct Interelectrode Capacitances (Approx.):		
Plate to Cathode and Heater	6.5	pF
Cathode to Plate and Heater	9	pF
Heater to Cathode	2.8	pF

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		5000	volts
Peak Plate Current		1100	mA
Average Plate Current		200	mA
Plate Dissipation		6	watts
Heater-Cathode Voltage:			
Peak value	+300	—5000	volts
Average value	+100	—900	volts

CHARACTERISTICS, *Instantaneous Value*

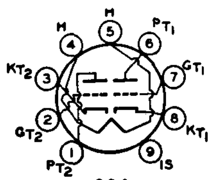
Tube Voltage Drop for plate current of 140 mA	12	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6BS8

4BS8

MEDIUM-MU TWIN TRIODE



9AJ

Miniature type used as a cascode amplifier in vhf color and black-and-white television tuners and in push-pull cathode-drive *rf* amplifiers. Outlines section, 6B; requires miniature 9-contact socket. Type 4BS8 is identical with type 6BS8 except for heater ratings.

	4BS8	6BS8	
Heater Voltage (ac/dc)	4.5	6.3	volts
Heater Current	0.6	0.4	ampere
Heater Warm-up Time (Average)	11	—	seconds
Peak Heater-Cathode Voltage	±200 max	±200 max	volts

Direct Interelectrode Capacitances:		
Grid to Plate (Each Unit)	1.15	pF
Grid to Cathode, Heater, and Internal Shield (Unit No.1)	2.6	pF
Plate to Cathode, Heater, and Internal Shield (Unit No.1)	1.2	pF
Plate to Cathode (Each Unit)	0.15 max	pF

Heater to Cathode (Each Unit)	2.6	pF
Cathode to Grid, Heater, and Internal Shield (Unit No.2)	5	pF
Plate to Grid, Heater, and Internal Shield (Unit No.2)	2.2	pF
Plate of Unit No.1 to Plate of Unit No.2	0.010 max	pF
Plate of Unit No.2 to Plate and Grid of Unit No.1	0.024 max	pF

Class A_1 Amplifier (Each Unit)

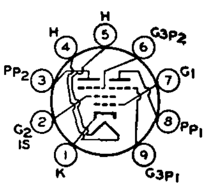
MAXIMUM RATINGS (Design-Center Values)		
Plate Voltage	150	volts
Cathode Current	20	mA
Plate Dissipation	2	watts
CHARACTERISTICS		
Plate-Supply Voltage	150	volts
Cathode-Bias Resistor	220	ohms
Amplification Factor	36	
Plate Resistance (Approx.)	5000	ohms
Transconductance	7200	μmhos
Plate Current	10	mA
Grid Voltage (Approx.) for plate current of 10 μA*	—7	volts
MAXIMUM CIRCUIT VALUE		
Grid-Circuit Resistance	0.5	megohm

* This value applies to Unit No.2 only.

6BU8

SHARP-CUTOFF TWIN PENTODE

3BU8/3GS8
4BU8/4GS8



9FG

Miniature type used as combined sync separator, sync clipper, and agc amplifier tube in color and black-and-white television receivers. Outlines section, 6E; requires miniature 9-contact socket. Types 3BU8/3GS8 and 4BU8/4GS8 are identical with type 6BU8 except for heater ratings.

	3BU8/3GS8	4BU8/4GS8	6BU8	
Heater Voltage (ac/dc)	3.15	4.2	6.3	volts
Heater Current	0.6	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts
Direct Interelectrode Capacitances:				
Grid No.3 to Plate (Each Unit)			1.9	pF
Grid No.1 to All Other Electrodes			6	pF
Grid No.3 to All Other Electrodes (Each Unit)			3.6	pF
Plate to All Other Electrodes (Each Unit)			3	pF
Grid No.3 of Unit No.1 to Grid No.3 of Unit No.2			0.015 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage (Each Unit)	300	volts
Grid-No.3 (Suppressor-Grid) Voltage (Each Unit):		
Peak positive value	50	volts
DC negative value	50	volts
DC positive value	3	volts
Grid-No.2 (Screen-Grid) Voltage	150	volts
Grid-No.1 (Control-Grid) Voltage, Negative bias value	50	volts
Cathode Current	12	mA
Plate Dissipation (Each Unit)	1.1	watts
Grid-No.2 Input	0.75	watt

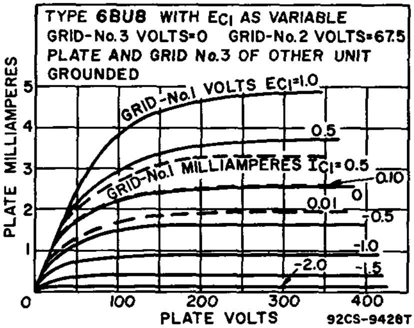
CHARACTERISTICS (With Both Units Operating)			
Plate Voltage (Each Unit)	100	100	volts
Grid-No.3 Voltage (Each Unit)	—10	0	volts
Grid-No.2 Voltage	67.5	67.5	volts
Grid-No.1 Voltage	*	*	volts
Plate Current (Each Unit)	—	2.2	mA
Grid-No.2 Current	6.5	3.3	mA
Cathode Current	6.6	7.8	mA
CHARACTERISTICS (With One Unit Operating)			
Plate Voltage	100	100	volts
Grid-No.3 Voltage	0	0	volts
Grid-No.2 Voltage	67.5	67.5	volts
Grid-No.1 Voltage	0	*	volts

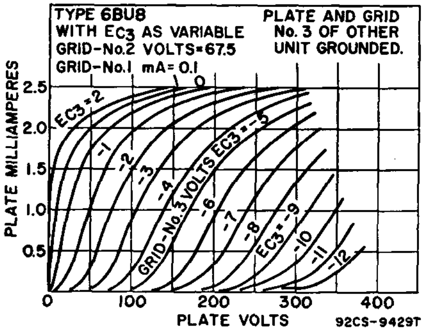
Grid-No.3 Transconductance	—	180	µmhos
Grid-No.1 Transconductance	1500	—	µmhos
Plate Current	—	2.2	mA
Grid-No.3 Voltage (Approx.) for plate current of 100 µA	—	—4.5	volts
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—	—2.3	volts

MAXIMUM CIRCUIT VALUES

Grid-No.3-Circuit Resistance (Each Unit)	0.5	megohm
Grid-No.1-Circuit Resistance	0.5	megohm

* Adjusted to provide a dc grid-No.1 current of 100 microamperes.
† With plate and grid No.3 of the other unit connected to ground.



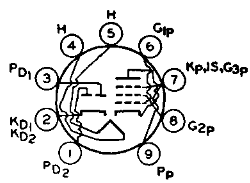


6BV8 Refer to chart at end of section.

6BW4 Refer to chart at end of section.

6BW8
5BW8

TWIN DIODE— SHARP-CUTOFF PENTODE



9HK

Miniature type used in television receiver applications; diodes are used as horizontal phase detectors; pentode is used as a sound if amplifier, sound limiter, and agc keyer. **Outlines section, 6B**; requires miniature 9-contact socket. Type 5BW8 is identical with type 6BW8 except for heater ratings.

	5BW8	6BW8	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:

Pentode Unit:		
Grid No.1 to Plate	0.02 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4.8	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.6	pF
Plate of Diode Unit No.1 to Cathode and Heater	1.3	pF
Plate of Diode Unit No.2 to Cathode and Heater	1.2	pF
Pentode Grid No.1 to Either Diode Plate	0.006 max	pF

Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	

Grid-No.1 (Control-Grid) Voltage:		
Positive-bias value	0	volts
Negative-bias value	55	volts
Plate Dissipation		
Grid-No.2 Input:	0.55	watt
For grid-No.2 voltages up to 165 volts	See curve page 98	
For grid-No.2 voltages between 165 and 330 volts	3	watts

CHARACTERISTICS

Plate Voltage	250	volts
Grid-No.2 Voltage	110	volts
Cathode-Bias Resistor	68	ohms
Plate Resistance (Approx.)	0.25	megohm
Transconductance	5200	μmhos
Plate Current	10	mA
Grid-No.2 Current	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—10	volts

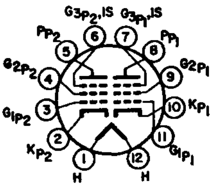
MAXIMUM CIRCUIT VALUES

Grid-No.1 Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

Diode Units (Each Unit)

MAXIMUM RATING (Design-Maximum Value)

Plate Current	5	mA



12HD

SHARP-CUTOFF DUAL PENTODE — 6BW11

Duodecar type used in color and black-and-white television receiver applications. Unit No. 1 is used as a video amplifier; unit No. 2 is used in bandpass amplifier, burst amplifier, or sound-if or video-if applications. Outlines section, 8B; requires duodecar 12-contact socket. Heater: volts (ac/dc), 6.3; amperes, 0.8; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1	Unit No.2	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	330	volts
Grid-No.2 Voltage	See curve page 98		
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	4	3.1	watts
Grid-No.2 Dissipation	0.8	0.65	watt

CHARACTERISTICS

Plate Voltage	125	125	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket		
Grid-No.2 Voltage	125	125	volts
Cathode-Bias Resistor	56	56	ohms
Plate Resistance (Approx.)	0.12	0.2	megohm
Transconductance	8500	13000	μmhos
Plate Current	22	11	mA
Grid-No.2 Current	4.8	3.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—9.5	—3	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:			
For cathode-bias operation	0.25	0.25	megohm

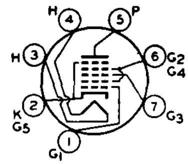
Refer to chart at end of section. **6BX7GT**

Refer to chart at end of section. **6BY5GA**

6BY6

3BY6

PENTAGRID AMPLIFIER



7CH

Miniature type used as a gated amplifier in color television receivers. In such service, it may be used as a combined sync separator and sync clipper. **Outlines section, 5C**; requires miniature 7-contact socket. Type 3BY6 is identical with type 6BY6 except for heater ratings.

	3BY6	6BY6	
Heater Voltage (ac/dc)	3.15	6.3	volts
Heater Current	0.6	0.3	ampere
Heater Warm-up Time (Average)	11	—	seconds
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.08 max	pF
Grid No.3 to Plate	0.35 max	pF
Grid No.1 to Grid No.3	0.22 max	pF
Grid No. 1 to All Other Electrodes	5.4	pF
Grid No.3 to All Other Electrodes	6.9	pF
Plate to All Other Electrodes	7.6	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	250	volts
Grids-No.2-and-No.4 Voltage	100	volts
Grid-No.3 Voltage	—2.5	volts
Grid-No.1 Voltage	—2.5	volts
Grid-No.3-to-Plate Transconductance	500	µmhos
Grid-No.1-to-Plate Transconductance	1900	µmhos
Plate Current	6.5	mA
Grids-No.2-and-No.4 Current	9	mA
Grid-No.3 Volts (Approx.) for plate current of 35 µA and grid-No.1 volts = —4	—15	volts
Grid-No.1 Volts (Approx.) for plate current of 35 µA and grid-No.3 volts = 0	—12	volts

Gated Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

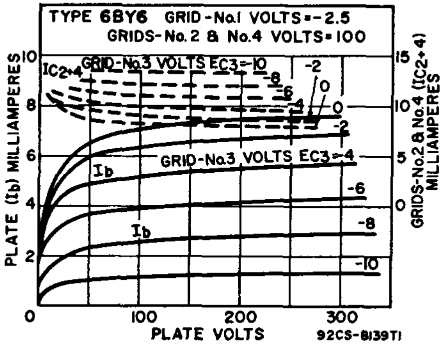
Plate Voltage	330	volts
Grids-No.2-and-No.4 Voltage	See curve page 98	
Grids-No.2-and-No.4 Supply Voltage	330	volts
Grid-No.3 Voltage:		
Negative-bias value	55	volts
Positive-bias value	0	volts
Positive peak value	27	volts
Grid-No.1 Voltage, Negative bias value	110	volts
Plate Dissipation	2.3	watts
Grid-No.3 Input	0.1	watt
Grids-No.2-and-No.4 Input:		
For grids-No.2-and-No.4 voltages up to 165 volts	1.1	watts
For grids-No.2-and-No.4 voltages between 165 and 330 volts	See curve page 98	
Grid-No.1 Input	0.1	watt

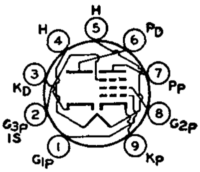
CHARACTERISTICS AS SYNC SEPARATOR AND SYNC CLIPPER

Plate Voltage	10	volts
Grid-No.3 Voltage	0	volts
Grids-No.2-and-No.4 Voltage	25	volts
Grid-No.1 Voltage	0	volts
Plate Current	1.4	mA
Grids-No.2-and-No.4 Current	3.5	mA
Grid-No.3 Volts (Approx.) for plate voltage of 25 volts, grids-No.2-and-No.4 voltage of 25 volts, grid-No.1 voltage of 0 volts, and plate current of 50 µA	—2.5	volts
Grid-No.1 Volts (Approx.) for plate voltage of 25 volts, grids-No.2-and-No.4 voltage of 25 volts, grid-No.3 voltage of 0 volts, and plate current of 50 µA	—2.3	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1 or Grid-No.3-Circuit Resistance:		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	1	megohm





9FN

DIODE— SHARP-CUTOFF PENTODE 6BY8

Miniature type used in television receiver applications. The pentode unit is used as an rf amplifier and the high-perveance diode as a limiter or detector. Outlines section, 6E; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.6	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:°		
Pentode Unit:		
Grid No.1 to Plate	0.0035 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5	pF
Diode Plate to All Other Electrodes	4.8▪	pF

° *With external shield connected to cathode of pentode unit (pin 9), except as noted.*
▪ *With external shield connected to ground.*

Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate Dissipation	3	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	0.65	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	100	250	volts
Grid No.3	Connected to cathode at socket		
Grid-No.2 Supply Voltage	100	150	volts
Cathode-Bias Resistor	150	68	ohms
Plate Resistance (Approx.)	0.5	1	megohm
Transconductance	3900	5200	μmhos
Plate Current	5	10.6	mA
Grid-No.2 Current	2.1	4.3	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—4.2	—6.5	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

Diode Unit

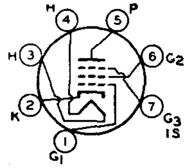
MAXIMUM RATINGS (Design-Center Values)

Peak Inverse Plate Voltage	430	volts
Peak Plate Current	180	mA
Average Plate Current	45	mA

6BZ3 Refer to type 6BE3.

6BZ6 SEMIREMOTE-CUTOFF PENTODE

3BZ6, 4BZ6, 12BZ6



7CM

Miniature type used in gain-controlled video if stages of color and black-and-white television receivers. Outlines section, 5C; requires miniature 7-contact socket. Types 3BZ6, 4BZ6, and 12BZ6 are identical with type 6BZ6 except for heater ratings.

	3BZ6	4BZ6	6BZ6	12BZ6	
Heater Voltage (ac/dc)	3.15	4.2	6.3	12.6	volts
Heater Current	0.6	0.45	0.3	0.15	ampere
Heater Warm-up Time (Average)	11	11	—	—	seconds
Heater-Cathode Voltage:					
Peak value	±200 max	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:	Unshielded	Shielded	
Grid No.1 to Plate	0.025 max	0.015 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	7	7	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2	3	pF

▵ With external shield connected to cathode.

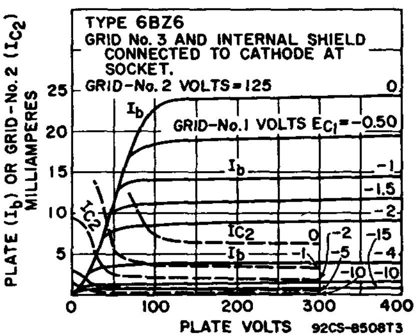
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	2.3	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	125	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.26	megohm
Transconductance	8000	μmhos
Plate Current	14	mA
Grid-No.2 Current	3.6	mA
Grid-No.1 Voltage (Approx.) for transconductance of 50 μmhos	—19	volts
Grid-No.1 Voltage (Approx.) for transconductance of 700 μmhos	—4.5	volts



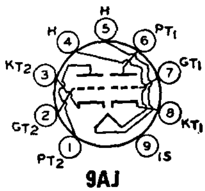
MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

6BZ7

MEDIUM-MU TWIN TRIODE

4BZ7



9AJ

Miniature type used as a cascode amplifier in vhf color and black-and-white television tuners and in push-pull cathode-drive rf amplifiers. Outlines section, 6B; requires miniature 9-contact socket. For typical operation as a resistance-coupled amplifier, refer to Resistance-Coupled Amplifier section. Type 4BZ7 is identical with type 6BZ7 except for heater ratings.

	4BZ7	6BZ7	
Heater Voltage (ac/dc)	4.2	6.3	volts
Heater Current	0.6	0.4	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200*max	±200*max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Grid to Plate (Each Unit)	1.2	pF
Grid to Cathode, Heater, and Internal Shield (Unit No.1)	2.6	pF
Plate to Cathode, Heater, and Internal Shield (Unit No.1)	1.2	pF
Plate to Cathode (Each Unit)	0.12	pF
Heater to Cathode (Each Unit)	2.6	pF
Cathode to Grid, Heater, and Internal Shield (Unit No.2)	5	pF
Plate to Grid, Heater, and Internal Shield (Unit No.2)	2.2	pF
Plate of Unit No.1 to Plate of Unit No.2	0.010 max	pF
Plate of Unit No.2 to Plate and Grid of Unit No.1	0.024 max	pF

* Rating may be as high as 300 volts under cutoff conditions, when tube is used as a cascode amplifier, the two units are connected in series, and heater is negative with respect to cathode.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

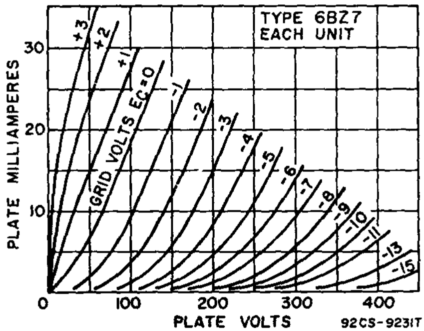
Plate Voltage	250*	volts
Cathode Current	20	mA
Plate Dissipation	2	watts

CHARACTERISTICS

Plate Supply Voltage	150	volts
Cathode-Bias Resistor	220	ohms
Amplification Factor	36	
Plate Resistance (Approx.)	5300	ohms
Transconductance	6800	µmhos
Plate Current	10	mA
Grid Voltage (Approx.) for plate current of 100 µA	—7	volts

MAXIMUM CIRCUIT VALUE

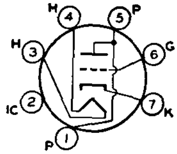
Grid-Circuit Resistance	0.5	megohm



Refer to chart at end of section. 6BZ8

6C4

POWER TRIODE



6BG

Miniature type used as a cascode amplifier in vhf color local oscillator in FM and other high-frequency circuits and as a class C rf amplifier. Outlines section, 5C; requires miniature 7-contact socket. For typical operation as a resistance-coupled amplifier, refer to Resistance-Coupled Amplifier section. For additional curve of plate characteristics, refer to type 12AU7A.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.15	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Direct Interelectrode Capacitances (Approx.)	Unshielded	Shielded▵	
Grid to Plate	1.6	1.4	pF
Grid to Cathode and Heater	1.8	1.8	pF
Plate to Cathode and Heater	1.3	2.5	pF

▵ With external shield connected to cathode.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300 max	volts
Plate Dissipation	3.5 max	watts

CHARACTERISTICS

Plate Voltage	100	250	volts
Grid Voltage*	0	−8.5	volts
Amplification Factor	19.5	17	
Plate Resistance (Approx.)	6250	7700	ohms
Transconductance	3100	2200	μmhos
Plate Current	11.8	10.5	mA
Grid Voltage (Approx.) for plate current of 10 μA	−10	−25	volts

MAXIMUM CIRCUIT VALUES

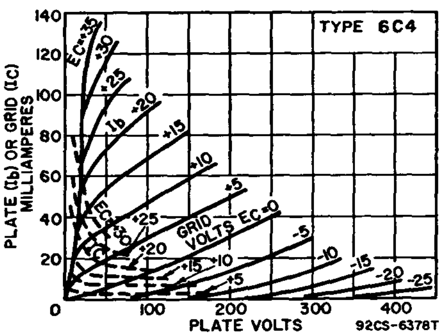
Grid-Circuit Resistance:		
For fixed bias operation	0.25	megohm
For cathode-bias operation	1	megohm

* Transformer- or impedance-type input coupling devices are recommended to minimize resistance in the grid circuit.

RF Power Amplifier and Oscillator—Class C Telegraphy

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Grid Voltage	−50	volts
Plate Current	25	mA
Grid Current	5	watts
Plate Dissipation	8	mA



TYPICAL OPERATION AT FREQUENCIES UP TO 50 MHz

Plate Voltage	300	volts
Grid Voltage	—27	volts
Plate Current	25	mA
Grid Current (Approx.)	7	mA
Driving Power (Approx.)	0.35	watt
Power Output (Approx.)•	5.5	watts

• Approximately 2.5 watts power output can be obtained when the 6C4 is used at 150 MHz as an oscillator with grid resistor of 10,000 ohms and with maximum rated input.

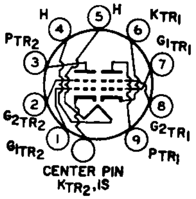
Refer to chart at end of section. **6C5**

Refer to chart at end of section. **6C5GT**

Refer to chart at end of section. **6C6**

Refer to chart at end of section. **6C7**

Refer to chart at end of section. **6C8G**



10F

SHARP-CUTOFF DUAL TETRODE

6C9

17C9

Miniature type used as vhf rf-amplifier and autodyne mixer tube. Outlines section, 6B; except center pin is added to base; requires miniature 10-contact socket. Type 17C9 is identical with type 6C9 except for heater ratings.

	6C9	17C9	
Heater Voltage (ac/dc)	6.3	16.8	volts
Heater Current	0.4	0.15	ampere
Peak Heater-Cathode Voltage	±100 max	±100 max	volts
Direct Interelectrode Capacitances:	Unit No. 1	Unit No. 2	
Grid No.1 to Plate	0.055 max	0.06 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Internal Shield	4.4	4.2	pF
Plate to Cathode, Heater, Grid No.2 and Internal Shield	2.2	2.2	pF
Heater to Cathode	4.2	4.8	pF
Plate of Unit No.1 to Plate of Unit No.2	0.003 max		pF
Grid No.1 of Unit No.1 to Grid No.1 of Unit No.2	0.001 max		pF
Grid No.1 of Unit No.1 to Plate of Unit No.2	0.001 max		pF
Grid No.1 of Unite No.2 to Plate of Unit No.1	0.032 max		pF

Class A_1 Amplifier (Each Unit)

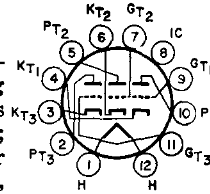
MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	180	volts
Grid-No.2 Voltage	See curve page 98	
Cathode Current	20	mA
Plate Dissipation:		
Either plate	1.5	watts
Both plates (both units operating)	2.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 90 volts	0.5	watt
For grid-No.2 voltages between 90 and 180 volts	See curve page 98	

CHARACTERISTICS

Plate Voltage	125	volts
Grid-No.2 Voltage	80	volts
Grid-No.1 Voltage	—1	volt
Plate Resistance (Approx.)	0.1	megohm
Transconductance	8000	μmhos
Plate Current	10	mA
Grid-No.2 Current	1.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—6	volts

6C10 HIGH-MU TRIPLE TRIODE



12BQ

Duodecar type used in resistance-coupled voltage amplifiers, phase inverters, and other circuits requiring high voltage gain. **Outlines section,** 8A; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.6; average warm-up time (for series heater operation), 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

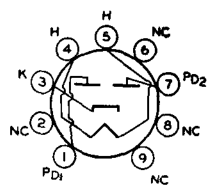
MAXIMUM RATINGS (Design-Maximum Values, Each Triode Unit)

Plate Voltage	330	volts
Grid Voltage:		
Positive-bias value	0	volts
Negative-bias value	50	volts
Plate Dissipation	1	volt
Total Plate Dissipation (All plates)	3	watts

CHARACTERISTICS (Each Triode Unit)

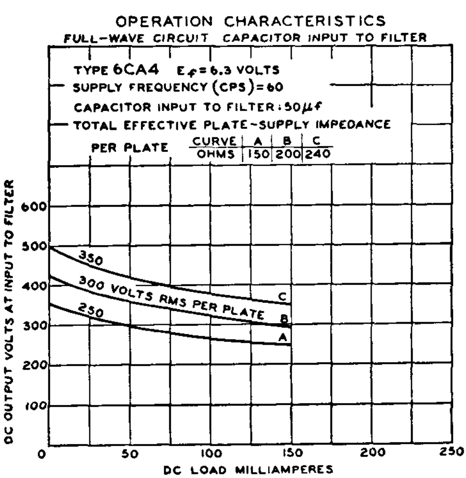
Plate Voltage	100	250	volts
Grid Voltage	—1	—2	volts
Amplification Factor	100	100	
Plate Resistance (Approx.)	80000	62500	ohms
Transconductance	1250	1600	μmhos
Plate Current	0.5	1.2	mA

6CA4 FULL-WAVE VACUUM RECTIFIER



9M

Miniature type used in power supply of compact audio equipment having moderate dc requirements. **Outlines section,** 6G; requires miniature 9-contact socket. This tube, like other power-handling tubes, should be adequately ventilated. **Heater:** volts (ac/dc), 6.3; amperes, 1.



Full-Wave Rectifier

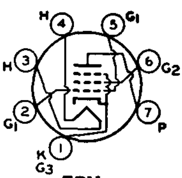
MAXIMUM RATINGS (Design-Center Values)

Peak Inverse Plate Voltage	1000	volts
Peak Plate Current (Per Plate)	450	mA
AC Plate Supply Voltage (Per Plate, rms) with Capacitor Input to Filter	350	volts
Average Output Current	150	mA
Hot Switching Transient Plate Current (Per Plate)	#	
Peak Heater-Cathode Voltage	—500	volts

TYPICAL OPERATION WITH CAPACITOR INPUT TO FILTER

AC Plate-to-Plate Supply Voltage (rms)	500	600	700	volts
Filter-Input Capacitor	50	50	50	μF
Total Effective Plate Supply Impedance per Plate	150	200	240	ohms
DC Output Voltage at Input to Filter (Approx.) For dc output current of 150 mA	245	293	347	volts

When capacitor-input circuits are used, a maximum peak current value per plate of 1 ampere during the initial cycles of the hot-switching transient should not be exceeded.



7CV

6CA5

BEAM POWER TUBE

12CA5, 25CA5

Miniature type used in af power output stage of radio and television receivers. Outlines section, 5D; requires miniature 7-contact socket. Types 12CA5 and 25CA5 are identical with type 6CA5 except for heater ratings.

	6CA5	12CA5		25CA5	
Heater Voltage (ac/dc)	6.3	12.6		25	volts
Heater Current	1.2	0.6		0.3	ampere
Heater Warm-up Time (Average)	—	11		—	seconds
Heater-Cathode Voltage:					
Peak value	±200 max	+200	—300 max	±200 max	volts
Average value	100 max	+100	—200 max	100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	130	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	5	watts
Grid-No.2 Input	1.4	watts
Bulb Temperature (At hottest point)	180	°C

TYPICAL OPERATION

Plate Voltage	110	125	volts
Grid-No.2 Voltage	110	125	volts
Grid-No.1 (Control-Grid) Voltage	—4	—4.5	volts
Peak AF Grid-No.1 Voltage	4	4.5	volts
Zero-Signal Plate Current	32	37	mA
Maximum-Signal Plate Current	31	36	mA
Zero-Signal Grid-No.2 Current (Approx.)	3.5	4	mA
Maximum-Signal Grid-No.2 Current (Approx.)	7.5	11	mA
Plate Resistance (Approx.)	16000	15000	ohms
Transconductance	8100	9200	μmhos
Load Resistance	3500	4500	ohms
Total Harmonic Distortion	5	6	per cent
Maximum-Signal Power Output	1.1	1.5	watts

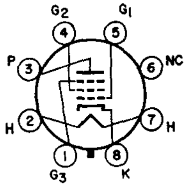
MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

6CA7

Refer to chart at end of section.

6CA7/EL34 POWER PENTODE



8ET

Glass octal types used in the output stage of audio-frequency amplifiers. Maximum dimensions: over-all length, 4 7/16 inches; seated height, 3 7/8 inches; diameter, 1 1/2 inches. Tube requires octal socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	1.5	amperes
Peak Heater-Cathode Voltage	±200 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	1	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	15.5	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	7.2	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	800	volts
Grid-No.2 (Screen-Grid) Voltage	425	volts
Grid-No.2 Input	8	watts
Cathode Current	150	mA
Plate Dissipation	25	watts

TYPICAL OPERATION

Plate Voltage	265	volts
Grid-No.2 Voltage	250	volts
Grid-No.1 (Control-Grid) Voltage	—13.5	volts
Peak AF Grid-No.1 Voltage	12.3	volts
Zero-Signal Plate Current	100	mA
Zero-Signal Grid-No.2 Current	15	mA
Transconductance	11000	µmhos
Plate Resistance	15000	ohms
Load Resistance	2000	ohms
Maximum-Signal Power Output	11	watts
Total Harmonic Distortion	10	per cent

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, for cathode-bias operation	0.7	megohm

Push-Pull Class AB_1 Amplifier

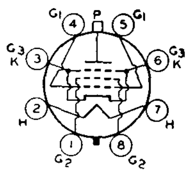
MAXIMUM RATINGS (Same as for Class A_1 Amplifier)

TYPICAL OPERATION (Values are for two tubes)

Plate Supply Voltage	450	volts
Grid-No.2 Supply Voltage	450	volts
Cathode-Bias Resistor	232	ohms
Grid-No.2 Resistor	1000	ohms
Peak AF Grid-No.1 to Grid-No.1 Voltage	38.2	volts
Zero-Signal Plate Current	120	mA
Maximum-Signal Plate Current	143	mA
Zero-Signal Grid-No.2 Current	20	mA
Maximum-Signal Grid-No.2 Current	44	mA
Effective Load Resistance (Plate-to-plate)	6500	ohms
Total Harmonic Distortion	5.1	per cent
Maximum-Signal Power Output	40	watts

6CB5 Refer to chart at end of section.

6CB5A BEAM POWER TUBE



8GD

Glass octal type used as horizontal-deflection amplifier in color and black-and-white television receivers. Outlines section, 21B; requires octal socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	2.5	amperes
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Direct Interelectrode Capacitances (Approx.):

Grid No.1 to Plate	0.4	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	22	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	10	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	75	175	volts
Grid-No.2 Voltage	150	175	volts
Grid-No.1 Voltage	0	—30	volts
Mu-Factor, Grid No.2 to Grid No.1	—	3.8	
Plate Resistance (Approx.)	—	5000	ohms
Transconductance	—	8800	µmhos
Plate Current	460•	90	mA
Grid-No.2 Current	42•	6	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—60	volts

• These values can be measured by a method involving a recurrent waveform such that the maximum rating of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

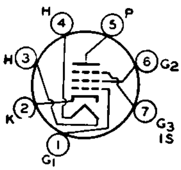
DC Plate Voltage	880	volts
Peak Positive-Pulse Plate Voltage#	6800	volts
Peak Negative-Pulse Plate Voltage	1650	volts
DC Grid-No.2 (Screen-Grid) Voltage	220	volts
DC Grid-No.1 (Control-Grid) Voltage	—55	volts
Peak Negative-Pulse Grid-No.1 Voltage	220	volts
Peak Cathode Current	850	mA
Average Cathode Current	240	mA
Grid-No.2 Input	4	watts
Plate Dissipation†	26	watts
Bulb Temperature (At hottest point)	220	°C

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	0.47	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
† A bias resistor or other means is required to protect the tube in absence of excitation.

Refer to chart at end of section. **6CB6**



7CM

6CB6A 6CB6A/ 6CF6

SHARP-CUTOFF PENTODE

3CB6, 3CB6/3CF6, 4CB6

Miniature types used in color and black-and-white television receivers as if amplifier at frequencies up to about 45 MHz and as rf amplifiers in vhf television tuners. **Outlines section, 5C**; requires miniature 7-contact socket. For typical operation as resistance-coupled amplifiers, refer to **Resistance-Coupled Amplifier** section. Types 3CB6, 3CB6/3CF6, and 4CB6 are identical with type 6CB6A except for heater ratings.

	3CB6 3CB6/3CF6	4CB6	6CB6A 6CB6A/6CF6	
Heater Voltage (ac/dc)	3.15	4.2	6.3	volts
Heater Current	0.6	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	11	11	seconds
Heater-Cathode Voltage:				
Peak value	+200 max —300 max	+200 max —300 max	±200 max	volts
Average value	100 max	+100 max —200 max	100 max	volts

Direct Interelectrode Capacitances:	Unshielded	Shielded▲	
Grid No.1 to Plate	0.025 max	0.015 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	6.5	6.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2	3	pF

▲ With external shield connected to cathode.

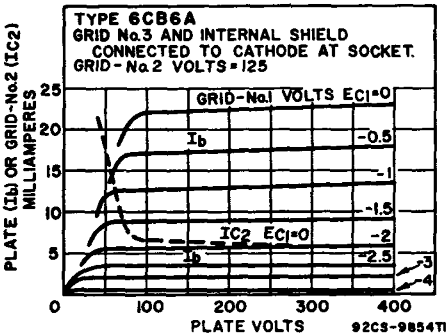
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Voltage	See curve page 98	
Grid-No.2 Supply Voltage	330	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	2.3	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	125	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.28	megohm
Transconductance	8000	µmhos
Plate Current	13	mA
Grid-No.2 Current	3.7	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—6.5	volts
Grid-No.1 Voltage (Approx.) for plate current of 2.8 mA	—3	volts



6CD3 Refer to type 6CG3.

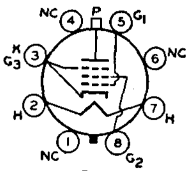
6CD6G Refer to chart at end of section.

6CD6GA

25CD6GB

BEAM POWER TUBE

Glass octal type used as horizontal-deflection amplifier in high-efficiency deflection circuits of color and black-and-white television receivers. Outlines section, 21B; requires octal socket. This type may be supplied with pins 1, 4, and 6 omitted. Vertical tube mounting is preferred, but horizontal operation is permissible if pins No.2 and 7 are in vertical plane. Type 25CD6GB is identical with type 6CD6GA except for heater ratings.



5BT

	6CD6GA	25CD6GB	
Heater Voltage (ac/dc)	6.3	25	volts
Heater Current	2.5	0.6	amperes
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	1.1	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	22	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	8.5	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	60	175	volts
Grid-No.2 (Screen-Grid) Voltage	100	175	volts
Grid-No.1 (Control-Grid) Voltage	0	—30	volts
Mu-Factor, Grid No.2 to Grid No.1	—	3.9	
Plate Resistance (Approx.)	—	7200	ohms
Transconductance	—	7700	μmhos
Plate Current	230•	5.5	mA
Grid-No.2 Current	21•	5.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—55	volts

• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)

DC Plate Voltage	700	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	7000▪	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.2 (Screen-Grid) Voltage	175	volts
Peak Negative-Pulse Grid-No.1 Voltage	700	volts
Peak Cathode Current	200	mA
Average Cathode Current	—200	mA
Plate Dissipation†	20	watts
Grid-No.2 Input	3	watts
Bulb Temperature (At hottest point)	225	°C

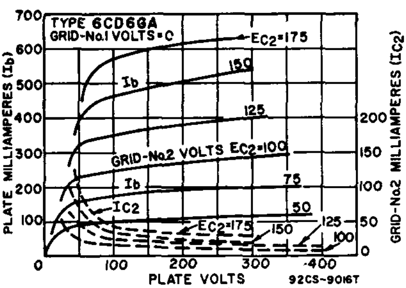
MAXIMUM CIRCUIT VALUE

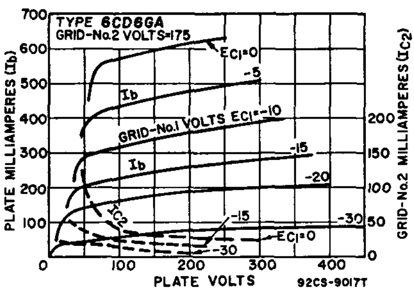
Grid-No.-Circuit Resistance, for grid-resistor-bias operation	0.47	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

▪ Under no circumstances should this absolute value be exceeded.

† A bias resistor or other means is required to protect the tube in absence of excitation.





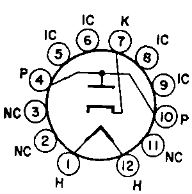
Refer to chart at end of section. **6CE5**

Refer to chart at end of section. **6CF6**

6CG3
6CG3/6CD3

19CG3, 25CG3

HALF-WAVE VACUUM RECTIFIER



12FX

Duodecar type used as damper diode in horizontal-deflection circuits of color and black-and-white television receivers. Outlines section, 8F; requires duodecar 12-contact socket. Types 19CG3 and 25CG3 are identical with type 6CG3 except for heater ratings.

	6CG3 6CG3/6CD3	19CG3	25CG3	
Heater Voltage (ac/dc)	6.3	19	25	volts
Heater Current	1.8	0.6	0.45	amperes
Heater Warm-up Time	—	11	11	seconds

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		5000	volts
Peak Plate Current		2100	mA
Average Plate Current		350	mA
Plate Dissipation		6.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—5000	volts
Average value	+100	—900	volts

CHARACTERISTICS, Instantaneous Value

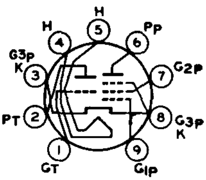
Tube Voltage Drop for plate current of 700 mA	25	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6CG8 Refer to chart at end of section.

6CG8A
5CG8

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE



9GF

Miniature type used as combined oscillator and mixer tube in color and black-and-white television receivers utilizing an intermediate frequency in the order of 40 MHz. When used in an AM/FM receiver, the triode unit is used as an oscillator for both sections. In the AM section, the pentode unit is used as a high-gain pentode mixer; in the FM section, the pentode unit is used either as a pentode mixer or as a triode-connected mixer depending on signal-to-noise considerations. **Outlines section,** 6B; requires miniature 9-contact socket. Type 5CG8 is identical with type 6CG8A except for heater ratings. These types are electrically identical with miniature type 6X8 except for interelectrode capacitances.

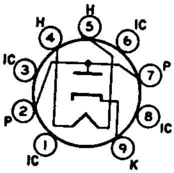
	5CG8	6CG8A	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:	Unshielded	Shielded°	
Triode Unit:			
Grid to Plate	1.5	1.5	pF
Grid to Cathode, Heater, and Pentode Grid No.3	2	2.4	pF
Plate to Cathode, Heater, and Pentode Grid No.3	0.5	1	pF
Pentode Unit:			
Grid No.1 to Plate	0.04 max	0.02 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	4.6	4.8	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	0.9	1.6	pF
Pentode Grid No.1 to Triode Plate	0.05 max	0.04 max	pF
Pentode Plate to Triode Plate	0.05 max	0.008 max	pF
Heater to Cathode	6.5	6.5•	pF

° With external shield connected to cathode, except as noted.

• *With external shield connected to plate.*

6CH8 Refer to chart at end of section.

HALF-WAVE VACUUM RECTIFIER — 6CJ3



9HP

Novar type used as damper tube in horizontal-deflection circuits of black-and-white television receivers. **Outlines section, 30F; requires novar 9-contact socket.** Socket terminals 1, 3, 6, and 8 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated. **Heater:** volts (ac/dc), 6.3; amperes, 1.8.

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		5500	volts
Peak Plate Current		2100	mA
Average Plate Current		350	mA
Plate Dissipation		6.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—5500	volts
Average value	+100	—900	volts

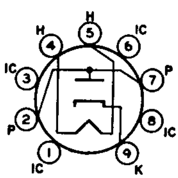
CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 700 mA	25	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

HALF-WAVE VACUUM RECTIFIER — 6CK3

12CK3, 17CK3



9HP

Novar type used as damper tube in horizontal-deflection circuits of black-and-white television receivers. **Outlines section, 30B; requires novar 9-contact socket.** Socket terminals 1, 3, 6, and 8 should not be used as tie points; it is recommended that socket tabs be removed to reduce the possibility of arc-over and to minimize leakage. This tube, like other power-handling tubes, should be adequately ventilated. Types 12CK3 and 17CK3 are identical with type 6CK3 except for heater ratings.

	6CK3	12CK3	17CK3	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds

Direct Interelectrode Capacitances (Approx.):		
Plate to Cathode and Heater	6.5	pF
Cathode to Plate and Heater	9	pF
Heater to Cathode	3	pF

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		5200	volts
Peak Plate Current		1200	mA
Average Plate Current		250	mA
Plate Dissipation		6.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—5200	volts
Average value	+100	—900	volts

CHARACTERISTICS, INSTANTANEOUS VALUE

Tube Voltage Drop for plate current of 350 mA	16	volts

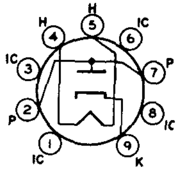
Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

Refer to chart at end of section. **6CK4**

6CL3

12CL3

HALF-WAVE VACUUM RECTIFIER



9HP

Novar type used as a damper tube in horizontal-deflection circuits of color and black-and-white television receivers. **Outlines section,** 30B; requires novar 9-contact socket. Socket terminals 1, 3, 6, and 8 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated. Type 12CL3 is identical with type 6CL3 except for heater ratings.

	6CL3	12CL3	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	1.2	0.6	amperes
Heater Warm-up Time (Average)	—	11	seconds
Direct Interelectrode Capacitances:			
Plate to Cathode and Heater		6.5	pF
Cathode to Plate and Heater		9	pF
Heater to Cathode		3	pF

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		5500	volts
Peak Plate Current		1300	mA
Average Plate Current		250	mA
Plate Dissipation		8.5	watts
Bulb Temperature (At hottest point)		220	°C
Heater-Cathode Voltage:			
Peak value	+300	—5000	volts
Average value	+100	—900	volts

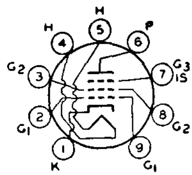
CHARACTERISTICS, INSTANTANEOUS VALUE

Tube Voltage Drop for plate current of 350 mA	16	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6CL6

POWER PENTODE



9BV

Miniature type used in output stage of video amplifier of color and black-and-white television receivers and as wide-band amplifier tube in industrial and laboratory equipment. **Outlines section,** 6E; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.65	ampere
Peak Heater-Cathode Voltage	±100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.12	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	11	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	150	volts
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate Dissipation	7.5	watts
Grid-No.2 Input	1.7	watts
Bulb Temperature (At hottest point)	200	°C

TYPICAL OPERATION

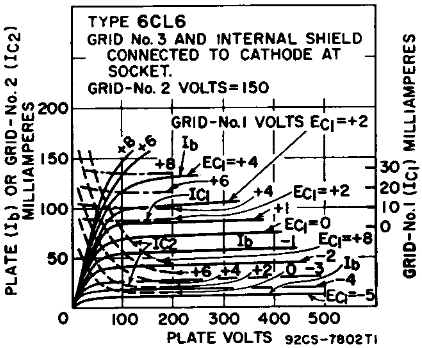
Plate Voltage	250	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Voltage	150	volts
Grid-No.1 Voltage	—3	volts
Peak AF Grid-No.1 Voltage	3	volts
Zero-Signal Plate Current	30	mA
Maximum-Signal Plate Current	31	mA
Zero-Signal Grid-No.2 Current	7	mA
Maximum-Signal Grid-No.2 Current	7.2	mA
Plate Resistance (Approx.)	0.09	megohm
Transconductance	11000	μmhos
Load Resistance	7500	ohms
Total Harmonic Distortion	8	per cent
Maximum-Signal Power Output	2.8	watts
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—14	volts

TYPICAL OPERATION IN MHz-BANDWIDTH VIDEO AMPLIFIER

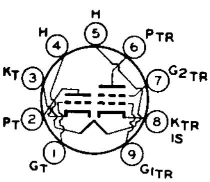
Plate Supply Voltage	300	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	300	volts
Grid-No.1 Bias Voltage	—2	volts
Grid-No.1 Signal Voltage (Peak to Peak)	3	volts
Grid-No.2 Resistor	24000	ohms
Grid-No.1 Resistor	0.1	megohm
Load Resistor	3900	ohms
Zero-Signal Plate Current	30	mA
Zero-Signal Grid-No.2 Current	7	mA
Voltage Output (Peak to Peak)	132	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1 Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm



Refer to chart at end of section. **6CL8**



MEDIUM-MU TRIODE— SHARP-CUTOFF TETRODE

6CL8A

5CL8A, 19CL8A

Miniature type used as combined vhf oscillator and mixer in color and black-and-white television receivers. Outlines section, 6B; requires miniature 9-contact socket. For maximum ratings as class A_1 amplifier, see type 6U8A. Types 5CL8A and 19CL8A are identical with type 6CL8A except for heater ratings.

	5CL8A	6CL8A	19CL8A	
Heater Voltage (ac/dc)	4.7	6.3	18.9	volts
Heater Current	0.6	0.45	0.15	ampere
Heater Warm-up Time (Average)	11	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:	Unshielded	Shielded	
Triode Unit:			
Grid to Plate	1.8	1.8	pF
Grid to Cathode, Tetrode Cathode, Heater, and Internal Shield	2.8	2.8	pF
Plate to Cathode, Tetrode Cathode, Heater, and Internal Shield	1.5	2	pF
Tetrode Unit:			
Grid No. 1 to Plate	0.02 max	0.01 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Internal Shield	5	5	pF
Plate to Cathode, Heater, Grid No.2, and Internal Shield	2	3	pF
Tetrode Grid No.1 to Triode Plate	0.015 max	0.01 max	pF
Tetrode Plate to Triode Plate	0.15 max	0.03 max	pF
Heater to Cathode (Each Unit)	3	3	pF

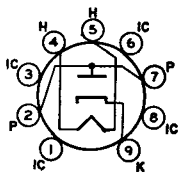
Class A_1 Amplifier

CHARACTERISTICS	Triode Unit	Tetrode Unit	
Plate Supply Voltage	125	125	volts
Grid-No.2 (Screen-Grid) Voltage	—	125	volts
Grid-No.1 Voltage	—1	—1	volt
Amplification Factor	40	—	
Plate Resistance (Approx.)	0.005	0.2	megohm
Transconductance	8000	6500	µmhos
Plate Current	14	12	mA
Grid-No.2 Current	—	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—9	—9	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

6CM3

25CM3, 34CM3

HALF-WAVE VACUUM RECTIFIER



9HP

Novar type used as damper tube in horizintal-deflection circuits of color and black-and-white television receivers. Outline section, 30B; requires novar 9-contact socket. Socket terminals 1, 3, 6, and 8 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated. Types 25CM3 and 34CM3 are identical with type 6CM3 except for heater ratings.

	6CM3	25CM3	34CM3	
Heater Voltage (ac/dc)	6.3	25	33.5	volts
Heater Current	2.4	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Direct Interelectrode Capacitances:				
Plate to Cathode and Heater			20	pF
Cathode to Plate and Heater			18	pF
Heater to Cathode			4	pF

Damper Service

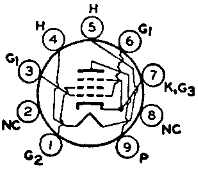
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)			
Peak Inverse Plate Voltage#		5500	volts
Peak Plate Current		1700	mA
Average Plate Current		400	mA
Plate Dissipation		12	watts
Heater-Cathode Voltage:			
Peak value	+300	—5500	volts
Average value	+100	—900	volts
CHARACTERISTICS, Instantaneous Value			
Tube Voltage Drop for plate current of 350 mA		10	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6CM6

BEAM POWER TUBE



9CK

Miniature type used as vertical-deflection amplifier in color and black-and-white television receivers and as audio power amplifier in radio and television receivers. Outlines section, 6E; requires miniature 9-contact socket. For typical operation and maximum circuit values as class A_1 amplifier, refer to type 6V6GTA. For curves of average plate characteristics, refer to type 6AQ5A.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.7	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	8	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	8.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)		
Plate Voltage	315	volts
Grid-No.2 (Screen-Grid) Voltage	285	volts
Plate Dissipation	12	watts
Grid-No.2 Input	2	watts
CHARACTERISTICS (Triode Connected)		
Plate Voltage	250	volts
Grid-No.1 Voltage	—12.5	volts
Amplification Factor	9.8	
Plate Resistance (Approx.)	1960	ohms
Transconductance	5000	μmhos
Plate Current	49.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 0.5 mA	—37	volts

Vertical-Deflection Amplifier

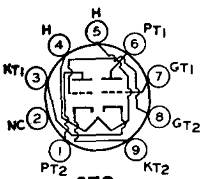
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)	Triode Connection•	Pentode Connection	
DC Plate Voltage	315	315	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	2000▵	2000▵	volts
DC Grid-No.2 (Screen-Grid) Voltage	—	285	volts
Peak Negative-Pulse Grid-No.1 (Control-Grid) Voltage	250	250	volts
Peak Cathode Current	120	120	mA
Average Cathode Current	40	40	mA
Plate Dissipation	9	8	watts
Grid-No.2 Input	—	1.75	watts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance, for cathode-bias operation	2.2	2.2	megohms

• Grid No.2 connected to plate.
Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).
▵ Under no circumstances should this absolute value be exceeded.

6CM7
8CM7

MEDIUM-MU DUAL TRIODE



9ES

Miniature type used as combined vertical-deflection oscillator and vertical-deflection amplifier in black-and-white television receivers. Unit No.1 is used as a conventional blocking oscillator in vertical-deflection circuits, and unit No.2 as a vertical-deflection amplifier. Outlines section, 6E; requires miniature 9-contact socket. Types 8CM7 is identical with type 6CM7 except for heater ratings.

	6CM7	8CM7	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	3.8	3	pF
Grid to Cathode and Heater	2	3.5	pF
Plate to Cathode and Heater	0.5	0.4	pF

Class A_1 Amplifier

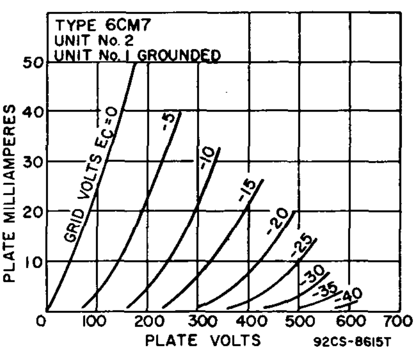
CHARACTERISTICS	Unit No.1	Unit No.2	
Plate Voltage	200	250	volts
Grid Voltage	—7	—8	volts
Amplification Factor	21	18	
Plate Resistance (Approx.)	10500	4100	ohms
Transconductance	2000	4400	µmhos
Plate Current	5	20	mA
Plate Current for grid voltage of —10 volts	1	—	mA
Grid Voltage (Approx.) for plate current of 10 µA	—14	—	volts

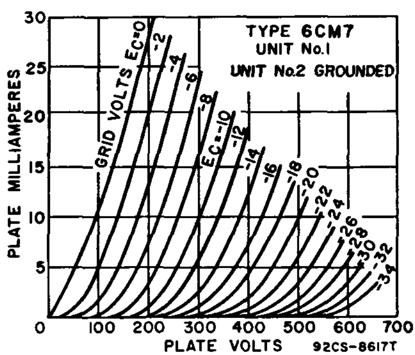
Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	550	550	volts
Peak Positive-Pulse Plate Voltage#	—	2200	volts
Peak Negative-Pulse Grid Voltage	220	220	volts
Peak Cathode Current	77	77	mA
Average Cathode Current	17	22	mA
Plate Dissipation	1.45	6	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For fixed-bias operation	2.2	1	megohms
For cathode-bias operation	2.2	2.5	megohms
For grid-resistor-bias operation	2.2	—	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

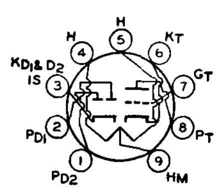




6CM8 Refer to chart at end of section.

6CN7
8CN7

TWIN DIODE—HIGH-MU TRIODE



9EN

Miniature type used as combined horizontal phase detector and reactance tube in color and black-and-white television receivers. The triode unit is used in sync-separator, sync-amplifier, or audio amplifier circuits. Outlines section, 6B; requires miniature 9-contact socket. For typical operation of triode unit as resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section. Type 8CN7 is identical with type 6CN7 except for heater ratings.

	6CN7	8CN7	
Heater Voltage (ac/dc):			
Series	6.3	8.4	volts
Parallel	3.15	4.2	volts
Heater Current:			
Series	0.3	0.225	ampere
Parallel	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	1.8	pF
Grid to Cathode and Heater	1.5	pF
Plate to Cathode and Heater	0.5	pF
Diode Units:		
Diode-No.1 Plate to Cathode of Diodes No.1 and No. 2, Heater, and Internal Shield	3.6	pF
Diode-No.2 Plate to Cathode of Diodes No.1 and No. 2, Heater, and Internal Shield	3.6	pF
Triode Grid to Either Diode Plate	0.006	pF

Triode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation	1.1	watt

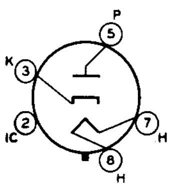
CHARACTERISTICS

Plate Voltage	100	250	volts
Grid Voltage	—1	—3	volts
Amplification Factor	70	70	
Plate Resistance (Approx.)	54000	58000	ohms
Transconductance	1300	1200	μmhos
Plate Current	0.8	1	mA

Diode Units

MAXIMUM RATINGS (Design-Maximum Values)

Plate Current (Each Unit)	5.5	mA



4CG

HALF-WAVE VACUUM RECTIFIER 6CQ4

Octal type used as damper tube in horizontal-deflection circuits of black-and-white television receivers .Outlines section, 13G; requires octal socket. Socket terminals 1, 2, 4, and 6 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated. **Heater:** volts (ac/dc), 6.3; amperes, 1.6.

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Current#		5500	volts
Peak Plate Current		1200	mA
Average Plate Current		190	mA
Plate Dissipation		6.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—5500	volts
Average value	+100	—900	volts

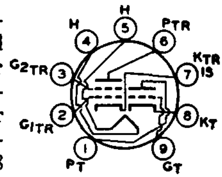
CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 250 mA	25	volts

\# Pulse duration must not exceed 15% of a horizontal cycle (10 microseconds).

6CQ8
5CQ8
MEDIUM-MU TRIODE— SHARP-CUTOFF TETRODE

Miniature type used in color and black-and-white television receiver applications. The tetrode unit is used as a mixer, video if amplifier, or sound if amplifier tube. The triode unit is used in vhf oscillator, phase-splitter, sync-clipper, sync-separator, and rf amplifier circuits. Outlines section, 6B; requires miniature 9-contact socket. Type 5CQ8 is identical with type 6CQ8 except for heater ratings.



9GE

	5CQ8	6CQ8	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:	Unshielded	Shielded▪	
Triode Unit:			
Grid to Plate	1.8	1.8	pF
Grid to Cathode and Heater	2.7	2.7	pF
Plate to Cathode and Heater	0.4	1.2	pF
Tetrode Unit:			
Grid No.1 to Plate	0.019 max	0.015 max	pF
Grid No.1 to Cathode, Heater, Grid No.2 and Internal Shield	5	5	pF
Plate to Cathode, Heater, Grid No.2, and Internal Shield	2.5	3.3	pF
Tetrode Plate to Triode Plate	0.07 max	0.01 max	pF
Heater to Cathode (Each Unit)	3	3†	pF

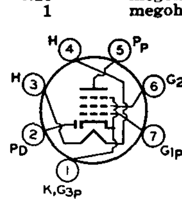
▪ With external shield connected to cathode of unit under test.

† With external shield connected to ground.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Tetrode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage		See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	3.1	3.2	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.7	watt
For grid-No.2 voltages between 165 and 330 volts		See curve page 98	
Grid Input	0.55	—	watt
CHARACTERISTICS			
Plate-Supply Voltage	125	125	volts
Grid-No.2 Supply Voltage	—	125	volts
Grid-No.1 Voltage	—	—1	volts
Cathode-Bias Resistor	56	—	ohms
Amplification Factor	40	—	
Plate Resistance (Approx.)	5000	140000	ohms
Transconductance	8000	5800	µmhos
Plate Current	15	12	mA
Grid-No.2 Current	—	4.2	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—7	—7	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

6CR6
12CR6
DIODE— REMOTE-CUTOFF PENTODE



7EA

Miniature type used as combined detector and audio amplifier in automobile and ac-operated radio receivers. The diode unit is used as an AM detector, and the pentode unit as an automatic-volume-controlled audio amplifier. Outlines section, 5C; requires miniature

7-contact socket. Type 12CR6 is identical with type 6CR6 except for heater ratings.

	6CR6	12CR6	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	0.3	0.15	ampere
Peak Heater-Cathode Voltage	±100 max	±100 max	volts

Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Grid-No.2 (Screen-Grid) Voltage	See curve page 98	
Grid-No.2 Supply Voltage	300	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	2.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	0.3	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

CHARACTERISTICS

Plate Voltage	250	volts
Grid-No.2 Voltage	100	volts
Grid-No.1 Voltage	—2	volts
Plate Resistance (Approx.)	0.8	megohm
Transconductance	2200	μmhos
Plate Current	9.6	mA
Grid-No.2 Current	2.6	mA
Grid-No.1 Voltage (Approx.) for transconductance of 10 μmhos	—32	volts

MAXIMUM CIRCUIT VALUES

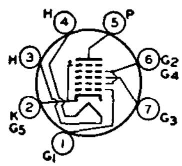
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

Diode Unit

MAXIMUM RATINGS (Design-Center Values)

Plate Current	1	mA

6CS6



7CH

PENTAGRID AMPLIFIER

3CS6, 4CS6, 12CS6

Miniature type used as a gated amplifier in color and black-and-white television receivers. In such service, it may be used as a combined sync separator and sync clipper. Outlines section, 5C; requires miniature 7-contact socket. Types 3CS6, 4CS6, and 12CS6 are identical with type 6CS6 except for heater ratings.

	3CS6	4CS6	6CS6	12CS6	
Heater Voltage (ac/dc)	3.15	4.2	6.3	12.6	volts
Heater Current	0.6	0.45	0.3	0.15	ampere
Heater Warm-up Time (Average)	11	11	11	—	seconds
Heater-Cathode Voltage:					
Peak value	±200 max	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.)		
Grid No.1 to Plate	0.07 max	pF
Grid No.3 to Plate	0.36 max	pF
Grid No.1 to Grid No.3	0.22 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, Grid No.4, and Grid No.5	5.5	pF
Grid No.3 to Cathode, Heater, Grid No.1, Grid No.2, Grid No.4, and Grid No.5	7	pF
Plate to Cathode, Heater, Grid No.1, Grid No.2, Grid No.3, Grid No.4, and Grid No.5	7.5	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	100	100	volts
Grids-No.2-and-No.4 Voltage	30	30	volts
Grid-No.3 Voltage	—1	0	volt
Grid-No.1 Voltage	0	—1	volt
Plate Resistance (Approx.)	0.7	1	megohm
Grid-No.3-to-Plate Transconductance	1500	—	μmhos
Grid-No.1-to-Plate Transconductance	—	1100	μmhos
Plate Current	0.8	1	mA
Grids-No.2-and-No.-4 Current	5.5	1.3	mA

Grid-No.3 Voltage (Approx.) for plate current of 50 μA	—2.2	—	volts
Grid-No.1 Voltage (Approx.) for plate current of 50 μA	—	—2.5	volts

Gated Amplifier Service

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Grids-No.2-and-No.4 Supply Voltage	300	volts
Grids-No.2-and-No.4 Voltage	See curve page 98	
Cathode Current	14	mA
Plate Dissipation	1	watt
Grids-No.2-and-No.4 Input:		
For grids-No.2-and-No.4 voltages up to 150 volts	1	watt
For grids-No.2-and-No.4 voltages between 150 and 300 volts	See curve page 98	

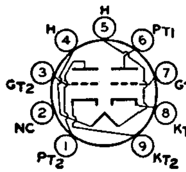
MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance	0.47	megohm
Grid-No.3-Circuit Resistance	2.2	megohms

6CS7

8CS7

MEDIUM-MU DUAL TRIODE



9EF

Miniature type used as combined vertical-deflection oscillator and vertical-deflection amplifier in television receivers. Unit No.1 is used as a conventional blocking oscillator in vertical-deflection circuits, and unit No.2 as a vertical-deflection amplifier. **Outline section,** ture 9-contact socket. Type 8CS7 is identical with type 6CS7 except for heater ratings.

	6CS7	8CS7	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	2.6	2.6	pF
Grid to Cathode and Heater	1.8	3	pF
Plate to Cathode and Heater	0.5	0.5	pF

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1 Oscillator	Unit No.2 Amplifier	
Plate Voltage	250	250	volts
Grid Voltage	—8.5	—10.5	volts
Amplification Factor	17	15.5	
Plate Resistance (Approx.)	7700	3450	ohms
Transconductance	2200	4500	μmhos
Plate Current	10.5	19	mA
Plate Current for grid voltage of —16 volts	—	3	mA
Grid Voltage (Approx.) for plate current of 10 μA	—24	—	volts
Grid Voltage (Approx.) for plate current of 50 μA	—	—22	volts

Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

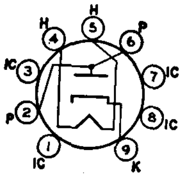
MAXIMUM RATINGS (Design-Center Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	500	500	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	—	2200▵	volts
Peak Negative-Pulse Grid Voltage	400	250	volts
Peak Cathode Current	70	105	mA
Average Cathode Current	20	30	mA
Plate Dissipation	1.25	6.5	watts

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance	2.2	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

▵ Under no circumstances should this absolute value be exceeded.



9RX

HALF-WAVE VACUUM RECTIFIER

6CT3

17CT3

Miniature type used as damper tube in horizontal-deflection circuits of black-and-white and small-screen color television receivers. **Outlines section, 6H;** requires miniature 9-contact socket. Socket terminals 1, 3, 7, and 8 should not be used as tie points for external-circuit components. This tube, like other power-handling tubes, should be adequately ventilated. Type 17CT3 is identical with type 6CT3 except for heater ratings.

	6CT3	17CT3	
Heater Voltage (ac/dc)	6.3	16.8	volts
Heater Current	1.2	0.45	amperes
Heater Warm-up Time (Average)	—	11	seconds
Direct Interelectrode Capacitances (Approx.):			
Plate to Cathode and Heater		12	pF
Cathode to Plate and Heater		9.5	pF
Heater to Cathode		2.8	pF

Damper Service

For operation in a 525-line, 30-frame system

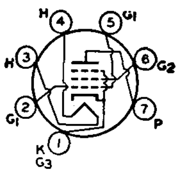
MAXIMUM RATINGS *(Design-Maximum Values)*

Peak Inverse Plate Voltage#		5000	volts
Peak Plate Current		1200	mA
Average Plate Current		250	mA
Plate Dissipation		4.75	watts
Heater-Cathode Voltage:			
Peak value	+300	−5000	volts
Average value	+100	−900	volts
Bulb Temperature (At hottest point)		220	°C

CHARACTERISTICS, *Instantaneous Value*

Tube Voltage Drop for plate current of 350 mA	16	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).



7CV

BEAM POWER TUBE

6CU5

12CU5/12C5, 17CU5, 17CU5/17C5

Miniature type used in the audio output stage of television receivers. **Outlines section, 5D;** requires miniature 7-contact socket. Types 12CU5/12C5, 17CU5, and 17CU5/17C5 are identical with type 6CU5 except for heater ratings.

	6CU5	12CU5/12C5	17CU5 17CU5/ 17C5	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts
Direct Interelectrode Capacitances (Approx.):				
Grid No.1 to Plate			0.6	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3			13	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3			8.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS *(Design-Maximum Values)*

Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	7	watts
Grid-No.2 Input	1.4	watts
Bulb Temperature (At hottest point)	220	°C

TYPICAL OPERATION

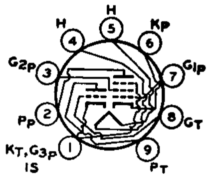
Plate Voltage	120	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 Voltage	—8	volts
Peak AF Grid-No.1 Voltage	8	volts
Zero-Signal Plate Current	49	mA
Maximum-Signal Plate Current	50	mA
Zero-Signal Grid-No.2 Current	4	mA
Maximum-Signal Grid-No.2 Current	8.5	mA
Plate Resistance (Approx.)	10000	ohms
Transconductance	7500	µmhos
Load Resistance	2500	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	2.3	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

6CU6 Refer to type 6BQ6GTB/6CU6.

6CU8 MEDIUM-MU TRIODE—SHARP-CUTOFF PENTODE



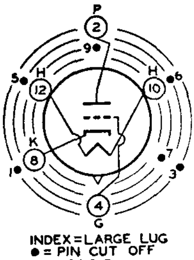
9GM

Miniature type used in color and black-and-white television receiver applications. The pentode unit is used as an if amplifier, video amplifier, agc amplifier, and reactance tube. The triode unit is used in low-frequency oscillator, sync-separator, sync-clipper, and phase-splitter circuits. Outlines section, 6B; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	1.6	pF
Grid to Cathode, Heater, Pentode Grid No.3, and Internal Shield	1.9	pF
Plate to Cathode, Heater, Pentode Grid No.3 and Internal Shield	1.6	pF
Pentode Unit:		
Grid No.1 to Plate	0.025 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, Triode Cathode, and Internal Shield	7	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, Triode Cathode, and Internal Shield	2.4	pF
Pentode Grid No.1 to Triode Plate	0.03 max	pF
Pentode Plate to Triode Plate	0.07 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS *(Design-Maximum Values)*	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 Supply Voltage	—	330	volts
Grid-No.2 (Screen-Grid) Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.8	2.3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	125	125	volts
Grid-No.2 Supply Voltage	—	125	volts
Grid-No.1 Voltage	—1	—	volts
Cathode-Bias Resistor	—	56	ohms
Amplification Factor	24	—	
Plate Resistance (Approx.)	4100	170000	ohms
Transconductance	5800	7800	µmhos
Plate Current	17	12	mA
Plate Current for grid-No.1 voltage of —3 volts	—	—1.6	mA
Grid-No.2 Current	—	3.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—19	—8	volts



INDEX = LARGE LUG
● = PIN CUT OFF
12AQ

6CW4

HIGH-MU TRIODE

2CW4, 13CW4

Nuvistor type used as a grounded-cathode, neutralized rf amplifier in vhf tuners of color and black-and-white television and FM receivers. **Outlines section, 1**; requires nuvistor socket. Types 2CW4 and 13CW4 are identical with type 6CW4 except for heater ratings.

	2CW4	6CW4	13CW4	
Heater Voltage (ac/dc)	2.1	6.3	13.5	volts
Heater Current	0.45	0.135	0.06	ampere
Heater Warm-up Time (Average)	8	—	—	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	±100 max	volts

Direct Interelectrode Capacitances (Approx.)		
Grid to Plate	0.92	pF
Grid to Cathode, Heater, and Shell	4.3	pF
Plate to Cathode, Heater, and Shell	1.8	pF
Plate to Cathode	0.18	pF
Heater to Cathode	1.6	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	300°	volts
Plate Voltage	135	volts
Grid Voltage:		
Negative-bias value	55	volts
Peak positive value	0	volts
Cathode Current	15	mA
Plate Dissipation	1.5	watt

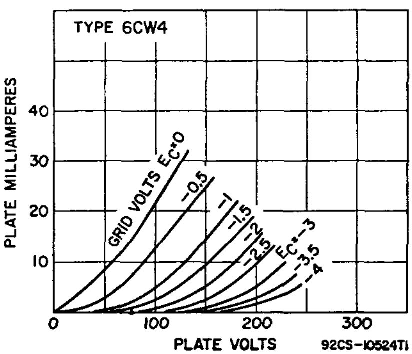
CHARACTERISTICS AND TYPICAL OPERATION	Characteristics	Typical Operation	
Plate Supply Voltage	110	70	volts
Grid Supply Voltage	0	0	volts
Cathode-Bias Resistor	130	—	ohms
Grid Resistor	—	47000	ohms
Amplification Factor	65	68	
Plate Resistance (Approx.)	6600	5440	ohms
Transconductance	9800	12500	μmhos
Plate Current	7	7.2	mA
Grid Voltage (Approx.) for plate current of 10 μA	—4	—	volts

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance:▪		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	2.2	megohms

° A plate supply voltage of 300 volts may be used provided a sufficiently large resistor is used in the plate circuit to limit the plate dissipation to 1.5 watts under any condition of operation.

▪ For operation at metal-shell temperatures up to 135° C.

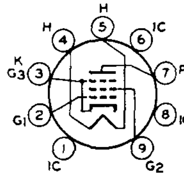


6CW5 Refer to chart at end of section.

6CW5/ EL86

8CW5/XL86, 10CW5/LL86, 15CW5/PL84

POWER PENTODE



9CV

Miniature type used for vertical-deflection amplifier service in color and black-and-white television receivers. Outlines section, 6G; requires miniature 9-contact socket. Types 8CW5/XL86, 10CW5/LL86, and 15CW5/PL84 are identical with type 6CW5/EL86 except for heater ratings.

	6CW5/EL86	8CW5/ XL86	10CW5/ LL86	15CW5/PL84	
Heater Voltage (ac/dc)	6.3	8	10.6	15	volts
Heater Current	0.76	0.6	0.45	0.3	ampere
Heater Warm-up Time	—	—	11	—	seconds
Heater-Cathode Voltage:					
Peak value	±330 max	±330 max	±330 max	±330 max	volts
Average value	±220 max	±220 max	±220 max	±220 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.6	pF
Grid No.1 to Heater	0.25 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	13	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	6.8	pF

Class A_1 or Class AB_1 Amplifier

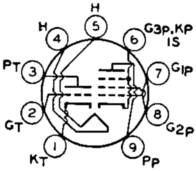
MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	275	volts
Plate Supply Voltage	600	volts
Grid-No.2 Voltage	220	volts
Grid-No.2 (Screen-Grid) Supply Voltage	600	volts
Cathode Current	110	mA
Plate Dissipation	14	watts
Grid-No.2 Input	2.1	watts
Peak Grid-No.2 Input	7	watts
CHARACTERISTICS		
Plate Voltage	170	volts
Grid-No.2 Voltage	170	volts
Grid-No.1 (Control-Grid) Voltage	—12.5	volts
Mu Factor (Grid No.2 to Grid No.1)	8	
Plate Resistance	26000	ohms
Transconductance	11000	μmhos
Plate Current	70	mA
Grid-No.2 Current	3.5	mA
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	1	megohm

Vertical-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	275	volts
Peak Positive-Pulse Plate Voltage#	2200	volts
Grid-No.2 Voltage	275	volts
Peak Negative-Pulse Grid-No.1 Voltage	250	volts
Peak Cathode Current	240	mA
Average Cathode Current	110	mA
Plate Dissipation	12	watts
Grid-No.2 Input	2.1	watts
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	2.2	megohms

Pulse duration must not exceed 6% of a vertical scanning cycle (1.2 milliseconds).



9DX

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

6CX8

8CX8

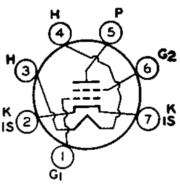
Miniature type used in television receiver applications. Pentode unit is used as video amplifier; triode unit is used in sound if-amplifier, sweep-oscillator, sync-separator, sync-amplifier, and sync-clipper circuits. **Outlines section, 6E;** requires miniature 9-contact socket. Type 8CX8 is identical with type 6CX8 except for heater ratings.

	6CX8	8CX8	
Heater Voltage (ac/dc)	6.3	8	volts
Heater Current	0.75	0.6	ampere
Heater Warm-up Time (Average)	—	11	volts
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	4.4	pF
Grid to Cathode and Heater	2.2	pF
Plate to Cathode and Heater	0.38	pF
Pentode Unit:		
Grid No.1 to Plate	0.06	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	9	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4.4	pF
Triode Grid to Pentode Plate	0.018 max	pF
Pentode Grid No.1 to Triode Plate	0.005 max	pF
Pentode Plate to Triode Plate	0.17 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2	5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	1.1	watts
For grid-No. voltages between 165 and 330 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	150	200	volts
Grid-No.2 Supply Voltage	—	125	volts
Cathode-Bias Resistor	150	68	ohms
Amplification Factor	40	—	
Plate Resistance (Approx.)	8700	70000	ohms
Transconductance	4600	10000	μmhos
Plate Current	9.2	24	mA
Grid-No.2 Current	—	52	mA
Grid-No.1 (Voltage Approx.) for plate current of 100 μA	—5	—8.5	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm



7EW

SHARP-CUTOFF TETRODE

6CY5

2CY5, 3CY5

Miniature type used as rf amplifier in vhf tuners of television receivers. **Outlines section,** 5C; requires miniature 7-contact socket. Types 2CY5 and 3CY5 are identical with type 6CY5 except for heater ratings.

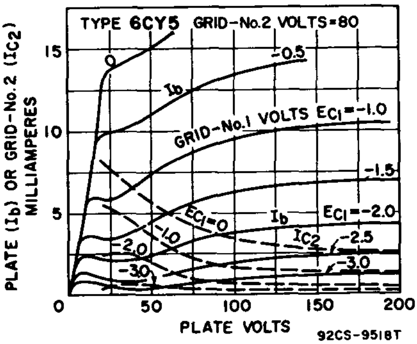
	2CY5	3CY5	6CY5	
Heater Voltage (ac/dc)	2.4	2.9	6.3	volts
Heater Current	0.6	0.45	0.2	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	±100 max	volts

Direct Interelectrode Capacitances (Approx.)*:		
Grid-No.1 to Plate	0.03	pF
Grid-No.1 to Cathode, Heater, Grid No.2 and Internal Shield	4.5	pF
Plate to Cathode, Heater, Grid No.2, and Internal Shield	3	pF

* With external shield connected to cathode.

Class A_1 Amplifier

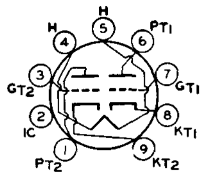
MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	180	volts
Grid-No.2 (Screen-Grid) Supply Voltage	180	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Cathode Current	20	mA
Plate Dissipation	2	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 90 volts	0.5	watt
For grid-No.2 voltages between 90 and 180 volts	See curve page 98	
CHARACTERISTICS		
Plate Voltage	125	volts
Grid-No.2 Voltage	80	volts
Grid-No.1 Voltage	—1	volt
Plate Resistance (Approx.)	0.1	megohm
Transconductance	8000	µmhos
Plate Current	10	mA
Grid-No.2 Current	1.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—6	volts
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	0.5	megohm



6CY7
11CY7

DUAL TRIODE

Miniature type used as combined vertical oscillator and vertical-deflection amplifier in television receivers. Unit No.1 is used as a blocking oscillator in vertical-deflection circuits, and unit No.2 is used as a vertical-deflection amplifier. **Outlines section,** 6E; requires miniature 9-contact socket. Type 11CY7 is identical with type 6CY7 except for heater ratings.



9LG

	6CY7	11CY7	
Heater Voltage (ac/dc)	6.3	11	volts
Heater Current	0.75	0.45	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2	
Plate Supply Voltage	250	150	volts
Grid Voltage	—3	—	volts

	Unit No.1	Unit No.2	
Cathode-Bias Resistor	—	620	ohms
Amplification Factor	68	5	
Plate Resistance (Approx.)	52000	920	ohms
Transconductance	1300	5400	µmhos
Plate Current	1.2	30	mA
Plate Current for grid voltage of —30 volts	—	3.5	mA
Grid Voltage (Approx.) for plate current of 10 µA	—5.5	—	volts
Grid Voltage (Approx.) for plate current of 200 µA	—	—40	volts

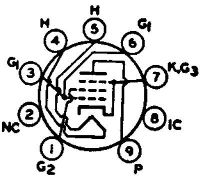
Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	350	350	volts
Peak Positive-Pulse Plate Voltage#	—	1800	volts
Peak Negative-Pulse Grid Voltage	—400	—250	volts
Peak Cathode Current	—	120	mA
Average Cathode Current	—	35	mA
Plate Dissipation	1	5.5	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance	2.2	2.2†	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

† For cathode-bias operation.



9HN

6CZ5

BEAM POWER TUBE

5CZ5

Miniature type used as a vertical-deflection amplifier in high-efficiency deflection circuits of color and black-and-white television receivers and in the audio output stage of television and radio receivers. Outlines section, 6G; requires miniature 9-contact socket. Type 5CZ5 is identical with type 6CZ5 except for heater ratings.

	5CZ5	6CZ5	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.4 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	9	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	6	pF

Class A_1 Amplifier

CHARACTERISTICS			
Plate Voltage	75	250	volts
Grid-No.2 Voltage	250	250	volts
Grid-No.1 Voltage	0	—15	volts
Plate Resistance	—	73000	ohms
Transconductance	—	4800	µmhos
Plate Current	130•	46	mA
Grid-No.2 Current	16•	4.6	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—	—40	volts

Vertical-Deflection Amplifier

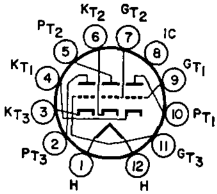
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Voltage	350	volts
Peak Positive-Pulse Plate Voltage#	2200	volts
Grid-No.2 (Screen-Grid) Voltage	315	volts
Peak Negative-Pulse Grid-No.1 (Control-Grid) Voltage	275	volts
Peak Cathode Current	155	mA
Average Cathode Current	45	mA
Plate Dissipation	10	watts
Grid-No.2 Input	2.2	watts
Bulb Temperature (At hottest point)	250	°C
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	1	megohm

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).
• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

6D6	Refer to chart at end of section.
6D7	Refer to chart at end of section.
6D8G	Refer to chart at end of section.

6D10 HIGH-MU TRIPLE TRIODE



12BQ

Duodecar type used in oscillator-mixer, grounded-grid amplifier, and automatic-frequency-control circuits. **Outlines section,** 8A; requires duodecar 12-contact socket. Heater: volts (ac/dc), 6.3; amperes, 0.45; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values, Each Triode Unit)

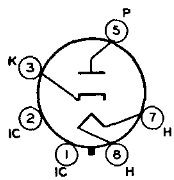
Plate Voltage	330	volts
Grid Voltage:		
Positive-bias value	0	volts
Negative-bias value	50	volts
Plate Dissipation	2	watts
Total Plate Dissipation (All plates)	6	watts

CHARACTERISTICS (Each Triode Unit)

Plate Voltage	125	volts
Grid Voltage	—1	volts
Amplification Factor	57	
Plate Resistance (Approx.)	13600	ohms
Transconductance	4200	μmhos
Plate Current	4.2	mA
Grid Voltage (Approx.) for plate current of 20 μA	—4	volts

6DA4 HALF-WAVE VACUUM RECTIFIER

12D4, 17D4



4CG

Glass octal type used as damper tube in horizontal-deflection circuits of television receivers. **Outlines section,** 13D; requires octal socket. May be supplied with pin No.1 omitted. Socket terminals 1, 2, 4, and 6 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated. Types 12D4 and 17D4 are identical with type 6DA4 except for heater ratings.

	6DA4	12D4	17D4	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds

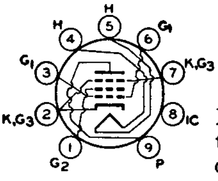
Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Current#		4400	volts
Peak Plate Current		900	mA
Average Plate Current		155	mA
Plate Dissipation		5.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—4400	volts
Average value	+100	—900	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).



9GR

6DB5

BEAM POWER TUBE

12DB5

Miniature type used as vertical-deflection-amplifier tube in television receivers. **Outlines section,** 6F; requires miniature 9-contact socket. Type 12DB5 is identical with type 6DB5 except for heater ratings.

	6DB5	12DB5	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	1.2	0.6	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)		
Plate Voltage	300	volts
Grid-No.2 (Screen-Grid) Voltage	150	volts
Plate Dissipation	10	watts
Grid-No.2 Input	1.25	watts
TYPICAL OPERATION		
Plate Supply Voltage	200	volts
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	180	ohms
Peak AF Grid-No.1 Voltage	8.5	volts
Zero-Signal Plate Current	46	mA
Maximum-Signal Plate Current	47	mA
Zero-Signal Grid-No.2 Current	2.2	mA
Maximum-Signal Grid-No.2 Current	8.5	mA
Plate Resistance (Approx.)	28000	ohms
Transconductance	8000	μmhos
Load Resistance	4000	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	3.8	watts
MAXIMUM CIRCUIT VALUES		
Grid-No.1 Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	2.2	megohms

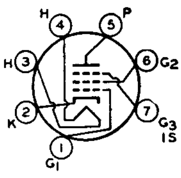
Vertical-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)		
DC Plate Voltage	300	volts
Peak Positive-Pulse Plate Voltage (Absolute Maximum)#	2000▪	volts
DC Grid-No.2 (Screen-Grid) Voltage	150	volts
Peak Negative-Pulse Grid-No.1 (Control-Grid) Voltage	250	volts
Peak Cathode Current	200	mA
Average Cathode Current	55	mA
Plate Dissipation	10	watts
Grid-No.2 Input	1.25	watts
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

▪ Under no circumstances should this absolute maximum value be exceeded.



7CM

6DC6

SHARP-CUTOFF PENTODE

Miniature type used in the gain-controlled picture if stages of color and black-and-white television receivers and as an rf amplifier in the tuners of such receivers. **Outlines section,** 5C; requires 7-contact miniature socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.3	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.02 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	6.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

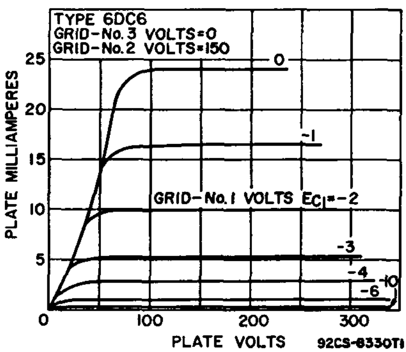
Plate Voltage	300	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 Supply Voltage	300	volts
Grid-No.2 (Screen-Grid) Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	2	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	0.5	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	200	volts
Grid No.3	*Connected to cathode at socket*	
Grid-No.2 Supply Voltage	150	volts
Cathode-Bias Resistor	180	ohms
Plate Resistance (Approx.)	0.5	megohm
Transconductance	5500	μmhos
Plate Current	9	mA
Grid-No.2 Current	3	mA
Grid-No.1 Voltage (Approx.) for transconductance of 50 μmhos	—12.5	volts

MAXIMUM CIRCUIT VALUES

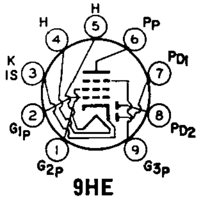
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm



6DC8 Refer to chart at end of section.

6DC8/ EBF89

TWIN DIODE—SEMIREMOTE-CUTOFF PENTODE



Miniature type used as rf- and if-amplifier tube in radio and television receivers. Outlines section, 6E; requires 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.3	ampere
Peak Heater-Cathode Voltage	±100 max	volts

Direct Interelectrode Capacitances:		
Pentode Unit:		
Grid No.1 to Plate	0.0025 max	pF
Grid No.1 to All Other Electrodes Except Plate	5	pF
Plate to All Other Electrodes Except Grid No.1	5.2	pF
Grid No.1 to Heater	0.05 max	pF
Plate of Each Diode Unit to All Other Electrodes	2.5	pF
Plate of Diode Unit No.1 to Plate of Diode Unit No.2	0.25 max	pF
Plate of Diode Unit No.1 to Heater	0.015 max	pF
Plate of Diode Unit No.2 to Heater	0.003 max	pF
Plate of Diode Unit No.1 to Pentode Grid No.1	0.0008 max	pF
Plate of Diode Unit No.2 to Pentode Grid No.1	0.001 max	pF
Plate of Diode Unit No.1 to Pentode Plate	0.15 max	pF
Plate of Diode Unit No.2 to Pentode Plate	0.025 max	pF

Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Supply Voltage	550	volts
Plate Voltage	300	volts
Grid-No.2 Voltage:		
With plate current greater than 8 mA	125	volts
With plate current less than 4 mA	300	volts
Cathode Current	16.5	mA
Plate Dissipation	2.25	watts
Grid-No.2 Input	0.45	watts

CHARACTERISTICS

Plate Voltage	200	250	volts
Grid No.3	Connected to cathode at socket		
Grid-No.2 Voltage	100	100	volts
Grid-No.1 Voltage	—1.5	—2	volts
Mu Factor, Grid No.2 to Grid No.1	20	20	
Plate Resistance (Approx.)	0.6	1	megohm
Transconductance	4500	3800	µmhos
Plate Current	11	9	mA
Grid-No.2 Current	3.3	2.7	mA
Transconductance, at grid-No.1 voltage of —20 volts	120	200	µmhos

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	3	megohms

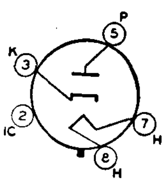
Diode Units (Each Unit)

MAXIMUM RATINGS (Design-Center Values)

Peak Inverse Plate Voltage	200	volts
Peak Plate Current	5	mA
Average Plate Current	0.8	mA

HALF-WAVE VACUUM RECTIFIER 6DE4

17DE4, 22DE4



4CG

Glass octal type used as damper tube in horizontal-deflection circuits of television receivers. Outlines section, 13G; requires octal socket. Socket terminals 1, 2, 4, and 6 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated. Types 17DE4 and 22DE4 are identical with type 6DE4 except for heater ratings.

	6DE4	17DE4	22DE4	
Heater Voltage (ac/dc)	6.3	17	22.4	volts
Heater Current	1.6	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds

Direct Interelectrode Capacitances (Approx.):		
Plate to Cathode and Heater	8.5	pF
Cathode to Plate and Heater	11.5	pF
Heater to Cathode	4	pF

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#	5500	volts
Peak Plate Current	1100	mA
Average Plate Current	180	mA
Plate Dissipation	6.5	watts

Heater-Cathode Voltage:			
Peak value	+300	—5500	volts
Average value	+100	—900	volts

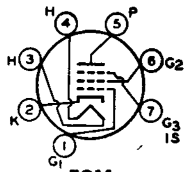
CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 350 mA	34	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6DE6
4DE6

SHARP-CUTOFF PENTODE



7CM

Miniature type used in the gain-controlled picture if stages of television receivers utilizing an intermediate frequency in the order of 40 MHz and as an rf amplifier in vhf television tuners. **Outlines section,** 5C; requires miniature 7-contact socket. Type 4DE6 is identical with type 6DE6 except for heater ratings.

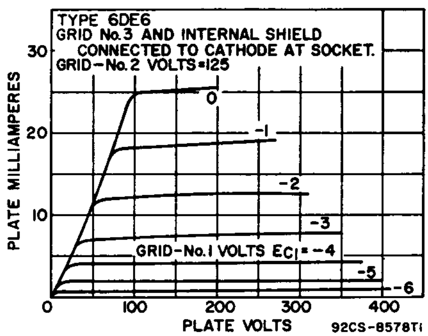
	4DE6	6DE6	
Heater Voltage (ac/dc)	4.2	6.3	volts
Heater Current	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:	Unshielded	Shielded▲	
Grid No.1 to Plate	0.025 max	0.015 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	6.5	6.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2	3	pF

▲ With external shield connected to cathode.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

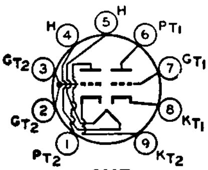
Plate Voltage	330	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	2.3	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	



CHARACTERISTICS

Plate Supply Voltage	125	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.25	megohm
Transconductance	8000	μmhos

Transconductance for grid-No.1 volts of —5.5 and cathode resistor of 0 ohms	700	μmhos
Plate Current	15.5	mA
Grid-No.2 Current	4.2	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—9	volts



9HF

6DE7

DUAL TRIODE

10DE7, 13DE7

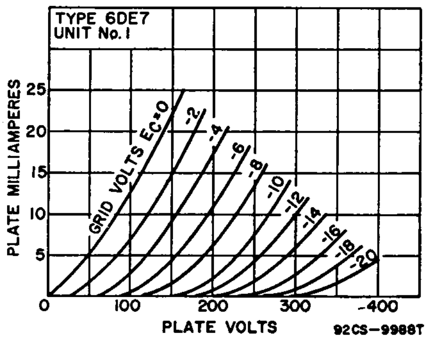
Miniature type used as combined vertical oscillator and vertical-deflection amplifier in television receivers. Unit No.1 is used as a blocking oscillator in vertical-deflection circuits, and unit No.2 is used as a vertical-deflection amplifier. **Outlines section, 6E;** requires miniature 9-contact socket. For curve of average plate characteristics, Unit No.2, refer to type 6DR7. Types 10DE7 and 13DE7 are identical with type 6DE7 except for heater ratings.

	6DE7	10DE7	13DE7	
Heater Voltage (ac/dc)	6.3	9.7	13	volts
Heater Current	0.9	0.6	0.45	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	4	8.5	pF
Grid to Cathode and Heater	2.2	5.5	pF
Plate to Cathode and Heater	0.52	1	pF

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2	
Plate Voltage	250	150	volts
Grid Voltage	—11	—17.5	volts
Amplification Factor	17.5	6	
Plate Resistance (Approx.)	8750	925	ohms
Transconductance	2000	6500	μmhos
Plate Current	5.5	35	mA
Plate Current for grid voltage of —24 volts	—	10	mA
Grid Voltage (Approx.) for plate current of 10 μA	—20	—	volts
Grid Voltage (Approx.) for plate current of 50 μA	—	—44	volts



Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

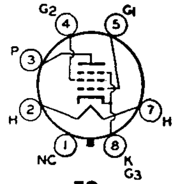
MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	330	275	volts
Peak Positive-Pulse Plate Voltage#	—	1500	volts
Peak Negative-Pulse Grid Voltage	400	250	volts
Peak Cathode Current	77	175	mA

Average Cathode Current	22	50	mA
Plate Dissipation	1.5	7	watts

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance:			
For grid-resistor bias or cathode-bias operation	2.2	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).



7S

6DG6GT BEAM POWER TUBE

Glass octal type used as output tube in audio-amplifier applications Outlines section, 13D; requires octal socket. This type may be supplied with pin 1 omitted.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	1.2	amperes
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.6	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	15	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	10	pF

Class A_1 Audio-Frequency Power Amplifier

MAXIMUM RATINGS (Design-Center Values)

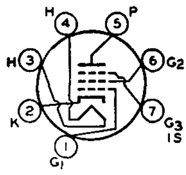
Plate Voltage	200	volts
Grid-No.2 (Screen-Grid) Voltage	125	volts
Plate Dissipation	10	watts
Grid-No.2 Input	1.25	watts

TYPICAL OPERATION

Plate Supply Voltage	110	200	volts
Grid-No.2 Supply Voltage	110	125	volts
Grid-No.1 (Control-Grid) Supply Voltage	—7.5	—	volts
Peak AF Grid-No.1 Voltage	7.5	8.5	volts
Cathode-Bias Resistor	—	180	ohms
Zero-Signal Plate Current	49	46	mA
Maximum-Signal Plate Current	50	47	mA
Zero-Signal Grid-No.2 Current	4	2.2	mA
Maximum-Signal Grid-No.2 Current	10	8.5	mA
Plate Resistance (Approx.)	13000	28000	ohms
Transconductance	8000	8000	μmhos
Load Resistance	2000	4000	ohms
Total Harmonic Distortion	10	10	per cent
Maximum-Signal Power Output	2.1	3.8	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm



7CM

6DK6

3DK6, 4DK6, 12DK6 SHARP-CUTOFF PENTODE

Miniature type used as if-amplifier tube in color and black-and-white television receivers. Outlines section, 5C; requires miniature 7-contact socket. Types 3DK6, 4DK6, and 12DK6 are identical with type 6DK6 except for heater ratings.

	3DK6	4DK6	6DK6	12DK6	
Heater Voltage (ac/dc)	3.15	4.2	6.3	12.6	volts
Heater Current	0.6	0.45	0.3	0.15	ampere
Heater Warm-up Time (Average)	11	11	—	—	seconds
Heater-Cathode Voltage:					
Peak value	+200 max / —300 max	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.025 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3 and Internal Shield	6.3	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3 ,and Internal Shield	1.9	pF

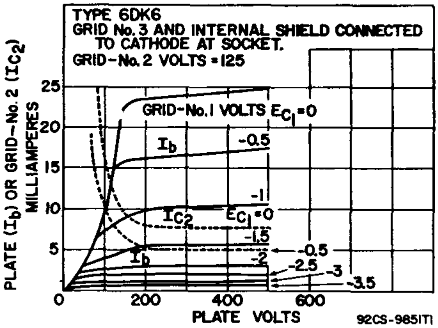
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

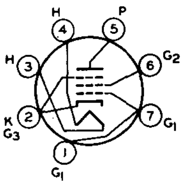
Plate Voltage	330	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	2.3	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	125	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.35	megohm
Transconductance	9800	μmhos
Plate Current	12	mA
Grid-No.2 Current	3.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—6.5	volts



Refer to chart at end of section. **6DL5**



7DQ

POWER PENTODE 6DL5/EL95

Miniature type used in audio output applications in automobile radios. **Outlines section**, 5E; requires miniature 7-contact socket. **Heater:** volts, 6.3; amperes, 0.2; maximum heater-cathode volts, ±100.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	550	volts
Plate Voltage	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	550	volts
Grid-No.2 Voltage	300	volts
Cathode Current	35	mA
Plate Dissipation	2.5	watts
Grid-No.2 Input	6	watts

TYPICAL OPERATION

Plate Voltage	200	250	volts
Grid-No.2 Voltage	200	250	volts
RMS AF Grid-No.1 (Control-Grid) Voltage	4.5	5	volts
Cathode-Bias Resistor	230	320	ohms
Plate Current	23	24	mA
Grid-No.2 Current	4.2	4.5	mA
Load Resistance	8000	10000	ohms
Total Harmonic Distortion	12	12	per cent
Power Output	2.3	3	watts

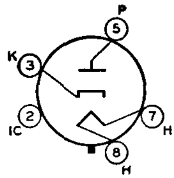
MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, for cathode-bias operation	2	megohms

6DM4 Refer to chart at end of section.

6DM4A
17DM4A

HALF-WAVE VACUUM RECTIFIER



4CG

Glass octal type used as damper tube in horizontal-deflection circuits of television receivers. Outlines section, 13G; requires octal socket. Socket terminals 1, 2, 4, and 6 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated. Type 17DM4A is identical with type 6DM4A except for heater ratings.

	6DM4A	17DM4A	
Heater Voltage (ac/dc)	6.3	16.8	volts
Heater Current	1.2	0.45	amperes
Heater Warm-up Time (Average)	—	11	seconds

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

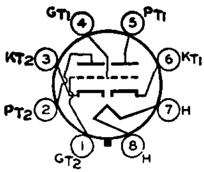
Peak Inverse Plate Voltage#		5000	volts
Peak Plate Current		1200	mA
Average Plate Current		200	mA
Plate Dissipation		6.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—5000	volts
Average value	+100	—900	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6DN6 Refer to chart at end of section.

6DN7

MEDIUM-MU DUAL TRIODE



8BD

Glass octal type used as combined vertical-deflection-oscillator and vertical-deflection-amplifier tube in television receivers. Outlines section, 13B; requires octal socket. Heater: volts (ac/dc), 6.3; amperes, 0.9; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2	
Plate Voltage	250	250	volts
Grid Voltage	—8	—9.5	volts
Amplification Factor	22.5	15.4	
Plate Resistance (Approx.)	9000	2000	ohms
Transconductance	2500	7700	μmhos
Plate Current	8	41	mA
Grid Voltage (Approx.) for plate current of 10 μA	—18	—	volts
Grid Voltage (Approx.) for plate current of 50 μA	—	—23	volts

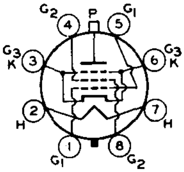
Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	350	550	volts
Peak Positive-Pulse Plate Voltage#	—	2500	volts
Peak Negative-Pulse Grid Voltage	400	250	mA
Peak Cathode Current	—	150	mA
Average Cathode Current	—	50	mA
Plate Dissipation	1	10	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For fixed-bias operation	2.2	2.2	megohms
For cathode-bias operation	2.2	—	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

Refer to chart at end of section. **6DQ4**



8JC

BEAM POWER TUBE 6DQ5

Glass octal type used as horizontal-deflection amplifier in color and black-and-white television receivers. Outlines section, 21B; requires octal socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	2.5	amperes
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.5	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	23	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	11	pF

Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection		Triode• Connection	
Plate Voltage	70	175	125	volts
Grid No.2 (Screen-Grid) Voltage	125	125	—	volts
Grid No.1 (Control-Grid) Voltage	0	—25	—25	volts
Amplification Factor	—	—	3.3	
Plate Resistance (Approx.)	—	5500	—	ohms
Transconductance	—	10500	—	μmhos
Plate Current	550*	110	—	mA
Grid-No.2 Current	42*	5	—	mA
Grid-No.1 Voltage (Approx.) for plate mA = 1	—	—55	—	volts

• Grid No.2 connected to plate.

* These values can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

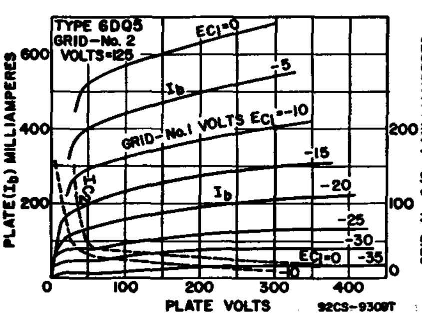
Horizontal-Deflection Amplifier

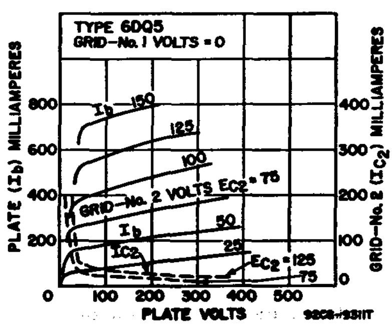
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Voltage	990	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1100	volts
DC Grid-No.2 (Screen-Grid) Voltage	190	volts
Peak Negative-Pulse Grid-No.1 (Control-Grid) Voltage	250	volts
Peak Cathode Current	1100	mA
Average Cathode Current	315	mA
Grid-No.2 Input	3.2	watts
Plate Dissipation▪	24	watts
Bulb Temperature (At hottest point)	220	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance, for grid-resistor-bias operation	0.47	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

▪ A bias resistor or other means is required to protect the tube in absence of excitation.





6DQ6A Refer to chart at end of section.

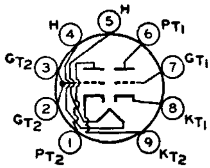
6DQ6B Refer to chart at end of section.

6DR7

10DR7, 13DR7

DUAL TRIODE

Miniature type containing high-mu and low-mu triodes; used as combined vertical-deflection-oscillator and vertical-deflection-amplifier tube in television receivers. **Outlines section,** 6E; requires miniature 9-contact socket. Types 10DR7 and 13DR7 are identical with type 6DR7 except for heater ratings.



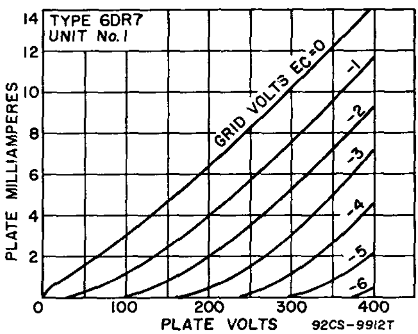
9HF

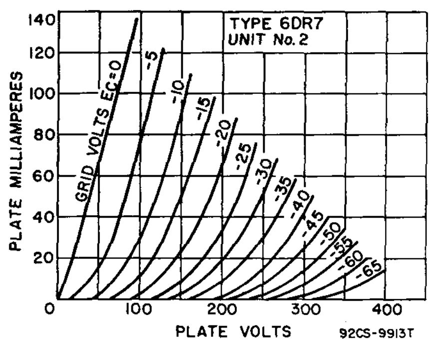
	6DR7	10DR7	13DR7	
Heater Voltage (ac/dc)	6.3	9.7	13	volts
Heater Current	0.9	0.6	0.45	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	4.5	8.5	pF
Grid to Cathode and Heater	2.2	5.5	pF
Plate to Cathode and Heater	0.34	1	pF

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2	
Plate Voltage	250	150	volts
Grid Voltage	—3	—17.5	volts
Amplification Factor	68	6	
Plate Resistance (Approx.)	40000	925	ohms
Transconductance	1600	6500	µmhos
Plate Current	1.4	35	mA
Plate Current for grid voltage of —24 volts	—	10	mA
Grid Voltage (Approx.) for plate current of 10 µA	—5.5	—	volts
Grid Voltage (Approx.) for plate current of 50 µA	—	—44	volts



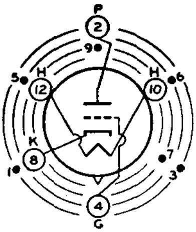


Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)			
DC Plate Voltage	330	275	volts
Peak Positive-Pulse Plate Voltage#	—	1500	volts
Peak Negative-Pulse Grid Voltage	400	250	volts
Peak Cathode Current	70	175	mA
Average Cathode Current	20	50	mA
Plate Dissipation	1	7	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For grid-resistance-bias or cathode-bias operation	2.2	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).



INDEX = LARGE LUG
● = PIN CUT OFF

12AQ

6DS4
2DS4

HIGH-MU TRIODE

Nuvistor type used as grounded-cathode, neutralized rf amplifier in vhf tuners of color and black-and-white television and FM receivers. **Outlines section,** 1; requires nuvistor socket. Type 2DS4 is identical with type 6DS4 except for heater ratings.

	2DS4	6DS4	
Heater Voltage (ac/dc)	2.1	6.3	volts
Heater Current	0.45	0.135	ampere
Heater Warm-up Time (Average)	8	—	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	volts

Direct Interelectrode Capacitances (Approx.)		
Grid to Plate	0.92	pF
Grid to Cathode, Heater, and Shell	4.3	pF
Plate to Cathode, Heater, and Shell	1.8	pF
Plate to Cathode	0.18	pF
Heater to Cathode	1.6	pF

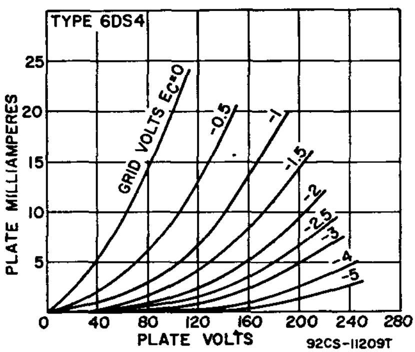
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	300°	volts
Plate Voltage	135	volts
Grid Voltage, Negative-bias value	55	volts
Grid Voltage, Peak positive value	0	volts
Cathode Current	15	mA
Plate Dissipation	1.5	watt

CHARACTERISTICS

Plate Supply Voltage	110	volts
Grid Supply Voltage	0	volts
Cathode-Bias Resistor	130	ohms
Amplification Factor	63	
Plate Resistance (Approx.)	7000	ohms
Transconductance	9000	μmhos
Plate Current	6.5	mA
Grid Voltage (Approx.) for plate current of 100 μA	—5	volts
Grid Voltage (Approx.) for plate current of 10 μA	—6.8	volts



TYPICAL OPERATION

Plate Voltage	70	volts
Grid Supply Voltage	0	volts
Grid Resistor	47000	ohms
Amplification Factor	68	
Plate Resistance (Approx.)	5440	ohms
Transconductance	12500	μmhos
Plate Current	7	mA

MAXIMUM CIRCUIT VALUES

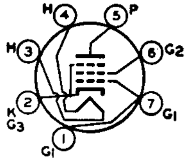
Grid-Circuit Resistance:▪		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	2.2	megohm

° A plate supply voltage of 300 volts may be used provided a sufficiently large resistor is used in the plate circuit to limit the plate dissipation to 1.5 watts under any condition of operation.

▪ For operation at metal-shell temperatures up to 125°C.

6DS5

BEAM POWER TUBE



7BZ

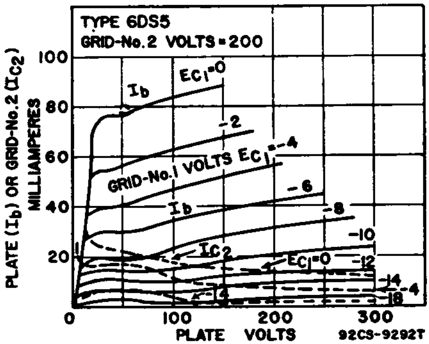
Miniature type used in the audio output stages of television and radio receivers. Outlines section, 5D; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.8	ampere
Peak Heater-Cathode Voltage	±200 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.19	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	9.5	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	6.3	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	275	volts
Grid-No.2 (Screen-Grid) Voltage	275	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	9	watts
Grid-No.2 Input	2.2	watts
Bulb Temperature (At hottest point)	250	°C



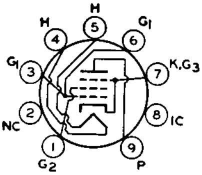
TYPICAL OPERATION AND CHARACTERISTICS

	Cathode-Bias Operation		Fixed-Bias Operation		
Plate Supply Voltage	200	200	200	250	volts
Grid-No.2 Supply Voltage	200	250	200	200	volts
Grid-No.1 Voltage	—	—	—7.5	—8.5	
Cathode-Bias Resistor	180	270	—	—	ohms
Peak AF Grid-No.1 Voltage	7.5	9.2	7.5	8.5	volts
Zero-Signal Plate Current	34.5	27	35	29	mA
Maximum-Signal Plate Current	32.5	25	36	32	mA
Zero-Signal Grid-No.2 Current	3.5	3	3	3	mA
Maximum-Signal Grid-No.2 Current	9	9	9	10	mA
Plate Resistance (Approx.)	28000	28000	28000	28000	ohms
Transconductance	6000	5800	6000	5800	µmhos

	Cathode-Bias Operation		Fixed-Bias Operation		
Load Resistance	6000	8000	6000	8000	ohms
Total Harmonic Distortion	10	10	9	10	per cent
Maximum-Signal Power Output	2.8	3.6	3	3.8	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	1	megohm



9HN

6DT5

BEAM POWER TUBE

12DT5

Miniature type used as a vertical-deflection-amplifier tube in television receivers employing 110-degree picture-tube systems. **Outlines section,** 6E; requires miniature 9-contact socket. Type 12DT5 is identical with type 6DT5 except for heater ratings.

	6DT5	12DT5	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	1.2	0.6	amprees
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	60	80	250	volts
Grid-No. 2 Voltage	150	250	250	volts
Grid-No.1 Voltage	0	0	—16.5	volts
Transconductance	—	—	6200	µmhos
Plate Current	95•	195•	44	mA
Grid-No.2 Current	8.5•	19•	1.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 mA	—	—	—35	volts

• These values can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Vertical-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

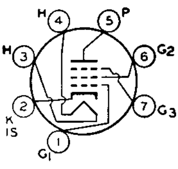
DC Plate Voltage	315	volts
Peak Positive-Pulse Plate Voltage#	2200	volts
Grid-No.2 (Screen-Grid) Voltage	285	volts
Peak Negative-Pulse Grid-No.1 (Control-Grid) Voltage	250	volts
Peak Cathode Current	190	mA
Average Cathode Current	55	mA
Plate Dissipation	9	watts
Grid-No.2 Input	2	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	1	megohm

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

Refer to chart at end of section. **6DT6**



7EN

6DT6A

SHARP-CUTOFF PENTODE

3DT6A, 4DT6A

Miniature type used as FM detector in color and black-and-white television receivers. **Outlines section,** 5C; requires miniature 7-contact socket. Types 3DT6A and 4DT6A are identical with type 6DT6A except for heater ratings.

	3DT6A	4DT6A	6DT6A	
Heater Voltage (ac/dc)	3.15	4.2	6.3	volts
Heater Current	0.6	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.)*		
Grid No.1 to Plate	0.02	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.8	pF
Grid No.3 to Plate	1.7	pF
Grid No.1 to Grid No.3	0.1	pF
Grid No.3 to Cathode, Heater, Grid No.1, Grid No.2, and Internal Shield	6.1	pF

* External shield connected to cathode.

Class A_1 Amplifier

CHARACTERISTICS

Plate Supply Voltage	150	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket	
Grid-No.2 (Screen-Grid) Supply Voltage	100	volts
Cathode-Bias Resistor	560	ohms
Plate Resistance (Approx.)	0.15	megohm
Transconductance, Grid No.1 to Plate	1350	μmhos
Transconductance, Grid No.3 to Plate	515	μmhos
Plate Current	1.55	mA
Grid-No.2 Current	1.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—5.2	volts
Grid-No.3 Voltage (Approx.) for plate current of 10 μA	—4.2	volts

FM Detector

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.3 Voltage	28	volts
Grid-No.2 Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	1.7	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

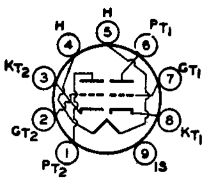
MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	0.5	megohm

6DT8

12DT8

HIGH-MU TWIN TRIODE



9AJ

Miniature type used in radio and television receiver applications and in push-pull rf amplifiers or as frequency converter in FM tuners. **Outlines section, 6B;** requires miniature 9-contact socket. Type 12DT8 is identical with type 6DT8 except for the heater ratings. Except for heater and heater-cathode ratings, interelectrode capacitances, and basing arrangement, these types are identical with miniature type 12AT7.

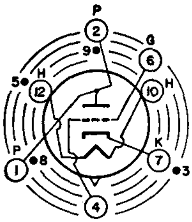
	6DT8	12DT8	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	0.3	0.15	ampere
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx., Each Unit Except as Noted:		
Grid to Plate	1.6*	pF
Grid to Cathode, Heater, and Internal Shield	2.7*	pF
Plate to Cathode, Heater, and Internal Shield	1.6*	pF
Heater to Cathode	3•	pF
Cathode to Grid, Heater, and Internal Shield (Unit No.2)	5.3†	pF
Plate to Grid, Heater, and Internal Shield (Unit No.2)	2.8†	pF

† With external shield connected to grid of unit under test.

• With external shield connected to ground.

* With external shield connected to cathode of unit under test.



12EA

HIGH-MU TRIODE

6DV4

2DV4

Nuvistor type used at frequencies up to 1000 MHz in uhf oscillator stages of color and black-and-white television receivers. Outlines section, 1; requires nuvistor socket. Type 2DV4 is identical with type 6DV4 except for heater ratings.

	2DV4	6DV4	
Heater Voltage (ac/dc)	2.1	6.3	volts
Heater Current	0.45	0.135	ampere
Heater Warm-up Time (Average)	8	—	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	volts

Direct Interelectrode Capacitance (Approx.):		
Grid to Plate	1.8	pF
Grid to Cathode, Heater, and Shell	4.4	pF
Plate to Cathode, Heater, and Shell	1.9	pF
Plate to Cathode	0.25	pF
Heater to Cathode	1.4	pF
Grid to Cathode	3.7	pF

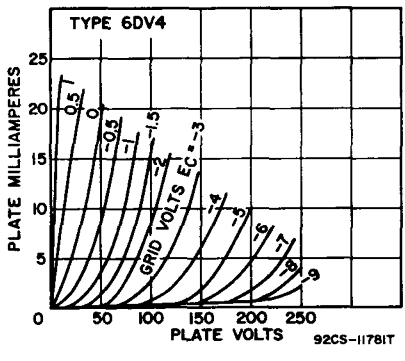
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	300	volts
Plate Voltage	125	volts
Grid Voltage:		
Negative-bias value	55	volts
Peak positive value	2	volts
Plate Dissipation	1	watt
Cathode Current	15	mA

CHARACTERISTICS

Plate Supply Voltage	75	volts
Cathode-Bias Resistor	100	ohms
Amplification Factor	35	
Plate Resistance (Approx.)	3100	ohms
Transconductance	11500	µmhos
Plate Current	10.5	mA
Grid Voltage (Approx.) for plate current of 10 µA	—7	volts



TYPICAL OPERATION AS OSCILLATOR AT 950 MHz

Plate Voltage	60	volts
Grid Voltage	—2	volts
Grid Resistor	5600	ohms
Plate Current	8	mA
Grid Current	350	µA

MAXIMUM CIRCUIT VALUES

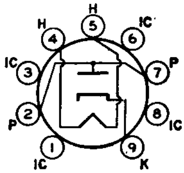
Grid-Circuit Resistance:°		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.2	megohm

° For operation at metal-shell temperatures up to 135°C.

6DW4 Refer to chart at end of section.

6DW4A
6DW4B
12DW4A

HALF-WAVE VACUUM RECTIFIER



9HP

Novar types used as damper tubes in horizontal-deflection circuits of color and black-and-white television receivers. Outlines section, 11D and 30B, respectively; require novar 9-contact socket. Socket terminals 1, 3, 6, and 8 should not be used as tie points; it is recommended that socket clips for these pins be removed to reduce the possibility of arc-over and to minimize leakage. These tubes, like other power-handling tubes, should be adequately ventilated. Type 12DW4A is identical with type 6DW4A except for heater ratings. Type 6DW4B is identical with type 6DW4A except for heater-cathode voltages.

	6DW4A 6DW4B	12DW4A	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	1.2	0.6	amperes
Heater Warm-up Time	—	11	seconds
Direct Interelectrode Capacitances (Approx.):			
Plate to Cathode and Heater		6.5	pF
Cathode to Plate and Heater		9	pF
Heater to Cathode		2.8	pF

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#			5500	volts
Peak Plate Current			1300	mA
Average Plate Current			250	mA
Plate Dissipation			8.5	watts
Heater-Cathode Voltage:		6DW4B	12DW4A 6DW4A	
Peak value	+300	—5000	—5500	volts
Average value	+100	—900	—900	volts

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 350 mA	25	volts

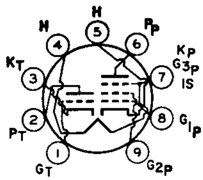
Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6DW5 Refer to chart at end of section.

6DX8 Refer to chart at end of section.

6DX8/ ECL84
10DX8/LCL84

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE



9HX

Miniature type used in color and black-and-white television-receiver applications. The triode unit is used as a sync-separator, sync-amplifier, keyed-agc, or noise-suppressor tube. The pentode unit is used as a video-output tube. Outlines section, 6E; requires miniature 9-contact socket. Type 10DX8/LCL84 is identical with type 6DX8/ECL84 except for heater ratings.

	6DX8/ECL84	10DX8/LCL84	
Heater Voltage (ac/dc)	6.3	10.2	volts
Heater Current	0.72	0.45	ampere
Peak Heater-Cathode Voltage	±200 max	±200 max	volts

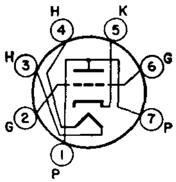
Class A₁ Amplifier

MAXIMUM RATINGS (Design-Center Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	550	volts
Peak Plate Voltage, with maximum plate current of 0.1 mA	600	—	volts
Plate Voltage	300	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	550	volts
Grid-No.2 Voltage	—	300	volts
Cathode Current	12	40	mA
Plate Dissipation	1	4	watts
Grid-No.2 Input	—	1.7	watts

CHARACTERISTICS	Triode Unit	Pentode Unit			
Plate Voltage	200	170	200	220	volts
Grid-No.2 Voltage	—	170	200	220	volts
Grid No.1 Voltage	—1.7	—2.1	—2.9	—3.4	volts
Amplification Factor	65	—	—	—	
Mu-Factor, Grid-No.2 to Grid-No.1	—	36	36	36	
Plate Resistance (Approx.)	—	0.1	0.13	0.15	megohm
Transconductance	4000	11000	10400	10000	μmhos
Plate Current	3	18	18	18	mA
Grid-No.2 Current	—	3	3	3	mA

MAXIMUM CIRCUIT VALUES	Triode Unit	Pentode Unit	
Grid-No.1- Circuit Resistance:			
For fixed-bias operation	1	1	megohm
For cathode-bias operation	3	2	megohms

• With maximum duty factor of 0.18 and maximum pulse duration of 18 microseconds.



7DK

MEDIUM-MU TRIODE 6DZ4

Miniature type used as a local-oscillator tube in uhf color and black-and-white television receivers covering the frequency range from 470 to 890 MHz. **Outlines section, 5B**; requires miniature 7-contact socket. For curve of average plate characteristics, refer to type 6AF4A.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.225	ampere
Heater-Cathode Voltage:		
Peak value	±50 max	volts
Average value	25 max	volts
Direct Interelectrode Capacitances (Approx.):°		
Grid to Plate	1.8	pF
Grid to Cathode and Heater	2.2	pF
Plate to Cathode and Heater	1.3	pF

° With external shield connected to cathode.

Class A₁ Amplifier

CHARACTERISTICS		
Plate Supply Voltage	80	volts
Plate Resistor	2700	ohms
Amplification Factor	14	
Plate Resistance (Approx.)	2000	ohms
Transconductance	6700	μmhos
Plate Current	15	mA
Grid Voltage (Approx.) for plate current of 20 μA	—11	volts

UHF Oscillator

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	135	volts
Grid Voltage, Negative-bias value	50	volts
Grid Current	2	mA
Cathode Current	20	mA
Plate Dissipation	2.3	watts

TYPICAL OPERATION AS OSCILLATOR AT 1000 MHz

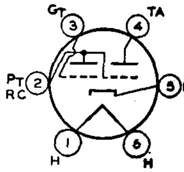
Plate Supply Voltage	135	volts
Plate-Circuit Resistance	2700	ohms
Grid Resistor	10000	ohms
Plate Current	15.5	mA
Grid Current (Approx.)	800	μA

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance:		
For fixed-bias operation	Not recommended	
For cathode-bias operation	0.5	megohm

6DZ7 Refer to chart at end of section.

6E5 ELECTRON-RAY TUBE



6R

Glass type used to indicate the effects of a change in a controlling voltage. It is used to indicate accurate radio-receiver tuning. **Outlines section,** 13H; requires 6-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.3. For additional considerations, refer to **Tuning Indication with Electron-Ray Tubes in Electron Tube Applications** section.

Tuning Indicator

MAXIMUM AND MINIMUM RATINGS (Design-Center Values)

Plate-Supply Voltage	250 max	volts
Target Voltage	250 max	volts
	125 min	volts

TYPICAL OPERATION

Plate and Target Supply Voltage	200	250	volts
Series Triode-Plate Resistor	1	1	megohm
Target Current*†	3	4	mA
Triode-Plate Current*	0.19	0.24	mA
Triode-Grid Voltage (Approx.):			
For shadow angle of 0°	—6.5	—8	volts
For shadow angle of 90°	0	0	volts

* For zero triode-grid voltage.

† Subject to wide variations.

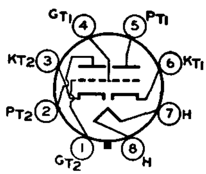
6E6 Refer to chart at end of section.

6E7 Refer to chart at end of section.

6EA4 Refer to chart at end of section.

6EA5 Refer to chart at end of section.

6EA7 DUAL TRIODE



8BD

Glass octal type used as a *combined* vertical-deflection oscillator and vertical-deflection amplifier in television receivers. **Outlines section,** 13B; requires octal socket. **Heater:** volts (ac/dc), 6.3; amperes, 1.05; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2		
Plate Voltage	250	60	175	volts
Grid Voltage	—3	0	—25	volts
Amplification Factor	66	—	5.5	
Plate Resistance (Approx.)	30000	—	920	ohms
Transconductance	2200	—	6000	μmhos
Plate Current	2	100•	40	mA

Grid Voltage (Approx.):				
For plate current of 20 μA	−5.3	—	—	volts
For plate current of 200 μA	—	—	−45	volts

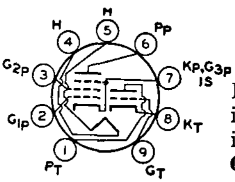
• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	350	550	volts
Peak Positive-Pulse Plate Voltage#	—	1500	volts
Peak Negative-Pulse Grid Voltage	400	250	volts
Peak Cathode Current	—	175	mA
Average Cathode Current	—	50	mA
Plate Dissipation	1	10	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For grid-resistor-bias operation	1	1	megohm
For cathode-bias operation	2.2	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).



9AE

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE 6EA8

5EA8, 9EA8, 19EA8

Miniature type used as combined oscillator and mixer in color and black-and-white television receivers utilizing an intermediate frequency in the order of 40 MHz. Outlines section, 6B; requires miniature 9-contact socket. Types 5EA8, 9EA8, and 19EA8 are identical with type 6EA8 except for heater ratings.

	5EA8	6EA8	9EA8	19EA8	
Heater Voltage (ac/dc)	4.7	6.3	9.5	18.9	volts
Heater Current	0.6	0.45	0.15	0.15	ampere
Heater Warm-up Time (Average)	11	11	11	11	seconds
Heater-Cathode Voltage:					
Peak value	±200 max	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	100 max	volts

	Unshielded	Shielded	
Direct Interelectrode Capacitances:			
Triode Unit:			
Grid to Plate	1.7	1.7	pF
Grid to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	3	3.2	pF
Plate to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	1.4	1.9	pF
Cathode to Heater	3	3▪	pF
Pentode Unit:			
Grid No.1 to Plate	0.02 max	0.01 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5	5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.6	3.4	pF
Heater to Cathode	3	3▪	pF

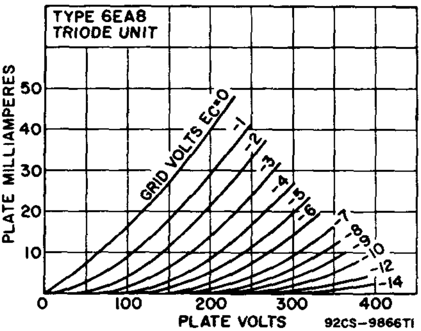
° With external shield connected to cathode of unit under test except as noted.

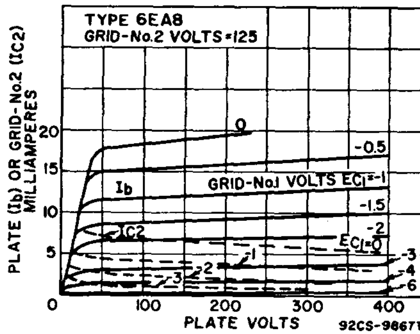
▪ With external shield connected to ground.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	3.1	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	150	125	volts
Grid-No.2 Voltage	—	125	volts
Grid-No.1 Voltage	—	−1	volt

	Triode Unit	Pentode Unit	
Cathode-Bias Resistor	56	—	ohms
Amplification Factor	40	—	
Plate Resistance (Approx.)	5000	200000	ohms
Transconductance	8500	6400	μmhos
Plate Current	18	12	mA
Grid-No.2 Current	—	4	mA
Grid-No.1 Voltage for plate current of 10 μA	—12	—9	volts

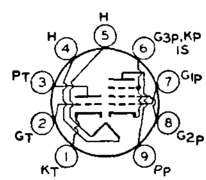




6EB8

8EB8

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE



9DX

Miniature type used in color and black-and-white television receiver applications. Pentode unit is used as video output amplifier; triode unit is used in sync-separator, sync-clipper, and phase-inverter circuits. **Outlines section, 6E**; requires miniature 9-contact socket. Type 8EB8 is identical with type 6EB8 except for heater ratings.

	6EB8	8EB8	
Heater Voltage (ac/dc)	6.3	8	volts
Heater Current	0.75	0.6	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	4.4	pF
Grid to Cathode and Heater	2.4	pF
Plate to Cathode and Heater	0.36	pF
Pentode Unit:		
Grid No.1 to Plate	0.1 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	11	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4.2	pF
Triode Grid to Pentode Plate	0.018 max	pF
Pentode Grid No.1 to Triode Plate	0.005 max	pF
Pentode Plate to Triode Plate	0.17 max	pF

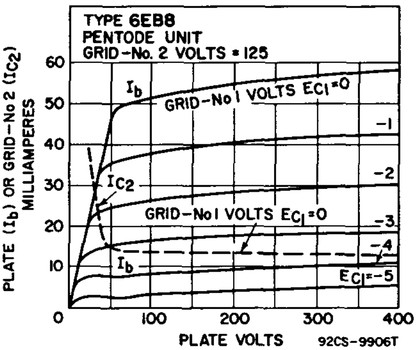
Class A₁ Amplifier

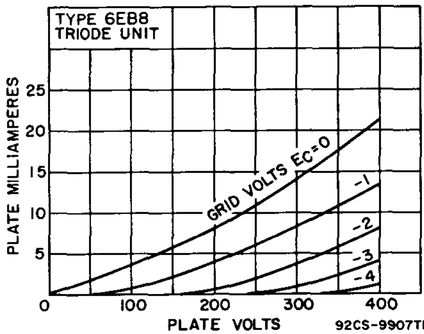
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	1	5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	250	200	volts
Grid-No.2 Supply Voltage	—	125	volts
Grid Voltage	—2	—	volts

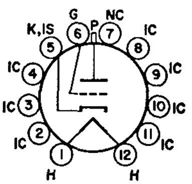
	Triode Unit	Pentode Unit	
Cathode-Bias Resistor	—	68	ohms
Amplification Factor	100	—	
Plate Resistance (Approx.)	37000	75000	ohms
Transconductance	2700	12500	µmhos
Plate Current	2	25	mA
Grid-No.2 Current	—	7	mA
Grid Voltage (Approx.) for plate current of 20 µA	—5	—	volts
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—	—9	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm







12FA

BEAM TRIODE 6EH4

Duodecar type used for the shunt regulation of low-current, high-voltage power supplies in color television receivers. Outlines section, 16E; requires duodecar 12-contact socket. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT. Heater: volts (ac/dc), 6.3; amperes, 0.2; maximum heater-cathode volts; + not recommended, —450.

Shunt Voltage-Regulator Service

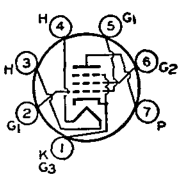
MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	27000	volts
Unregulated Plate-Supply Voltage	60000	volts
DC Grid Voltage	—135	volts
Peak Grid Voltage#	—440	volts
Average DC Plate Current	1.6	mA
Plate Dissipation	30	watts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance:		
For use with "Flyback Transformer" high-voltage supply	3	megohms

Peak value for duration of 20 seconds maximum during equipment warm-up.



7CV

POWER PENTODE 6EH5

25EH5, 50EH5

Miniature type used in the audio output stage of radio and television receivers and in phonographs. Outlines section, 5D; requires miniature 7-contact socket. Types 25EH5 and 50EH5 are identical with type 6EH5 except for heater ratings.

	6EH5	25EH5	50EH5	
Heater Voltage (ac/dc)	6.3	25	50	volts
Heater Current	1.2	0.3	0.15	ampere
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.65	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	17	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	9	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Plate Dissipation	5.5	watts
Grid-No.2 Input	2	watts
Bulb Temperature (at hottest point)	220	°C

TYPICAL OPERATION

Plate Supply Voltage	110	volts
Grid-No.2 Supply Voltage	115	volts
Cathode-Bias Resistor	62	ohms
Peak AF Grid-No.1 Voltage	3	volts
Zero-Signal Plate Current	42	mA
Maximum-Signal Plate Current	42	mA
Zero-Signal Grid-No.2 Current	11.5	mA
Maximum-Signal Grid-No.2 Current	14.5	mA
Plate Resistance (Approx.)	11000	ohms
Transconductance	14600	μmhos
Load Resistance	3000	ohms
Total Harmonic Distortion	7	per cent
Maximum-Signal Power Output	1.4	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

Push-Pull Class AB_1 Audio-Frequency Power Amplifier

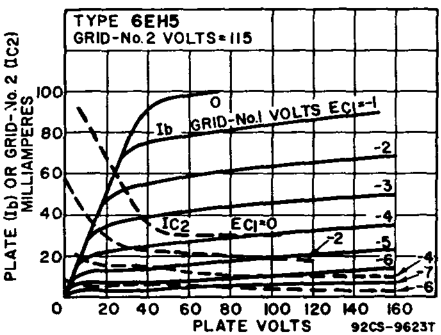
MAXIMUM RATINGS (Same as for Class A_1 audio-frequency power amplifier)

TYPICAL OPERATION (Values are for two tubes)

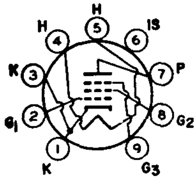
Plate Supply Voltage	140	volts
Grid-No.2 Supply Voltage	120	volts
Cathode-Bias Resistor	68	ohms
Peak AF Grid-No.1 Voltage	9.4	volts
Zero-Signal Plate Current	47	mA
Maximum-Signal Plate Current	51	mA
Zero-Signal Grid-No.2 Current	11	mA
Maximum-Signal Grid-No.2 Current	17.7	mA
Effective Load Resistance (Plate-to-plate)	6000	ohms
Total Harmonic Distortion	5	per cent
Maximum-Signal Power Output	3.8	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm



Refer to chart at end of section. **6EH7**



9AQ

SEMIREMOTE-CUTOFF PENTODE

6EH7/ EF183

3EH7/XF183, 4EH7

Miniature types used as if-amplifier tubes in color and black-and-white television receivers. Outlines section, 6C; requires miniature 9-contact socket. Types 3EH7/XF183 and 4EH7 are identical with type 6EH7/EF183 except for heater ratings.

	3EH7/XF183	4EH7	6EH7/EF183	
Heater Voltage (ac/dc)	3.4	4.4	6.3	volts
Heater Current	0.6	0.45	0.3	ampere
Peak Heater-Cathode Voltage	±150 max	±150 max	±150 max	volts
Direct Interelectrode Capacitances:				
Grid No.1 to Plate			0.005 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield			9	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield			3	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Supply Voltage	550	volts
Plate Voltage	250	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	550	volts
Grid-No.2 Voltage	250	volts
Cathode Current	20	mA
Plate Dissipation	2.5	watts
Grid-No.2 Input	0.65	watt

CHARACTERISTICS

Plate Voltage	200	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Voltage	90	volts
Grid-No.1 Voltage	—2	volts
Plate Resistance (Approx.)	0.5	megohm
Transconductance	12500	µmhos
Plate Current	12	mA
Grid-No.2 Current	4.5	mA

TYPICAL OPERATION

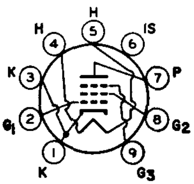
Plate Voltage	200	200	200	200	volts
Grid No.3	Connected to cathode at socket				
Grid-No.2 Supply Voltage	200	200	200	200	volts
Grid-No.2 Series Resistor	22000	22000	22000	22000	ohms
Grid-No.1 Voltage	—19.5	—9.5	—6.5	—2	volts
Transconductance	125	625	1250	12500	µmhos
RMS Grid-No.1 Voltage, for cross-modulation factor of 0.01	450	160	100	—	mV

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	1	megohm

Refer to chart at end of section. **6EH8**

Refer to chart at end of section. **6EJ7**



9AQ

SHARP-CUTOFF PENTODE

6EJ7/ EF184

3EJ7/XF184, 4EJ7

Miniature types used as if-amplifier tubes in color and black-and-white television receivers. Outlines section, 6C; requires miniature 9-contact socket. Types 3EJ7/XF184 and 4EJ7 are identical with type 6EJ7/EF184 except for heater ratings.

	3EJ7/XF184	4EJ7	6EJ7/EF184	
Heater Voltage (ac/dc)	3.4	4.4	6.3	volts
Heater Current	0.6	0.45	0.3	ampere
Peak Heater-Cathode Voltage	±150 max	±150 max	±150 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.005 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	10	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Supply Voltage	550	volts
Plate Voltage	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	550	volts
Grid-No.2 Voltage	250	volts
Cathode Current	25	mA
Plate Dissipation	2.5	watts
Grid-No.2 Input	0.9	watt

CHARACTERISTICS

Plate Voltage	190	200	volts
Grid No.3	Connected to cathode at socket		
Grid-No.2 Voltage	190	200	volts
Grid-No.1 Voltage	—2.35	—2.5	volts
Plate Resistance (Approx.)	0.35	0.35	megohm
Transconductance	15000	15000	μmhos
Plate Current	10	10	mA
Grid-No.2 Current	4.1	4.1	mA

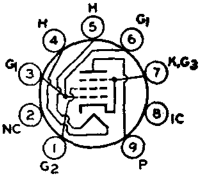
MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	1	megohm

6EM5

8EM5

BEAM POWER TUBE



9HN

Miniature type used as vertical-deflection amplifier in television receivers utilizing picture tubes having diagonal deflection angles of 110 degrees. Outlines section, 6G; requires miniature 9-contact socket. Type 8EM5 is identical with type 6EM5 except for heater ratings.

	6EM5	8EM5	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.8	0.6	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.7 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	10	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	5.1	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	60	250	volts
Grid-No.2 Voltage	250	250	volts
Grid-No.1 Voltage	0	—18	volts
Mu Factor, Grid No.1 to Grid No.2	—	8.7	
Plate Resistance	—	0.05	megohm
Transconductance	—	5100	μmhos
Plate Current	180•	40	mA
Grid-No.2 Current	30•	3	mA
Grid-No.1 Voltage (Approx.) for plate current of 0.2 mA	—	—37	volts

• These values can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

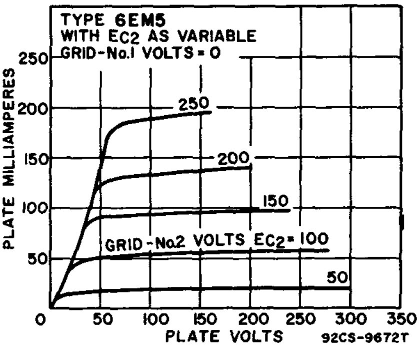
Vertical-Deflection Oscillator and Amplifier

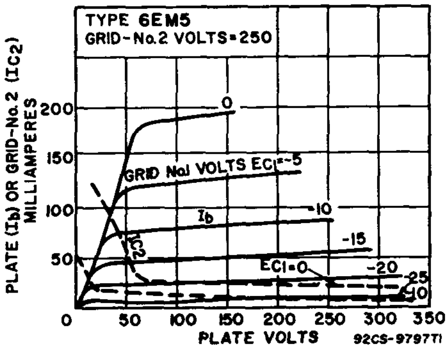
For operation in a 525-line, 30-frame system

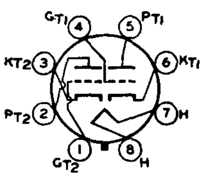
MAXIMUM RATINGS (Design-Center Values)

DC Plate Voltage	315	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	2200▲	volts
Grid-No.2 (Screen-Grid) Voltage	285	volts
Peak Negative-Pulse Grid-No.1 (Control-Grid) Voltage	250	volts
Peak Cathode Current	210	mA
Average Cathode Current	60	mA
Plate Dissipation	10	watts
Grid-No.2 Input	1.5	watts
Bulb Temperature (At hottest point)	250	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).
▲ Under no circumstances should this absolute value be exceeded.







8BD

6EM7

DUAL TRIODE

10EM7, 13EM7/15EAY

Glass octal type used as combined vertical-deflection amplifier and vertical-deflection oscillator in color and black-and-white television receivers. Outlines section, 13A; requires octal socket. For curve of average plate characteristics, Unit No.1, refer to type 6DR7 (Unit No.1). Types 10EM7 and 13EM7/15EAY are identical with type 6EM7 except for heater ratings.

	6EM7	10EM7	13EM7/15EAY	
Heater Voltage (ac/dc)	6.3	9.7	13	volts
Heater Current	0.925	0.6	0.45	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	4.8	10	pF
Grid to Cathode and Heater	2.2	7	pF
Plate to Cathode and Heater	0.6	1.8	pF

Class A₁ Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2	
Plate Voltage	250	150	volts
Grid Voltage	—3	—20	volts
Amplification Factor	64	5.4	
Plate Resistance (Approx.)	40000	750	ohms
Transconductance	1600	7200	μmhos
Plate Current	1.4	50	mA
Plate Current, for plate voltage of 60 volts and zero grid voltage	—	95	mA

	Unit No.1	Unit No.2	
Plate Current, for grid voltage of —28 volts	—	10	mA
Grid Voltage (Approx.):			
For plate current of 10 µA	—5.5	—	volts
For plate current of 100 µA	—	—45	volts

Vertical-Deflection Oscillator and Amplifier

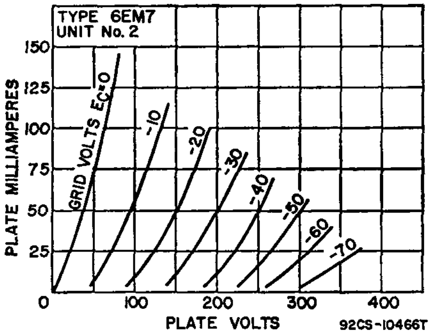
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	330	330	volts
Peak Positive-Pulse Plate Voltage#	—	1500	volts
Peak Negative-Pulse Grid Voltage	400	250	volts
Peak Cathode Current	77	175	mA
Average Cathode Current	22	50	mA
Plate Dissipation	1.5	10	watts

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance:	Unit No.1	Unit No.2	
For grid-resistor-bias operation	2.2	2.2	megohms
For cathode-bias operation	2.2	2.2	megohms

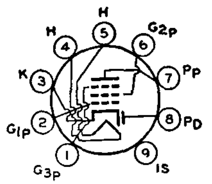
Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).



6EQ7

12EQ7

DIODE— REMOTE-CUTOFF PENTODE



9LQ

Miniature type used as combined if amplifier and AM detector in AM and AM/FM radio receivers. **Outlines section,** 6E; requires miniature 9-contact socket. Type 12EQ7 is identical with type 6EQ7 except for heater ratings.

	6EQ7	12EQ7	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	0.3	0.15	ampere
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Pentode Unit:		
Grid No.1 to Plate	0.002 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5	pF
Pentode Grid No.1 to Diode Plate	0.0015 max	pF
Pentode Plate to Diode Plate	0.095	pF

Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	300	volts
Grid-No.3 (Suppressor-Grid) Voltage:		
Positive value	300	volts
Negative value	300	volts

Grid-No.2 (Screen-Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:		
Positive-bias value	0	volts
Negative-bias value	50	volts
Plate Dissipation	3	watts
Grid-No.3 Input	0.2	watt
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	0.6	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	
Bulb Temperature (At hottest point)	150	°C

CHARACTERISTICS

Plate Voltage	100	volts
Grid No.3	Connected to cathode at socket	
Internal Shield	Connected to cathode at socket	
Grid-No.2 Voltage	100	volts
Grid-No.1 Supply Voltage	0	volts
Grid-No.1 Resistor (Bypassed)	2.2	megohms
Plate Resistance (Approx.)	0.25	megohm
Transconductance	3800	µmhos
Plate Current	9	mA
Grid-No.2 Current	3.5	mA
Grid-No.1 Voltage (Approx.) for transconductance of 40 µmhos	—20	volts

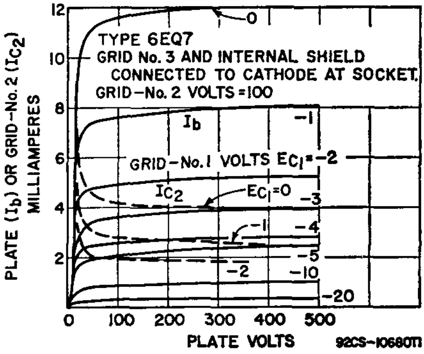
Diode Unit

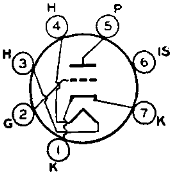
MAXIMUM RATINGS (Design-Maximum Values)

Plate Current	1	mA

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 2 mA	10	volts





7FP

6ER5

HIGH-MU TRIODE

2ER5, 3ER5

Miniature type with frame grid used in vhf tuners of color and black-and-white television receivers. **Outlines section, 5C;** requires miniature 7-contact socket. Types 2ER5 and 3ER5 are identical with type 6ER5 except for heater ratings.

	2ER5	3ER5	6ER5	
Heater Voltage (ac/dc)	2.3	2.8	6.3	volts
Heater Current	0.6	0.45	0.18	ampere
Peak Heater-Cathode Voltage	±100 max	±100 max	±100 max	volts

Direct Interelectrode Capacitances:	Unshielded	Shielded°	
Grid to Plate	0.38	0.36	pF
Grid to Cathode, Heater, and Internal Shield	4.4	4.4	pF
Plate to Cathode, Heater, and Internal Shield	3	4	pF
Grid to Heater	0.28 max	0.28 max	pF
Plate to Cathode	0.24	0.2▵	pF
Cathode to Grid	3.1	3.1▵	pF
Heater to Cathode	2.5	2.5▵	pF

° With external shield connected to cathode except as noted.

▵ With external shield connected to ground.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	250	volts
Grid Voltage, *Negative-bias* value	50	volts
Cathode Current	20	mA
Plate Dissipation	2.2	watts

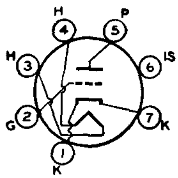
CHARACTERISTICS

Plate Voltage	200	volts
Grid Voltage	—1.2	volts
Amplification Factor	80	
Plate Resistance (Approx.)	8000	ohms
Transconductance	10500	μmhos
Plate Current	10	mA
Grid Voltage (Approx.) for transconductance of 500 μmhos	—3.8	volts
Grid Voltage (Approx.) for transconductance of 100 μmhos	—5.6	volts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance	1	megohm

6ES5 HIGH-MU TRIODE



7FP

Miniature type used as grounded-cathode rf amplifier in vhf television receivers. Outlines section, 5C; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.2	ampere
Peak Heater-Cathode Voltage	±100 max	volts

Direct Interelectrode Capacitances:	Unshielded	Shielded	
Grid to Plate	0.5 max	0.5 max	pF
Grid to Cathode, Heater, and Internal Shield	3.2	3.2	pF
Plate to Cathode, Heater, and Internal Shield	3.2	4	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	250	volts
Grid Voltage, Positive-bias value	0	volts
Cathode Current	22	mA
Plate Dissipation	2.2	watts

CHARACTERISTICS

Plate Voltage	200	volts
Grid Voltage	—1	volt
Amplification Factor	75	
Plate Resistance (Approx.)	8000	ohms
Transconductance	9000	μmhos
Plate Current	10	mA
Grid Voltage (Approx.) for plate current of 100 μA	—6	volts

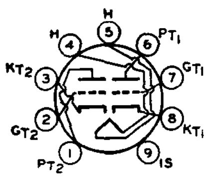
MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance	1	megohm

6ES8 Refer to chart at end of section.

6ES8/ ECC189 VARIABLE-MU TWIN TRIODE

4ES8/XCC189



9AJ

Miniature type used as cascode-type amplifier in tuners of television receivers. Outlines section, 6B; requires miniature 9-contact socket. Type 4ES8/XCC189 is identical with type 6ES8/ECC189 except for heater ratings.

	4ES8/ XCC189	6ES8/ECC189	
Heater Voltage (ac/dc)	4	6.3	volts
Heater Current	0.6	0.365	ampere
Heater Warm-up Time (Average)	11	—	seconds

Direct Interelectrode Capacitances:	Unshielded	Shielded*	
Grid to Plate (Each Unit)	1.9	1.9	pF
Plate to Cathode (Each Unit)	0.18	0.17	pF
Heater to Cathode (Each Unit)	3	3▵	pF
Plate of Unit No.2 to Plate of Unit No.1	0.04 max	0.015 max	pF
Plate of Unit No.2 to Grid of Unit No.1	0.003 max	0.003 max	pF
Grid of Unit No.1 to Cathode of Unit No.2	0.002 max	0.002 max	pF

* With external shield connected to cathode of unit under test except as noted.

▵ With external shield connected to ground.

Class A_1 Amplifier (Each Unit)

CHARACTERISTICS

Plate Voltage	90	90	90	volts
Grid Voltage	—1.2	—5	—9	volts
Plate Resistance (Approx.)	2500	—	—	ohms
Transconductance	12500	625	125	µmhos
Plate Current	15	—	—	mA

Cascode-Type Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage with plate current of 0 mA	550	volts
Plate Voltage (Each unit)	130	volts
Grid Voltage, Negative-bias value (Each unit)	50	volts
Cathode Current (Each unit)	22	mA
Plate Dissipation (Each unit)	1.8	watts
Heater-Cathode Voltage:		
Unit No.1:°		
RMS voltage between cathode and heater	50	volts
Unit No.2:▪		
RMS voltage between cathode and heater•	50	volts
DC voltage between cathode and heater•	130	volts

TYPICAL OPERATION in a cascode-type circuit■

Supply Voltage	180	volts
Plate Current	15	mA
Transconductance	12500	µmhos
Noise Figure*	6.5	dB
Grid Voltage (Approx.) for transconductance of 125 µmhos	—9	volts
Input Voltage for cross-modulation factor of 0.01 and transconductance of 125 µmhos	500	mV

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance (Each unit)	1	megohm

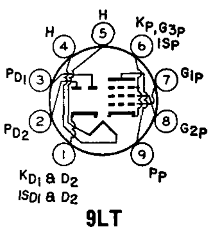
° Grounded-cathode input unit—pins 6, 7, and 8.

▪ Grounded-grid output unit—pins 1, 2, and 3.

• Cathode positive with respect to heater.

■ With grid of output unit connected to a voltage divider.

* Measured with tube operating in a television tuner.



9LT

TWIN DIODE— SHARP-CUTOFF PENTODE 6ET7

8ET7

Miniature type used in television receiver applications. The pentode unit is used as a video amplifier and the diodes are used as a horizontal phase inverter. Outlines section, 6E; requires miniature 9-contact socket. Type 8ET7 is identical with type 6ET7 except for heater ratings.

	6ET7	8ET7	
Heater Voltage (ac/dc)	6.3	8	volts
Heater Current	0.75	0.6	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	

Grid-No.1 (Control-Grid) Voltage, Positive-bias value		0	volts
Plate Dissipation		5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts		1.1	watts
For grid-No.2 voltages between 165 and 330 volts		See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	60	200	volts
Grid-No.2 Supply Voltage	150	150	volts
Grid-No.1 Voltage	0	—	volts
Cathode-Bias Resistor	—	100	ohms
Plate Resistance (Approx.)	—	60000	ohms
Transconductance	—	11500	μmhos
Plate Current	55▪	25	mA
Grid-No.2 Current	18▪	5.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—10	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.25	megohm

▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Diode Units (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

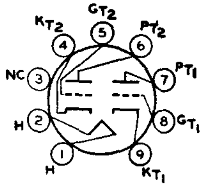
Average Plate Current	3	mA

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 1.5 mA	10	volts

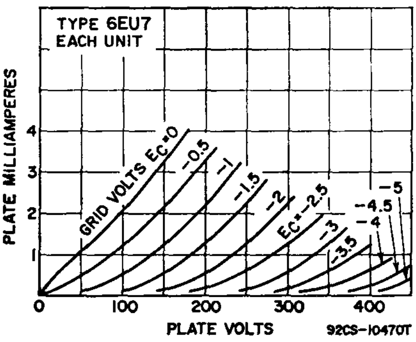
6EU7 HIGH-MU TWIN TRIODE

Miniature type used in high-gain, resistance-coupled, low-level audio-amplifier applications where low-hum and non-microphonic characteristics are important, such as microphone amplifiers and pre-amplifiers for phonographs. **Outlines section, 6B**; requires miniature 9-contact socket. For typical operation as a resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section.



9LS

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.3	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Each Unit, Approx.):		
Grid to Plate	1.5	pF
Grid to Cathode and Heater	1.6	pF
Plate to Cathode and Heater	0.2	pF



Equivalent Noise and Hum Voltage (Referenced to Grid, Each Unit):

Average Value*	1.8 microvolts rms	

* Measured in "true rms" units under the following conditions: Heater volts (ac), 6.3; center-tap of heater transformer grounded; plate supply volts, 250; plate load resistor, 100000 ohms; cathode resistor, 2700 ohms; cathode bypass capacitor, 100 μF; grid resistor, 0 ohms; amplifier frequency range, 25 to 10000 Hz.

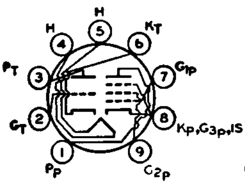
Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid Voltage:		
Negative-bias value	55	volts
Positive-bias value	0	watts
Plate Dissipation	1.2	watts

CHARACTERISTICS

Plate Voltage	100	250	volts
Grid Voltage	—1	—2	volts
Amplification Factor	100	100	
Plate Resistance (Approx.)	80000	62500	ohms
Transconductance	1250	1600	μmhos
Plate Current	0.5	1.2	mA



9JF

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

6EU8

5EU8

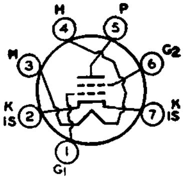
Miniature type used as combined triode oscillator and pentode mixer in television receivers. **Outlines section, 6B**; requires miniature 9-contact socket. Type 5EU8 is identical with type 6EU8 except for heater ratings.

	5EU8	6EU8	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	3	3.1	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	150	125	volts
Grid-No.2 Supply Voltage	—	125	volts
Grid-No.1 Voltage	—	—1	volt
Cathode-Bias Resistor	56	—	ohms
Amplification Factor	40	—	
Plate Resistance (Approx.)	5000	80000	ohms
Transconductance	8500	6400	μmhos
Plate Current	18	12	mA
Grid-No.2 Current	—	4	mA
Cathode Warm-up Time■	35	—	seconds
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—12	—9	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance	0.1	0.1	megohm

■ The cathode warm-up time is defined as the time required for the transconductance to reach 6500 μmhos when the tube is operated from a cold start with dc plate volts = 100, grid volts = 0, and heater volts = 5.5.



6EV5

SHARP-CUTOFF TETRODE

7EW

Miniature type used as rf amplifier in vhf tuners of television receivers. **Outlines section,** 5C; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.2	ampere
Heater-Cathode Voltage:		
Peak value	±100 max	volts
Average value	50 max	volts
Direct Interelectrode Capacitances:△		
Grid No.1 to Plate	0.035 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Internal Shield	4.5	pF
Plate to Cathode, Heater, Grid No.2, and Internal Shield	2.9	pF

△ With external shield connected to cathode.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	275	volts
Grid-No.2 (Screen-Grid) Supply Voltage	180	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Cathode Current	20	mA
Plate Dissipation	3.25	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 90 volts	0.2	watt
For grid-No.2 voltages between 90 and 180 volts	See curve page 98	

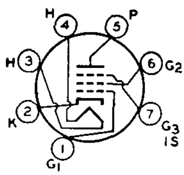
CHARACTERISTICS

Plate Voltage	250	volts
Grid-No.2 Voltage	80	volts
Grid-No.1 Voltage	—1	volt
Plate Resistance (Approx.)	0.15	megohm
Transconductance	8800	µmhos
Plate Current	11.5	mA
Grid-No.2 Current	0.9	mA
Grid-No.1 Voltage (Approx.) for transconductance of 100 µmhos	—4.5	volts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	0.5	megohm

6EV7 Refer to chart at end of section.



6EW6

4EW6, 5EW6

SHARP-CUTOFF PENTODE

7CM

Miniature type used in the gain-controlled picture-if stages of vhf color and black-and-white television receivers operating at an interemediate frequency in the order of 40 MHz. **Outlines section,** 5C; requires miniature 7-contact socket. Types 4EW6 and 5EW6 are identical with type 6EW6 except for heater ratings.

	4EW6	5EW6	6EW6	
Heater Voltage (ac/dc)	4.2	5.6	6.3	volts
Heater Current	0.6	0.45	0.4	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

	Unshielded	Shielded*	
Direct Interelectrode Capacitances:			
Grid No.1 to Plate	0.04 max	0.03 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	10	10	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.4	3.4	pF

* With external shield connected to cathode.

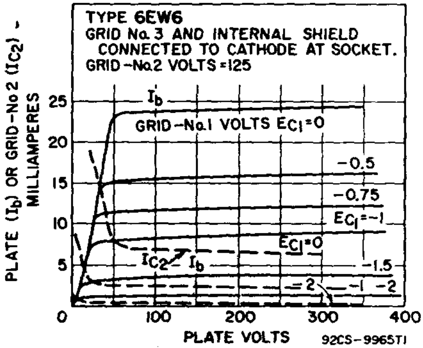
Class A_1 Amplifier

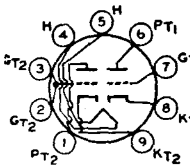
MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	3.1	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.65	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	125	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.2	megohm
Transconductance	14000	μmhos
Plate Current	11	mA
Grid-No.2 Current	3.2	mA
Grid-No 1 Voltage (Approx.) for plate current of 20 μA	—3.5	volts





9HF

6EW7

DUAL TRIODE

10EW7, 15EW7

Neonoval type used as combined vertical-deflection oscillator and vertical-deflector amplifier in television receivers. **Outlines section, 10C; requires neonoval 9-**contact socket. For curve of average plate characteristics, Unit No.1, refer to type 6DE7 (Unit No.1). Types 10EW7 and 15EW7 are identical with type 6EW7 except for heater ratings.

	6EW7	10EW7	15EW7	
Heater Voltage (ac/dc)	6.3	9.7	14.8	volts
Heater Current	0.9	0.6	0.45	ampere
Heater Warm-up Time	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	4.2	9	pF
Grid to Cathode and Heater	2.2	7	pF
Plate to Cathode and Heater	0.4	1.2	pF

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2	
Plate Voltage	250	150	volts
Grid Voltage	—11	—17.5	volts
Amplification Factor	17.5	6	
Plate Resistance (Approx.)	8750	800	ohms
Transconductance	2000	7500	μmhos
Plate Current	5.5	45	mA

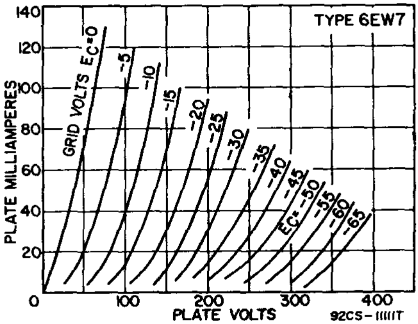
	Unit No.1	Unit No.2	
Plate Current for plate voltage of 60 volts and zero grid voltage	—	95	mA
Plate Current for grid voltage of —25 volts	—	8	mA
Grid Voltage (Approx.) for plate current of 10 μA	—20	—	volts
Grid Voltage (Approx.) for plate current of 100 μA	—	—40	volts

Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	330	330	volts
Peak Positive-Pulse Plate Voltage#	—	1500	volts
Peak Negative-Pulse Grid Voltage	400	250	volts
Peak Cathode Current	77	175	mA
Average Cathode Current	22	50	mA
Plate Dissipation	1.5	10	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For cathode-bias operation	2.2	2.2	megohms
For grid-resistor-bias operation	2.2	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).



6EX6 Refer to chart at end of section.

6EY6 Refer to chart at end of section.

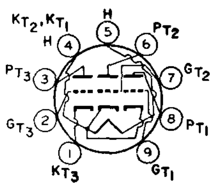
6EZ5 Refer to chart at end of section.

6EZ8

19EZ8

HIGH-MU TRIPLE TRIODE

Miniature type used in oscillator-mixer and afc service in FM receivers. **Outlines section, 6B**; requires miniature 9-contact socket. Type 19EZ8 is identical with type 6EZ8 except for heater ratings.



9KA

	6EZ8	19EZ8	
Heater Voltage (ac/dc)	6.3	18.9	volts
Heater Current	0.45	0.15	ampere
Peak Heater-Cathode Voltage		±100 max	volts

Class A_1 Amplifier (Each Unit Unless Otherwise Specified)

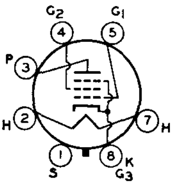
MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	330	volts
Grid Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate Dissipation	2	watts
Total Plate Dissipation (All plates)	5	watts

CHARACTERISTICS

Plate Voltage	125	volts
Grid Voltage	1	volt
Amplification Factor	57	
Plate Resistance (Approx.)	13600	ohms
Transconductance	4200	μmhos
Plate Current	4.2	mA
Grid Voltage (Approx.) for plate current of 20 μA	—4	volts

Refer to chart at end of section. **6F5**

Refer to chart at end of section. **6F5GT**



7S

POWER PENTODE **6F6**

Metal type used in the audio output stage of ac receivers. **Outlines section,** 2B; requires octal socket. This tube, like other power-handling tubes, should be adequately ventilated. **Heater:** volts (ac/dc), 6.3; amperes, 0.7; maximum heater-cathode volts, 90 peak.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Pentode Connection	Triode Connection▲	
Plate Voltage	375	350	volts
Grid-No.2 (Screen-Grid) Voltage	285	—	volts
Plate Dissipation	11	10	watts
Grid-No.2 Input	3.75	—	watts

TYPICAL OPERATION

Plate Voltage	250	285	250	volts
Grid-No.2 Voltage	250	285	—	volts
Grid-No.1 (Control-Grid) Voltage	—16.5	—20	—20	volts
Peak AF Grid-No.1 Voltage	16.5	20	20	volts
Zero-Signal Plate Current	34	38	31	mA
Maximum-Signal Plate Current	36	40	34	mA
Zero-Signal Grid-No.2 Current	6.5	7	—	mA
Maximum-Signal Grid-No.2 Current	10.5	13	—	mA
Amplification Factor	—	—	6.8	
Plate Resistance (Approx.)	80000	78000	2600	ohms
Transconductance	2500	2550	2600	μmhos
Load Resistance	7000	7000	4000	ohms
Total Harmonic Distortion	8	9	6.5	per cent
Maximum-Signal Power Output	3.2	4.8	0.85	watts

▲ Grid No.2 connected to plate.

Push-Pull Class AB_1 Amplifier

MAXIMUM RATINGS (Same as for class A_1 amplifier)

TYPICAL OPERATION (Values are for two tubes)

Plate Voltage	315	volts
Grid-No.2 Voltage	285	volts
Grid-No.1 Voltage	—24	volts
Peak AF Grid-No.1-to-Grid-No.1 Voltage	48	volts
Zero-Signal Plate Current	62	mA
Maximum-Signal Plate Current	80	mA
Zero-Signal Grid-No.2 Current	12	mA
Maximum-Signal Grid-No.2 Current	19.5	mA
Effective Load Resistance (Plate-to-plate)	10000	ohms
Total Harmonic Distortion	4	per cent
Maximum-Signal Power Output	11	watts

MAXIMUM CIRCUIT VALUES

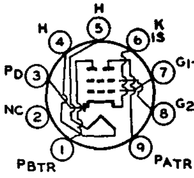
Grid-No.1 Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

Refer to chart at end of section. **6F6G** **6F6GT**

Refer to chart at end of section. **6F7**

Refer to chart at end of section. **6F8G**

6FA7 DIODE—SHARP-CUTOFF, TWIN-PLATE TETRODE



9MR

Miniature type used in television receivers and in frequency-divider and complex-wave generator circuits of electronic musical instruments. Outlines section, 6E; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.3	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Tetrode Unit:		
Grid No.1 to Plate A	0.040	pF
Grid No.1 to Plate B	0.030 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Internal Shield	5.5	pF
Plate A to Cathode, Heater, Grid No.2, and Internal Shield	1.8	pF
Plate B to Cathode, Heater, Grid No.2, and Internal Shield	1.8	pF
Tetrode Grid No.1 to Diode Plate	0.022	pF
Tetrode Plate A to Diode Plate	0.020 max	pF
Tetrode Plate B to Diode Plate	0.055	pF

Class A_1 Amplifier

CHARACTERISTICS (Tetrode Unit)

Plate A and Plate B connected together

Plate Voltage	100	volts
Grid-No.2 Voltage	100	volts
Grid-No.1 Supply Voltage	0	volts
Grid-No.1 Resistor (Bypassed)	2.2	megohms
Plate Resistance (Approx.)	90000	ohms
Transconductance	3200	μmhos
Plate Current	3.8	mA
Grid-No.2 Current	1.7	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—4	volts

Using either Plate A or B, with unused plate grounded

Plate Voltage	100	volts
Grid-No.2 Voltage	100	volts
Grid-No.1 Supply Voltage	0	volts
Grid-No.1 Resistor (Bypassed)	2.2	megohms
Plate Resistance (Approx.)	130000	ohms
Transconductance	1900	μmhos
Plate Current	2.2	mA
Grid-No.2 Current	3	mA

Frequency Divider and Complex-Wave Generator

Tetrode Unit

MAXIMUM RATINGS (Design-Maximum Values)

Plate-A Voltage	330	volts
Plate-B Voltage	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate-A Dissipation	1.5	watts
Plate-B Dissipation	1.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.65	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, for grid-No.1-resistor-bias operation	2.2	megohms

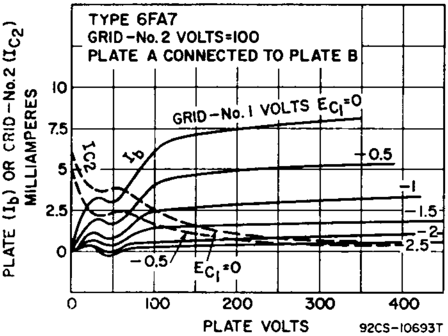
Diode Unit

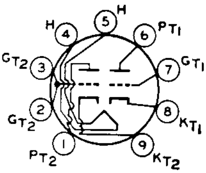
MAXIMUM RATINGS (Design-Maximum Values)

Plate Current	1	mA

CHARACTERISTICS, *Instantaneous Value*

Tube Voltage Drop for plate current of 2 mA	10	volts





9HF

6FD7
13FD7

DUAL TRIODE

Glass type containing high-mu and low-mu triode units used as combined vertical-deflection oscillator and vertical-deflection amplifier in television receivers. Outlines section, 10B; requires miniature 9-contact socket. Type 13FD7 is identical with type 6FD7 except for heater ratings.

	6FD7	13FD7	
Heater Voltage (ac/dc)	6.3	13	volts
Heater Current	0.925	0.45	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:	±200 max	±200 max	volts
Peak value	100 max	100 max	volts
Average value			

Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	4.5	10	pF
Grid to Cathode and Heater	2.2	6.5	pF
Plate to Cathode and Heater	0.4	0.2	pF

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2		
Plate Voltage	250	60	150	volts
Grid Voltage	—3	0	—17.5	volts
Amplification Factor	64	—	6	
Plate Resistance (Approx.)	40000	—	800	ohms
Transconductance	1600	—	7500	μmhos
Plate Current	1.5	95▪	40	mA
Grid Voltage (Approx.):				
For plate current of 10 μA	—5.5	—	—	volts
For plate current of 100 μA	—	—	—40	volts
Transconductance, For plate current of 1 mA	—	—	500	μmhos
Plate Current, For grid voltage of —25 volts	—	—	6	mA

▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	330	330	volts
Peak Positive-Pulse Plate Voltage#	—	1500	volts
Peak Negative-Pulse Grid Voltage	400	250	volts
Peak Cathode Current	70	175	mA
Average Cathode Current	20	50	mA
Plate Dissipation	1.5	10	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For grid-resistor-bias or cathode-bias operation	2.2	2.2	megohms

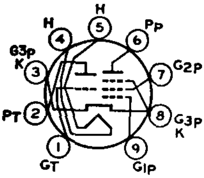
Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

6FE5 Refer to chart at end of section.

6FG6 Refer to chart at end of section.

6FG7
5FG7

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE



9GF

Miniature type used as combined oscillator and mixer tube in vhf color and black-and-white television receivers. Outlines section, 6B; requires miniature 9-contact socket. Type 5FG7 is identical with type 6FG7 except for heater ratings.

	5FG7	6FG7	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:			
Triode Unit:			
Grid to Plate	1.8	1.8	pF
Grid to Cathode, Pentode Grid No.3, and Heater	3	3	pF
Plate to Cathode, Pentode Grid No.3, and Heater	1.3	1.9	pF
Pentode Unit:			
Grid No.1 to Plate	0.02 max	0.01 max	pF
Grid No.1 to Cathode, Grid No.3, Grid No.2, and Heater	5	5	pF
Plate to Cathode, Grid No.3, Grid No.2, and Heater	2.4	3.4	pF
Heater to Cathode, and Pentode Grid No.3	6	6▪	pF

• With external shield connected to cathode except as noted.
▪ With external shield connected to ground.

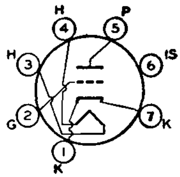
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	See curve page 98	
For grid-No.2 voltages between 165 and 330 volts	—	0.55	watt

CHARACTERISTICS	Triode Unit	Pentode Unit		
Plate Voltage	125	100	125	volts
Grid-No.2 Voltage	—	100	125	volts
Grid-No.1 Voltage	—1	0	—1	volts
Amplification Factor	43	—	—	
Plate Resistance (Approx.)	5700	—	180000	ohms
Transconductance	7500	7400	6000	µmhos
Plate Current	13	—	11	mA
Grid-No.2 Current	—	—	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 30 µA	—6.5	—	—7.5	volts

6FH5
2FH5, 3FH5

HIGH-MU TRIODE



7FP

Miniature type used as an rf amplifier in vhf tuners of color and black-and-white television receivers. Outlines section, 5C; requires 7-contact socket. Types 2FH5 and 3FH5 are identical with type 6FH5 except for heater ratings.

	2FH5	3FH5	6FH5	
Heater Voltage (ac/dc)	2.35	3	6.3	volts
Heater Current	0.6	0.45	0.2	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	±100 max	volts

Direct Interelectrode Capacitances (Approx.):	Unshielded	Shielded°	
Grid to Plate	0.52	0.52	pF
Grid to Cathode, Heater, and Internal Shield	3.2	3.2	pF
Plate to Cathode, Heater, and Internal Shield	3.2	4	pF

Class A₁ Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

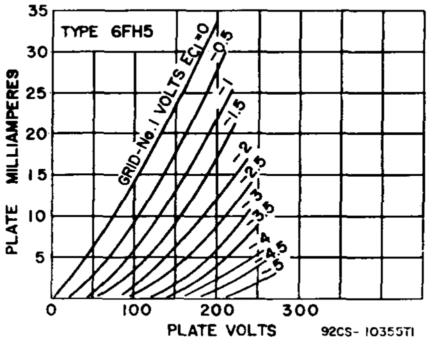
Plate Voltage	150	volts
Grid Voltage, Positive-bias value	0	volts
Cathode Current	22	mA
Plate Dissipation	2.2	watts

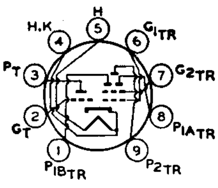
CHARACTERISTICS

Plate Voltage	135	volts
Grid Voltage	—1	volts
Plate Resistance (Approx.)	5600	ohms
Transconductance	9000	μmhos
Plate Current	11	mA
Grid Voltage (Approx.) for plate current of 100 μA	—5.5	volts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance, for cathode-bias operation	1	megohm





9KP

MEDIUM-MU TRIODE— THREE-PLATE TETRODE 6FH8

Miniature type used in complex-wave generator applications and in television receiver applications. Sharp-cutoff tetrode unit has pair of additional plates. Outlines section, 6B; requires 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Direct Interelectrode Capacitances:°		
Triode Unit:		
Grid to Plate	1.4	pF
Grid to Cathode and Heater	2.6	pF
Plate to Cathode and Heater	1	pF
Tetrode Unit:		
Grid No.1 to Plate No.2	0.06 max	pF
Grid *No.1 to* Cathode, Heater, Grid No.2, Plate No.1A, and Plate No.1B	4.5	pF
Plate No.2 to Cathode, Heater, Grid No.2, Plate No.1A, and Plate No.1B	1.4	
Tetrode Grid No.1 to Triode Plate	0.35 max	pF
Tetrode Plate No.2 to Triode Plate	0.008 max	pF

° With external shield connected to cathode.

Class A_1 Amplifier

CHARACTERISTICS	Triode Unit	
Plate Voltage	100	volts
Grid Voltage	—1	volt
Amplification Factor	40	
Plate Resistance (Approx.)	7400	ohms
Transconductance	5400	µmhos
Plate Current	7.9	mA
Grid Voltage (Approx.) for plate current of 100 µA	—7	volts

Tetrode Unit with Plates No.1A and No.1B Connected to Cathode at Socket

MAXIMUM RATINGS (Design-Maximum Values)		
Plate-No.2 Voltage	250	volts
Grid-No.2 Voltage	250	volts
Grid-No.1 Voltage	—2	volts
Plate-No.2 Resistance (Approx.)	0.75	megohm
Transconductance, Grid No.1 to Plate No.2	4400	µmhos
Plate-No.2 Current	7.3	mA
Grid-No.2 Current	1.4	mA
Grid-No.1 Voltage (Approx.) for plate-No.2 current of 100 µA	—7	volts

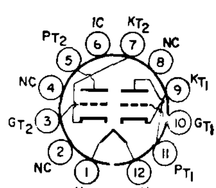
Complex-Wave Generator

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Tetrode Unit	
Plate Voltage	275	—	volts
Plate-No.1A Voltage	—	200	volts
Plate-No.1B Voltage	—	200	volts
Plate-No.2 Voltage		275	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	275	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:			
Negative-bias value	—40	—40	volts
Positive-bias value	0	0	volts
Plate Dissipation	1.7	—	watts
Plate-No.1A Dissipation	—	0.3	watt
Plate-No.1B Dissipation	—	0.3	watt
Plate-No.2 Dissipation	—	2.3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 137.5 volts	—	0.45	watt
Fro grid-No.2 voltages between 137.5 and 275 volts	—	See curve page 98	

TYPICAL OPERATION WITH SEPARATE PLATE OPERATION	Tetrode Unit	
Plates-No.1A, No.1B, and No.2 Voltage	100	volts
Grid-No.2 Voltage	50	volts
Grid-No.1 Voltage	—1	volts
Plate-No.1A Current	0.04	mA
Plate-No.1B Current	0.04	mA
Plate-No.2 Current	1.6	mA
Grid-No.2 Current	0.3	mA
Transconductance (Approx.):		
Grid No.1 to Plate No.1A	70	µmhos
Grid No.1 to Plate No.1B	70	µmhos
Grid No.1 to Plate No.2	2500	µmhos

MAXIMUM CIRCUIT VALUES	Triode Unit	Tetrode Unit	
Grid-No.1-Circuit Resistance, for fixed-bias operation	0.5	0.5	megohm

6FJ7 MEDIUM-MU DUAL TRIODE



12BM

Duodecar type used as combined vertical-deflection-oscillator and vertical-deflection-amplifier tube in color and black-and-white television receivers. **Outlines section,** 8B; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.9; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2		
Plate Voltage	250	150	250	volts
Grid Voltage	—8	0	—9.5	volts
Amplification Factor	22.5	—	15.4	
Plate Resistance (Approx.)	9000	—	2000	ohms

	Unit No.1	Unit No.2		
Transconductance	2500	—	7700	µmhos
Plate Current	8	68■	41	mA
Grid Voltage (Approx.) for plate current of 10 µA	—18	—	—	volts
Grid Voltage (Approx.) for plate current of 50 µA	—	—	—23	volts

■ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

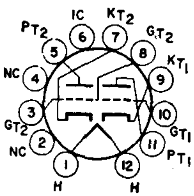
MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	350	550	volts
Peak Positive-Pulse Plate Voltage#	—	2500	volts
Peak Negative-Pulse Grid Voltage	400	250	volts
Peak Cathode Current	—	150	mA
Average Cathode Current	—	50	mA
Plate Dissipation	1	10	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For fixed-bias operation	2.2	2.2	megohms
For cathode-bias operation	2.2	—	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

6FM7

DUAL TRIODE

13FM7, 15FM7



12EJ

Duodecar type used as combined vertical-deflection oscillator and vertical-deflection amplifier in color and black-and-white television receivers. Triode unit No.1 is used as an oscillator, and triode unit No.2 is used as an amplifier. Outlines section, 8C; requires duodecar 12-contact socket. Types 13FM7 and 15FM7 are identical with type 6FM7 except for heater ratings.

	6FM7	13FM7	15FM7	
Heater Voltage (ac/dc)	6.3	13	14.8	volts
Heater Current	1.05	0.45	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Average value	±200 max	±200 max	±200 max	volts
Peak value	100 max	100 max	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2	
Plate Voltage	250	175	volts
Grid Voltage	—3	—25	volts
Amplification Factor	66	5.5	
Plate Resistance (Approx.)	30000	920	ohms
Transconductance	2200	6000	µmhos
Plate Current	2	40	mA
Grid Voltage (Approx.) for plate current of 20 µA	—5.3	—	volts
Grid Voltage (Approx.) for plate current of 200 µA	—	—45	volts

Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

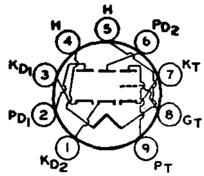
MAXIMUM RATINGS (Design-Maximum Values)			
DC Plate Voltage	350	500	volts
Peak Positive-Pulse Plate Voltage#	—	1500	volts
Peak Negative-Pulse Plate Voltage	400	250	volts
Peak Cathode Current	—	175	mA
Average Cathode Current	—	50	mA
Plate Dissipation†	1	10	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For fixed-bias operation	1	1	megohm
For cathode-bias operation	2.2	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

† A bias resistor or other means is required to protect the tube in absence of excitation.

6FM8

TWIN DIODE— HIGH-MU TRIODE



Miniature type used in television receiver applications and as combined FM detector and af voltage amplifier in FM receivers. **Outlines section,** 6B; requires miniature 9-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.45; maximum heater-cathode volts, ±200 peak, 100 average.

Triode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation	1.1	watts

CHARACTERISTICS

Plate Voltage	250	volts
Grid Voltage	—3	volts
Amplification Factor	70	
Plate Resistance (Approx.)	58000	ohms
Transconductance	1200	μmhos
Plate Current	1	mA

Diode Units (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

Plate Current	5	mA

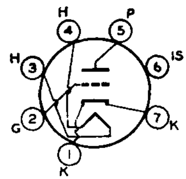
CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 20 mA	5	volts

6FQ5A

2FQ5A

HIGH-MU TRIODE



Miniature type with frame grid used as rf-amplifier tube in vhf tuners of television receivers. **Outlines section,** 5C; requires miniature 7-contact socket. Type 2FQ5A is identical with type 6FQ5A except for heater ratings.

	2FQ5A	6FQ5A	
Heater Voltage (ac/dc)	2.3	6.3	volts
Heater Current	0.6	0.18	ampere
Heater Warm-up Time (Average)	11	—	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	volts

Direct Interelectrode Capacitances:°

Grid to Plate	0.52	pF
Grid to Cathode, Heater, and Internal Shield	5	pF
Plate to Cathode, Heater, and Internal Shield	3.5	pF
Heater to Cathode	2.5	pF

° With external shield connected to cathode except as noted.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	200	volts
Grid Voltage, Negative-bias value	50	volts
Average Cathode Current	22	mA
Plate Dissipation	2.5	watts

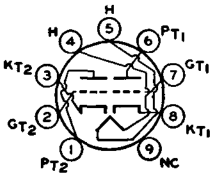
CHARACTERISTICS

Plate Voltage	135	volts
Grid Voltage	—1.2	volts
Amplification Factor	74	
Plate Resistance (Approx.)	6300	ohms
Transconductance	12000	μmhos
Plate Current	8.9	mA
Grid Voltage (Approx.) for plate current of 100 μA	—4.5	volts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance, for cathode-bias operation	1	megohm

6FQ7 Refer to chart at end of section.



9LP

MEDIUM-MU TWIN TRIODE

6FQ7/ 6CG7

8FQ7/8CG7

Miniature type used as combined vertical- and horizontal-deflection oscillator in color and black-and-white television receivers. **Outlines section,** 6E; requires miniature 9-contact socket. Type 8FQ7/8CG7 is identical with type 6FQ7/6CG7 except for heater ratings. For typical operation as a resistance-coupled amplier, refer to **Resistance-Coupled Amplifier** section.

	6FQ7/6CG7	8FQ7/8CG7	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	3.6	3.8	pF
Grid to Cathode and Heater	2.4	2.4	pF
Plate to Cathode and Heater	0.34	0.26	pF
Plate of Unit No.1 to Plate of Unit No.2	1		pF

MAXIMUM RATINGS (Design-Maximum Values)

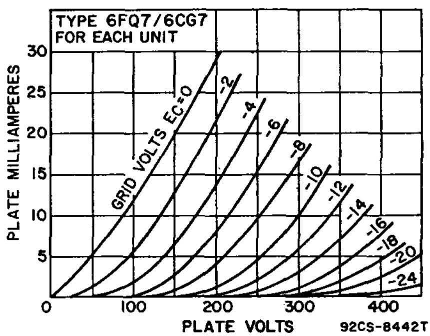
Plate Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Cathode Current	22	mA
Plate Dissipation:		
For either plate	4	watts
For both plates with both units operating	5.7	watts

CHARACTERISTICS

Plate Voltage	90	250	volts
Grid Voltage	0	—8	volts
Amplification Factor	20	20	
Plate Resistance (Approx.)	6700	7700	ohms
Transconductance	3000	2600	μmhos
Plate Current	10	9	mA
Grid Voltage (Approx.) for plate current of 10 μA	—7	—18	volts
Plate Current for grid voltage of —12.5 volts	—	1.3	mA

MAXIMUM CIRCUIT VALUE

Grid Circuit Resistance, for fixed-bias operation	1	megohm



Oscillator

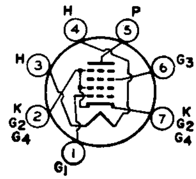
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Vertical-Deflection Oscillator	Horizontal-Deflection Oscillator	
DC Plate Voltage	330	330	volts
Peak Negative-Pulse Grid Voltage	440	660	volts

	Vertical-Deflection Oscillator	Horizontal-Deflection Oscillator	
Peak Cathode Current	77	330	mA
Average Cathode Current	22	22	mA
Plate Dissipation:			
For either plate	4	4	watts
For both plates with both units operating	5.7	5.7	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance	2.2	2.2	megohms

6FS5 BEAM HEXODE

2FS5, 3FS5



7GA

Miniature type used as rf-amplifier tube in vhf television receivers. In this tube, grid No.1 is the control grid, grid No.2 is a focusing grid, grid No.3 is the screen grid, and grid No.4 is the suppressor grid. Grid No.2 is internally connected to the cathode and grid No.4 and aligned with grid No.3 **Outlines section,** 5C; requires miniature 7-contact socket. Types 2FS5 and 3FS5 are identical with type 6FS5 except for heater ratings.

	2FS5	3FS5	6FS5	
Heater Voltage (ac/dc)	2.4	2.9	6.3	volts
Heater Current	0.6	0.45	0.2	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

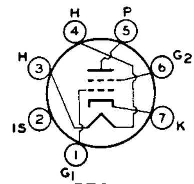
Direct Interelectrode Capacitances:	Shielded	Unshielded▪	
Grid No.1 to Plate	0.03	0.016	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Grid No.4	4.8	4.8	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Grid No.4	2	2.8	pF

▪ With external shield connected to pin 7.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	300	volts
Grid-No.3 (Screen-Grid) Voltage	150	volts
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Cathode Current	20	mA
Plate Dissipation	3.25	watts
Grid-No.3 Input	0.15	watt
CHARACTERISTICS		
Plate Voltage	275	volts
Grid-No.3 Voltage	135	volts
Grid-No.1 Voltage	—0.2	volt
Plate Resistance (Approx.)	0.24	megohm
Transconductance	10000	µmhos
Plate Current	9	mA
Grid-No.3 Current	0.17	mA
Grid-No.1 Voltage (Approx.) for transconductance of 100 µmhos	—5	volts
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance, for fixed-bias operation	0.5	megohm

6FV6 SHARP-CUTOFF TETRODE



7FQ

Miniature type used as rf amplifier in vhf tuners of television receivers. **Outlines section,** 5C; requires 7-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.2	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:°		
Grid No.1 to Plate	0.03 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Internal Shield	4.5	pF
Plate to Cathode, Heater, Grid No.2, and Internal Shield	3	pF
Cathode to Heater	2.7•	pF

° With external shield connected to cathode except as noted.

• With external shield connected to ground.

Class A_1 Amplifier

MAXIMUM RATINGS *(Design-Maximum Values)*

Plate Voltage	275	volts
Grid-No.2 (Screen-Grid) Supply Voltage	180	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Cathode Current	20	mA
Plate Dissipation	2	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 90 volts	0.5	watt
For grid-No.2 voltages between 90 and 180 volts	See curve page 98	

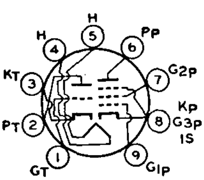
CHARACTERISTICS

Plate Voltage	125	volts
Grid-No.2 Voltage	80	volts
Grid-No.1 Voltage	—1	volt
Plate Resistance (Approx.)	0.1	megohm
Transconductance	8000	μmhos
Plate Current	10	mA
Grid-No.2 Current	1.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—6	volts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	0.5	megohm

Refer to chart at end of section. **6FV8**



9FA

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

6FV8A

5FV8

Miniature type used in television receivers as combined oscillator and amplifier. Triode unit is used as vertical-deflection oscillator; pentode unit is used as if or general-purpose amplifier. **Outlines section, 6B;** requires 9-contact socket. Type 5FV8 is identical with type 6FV8A except for heater ratings.

	5FV8	6FV8A	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:	**Shielded**	**Unshielded**	
Triode Unit:			
Grid to Plate	1.8	1.8	pF
Grid to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	2.8	2.8	pF
Plate to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	1.5	2	pF
Pentode Unit:			
Grid No.1 to Plate	0.02 max	0.01 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5	5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2	3	pF
Pentode Plate to Triode Plate	0.15 max	0.03 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Pentode Unit	
Plate Voltage	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, *Positive-bias value*	0	volts
Plate Dissipation	2.3	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

CHARACTERISTICS	Triode Unit	Pentode Unit	
Plate Voltage	125	125	volts
Grid-No.2 Voltage	—	125	volts
Grid-No.1 Voltage	—1	—1	volt
Amplification Factor	45	—	
Plate Resistance (Approx.)	5600	200000	ohms
Transconductance	8000	6500	μmhos
Plate Current	12	12	mA
Grid-No.2 Current	—	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—7.5	—9	volts

Vertical-Deflection Oscillator—Triode Unit

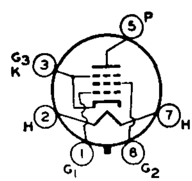
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Voltage	330	volts
Peak Negative-Pulse Grid Voltage	250	volts
Peak Cathode Current	70	mA
Average Cathode Current	20	mA
Plate Dissipation	2	watts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance, for cathode-bias operation	3	megohms

6FW5 BEAM POWER TUBE



6CK

Glass octal type used as horizontal-deflection amplifier in television receivers. **Outlines section, 19B; requires octal socket. Heater:** volts (ac/dc), 6.3; amperes, 1.2; maximum heater-cathode volts, ±200 peak, 100 average.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
DC Grid-No.2 (Screen-Grid) Voltage	220	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
DC Grid-No.1 (Control-Grid) Voltage	—55	volts
Peak Cathode Current	610	mA
Average Cathode Current	175	mA
Plate Dissipation•	18	watts
Grid-No.2 Input	3.6	watts
Bulb Temperature (At hottest point)	220	°C

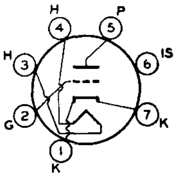
MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

• A bias resistor or other means is required to protect the tube in absence of excitation.

6FW8 Refer to chart at end of section.



7FP

6FY5/ EC97

HIGH-MU TRIODE

Miniature type with frame grid used for rf-amplifier applications in vhf tuners of television receivers. Outlines section, 5C; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.2	ampere
Peak Heater-Cathode Voltage	±100 max	volts

Direct Interelectrode Capacitances:	Unshielded	Shielded	
Grid to Plate	0.50	0.48	pF
Grid to Cathode, Heater, and Internal Shield	4.75	4.75	pF
Plate to Cathode, Heater, and Internal Shield	3.3	4.3	pF
Grid to Heater	0.28 max	0.28 max	pF
Plate to Cathode	0.25	0.21	pF
Cathode to Grid	3.2	3.2	pF
Heater to Cathode	2.5	2.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Absolute-Maximum Values)

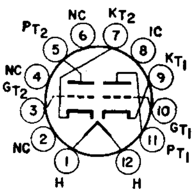
Plate Supply Voltage	550	volts
Plate Voltage	200	volts
Grid Voltage, Negative-bias value	50	volts
Cathode Current	20	mA
Plate Dissipation	2.2	watts

CHARACTERISTICS

Plate Voltage	135	135	135	135	volts
Grid Voltage	—1	—3.1	—5	—4.5	volts
Transconductance	13000	625	125	—	μmhos
Amplification Factor	70	—	—	—	
Plate Current	11	—	—	0.1	mA

MAXIMUM CIRCUIT VALUES

Cathode-Heater Circuit Resistance	0.02	megohm
Grid-Circuit Resistance	1	megohm



12EO

6FY7

DUAL TRIODE

11FY7, 15FY7

Duodecar type used as combined vertical-deflection oscillator and vertical-deflection amplifier in television receivers. Triode unit No.1 is used as an oscillator, and triode unit No.2 is used as an amplifier. Outlines section, 8D; requires duodecar 12-contact socket. Types 11FY7 and 15FY7 are identical with type 6FY7 except for heater ratings.

	6FY7	11FY7	15FY7	
Heater Voltage (ac/dc)	6.3	11	14.7	volts
Heater Current	1.05	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2	
Plate Voltage	250	150	volts
Grid Voltage	—3	—17.5	volts
Amplification Factor	65	6	
Plate Resistance (Approx.)	40500	920	ohms
Transconductance	1600	6500	μmhos
Plate Current	1.4	35	mA
Grid Voltage (Approx.) for plate current of 30 μA	—5.5	—	volts
Grid Voltage (Approx.) for plate current of 50 μA	—	—36	volts
Plate Current (Approx.) for grid voltage of —25 volts	—	6	mA

Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	330	275	volts
Peak Positive-Pulse Plate Voltage#	—	2000	volts
Peak Negative-Pulse Plate Voltage	400	250	volts
Peak Cathode Current	70	175	mA
Average Cathode Current	20	50	mA
Plate Dissipation	1	7†	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance	2.2	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).
† A bias resistor or other means is required to protect the tube in absence of excitation.

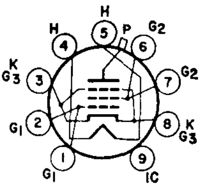
6G6G Refer to chart at end of section.

6G11 Refer to chart at end of section.

6GB5 Refer to chart at end of section.

6GB5/ EL500 BEAM POWER TUBE

13GB5/XL500, 18GB5, 27GB5/PL500



9NH

Neonoval type used as horizontal-deflection amplifier in television receivers. Outlines section, 10E; requires neonoval 9-contact socket. Typical instantaneous characteristics (measured with recurrent waveform such that maximum ratings are not exceeded): plate volts, 75; grid-No.2 volts, 200; grid-No.1 volts, —10; plate mA, 440; grid-No.2 mA, 37. Types 13GB5/XL500, 18GB5, and 27GB5/PL500 are identical with type 6GB5/EL500 except for heater ratings.

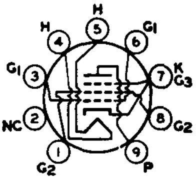
	6GB5/ EL500	13GB5/ XL500	18GB5	27GB5/ PL500	
Heater Voltage (ac/dc)	6.3	13.3	18	27	volts
Heater Current	1.38	0.6	0.45	0.3	amperes
Heater-Cathode Voltage:					
Peak value	±250 max	±250 max	±250 max	±250 max	volts
Average value	125 max	125 max	125 max	125 max	volts

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Voltage	275	volts
Peak Positive-Pulse Plate Voltage#	7700	volts
DC Grid-No.2 (Screen-Grid) Voltage	275	volts
Average Cathode Current	275	mA
Plate Dissipation▲	17	watts
Grid-No.2 Input■	5	watts
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance:		
Without grid current	0.5	megohm
With grid current (horizontal-output service only)	2.2	megohms

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
▲ A bias resistor or other means is required to protect the tube in absence of excitation.
■ Grid-No.2 input may reach 6 watts for plate-dissipation values below 11 watts.



9EU

BEAM POWER TUBE 6GC5

Neonoval type used in color and black-and-white television receiver applications and as output tube in audio-amplifier applications. Outlines section, 10D; requires neonoval 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	1.2	amperes
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.9	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	18	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	7	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

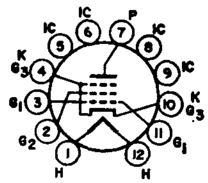
Plate Voltage	220	volts
Grid-No.2 (Screen-Grid) Voltage	140	volts
Plate Dissipation	12	watts
Grid-No.2 Input	1.4	watts

TYPICAL OPERATION AND CHARACTERISTICS

Plate Voltage	110	200	volts
Grid-No.2 Voltage	110	125	volts
Grid-No.1 Voltage	—7.5	—	volts
Cathode-Bias Resistor	—	180	ohms
Peak AF Grid-No.1 Voltage	7.5	8.5	volts
Zero-Signal Plate Current	49	46	mA
Maximum-Signal Plate Current	50	47	mA
Zero-Signal Grid-No.2 Current	4	2.2	mA
Maximum-Signal Grid-No.2 Current	10	8.5	mA
Plate Resistance (Approx.)	13000	28000	ohms
Transconductance	8000	8000	µmhos
Load Resistance	2000	4000	ohms
Total Harmonic Distortion	10	10	per cent
Maximum-Signal Power Output	2.1	3.8	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm



12BJ

BEAM POWER TUBE 6GE5

12GE5, 17GE5

Duodecar type used as horizontal-deflection-amplifier tube in television receivers. Outlines section, 15A; requires duodecar 12-contact socket. Types 12GE5 and 17GE5 are identical with type 6GE5 except for heater ratings.

	6GE5	12GE5	17GE5	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection		Triode* Connection	
Plate Voltage	60	250	150	volts
Grid-No.2 (Screen-Grid) Voltage	150	150	150	volts
Grid-No.1 (Control-Grid) Voltage	0	—22.5	—	volts
Amplification Factor	—	—	4.4	
Plate Resistance (Approx.)	—	18000	—	ohms
Transconductance	—	7300	—	µmhos
Plate Current	345•	65	—	mA

	Pentode Connection		Triode* Connection	
Grid-No.2 Current	27•	1.8	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—42	—	volts

* Grid No.2 tied to plate.

• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

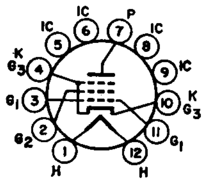
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.2 Voltage	220	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
DC Grid-No.1 Voltage	—55	volts
Peak Cathode Current	550	mA
Average Cathode Current	175	mA
Plate Dissipation†	17.5	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	200	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1 Circuit Resistance	1	megohm

\# Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

† A bias resistor or other means is required to protect the tube in absence of excitation.

6GF5 BEAM POWER TUBE



12BJ

Duodecar type used as horizontal-deflection amplifier in television receivers. **Outlines section, 8D; requires duodecar 12-contact socket. Heater:** volts (ac/dc), 6.3; amperes, 1.2; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection		Triode* Connection	
Plate Voltage	60	250	150	volts
Grid-No.2 (Screen-Grid) Voltage	150	150	150	volts
Grid-No.1 (Control-Grid) Voltage	0	—26.5	—	volts
Amplification Factor	—	—	4.2	
Plate Resistance (Approx.)	—	0.26	—	megohm
Transconductance	—	4700	—	μmhos
Plate Current	345•	34	—	mA
Grid-No.2 Current	33•	1.6	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—46	—	volts

* Grid No.2 connected to plate.

• These values can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

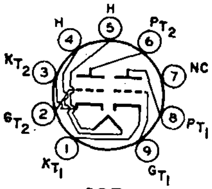
MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	5000	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.2 Voltage	220	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Negative DC Grid-No.1 Voltage	—55	volts
Peak Cathode Current	500	mA
Average Cathode Current	160	mA
Plate Dissipation†	9	watts

Grid-No.2 Input	2.5	watts
Bulb Temperature (At hottest point)	200	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

† A bias resistor or other means is required to protect the tube in absence of excitation.

Refer to chart at end of section. **6GF7**



9QD

DUAL TRIODE

6GF7A

10GF7A, 13GF7A

Novar types used as combined vertical-deflection oscillator and vertical-deflection amplifiers in color and black-and-white television receivers. **Outlines section, 30A;** requires novar 9-contact socket. For curves of average plate characteristics for Unit No.1 and Unit No.2, refer to types 6DR7 (Unit No.1) and 6EM7, respectively. Types 10GF7A and 13GF7A are identical with type 6GF7A except for heater ratings.

	6GF7A	10GF7A	13GF7A	
Heater Voltage (ac/dc)	6.3	9.7	13	volts
Heater Current	0.985	0.6	0.45	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	4.6	9	pF
Grid to Cathode and Heater	2.4	6.5	pF
Plate to Cathode and Heater	0.26	1.4	pF

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2	
Plate Voltage	250	150	volts
Grid Voltage	—3	—20	volts
Amplification Factor	64	5.4	
Plate Resistance (Approx.)	40000	750	ohms
Transconductance	1600	7200	μmhos
Grid Voltage (Approx.):			
For plate current of 10 μA	—5.5	—	volts
For plate current of 100 μA	—	—45	volts
Plate Current	1.4	50	mA
For plate voltage of 60 volts and zero grid voltage	—	95	mA
For grid voltage of —28 volts	—	10	mA

Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	330	330	volts
Peak Positive-Pulse Plate Voltage (Absolute Maximum)#	—	1500•	volts
Peak Negative-Pulse Grid Voltage	400	250	volts
Peak Cathode Current	77	175	mA
Average Cathode Current	22	50	mA
Plate Dissipation	1.5	11	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For grid-resistor-bias or cathode-bias operation	2.2	2.2	megohms

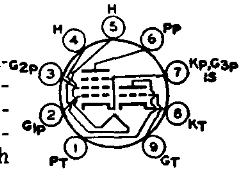
• Under no circumstances should this absolute value be exceeded.

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

Refer to chart at end of section. **6GH8**

6GH8A
5GH8A

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE



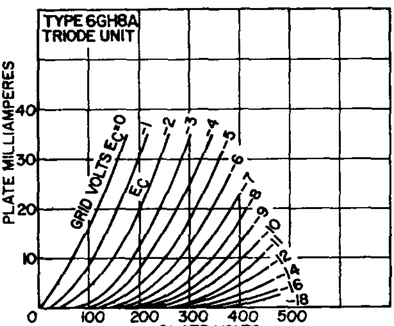
9AE

Miniature type used in multivibrator-type horizontal-deflection circuits and for agc-amplifier or sync-separator applications in color and black-and-white television receivers. Outlines section, 6B; requires miniature 9-contact socket. Type 5GH8A is identical with type 6GH8A except for heater ratings.

	5GH8A	6GH8A	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:	Unshielded	Shielded	
Triode Unit:			
Grid to Plate	1.7	1.7	pF
Grid to Cathode, Heater, Pentode Grid No.3, Pentode Cathode, and Internal Shield	3	3.2	pF
Plate to Cathode, Heater, Pentode Grid No.3, Pentode Cathode, and Internal Shield	1.4	1.9	pF
Heater to Cathode	3	3	pF
Pentode Unit:			
Grid No.1 to Plate	0.02 max	0.01 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5	5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.6	3.4	pF
Heater to Cathode, Grid No.3, and Internal Shield	3	3	pF

Class A_1 Amplifier

CHARACTERISTICS	Triode Unit	Pentode Unit	
Plate Voltage	125	125	volts
Grid-No.2 Voltage	—	125	volts
Grid-No.1 Voltage	—1	—1	volts
Amplification Factor	46	—	
Plate Resistance (Approx.)	5400	200000	ohms
Transconductance	8500	7500	μmhos
Plate Current	13.5	12	mA
Grid-No.2 Current	—	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—8	—8	volts



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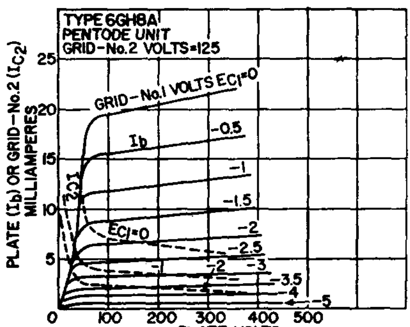

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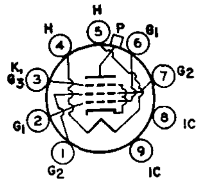
Horizontal-Deflection Oscillator

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	350	volts
Grid-No.2 (Screen-Grid) Voltage	—	330	volts

	Triode Unit	Pentode Unit	
Grid-No.1 (Control-Grid) Voltage:			
Positive-bias value	0	0	volts
Peak negative value	—	175	volts
Peak Cathode Current	—	300	mA
Average Cathode Current	—	20	mA
Plate Dissipation	2.5	2.5	watts
Grid-No.2 Input	—	0.55	watt
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	2.2	2.2	megohms
For cathode-bias operation	2.2	2.2	megohms

Refer to chart at end of section. **6GJ5**



9QK

BEAM POWER TUBE 6GJ5A

12GJ5A, 17GJ5A

Novar type used in high-efficiency horizontal-deflection-amplifier circuits of television receivers. **Outlines section, 32A; requires novar 9-contact socket. For curve** of average characteristics see type 6GW6. Types 12GJ5A and 17GJ5A are identical with type 6GJ5A except for heater ratings.

	6GJ5A	12GJ5A	17GJ5A	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts
Direct Interelectrode Capacitances (Approx.):				
Grid No.1 to Plate			0.26	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3			15	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3			6.5	pF

Class A_1 Amplifier

CHARACTERISTICS	Triode Connection	Pentode Connection		
Plate Voltage	150	60	250	volts
Grid-No.2 Voltage	150	150	150	volts
Grid-No.1 Voltage	—22.5	0	—22.5	volts
Mu-Factor, Grid No.2 to Grid No.1	4.4	—	—	
Plate Resistance (Approx.)	—	—	15000	ohms
Transconductance	—	—	7100	µmhos
Plate Current	—	390▪	70	mA
Grid-No.2 Current	—	32▪	2.1	mA
Grid-No.1 Voltage for plate current of 1 mA	—	—	—42	volts

▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage	—55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	550	mA
Average Cathode Current	175	mA
Plate Dissipation•	17.5	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (at hottest point)	240	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance:		
For grid-resistor-bias operation•	1	megohm

\# Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

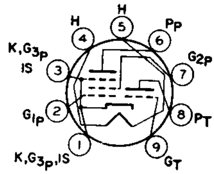
• A bias resistor or other means is required to protect the tube in absence of excitation.

6GJ7 Refer to chart at end of section.

6GJ7/ ECF801

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

5GJ7, 8GJ7/PCF801



9QA

Miniature types used as combined oscillator and mixer tubes in color and black-and-white television receivers utilizing an intermediate frequency in the order of 40 MHZ. Outlines section, 6J; requires miniature 9-contact socket. Types 5GJ7 and 8GJ7/PCF801 are identical with type 6GJ7/ECF801 except for heater ratings.

	5GJ7	6GJ7/ECF801	8GJ7/PCF801	
Heater Voltage (ac/dc)	5.6	6.3	8	volts
Heater Current	0.45	0.41	0.3	ampere
Peak Heater-Cathode Voltage▲	±110 max	±110 max	±110 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate-Supply Voltage	600	600	volts
DC Plate Voltage	140	275	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	600	volts
DC Grid-No.2 Voltage	—	275	volts
DC Grid-No.1 (Control-Grid) Voltage	—	—50	volts
Cathode Current	22	20	mA
Plate Dissipation	1.8	2.4	watts
Grid-No.2 Input•	—	0.55	watt
CHARACTERISTICS			
DC Plate Voltage	100	170	volts
DC Grid-No.2 Voltage	—	120	volts
DC Grid-No.1 Voltage	—3	—1.2	volts
Amplification Factor	20	55*	
Plate Resistance (Approx.)	—	0.35	megohm
Transconductance	9000	11000	μmhos
Plate Current	15	10	mA
Grid-No.2 Current	—	3	mA
Grid-No.1 Voltage for grid-No.1 current of 0.3 μA	—1.3 max	—1.3 max	volts
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	1	megohm
For cathode-bias operation	0.5	2.2	megohms

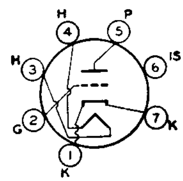
▲ The hum should be minimized in intercarrier applications by limiting the heater-cathode voltage to 100 volts rms, and in AM receivers to 50 volts rms.
* Grid No.2 to grid No.1, approximate value.
• When control-grid bias is between —1.5 and —2 volts, screen-grid dissipation is limited to 0.50 watt. When this bias is greater than —2 volts, maximum screen-grid dissipation is 0.36 watt.

6GJ8 Refer to chart at end of section.

6GK5

HIGH-MU TRIODE

2GK5, 3GK5, 4GK5



7FP

Miniature type with frame grid used as grounded-cathode rf-amplifier tube in vhf tuners of color and black-and-white television receivers. Outlines section, 5C; requires miniature 7-contact socket. Types 2GK5, 3GK5, and 4GK5 are identical with type 6GK5 except for heater ratings.

	2GK5	3GK5	4GK5	6GK5	
Heater Voltage (ac/dc)	2.3	2.8	4	6.3	volts
Heater Current	0.6	0.45	0.3	0.18	ampere
Heater Warm-up Time (Average)	11	11	11	—	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	±100 max	±100 max	volts
Direct Interelectrode Capacitances (Approx.):°					
Grid to Plate				0.52	pF
Grid to Cathode, Heater, and Internal Shield				5	pF
Plate to Cathode, Heater, and Internal Shield				3.5	pF
Heater to Cathode				2.5•	pF

° With external shield connected to cathode, except as noted.
▪ With external shield and internal shield connected to ground.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	200	volts
Grid Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Average Cathode Current	22	mA
Plate Dissipation	2.5	watts

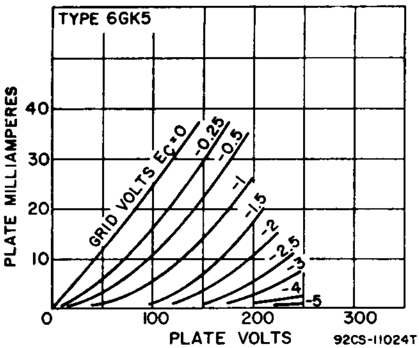
CHARACTERISTICS

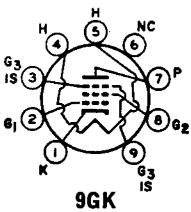
Plate Voltage	135	volts
Grid Voltage	—1	volts
Amplification Factor	78	
Plate Resistance (Approx.)	5400	ohms
Transconductance	15000	μmhos
Plate Current	11.5	mA
Input Resistance•	275	ohms
Input Capacitance•	11.2	pF
Noise Figure†	4.7	dB
Grid Voltage (Approx.) for transconductance of 150 μmhos	—4.2	volts
Grid Voltage (Approx.) for transconductance of 1500 μmhos	—2.5	volts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance, for cathode-bias operation	1	megohm

• Measured at 200 MHz with heater volts = 6.3 and plate effectively grounded for rf voltages.
† For a neutralized triode amplifier at a frequency of 200 MHz with signal source impedance adjusted for minimum noise output.





9GK

POWER PENTODE 6GK6

10GK6, 16GK6

Miniature type used in the output stage of audio amplifying equipment and also in the video output stage of color and black-and-white television receivers. Outlines section, 6G; requires miniature 9-contact socket. Types 10GK6 and 16GK6 are identical with type 6GK6 except for heater ratings.

	6GK6	10GK6	16GK6	
Heater Voltage (ac/dc)	6.3	10.6	16	volts
Heater Current	0.76	0.45	0.3	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	±100 max	volts
Direct Interelectrode Capacitances:				
Grid No.1 to Plate			0.14 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield			10	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield			7	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	605	volts
Plate Voltage	330	volts
Grid-No.2 Supply Voltage	605	volts
Grid-No.2 (Screen-Grid) Voltage	330	volts
Grid-No.1 (Control-Grid) Voltage, Negative-bias value	100	volts
Cathode Current	65	mA
Plate Dissipation	13.2	watts
Grid-No.2 Input, Peak	4	watts
Grid-No.-2 Input, Average	2	watts

CHARACTERISTICS AND TYPICAL OPERATION

Plate Supply Voltage	250	volts
Grid-No.2 Supply Voltage	250	volts
Cathode-Bias Resistor	135	ohms
Mu-Factor, Grid No.2 to Grid No.1	19	
Plate Resistance (Approx.)	38000	ohms
Transconductance	11300	µmhos
Peak AF Grid-No.1 Voltage	7.3	volts
Zero-Signal Plate Current	48	mA
Maximum-Signal Plate Current	50.6	mA
Zero-Signal Grid-No.2 Current	5.5	mA
Maximum-Signal Grid-No.2 Current	10	mA
Effective Load Resistance	5200	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	5.7	watts

Push-Pull Class AB_1 and Class B Amplifier

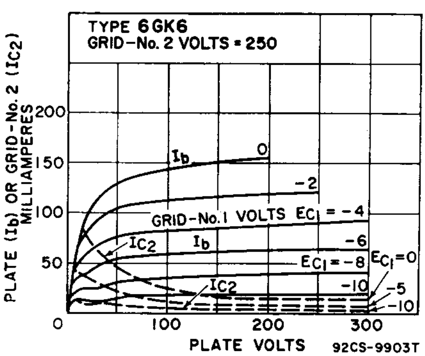
MAXIMUM RATINGS (Same as for Class A_1 Amplifier)

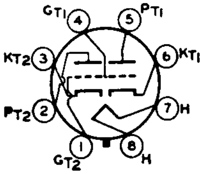
TYPICAL OPERATION (Values are for two tubes)

	Class AB_1		Class B		
Plate Voltage	250	300	250	300	volts
Grid-No.2 Voltage	250	300	250	300	volts
Grid-No.1 Voltage	—	—	—11.6	—14.7	volts
Cathode-Bias Resistor	130	130	—	—	ohms
Peak AF Grid-No.1-to-Grid-No.1 Voltage	22.4	28	22.4	28	volts
Zero-Signal Plate Current	62	72	20	15	mA
Maximum-Signal Plate Current	75	92	75	92	mA
Zero-Signal Grid-No.2 Current	7	8	2.2	1.6	mA
Maximum-Signal Grid-No.2 Current	15	22	15	22	mA
Effective Load Resistance (plate to plate)	8000	8000	8000	8000	ohms
Total Harmonic Distortion	3	4	3	4	per cent
Maximum-Signal Power Output	11	17	11	17	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.3	megohm
For cathode-bias operation	1	megohm





8BD

DUAL TRIODE 6GL7

Glass type used as combined vertical-deflection-oscillator and vertical-deflection-amplifier tube in color and black-and-white television receivers. **Outlines section,** 13B; requires octal socket. **Heater:** volts (ac/dc), 6.3; amperes, 1.05; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

CHARACTERISTICS	Unit No.1	Unit No.2	
Plate Voltage	250	175	volts
Grid Voltage	—3	—25	volts
Amplification Factor	66	5	
Plate Resistance (Approx.)	30000	780	ohms
Transconductance	2200	6400	µmhos
Plate Current	2	46	mA
Grid Voltage (Approx.):			
For plate current of 20 µA	—5.3	—	volts
For plate current of 200 µA	—	—60	volts

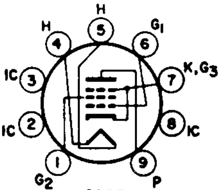
Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 Amplifier	
DC Plate Voltage	350	550	volts
Peak Positive-Pulse Plate Voltage	—	1500#	volts
Peak Negative-Pulse Grid Voltage	400	250	volts
Peak Cathode Current	—	175	mA
Average Cathode Current	—	50	mA
Plate Dissipation▪	1	10	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For fixed-bias operation	1	1	megohm
For cathode-bias operation	2.2	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

▪ A bias resistor or other means is required to protect the tube in absence of excitation.



9MQ

POWER PENTODE 6GM5

Neonoval type used in television receivers and as power amplifier in radio receivers and audio amplifiers. **Outlines section,** 10D; requires neonoval 9-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.8; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

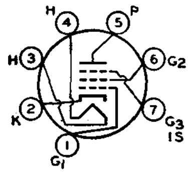
MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	550	volts
Grid-No.2 (Screen-Grid) Voltage	440	volts
Cathode Current	85	mA
Plate Dissipation	19	watts
Grid-No.2 Input	3.3*	watts
TYPICAL OPERATION AND CHARACTERISTICS		
Plate Voltage	300	volts
Grid-No.2 Voltage	300	volts
Grid-No.1 (Control-Grid) Voltage	—10	volts
Peak AF Grid-No.1 Voltage	10	volts
Zero-Signal Plate Current	60	mA
Maximum-Signal Plate Current	75	mA
Zero-Signal Grid-No.2 Current	8	mA
Maximum-Signal Grid-No.2 Current	15	mA
Plate Resistance (Approx.)	29000	ohms
Transconductance	10200	µmhos
Load Resistance	3000	ohms
Total Harmonic Distortion	13	per cent
Maximum-Signal Power Output	11	watts

* Grid-No.2 input may reach 6 watts during peak levels of speech and music signals.

6GM6

4GM6, 5GM6

SEMIREMOTE-CUTOFF PENTODE



7CM

Miniature type used in gain-controlled picture-if stages of color and black-and-white television receivers operating at intermediate frequencies in the order of 40 MHz. Outlines section, 5C; requires 7-contact socket. Types 4GM6 and 5GM6 are identical with type 6GM6 except for heater ratings.

	4GM6	5GM6	6GM6	
Heater Voltage (ac/dc)	4.2	5.6	6.3	volts
Heater Current	0.6	0.45	0.4	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts
Direct Interelectrode Capacitances:		Unshielded	Shielded°	
Grid No.1 to Plate		0.036 max	0.026 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield		10	10	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield		2.4	3.4	pF

° With external shield connected to cathode.

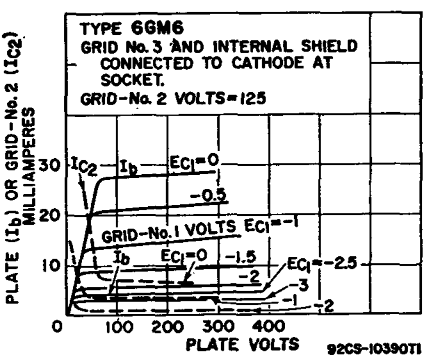
Class A_1 Amplifier

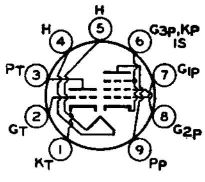
MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	3.1	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.65	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	125	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.2	megohm
Transconductance	13000	μmhos
Plate Current	14	mA
Grid-No.2 Current	3.4	mA
Grid-No.1 Voltage (Approx.) for transconductance of 60 μmhos	—15	volts





9DX

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE

6GN8

8GN8, 10GN8

Miniature type used in color and black-and-white television receiver applications. Triode unit is used as sync-separator, sync-clipper, phase inverter, or sound-if amplifier. Pentode unit is used in output stage of video amplifier. **Outlines section, 6E**; requires miniature 9-contact socket. For direct interelectrode capacitances, refer to type 6EB8; curve for average plate characteristics of triode unit is same as for type 6EB8. Types 8GN8 and 10GN8 are identical with type 6GN8 except for heater ratings.

	6GN8	8GN8	10GN8	
Heater Voltage (ac/dc)	6.3	8	10.5	volts
Heater Current	0.75	0.6	0.45	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

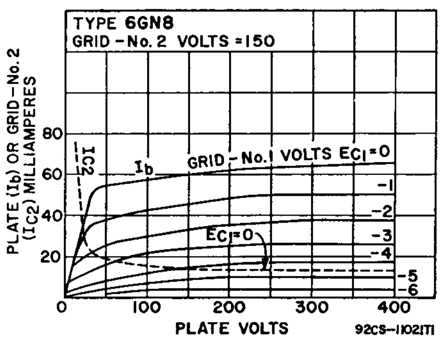
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volt
Plate Dissipation	1	5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	

CHARACTERISTICS	Triode Unit	Pentode Unit		
Plate Supply Voltage	250	60	200	volts
Grid-No.2 Supply Voltage	—	150	150	volts
Grid-No.1 Voltage	—2	0	—	volts
Cathode-Bias Resistor	—	—	100	ohms
Amplification Factor	100	—	—	
Plate Resistance (Approx.)	37000	—	60000	ohms
Transconductance	2700	—	11500	µmhos
Plate Current	2	55▪	25	mA
Grid-No.2 Current	—	18▪	5.5	mA
Grid Voltage (Approx.) for plate current of 20 µA	—5	—	—	volts
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—	—	—10	volts

MAXIMUM CIRCUIT VALUES	Triode Unit	Pentode Unit	
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

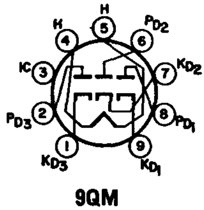
▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.



6GQ7

19GQ7

TRIPLE DIODE



Miniature type used in AM/FM radio receivers as a combination FM discriminator and AM detector tube. **Outlines section,** 6B; requires miniature 9-contact socket. Type 19GQ7 is identical with type 6GQ7 except for heater ratings.

	6GQ7	19GQ7	
Heater Voltage (ac/dc)	6.3	18.9	volts
Heater Current	0.45	0.15	ampere
Heater-Cathode Voltage:			
Peak value	+200, —300 max		volts
Average value	100 max		volts

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Voltage	330	volts
AC Plate Voltage	117	volts
AC Plate Current	54	mA
DC Output Current	9	mA
Minimum Total Effective Plate Supply Impedance	300	ohms

CHARACTERISTICS (Each Diode Unit)

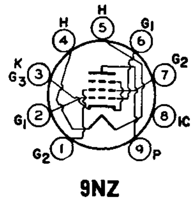
Tube Voltage Drop for plate current of 60 mA	10	volts

6GT5 Refer to chart at end of section.

6GT5A

17GT5A

BEAM POWER TUBE



Novar type used as horizontal-deflection amplifier in television receivers. **Outlines section,** 31A; requires novar 9-contact socket. For curve of average characteristics, refer to type 6GW6. Type 17GT5A is identical with type 6GT5A except for heater ratings.

	6GT5A	17GT5A	
Heater Voltage (ac/dc)	6.3	16.8	volts
Heater Current	1.2	0.45	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances (Approx.):			
Grid No.1 to Plate		0.26	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3		15	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3		6.5	pF

Class A_1 Amplifier

CHARACTERISTICS	Triode Connection	Pentode Connection		
Plate Voltage	150	60	250	volts
Grid-No.2 (Screen-Grid) Voltage	150	150	150	volts
Grid-No.1 (Control-Grid) Voltage	—22.5	0	—22.5	volts
Mu Factor, Grid No.2 to Grid No.1	4.4	—	—	
Plate Resistance (Approx.)	—	—	15000	ohms
Transconductance	—	—	7100	μmhos
Plate Current	—	390*	70	mA
Grid-No.2 Current	—	32*	2.1	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—	—42	volts

* This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

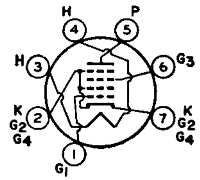
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts

Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage	—55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	550	mA
Average Cathode Current	175	mA
Plate Dissipation•	17.5	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	240	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance, for grid-resistor-bias operation•	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
• A bias resistor or other means is required to protect the tube in absence of excitation.



7GA

BEAM HEXODE 6GU5

2GU5

Miniature type used as rf amplifier in vhf television receivers. Outlines section, 5C; requires miniature 7-contact socket. Type 2GU5 is identical with type 6GU5 except for heater ratings.

	2GU5	6GU5	
Heater Voltage (ac/dc)	2.4	6.3	volts
Heater Current	0.6	0.22	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:			
Grid No.1 to Plate		0.018	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Grid No.4		7	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Grid No.4		3.2	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

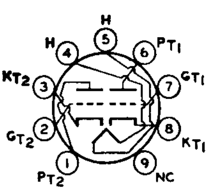
Plate Voltage	300	volts
Grid-No.3 (Screen-Grid) Voltage:	150	volts
DC Grid-No.1 (Control-Grid) Voltage:		
Positive-bias value	0	volts
Negative-bias value	50	volts
Average Cathode Current	20	mA
Plate Dissipation	3	watts
Grid-No.3 Input	0.15	watts

CHARACTERISTICS

Plate Voltage	135	275	volts
Grid-No.3 Voltage	135	135	volts
Grid-No.1 Voltage	—0.4	—0.4	volts
Plate Resistance (Approx.)	0.67	0.165	megohms
Transconductance	15000	15500	μmhos
Plate Current	9	10	mA
Grid-No.3 Current	0.25	0.17	mA
Grid-No.1 Voltage (Approx.) for transconductance of 100 μmhos	—6.2	—6.5	volts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, for fixed-bias operation	0.5	megohm



9LP

MEDIUM-MU TWIN TRIODE 6GU7

8GU7

Miniature type used in the matrixing circuits of color and black-and-white television receivers and in phase-inverter, multivibrator, and general-purpose amplifier applications. Outlines section, 6E; requires miniature 9-contact socket. Type 8GU7 is identical with type 6GU7 except for heater ratings.

	6GU7	8GU7	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	3	3	pF
Grid to Cathode and Heater	3.4	3.6	pF
Plate to Cathode and Heater	0.44	0.34	pF
Plate of Unit No.1 to Plate of Unit No.2	1		pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

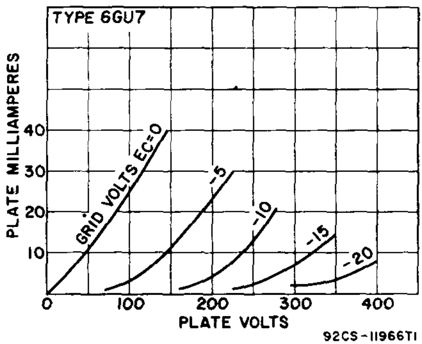
Plate Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation	3	watts

CHARACTERISTICS

Plate Voltage	250	volts
Grid Voltage	—10.5	volts
Amplification Factor	17	
Plate Resistance (Approx.)	5500	ohms
Transconductance	3100	μmhos
Plate Current	11.5	mA
Grid Voltage (Approx.) for plate current of 50 μA	—23	volts
Plate Current for grid voltage of —14 volts	4	mA

MAXIMUM CIRCUIT VALUE

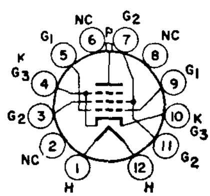
Grid-Circuit Resistance, for fixed-bias operation	1	megohm



6GV5

17GV5

BEAM POWER TUBE



12DR

Duodecar type used as horizontal-deflection amplifier in television receivers. Outlines section, 16A; requires duodecar 12-contact socket. Type 17GV5 is identical with type 6GV5 except for heater ratings.

	6GV5	17GV5	
Heater Voltage (ac/dc)	6.3	16.8	volts
Heater Current	1.2	0.45	amperes
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection			Triode* Connection	
Plate Voltage	5000	60	250	150	volts
Grid-No.2 (Screen-Grid) Voltage	150	150	150	150	volts
Grid-No.1 (Control-Grid) Voltage	—	0	—22.5	—22.5	volts

	Pentode Connection			Triode* Connection	
Plate Resistance (Approx.)	—	—	18000	—	ohms
Transconductance	—	—	7300	—	µmhos
Amplification Factor	—	—	—	4.4	
Plate Current	—	345▪	65	—	mA
Grid-No.2 Current	—	27▪	1.8	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—100	—	—42	—	volts

* Grid No.2 tied to plate.

▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.2 Voltage	220	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
DC Grid-No.1 Voltage	—55	volts
Peak Cathode Current	550	mA
Average Cathode Current	175	mA
Plate Dissipation†	17.5	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	220	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	1	megohm

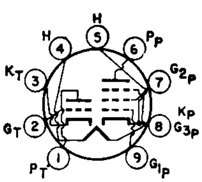
Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
† A bias resistor or other means is required to protect the tube in absence of excitation.

Refer to chart at end of section. **6GV8**

6GV8/ EL500

9GV8/XCL85

HIGH-MU TRIODE— POWER PENTODE



9LY

Miniature type used for sync-amplifier and video-output applications in television receivers. Outlines section, 6G; requires miniature 9-contact socket. Type 9GV8/XCL85 is identical with type 6GV8/EL500 except for heater ratings.

	6GV8/ EL500	9GV8/ XCL85	
Heater Voltage (ac/dc)	6.3	9.5	volts
Heater Current	0.9	0.6	ampere
Peak Heater-Cathode Voltage	±220 max	±220 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Absolute-Maximum Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	550	volts
Peak Plate Voltage°	—	2000	volts
DC Plate Voltage	250	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	550	volts
Grid-No.2 Voltage	—	250	volts
Peak Cathode Current▪	200	—	mA
Average Cathode Current	15	75	mA
Plate Dissipation	0.5	7	watts
Grid-No.2 Input	—	2	watts

CHARACTERISTICS

Plate Voltage	100	50	65	170	volts
Grid-No.2 Voltage	—	170	210	170	volts
Grid-No.1 Voltage	—0.8	—1	—1	—15	volts
Amplification Factor	50	—	—	—	
Mu-Factor, Grid No.1 to Grid No.2	—	—	—	7	
Plate Resistance (Approx.)	7600	—	—	25000	ohms
Transconductance	6500			7500	µmhos
Plate Current	5	200•	240•	41	mA
Grid-No.2 Current	—	40•	50•	2.7	mA

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:			
For fixed-bias operation	1	1	megohm
For cathode-bias operation	3.3	2.2	megohms

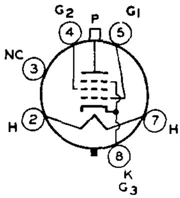
° Maximum pulse duration 5 per cent of a cycle with a maximum of 1 millisecond.
▪ Maximum pulse duration 200 microseconds. If a larger flyback is required, this value may be reduced to 100 mA with a maximum pulse duration of 400 microseconds.
• This value can be measured by a method involving a recurrent waveform such that the maximum tube ratings will not be exceeded.

6GW6 Refer to chart at end of section.

6GW6/ 6DQ6B

BEAM POWER TUBE

12GW6/12DQ6B
17GW6/17DQ6B



6AM

Glass octal type used as horizontal-deflection amplifier in high-efficiency deflection circuits of television receivers. **Outlines section,** 20; requires octal socket. Types 12GW6/12DQ6B and 17GW6/17DQ6B are identical with type 6GW6/6DQ6B except for heater ratings.

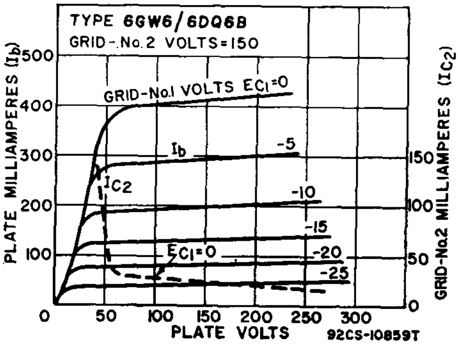
	6GW6/ 6DQ6B	12GW6/ 12DQ6B	17GW6/ 17DQ6B	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.5	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	17	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	7	pF

Class A_1 Amplifier

CHARACTERISTICS	Triode Connection	Pentode Connection		
Plate Voltage	150	60	250	volts
Grid-No.2 Voltage	150	150	150	volts
Grid-No.1 Voltage	—22.5	0	—22.5	volts
Mu-Factor, Grid No.2 to Grid No.1	4.4	—	—	
Plate Resistance (Approx.)	—	—	15000	ohms
Transconductance	—	—	7100	μmhos
Plate Current	—	390*	70	mA
Grid-No.2 Current	—	32*	2.1	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—	—42	volts

* This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

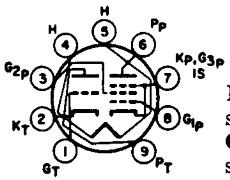


Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.2 (Screen-Grid) Voltage	220	volts
DC Grid-No.1 (Control-Grid) Voltage	—55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	550	mA
Average Cathode Current	175	mA
Plate Dissipation•	17.5	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	240	°C
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance, for grid-resistor-bias operation	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
• A bias resistor or other means is required to protect the tube in absence of excitation.



9LZ

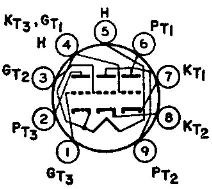
HIGH-MU TRIODE—SHARP-CUTOFF PENTODE 6GW8/ECL86

Miniature type used in preamplifier and audio output stages of audio equipment and television receivers. Outlines section, 6G; requires miniature 9-contact socket. Heater: volts (ac/dc), 6.3; amperes, 0.7; maximum heater-cathode volts, 100 peak.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	550	volts
Plate Voltage	300	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	550	volts
Grid-No.2 Voltage	—	300	volts
Grid-No.1 (Control-Grid) Voltage, Negative-bias value	1.3	1.3	volts
Cathode Current	4	55	mA
Plate Dissipation	0.5	9	watts
Grid-No.2 Input	—	1.5	watts
CHARACTERISTICS			
Plate Voltage	250	250	volts
Grid-No.2 Voltage	—	250	volts
Grid-No.1 Voltage	—1.9	—7	volts
Amplification Factor	100	21*	
Plate Resistance (Approx.)	—	45000	ohms
Transconductance	1600	10000	µmhos
Plate Current	1.2	36	mA
Grid-No.2 Current	—	6	mA
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance, for fixed-bias operation	1	0.5	megohm

* Grid No.2 to grid No.1.



9MB

HIGH-MU TRIPLE TRIODE 6GY8

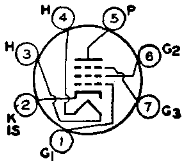
Miniature type used for rf-amplifier, autodyne mixer, and af-control service in FM receivers. Outlines section, 6B; requires miniature 9-contact socket. Heater: volts (ac/dc), 6.3; amperes, 0.45; maximum heater cathode volts, ±100.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values, Each Unit)		
Plate Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation (Each unit)	2	watts
Plate Dissipation (All plates)	5	watts

CHARACTERISTICS	Unit No.1	Units No.2 and No.3	
Plate Voltage	125	125	volts
Grid Voltage	—	—1	volt
Cathode-Bias Resistor	220	—	ohms
Amplification Factor	63	63	
Plate Resistance (Approx.)	14000	14000	ohms
Transconductance	4500	4500	μmhos
Plate Current	4.5	4.5	mA
Grid Voltage (Approx.) for plate current of 20 μA	—	—4	volts

6GX6 SHARP-CUTOFF PENTODE



7EN

Miniature type used in color and black-and-white television receivers and for FM sound-detector service in locked-oscillator, quadrature-grid FM detector circuits as combined detector, limiter, and audio-voltage driver. Tube has two independent control grids. Outlines section, 5C; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.026	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	8	pF
Grid No.1 to Grid No.3	0.12	pF
Grid No.3 to Plate	1.6	pF
Grid No.3 to Cathode, Heater, Grid No.1, Grid No.2, Plate, and Internal Shield	6.5	pF

Class A_1 Amplifier

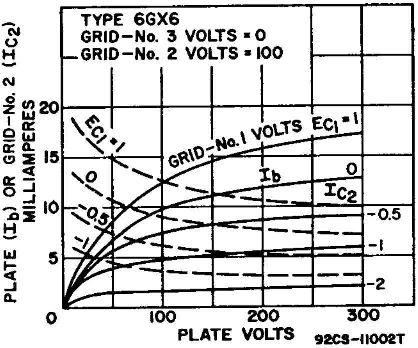
CHARACTERISTICS		
Plate Supply Voltage	150	volts
Grid-No.3 Supply Voltage	0	volts
Grid-No.2 Supply Voltage	100	volts
Grid-No.1 Supply Voltage	0	volts
Cathode-Bias Resistor	180	ohms
Plate Resistance (Approx.)	0.14	megohm
Transconductance, grid No.1 to plate	3700	μmhos
Transconductance, grid No.3 to plate	750	μmhos
Plate Current	3.7	mA
Grid-No.2 Current	3	mA
Grid-No.3 Supply Voltage (Approx.) for plate current of 20 μA	—7	volts
Grid-No.1 Supply Voltage (Approx.) for plate current of 20 μA	—4.5	volts

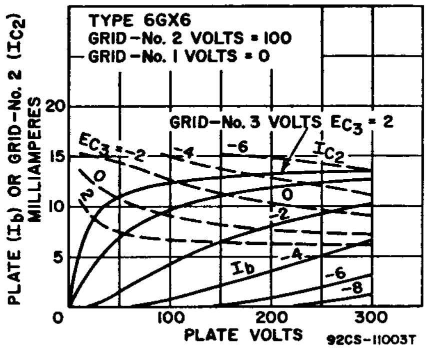
FM Sound Detector

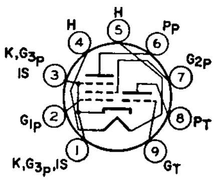
MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	300	volts
Grid-No.3 (Control-Grid) Voltage:		
Negative value (dc and peak ac)	100	volts
Positive value (dc and peak ac)	25	volts
Grid-No.2 (Screen-Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate Dissipation	1.7	watts
Grid-No.3 Input	0.1	watt
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	1	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

MAXIMUM CIRCUIT VALUES

Grid-No.3-Circuit Resistance	0.68	megohm
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.22	megohm
For cathode-bias operation	0.47	megohm







9QA

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE 6GX7

Miniature type used as combined oscillator-mixer tube in vhf tuner circuits of color and black-and-white television receivers. **Outlines section, 6B**; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.4	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:**		
Triode Unit:		
Grid to Plate	1.2	pF
Grid to Cathode, Heater, Pentode Cathode, Grid No.3, and Internal Shield	2.3	pF
Plate to Cathode, Heater, Pentode Cathode, Grid No.3, and Internal Shield	1.9	pF
Pentode Unit:		
Grid No.1 to Plate	0.005	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.4	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3.3	pF
Grid No.1 to Grid No.2	1.6	pF

** With external shield connected to cathode.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	**Triode Unit**	**Pentode Unit**	
Plate Voltage	275	275	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	275	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:			
Positive-bias value	0	0	
Negative-bias value	40	40	volts
Cathode Current	20	20	mA
Plate Dissipation	1.5	2.2	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 137.5 volts	—	0.45	watts
For grid-No.2 voltages between 137.5 and 275 volts	—	See curve page 98	

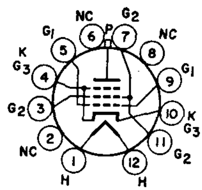
CHARACTERISTICS	Triode Unit		Pentode Unit		
Plate Voltage	100	125	120	125	volts
Grid-No.2 Voltage	—	—	90	125	volts
Grid-No.1 Voltage	—	—1	—	—1	volt
Grid-No.1-Circuit Resistance	0.1	—	0.1	—	megohm
Amplification Factor	40	—	—	—	
Plate Resistance	—	4700	—	200000	ohms
Transconductance	8700	8500	13000	11000	µmhos
Plate Current	12.5	13	8.5	8	mA
Grid-No.2 Current	—	—	2.8	2.5	mA
Grid-No.1 Voltage for plate current of 20 µA	—6	—	—2.5	—	volts

MAXIMUM CIRCUIT VALUES	Triode Unit	Pentode Unit	
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	0.5	megohm

6GY5

16GY5, 21GY5

BEAM POWER TUBE



12DR

Duodecar type used as horizontal-deflection amplifier in television receivers. **Outlines section, 16A; requires duodecar 12-contact socket. Types 16GY5 and 21GY5 are identical with type 6GY5 except for heater ratings.**

	6GY5	16GY5	21GY5	
Heater Voltage (ac/dc)	6.3	15.8	21	volts
Heater Current	1.5	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection			Triode† Connection	
Plate Voltage	*5000*	*60*	*130*	130	volts
Grid-No.2 (Screen-Grid) Voltage	130	130	130	130	volts
Grid-No.1 (Control-Grid) Voltage	—	0	—20	—20	volts
Amplification Factor	-	—	—	4.7	
Plate Resistance (Approx.)	—	—	11000	—	ohms
Transconductance	—	—	9100	—	µmhos
Plate Current	—	410**	50	—	mA
Grid-No.2 Current	—	24**	1.75	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 µA	—66	—	—33	—	volts

** This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

† Grid No.2 tied to plate.

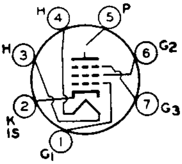
Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage	—55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	800	mA
Average Cathode Current	230	mA
Plate Dissipation††	18	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	220	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

†† A bias resistor or other means is required to protect the tube in absence of excitation.



7EN

SHARP-CUTOFF PENTODE 6GY6

Miniature type used in gated-agc-amplifier circuits and as a noise-inverter tube in color and black-and-white television receivers. Tube has two independent control grids. **Outlines section,** 5C; requires miniature 7-contact socket. For curves of average characteristics, refer to type 6GX6.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.026	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	8	pF
Grid No.1 to Grid No.3	0.12	pF
Grid No.3 to Plate	1.6	pF
Grid No.3 to Cathode, Heater, Plate, Grid No.1, Grid No.2, and Internal Shield	6.5	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Supply Voltage	150	volts
Grid-No.3 Supply Voltage	0	volts
Grid-No.2 Supply Voltage	100	volts
Grid-No.1 Supply Voltage	0	volts
Cathode-Bias Resistor	180	ohms
Plate Resistance (Approx.)	0.14	megohm
Transconductance, Grid No.1 to Plate	3700	μmhos
Transconductance, Grid No.3 to Plate	750	μmhos
Plate Current	3.7	mA
Grid-No.2 Current	3	mA
Grid-No.3 Supply Voltage (Approx.) for plate current of 20 μA	—7	volts
Grid-No.1 Supply Voltage (Approx.) for plate current of 20 μA	—4.5	volts

Gated AGC Amplifier and Noise Inverter

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	300	volts
Peak Positive-Pulse Plate Voltage#	600	volts
Grid-No.3 (Control-Grid) Voltage:		
Negative-bias value	100	volts
Positive-bias value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate Dissipation	1.7	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	1	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

MAXIMUM CIRCUIT VALUES

Grid-No.3-Circuit Resistance	0.68	megohm
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.22	megohm
For cathode-bias operation	0.47	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

Refer to chart at end of section. **6GZ5**

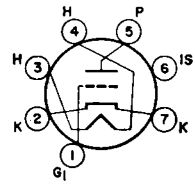
Refer to chart at end of section. **6H6**

Refer to chart at end of section. **6H6GT**

6HA5

HIGH-MU TRIODE

2HA5, 4HA5/PC900

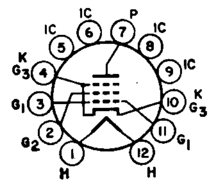


7GM

Miniature type used as rf-amplifier tube in vhf television tuners. **Outlines section, 5A**; requires miniature 7-contact socket. Type 6HA5 is electrically identical with type 6HM5/6HA5. Related types 2HA5 and 4HA5/PC900 are electrically identical with type 6HA5 except for heater voltages of 2.2 and 3.9 volts and heater currents of 0.6 and 0.3 ampere, respectively.

6HB5

BEAM POWER TUBE



12BJ

Duodecar type used as horizontal-deflection amplifier in television receivers. **Outlines section, 15B**; requires duodecar 12-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	1.5	amperes
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection			Triode* Connection	
Plate Voltage	5000	60	130	130	volts
Grid-No.2 (Screen-Grid) Voltage	130	130	130	130	volts
Grid-No.1 (Control-Grid) Voltage	—	0	—20	—20	volts
Amplification Factor	—	—	—	4.7	
Plate Resistance (Approx.)	—	—	11000	—	ohms
Transconductance	—	—	9100	—	µmhos
Plate Current	—	410▪	50	—	mA
Grid-No.2 Current	—	24▪	1.75	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—66	—	—33	—	volts

* Grid No.2 tied to plate.

▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

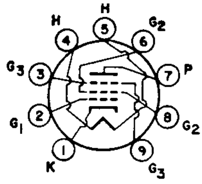
Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6000	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage	—55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	800	mA
Average Cathode Current	230	mA
Plate Dissipation†	18	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	220	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

† A bias resistor or other means is required to protect the tube in absence of excitation.



9NW

POWER PENTODE 6HB6

Miniature type used as vertical-deflection amplifier in color and black-and-white television receivers. **Outlines section, 6G; requires 9-contact socket.**

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.76	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

CHARACTERISTICS

Plate Supply Voltage	60	250	250	volts
Grid No.3		Connected to cathode at socket		
Grid-No.2 Supply Voltage	250	125	250	volts
Grid-No.1 Voltage	0	—	—	volts
Cathode-Bias Resistor	—	33	100	ohms
Mu-Factor, Grid No.2 to Grid No.1	—	—	33	
Plate Resistance (Approx.)	—	28000	24000	ohms
Transconductance	—	24000	20000	µmhos
Plate Current	150•	40	40	mA
Grid-No.2 Current	37•	4.2	6.2	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—	—6.4	—13	volts

• This value can be measured by a method involving a recurrent waveform such that the maximum ratings will not be exceeded.

Vertical-Deflection Amplifier

For operation in a 525-line, 30-frame system

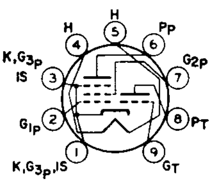
MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Voltage	350	volts
Peak Positive-Pulse Plate Voltage#	2500	volts
DC Grid-No.2 (Screen-Grid) Voltage	300	volts
DC Grid-No.1 (Control-Grid) Voltage	—100	volts
Plate Dissipation	10	watts
Grid-No.2 Input	2	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	1	megohm
For cathode-bias operation	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).



9QA

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE 6HB7

5HB7

Miniature type used as combined oscillator and mixer tube in color and black-and-white television receivers utilizing an intermediate frequency in the order of 40 MHz. **Outlines section, 6B;** requires miniature 9-contact socket. Type 5HB7 is identical with type 6HB7 except for heater ratings.

	5HB7	6HB7	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:△		
Triode Unit:		
Grid to Plate	1.9	pF
Grid to Cathode, Heater, Pentode Grid No.3, and Internal Shield	3	pF
Plate to Cathode, Heater, Pentode Grid No.3, and Internal Shield	1.9	pF

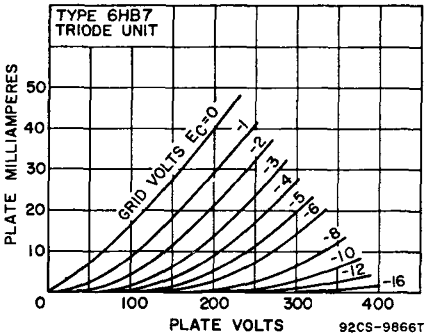
Pentode Unit:		
Grid No.1 to Plate	0.010 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3.4	pF
Heater to Cathode■	3.8	pF

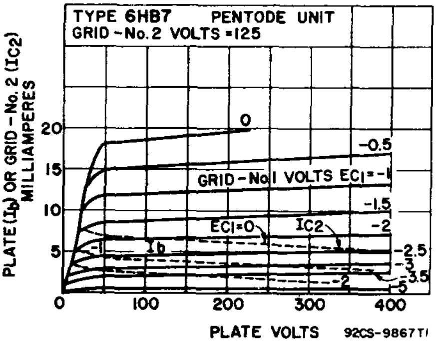
▵ With external shield connected to cathode except as noted.

■ With external shield connected to ground.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:			
Positive-bias value	0	0	volts
Plate Dissipation	2.5	3.1	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	—	*See curve page 98*	
CHARACTERISTICS			
Plate Supply Voltage	150	125	volts
Grid-No.2 Supply Voltage	—	125	volts
Grid-No.1 Supply Voltage	0	—1	volts
Cathode-Bias Resistor	56	—	ohms
Amplification Factor	40	—	
Plate Resistance (Approx.)	0.005	0.2	megohm
Transconductance	8500	6400	μmhos
Plate Current	18	12	mA
Grid-No.2 Current	—	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—12	—9	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	0.5	megohm

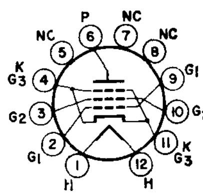




6HE5

BEAM POWER TUBE

Duodecar type used as vertical-deflection amplifier in television receivers. **Outlines section,** 8D; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.8; maximum heater-cathode volts, ±200 peak, 100 average.



12EY

Class A_1 Amplifier

CHARACTERISTICS			
Plate Voltage	60	250	volts
Grid-No.2 (Screen-Grid) Voltage	250	250	volts
Grid-No.1 (Control-Grid) Voltage	0	—20	volts

Plate Resistance (Approx.)	—	50000	ohms
Transconductance	—	4100	µmhos
Plate Current	180▪	43	mA
Grid-No.2 Current	20▪	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—	—50	volts

▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Vertical-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

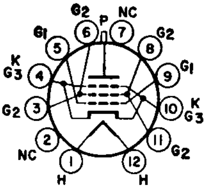
DC Plate Voltage	350	volts
Peak Positive-Pulse Plate Voltage#	2500	volts
Grid-No.2 Voltage	300	volts
Peak Cathode Current	260	mA
Average Cathode Current	75	mA
Plate Dissipation†	12	watts
Grid-No.2 Input†	2.75	watts
Bulb Temperature (At hottest point)	200	°C

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	1	megohm
For cathode-bias operation	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

† A resistor or other means is required to protect the tube in absence of excitation.



12FB

BEAM POWER TUBE 6HF5

Duodecar type used as horizontal-deflection amplifier in color and black-and-white television receivers. Outlines section, 16B; requires duodecar 12-contact socket. Heater: volts (ac/dc), 6.3; amperes, 2.25; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection			Triode* Connection	
Plate Voltage	5000	70	175	125	volts
Grid-No.2 (Screen-Grid) Voltage	125	125	125	125	volts
Grid-No.1 (Control-Grid) Voltage	—	0	—25	—25	volts
Amplification Factor	—	—	—	3	
Plate Resistance (Approx.)	—	—	5600	—	ohms
Transconductance	—	—	11300	—	µmhos
Plate Current	—	570▪	125	—	mA
Grid-No.2 Current	—	34▪	4.5	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—140	—	—54	—	volts

* Grid No.2 tied to plate.

▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	900	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	7500▲	volts
Peak Negative-Pulse Plate Voltage	1100	volts
DC Grid-No.2 Voltage	190	volts
Peak Negative-Pulse Grid-No.1 Voltage	250	volts
Peak Cathode Current	1100	mA
Average Cathode Current	315	mA
Plate Dissipation†	28	watts
Grid-No.2 Input	5.5	watts
Bulb Temperature (At hottest point)	225	°C

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	1	megohm

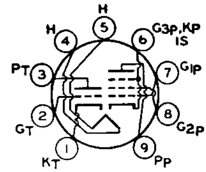
Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

▲ Under no circumstances should this absolute value be exceeded.

† A bias resistor or other means is required to protect the tube in absence of excitation.

6HF8
10HF8

HIGH-MU TRIODE—SHARP-CUTOFF PENTODE



9DX

Miniature type used in color and black-and-white television receiver applications. The triode unit is used in high-gain, sound-if stages and in sync-separator, sync-clipper, and phase-inverter circuits; the pentode unit is used as a video-output amplifier. **Outlines section, 6E;** requires miniature 9-contact socket. For curves of average characteristics, refer to type 6AW8A for the triode unit and to type 6EB8 for the pentode unit. Type 10HF8 is identical with type 6HF8 except for heater ratings.

	6HF8	10HF8	
Heater Voltage (ac/dc)	6.3	10.5	volts
Heater Current	0.75	0.45	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	3.5	pF
Grid to Cathode, Heater, Pentode Cathode, Grid No.3, and Internal Shield	2.8	pF
Plate to Cathode, Heater, Pentode Cathode, Grid No.3, and Internal Shield	2.6	pF
Pentode Unit:		
Grid No.1 to Plate	0.1 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	10	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4.2	pF
Triode Grid to Pentode Plate	0.015 max	pF

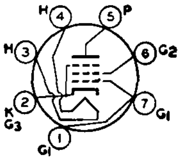
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	1	5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	

CHARACTERISTICS	Triode Unit	Pentode Unit		
Plate Supply Voltage	200	45	200	volts
Grid-No.2 Supply Voltage	—	125	125	volts
Grid-No.1 Voltage	—2	0	—	volts
Cathode-Bias Resistor	—	—	68	ohms
Amplification Factor	70	—	—	
Plate Resistance (Approx.)	17500	—	75000	ohms
Transconductance	4000	—	12500	μmhos
Plate Current	4	40•	25	mA
Grid-No.2 Current	—	15•	7	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—	—9	volts
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—6	—	—	volts

MAXIMUM CIRCUIT VALUES	Triode Unit	Pentode Unit	
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.



7BZ

BEAM POWER TUBE 6HG5

Miniature type used in the audio output stages of television receivers. This type has a controlled cathode warm-up time to minimize extraneous sound during receiver warm-up. **Outlines section, 5D; requires miniature 7-contact socket.**

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Cathode Warm-up Time#	14 min	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.4	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	8	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	8.5	pF

Time interval between application of voltages and rise of plate current to 1 mA; heater volts, 6.3; plate and grid-No.2 volts, 250; cathode-bias resistor, 680 ohms.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

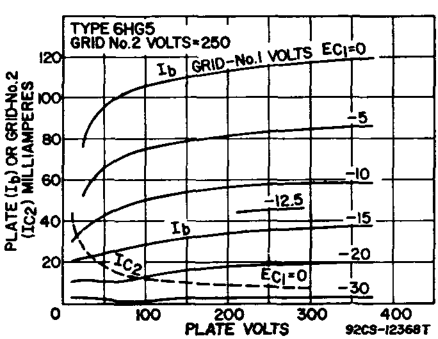
Plate Voltage	275	volts
Grid-No.2 (Screen-Grid) Voltage	275	volts
Plate Dissipation	12	watts
Grid-No.2 Input	2	watts
Bulb Temperature (At hottest point)	250	°C

TYPICAL OPERATION AND CHARACTERISTICS

Plate Voltage	180	250	volts
Grid-No.2 Voltage	180	250	volts
Grid-No.1 (Control-Grid) Voltage	—8.5	—12.5	volts
Peak AF Grid-No.1 Voltage	8.5	12.5	volts
Zero-Signal Plate Current	29	45	mA
Maximum-Signal Plate Current	30	47	mA
Zero-Signal Grid-No.2 Current	3	4.5	mA
Maximum-Signal Grid-No.2 Current	4	7	mA
Plate Resistance (Approx.)	58000	52000	ohms
Transconductance	3700	4100	µmhos
Load Resistance	5500	5000	ohms
Total Harmonic Distortion	8	8	per cent
Maximum-Signal Power Output	2	4.5	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

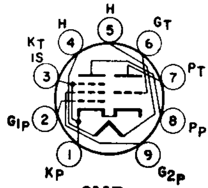


6HG8 Refer to chart at end of section.

6HG8/ ECF86

7HG8/PCF86

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE



9MP

Miniature type with frame-grid pentode unit used as combined oscillator and mixer tubes in vhf color and black-and-white television receivers. **Outlines section, 6B;** requires miniature 9-contact socket. Type 7HG8/PCF86 is identical with type 6HG8/ECF86 except for slightly higher current and dissipation ratings and for heater ratings.

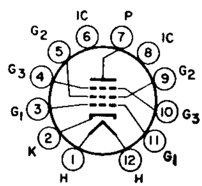
	6HG8/ ECF86	7HG8/ PCF86	
Heater Voltage (ac/dc)	6.3	7.2	volts
Heater Current	0.34	0.3	ampere
Peak Heater-Cathode Voltage	±100 max	±100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	125	250	volts
Grid-No.2 (Screen-Grid) Voltage	—	150	volts
Cathode Current	15	18	mA
Plate Dissipation	1.5	2	watts
Grid-No.2 Input	—	0.5	watt
CHARACTERISTICS			
Plate Voltage	100	170	volts
Grid-No.2 Voltage	—	150	volts
Grid-No.1 (Control-Grid) Voltage	—3	—1.2	volts
Amplification Factor	17	—	
Mu-Factor, Grid No.2 to Grid No.1	—	70	
Plate Resistance (Approx.)	—	0.35	megohm
Transconductance	5500	12000	µmhos
Plate Current	14	10	mA
Grid-No.2 Current	—	3.3	mA
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	—	0.25	megohm
For cathode-bias operation	0.5	0.5	megohm

6HJ5

BEAM POWER TUBE



12FL

Duodecar type used as horizontal-deflection amplifier in television receivers. **Outlines section, 15C;** requires duodecar 12-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	2.25	amperes
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS					
Plate Voltage	20	40	60	135	volts
Grid-No.2 (Screen-Grid) Voltage	110	110	135	135	volts
Grid No.3	Connected to cathode at socket				
Grid-No.1 (Control-Grid) Voltage	0	0	0	—22	volts

Triode Amplification Factor	—	—	—	4.2	
Plate Resistance (Approx.)	—	—	—	5000	ohms
Transconductance	—	—	—	10000	µmhos
Plate Current	240▪	400▪	540▪	80	mA
Grid-No.2 Current	160▪	42▪	48▪	5.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—	—	—70	volts

▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	7000	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage	70	volts
DC Grid-No.2 Voltage	220	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	1000	mA
Average Cathode Current	280	mA
Plate Dissipation†	24	watts
Grid-No.2 Input	6	watts
Grid-No.2 Input (Warm-up Surge)*	12	watts
Bulb Temperature (At hottest point)	240	°C

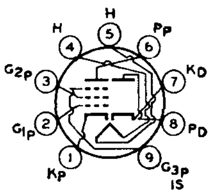
MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

† *A bias resistor or other means is required to protect the tube in absence of excitation.*

* Surge not to exceed 15-second duration.



9CY

DIODE—SHARP-CUTOFF PENTODE 6HJ8

Miniature type used as combined video-detector and if-amplifier tube in television receivers. Outlines section, 6B; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater Warm-up Time (Average)	11	seconds
Peak Heater-Cathode Voltage	±200 max	volts
Direct Interelectrode Capacitances:		
Diode Unit:		
Plate to Cathode and Heater	2.4	pF
Cathode to Plate and Heater	3	pF
Pentode Unit:		
Grid No.1 to Plate	0.015 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	7	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3.2	pF
Diode Plate to Pentode Grid No.1	0.005 max	pF
Diode Cathode to Pentode Plate	0.15 max	pF
Diode Plate to Pentode Plate	0.035 max	pF

Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	**See curve page 98**	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	3.2	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	**See curve page 98**	

CHARACTERISTICS

Plate Dissipation		
Plate Supply Voltage	125	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts

Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.2	megohm
Transconductance	9300	µmhos
Plate Current	11.5	mA
Grid-No.2 Current	3.6	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—6	volts
Grid-No.1 Voltage (Approx.) for plate current of 2 mA	—3	volts

Diode Unit

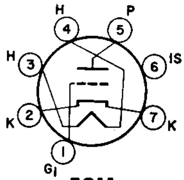
MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Current	5	mA

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 50 mA	10	volts

6HK5 HIGH-MU TRIODE



7GM

Miniature type with frame grid used in vhf tuners of television receivers. Outlines section, 5C; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.19	ampere
Peak Heater-Cathode Voltage	±100 max	volts
Direct Interelectrode Capacitances:*		
Grid to Plate	0.29	pF
Grid to Cathode, Heater, and Internal Shield	4.4	pF
Plate to Cathode, Heater, and Internal Shield	2.6	pF
Heater to Cathode	2.5	pF

* With external shield.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	200	volts
Grid Voltage, Negative-bias value	50	volts
Cathode Current	22	mA
Plate Dissipation	2.3	watts

CHARACTERISTICS

Plate Voltage	135	volts
Grid Voltage	—1	volt
Amplification Factor	75	
Plate Resistance (Approx.)	5000	ohms
Transconductance	15000	µmhos
Plate Current	12.5	mA
Input Resistance**	600	ohms
Input Capacitance**	9	pF
Noise Figure#	4.2	dB
Grid Voltage (Approx.) for transconductance of 150 µmhos	—5	volts
Grid Voltage (Approx.) for transconductance of 1500 µmhos	—2.6	volts

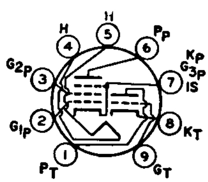
MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance, for cathode-bias operation	1	megohm

** Measured at 200 MHz with plate effectively grounded for rf voltages.

For a neutralized triode amplifier at a frequency of 200 MHz with signal source impedance adjusted for minimum noise output.

6HL8 MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE



9AE

Miniature type used in color and black-and-white television receiver applications. The triode unit is used as a sync-separator or voltage-amplifier tube, and the pentode unit is used as a video if-amplifier, agc-amplifier, or reactance tube. **Outlines section,** 6B; requires miniature 9-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.6; warm-up time (average), 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

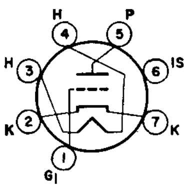
	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage		See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	2.5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	

CHARACTERISTICS

	Triode Unit	Pentode Unit	
Plate Voltage	125	125	volts
Grid-No.2 Voltage	—	125	volts
Grid-No.1 Voltage	—1	—1	volt
Amplification Factor	40	—	
Plate Resistance (Approx.)	5000	150000	ohms
Transconductance	7000	10000	μmhos
Plate Current	12.5	12	mA
Grid-No.2 Current	—	4.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—	—7	volts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	1	—	megohm



7GM

HIGH-MU TRIODE

6HM5/ 6HA5

3HM5/3HA5

Miniature type used as rf-amplifier tube in vhf color and black-and-white television tuners. Outlines section, 5C; requires miniature 7-contact socket. Type 3HM5/3HA5 is identical with type 6HM5/6HA5 except for heater ratings.

	3HM5/3HA5	6HM5/6HA5	
Heater Voltage (ac/dc)	2.7	6.3	volts
Heater Current	0.45	0.18	ampere
Peak Heater-Cathode Voltage	±110 max	±110 max	volts

Direct Interelectrode Capacitances:

Grid to Plate	0.36	pF
Grid to Cathode, Heater, Internal Shield, and External Shield	4.3	pF
Plate to Cathode, Heater, Internal Shield, and External Shield	0.080	pF
Cathode to Plate	2.9	pF
Cathode to Heater, Grid, Internal Shield, and External Shield	3.1	pF
Heater to Cathode	2.3	pF
Heater to Grid	0.070 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

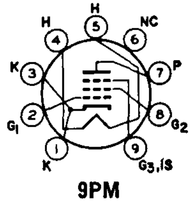
DC Plate Voltage	220	volts
DC Plate Supply Voltage	600	volts
Grid Voltage	—50	volts
Cathode Current	22	mA
Plate Dissipation	2.6	watts

CHARACTERISTICS AND TYPICAL OPERATION

	Fixed Bias		Cathode Bias		
DC Plate Supply Voltage	135	135	135	135	volts
Plate-Load Resistor	—	—	1000	5600	ohms
Internal-Shield Voltage	0	0	0	0	volts
DC Grid Voltage	—1	—2.7	—	—	volts
Cathode-Bias Resistor	—	—	0	87	ohms
Amplification Factor	72	—	80	72	
Transconductance	14500	1500	20000	14500	μmhos
Plate Current	11.5	—	19	11.5	mA
DC Grid Current	—	—	10	—	μA
Grid-No.1 Voltage for one-per-cent transconductance	—	—	—5.3	—8.1	volts

6HM6 SHARP-CUTOFF PENTODE

4HM6



9PM

Miniature type with frame grid used in the if-amplifier stages of television receivers. **Outlines** section, 6B; requires miniature 9-contact socket. Type 4HM6 is identical with type 6HM6 except for heater ratings.

	4HM6	6HM6	
Heater Voltage (ac/dc)	4.2	6.3	volts
Heater Current	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:	Unshielded	Shielded	
Grid No.1 to Plate	0.031	0.024	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	8.7	8.7	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.15	3	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	250	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Negative-bias value	50	volts
Cathode Current	25	mA
Plate Dissipation	2.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 125 volts	2.5	watts
For grid-No.2 voltages between 125 and 250 volts	See curve page 98	

CHARACTERISTICS

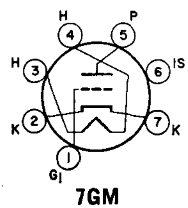
Plate Supply Voltage	125	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.156	megohm
Transconductance	15000	μmhos
Plate Current	13	mA
Grid-No.2 Current	3.2	mA
Grid-No.1 Voltage (Approx.) for transconductance of 100 μmhos	—3	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

6HQ5 HIGH-MU TRIODE

2HQ5, 3HQ5, 4HQ5



7GM

Miniature type used as grounded-cathode rf-amplifier tube in vhf tuners of television receivers. **Outlines** section, 5C; requires miniature 7-contact socket. Types 2HQ5, 3HQ5, and 4HQ5 are identical with type 6HQ5 except for heater ratings.

	2HQ5	3HQ5	4HQ5	6HQ5	
Heater Voltage (ac/dc)	2.4	3	4.2	6.3	volts
Heater Current	0.6	0.45	0.3	0.2	ampere
Heater Warm-up Time (Average)	11	11	11	—	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	±100 max	±100 max	volts
Direct Interelectrode Capacitances (Approx.):*					
Grid to Plate				0.52	pF
Grid to Cathode, Heater, and Internal Shield				5	pF
Plate to Cathode, Heater, and Internal Shield				3.5	pF
Heater to Cathode				2.5	pF

With external shield connected to cathode.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	200	volts
Grid Voltage, Negative-bias Value	50	volts
Cathode Current	22	mA
Plate Dissipation	2.5	watts

CHARACTERISTICS

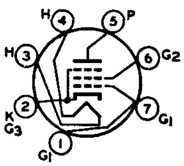
Plate Voltage	135	volts
Grid Voltage	—1	volt
Amplification Factor	78	
Plate Resistance	5400	ohms
Transconductance	15000	μmhos
Plate Current	11.5	mA
Input Resistance**	275	ohms
Input Capacitance**	11.2	pF
Noise Figure#	4.7	dB
Grid Voltage (Approx.) for transconductance of 150 μmhos	—4.2	volts
Grid Voltage (Approx.) for transconductance of 1500 μmhos	—2.5	volts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance, for cathode-bias operation	1	megohm

** Measured at 200 MHz with heater volts = 6.3 volts and plate effectively grounded for rf voltages.

For a neutralized triode amplifier at a frequency of 200 MHz with signal source impedance adjusted for minimum noise output.



7BZ

BEAM POWER TUBE 6HR5

Miniature type used as vertical-deflection amplifier in television receivers. **Outlines section,** 5D; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.35	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	8.3	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	8.2	pF

Vertical-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	260	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	1500	volts
Grid-No.2 (Screen-Grid) Voltage	270	volts
Peak Negative-Pulse Grid-No.1 Voltage	150	volts
Plate Dissipation	8	watts
Peak Cathode Current	125	mA
Average Cathode Current	35	mA
Grid-No.2 Input	2	watts

CHARACTERISTICS

Plate Voltage	50	260	volts
Grid-No.2 Voltage	250	270	volts
Grid-No.1 Voltage	0	—19	volts
Transconductance	—	3600	μmhos
Plate Current	105	30	mA
Grid-No.2 Current	25	2.3	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—43	volts

MAXIMUM CIRCUIT VALUE

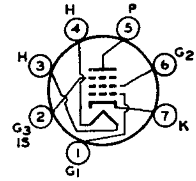
Grid-No.1-Circuit Resistance	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

6HR6

19HR6

SEMIREMOTE-CUTOFF PENTODE



7BK

Miniature type used as if-amplifier tube in FM receivers. Outlines section, 5C; requires miniature 7-contact socket. Type 19HR6 is identical with type 6HR6 except for heater ratings.

	6HR6	19HR6	
Heater Voltage (ac/dc)	6.3	18.9	volts
Heater Current	0.45	0.15	ampere
Heater Warm-up Time (Average)	11	17	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:			
Grid No.1 to Plate		0.006 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield		8.8	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield		5.2	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

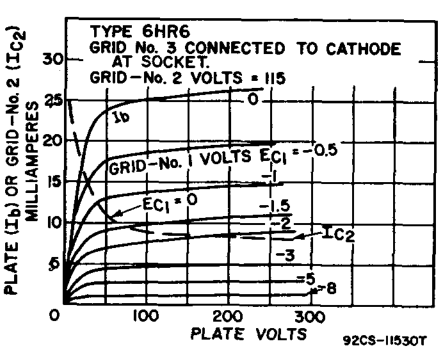
Plate Supply Voltage	300	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate Dissipation	3	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	1	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

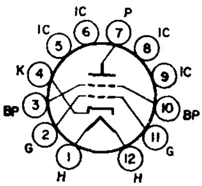
CHARACTERISTICS

Plate Supply Voltage	200	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	115	volts
Grid-No.1 Supply Voltage	0	volts
Cathode-Bias Resistor	68	ohms
Plate Resistance (Approx.)	0.5	megohm
Transconductance	8500	μmhos
Plate Current	13.2	mA
Grid-No.2 Current	4.3	mA
Grid-No.1 Voltage (Approx.) for transconductance of 60 μmhos	—15	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	1	megohm





12GY

BEAM TRIODE 6HS5

Duodecar type used as a pulse-type regulator in the high-voltage power supply of color television receivers. **Outlines section, 15E; requires duodecar 12-contact socket. Heater:** volts (ac/dc), 6.3; amperes, 1.5.

Class A_1 Amplifier

CHARACTERISTICS

Pulse Plate Voltage*	3500	volts
Grid No.2 (Beam Plate)	Connected to cathode at socket	
Grid-Voltage, Negative-bias value	4.4	volts
Peak Plate Current	300	mA
Amplification Factor	300	
Transconductance	65000	µmhos
Plate Resistance (Approx.)	4600	ohms
Grid Voltage (Approx.) for plate current of 1 mA	—13	volts

* Duty cycle of the pulse must be less than 2.5%.

High-Voltage Regulator Service

For operation in a 525-line, 30-frame system

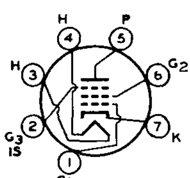
MAXIMUM RATINGS (Design-Maximum Values)

Peak Plate Voltage#		5500	volts
Plate Dissipation		30	watts
Peak Plate Current		325	mA
Heater-Cathode Voltage:			
Peak value	+200	—450	volts
Average value		100	volts
Bulb Temperature (At hottest point)		220	°C

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance▲	0.1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
▲ Larger values of grid-circuit resistance may be used if provisions are made to protect the tube.



7BK

SHARP-CUTOFF PENTODE 6HS6

Miniature type used as if-amplifier and limiter tube in FM receivers. **Outlines section, 5C; requires miniature** *7-contact* socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.006 max	volts
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	8.8	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.2	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	300	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive Value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	

Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate Dissipation	3	volts
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	1	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

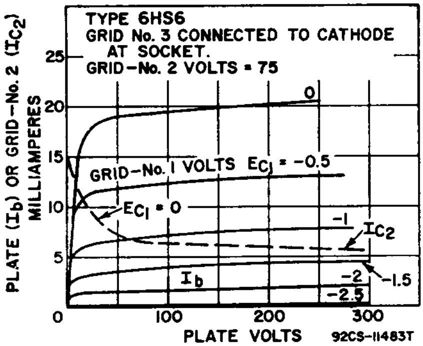
CHARACTERISTICS

Plate Supply Voltage	75	150	volts
Grid No.3	Connected to cathode at socket		
Grid-No.2 Supply Voltage	75	75	volts
Grid-No.1 Supply Voltage	0	0	volts
Cathode-Bias Resistor	68	68	ohms
Amplification Factor•	50	—	
Plate Resistance (Approx.)	—	0.5	megohm
Transconductance	—	9500	μmhos
Plate Current	—	8.8	mA
Grid-No.2 Current	—	2.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—	—4	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	1	megohm

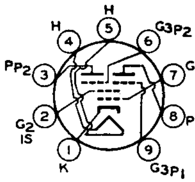
• Grid No.2 connected to plate.



6HS8
4HS8

SHARP-CUTOFF TWIN PENTODE

Miniature type used in agc amplifier, sync, and noise-limiting circuits of color and black-and-white television receivers. One pentode unit is used as combined sync separator and sync clipper; second pentode unit is used as agc amplifier. **Outlines section, 6E;** requires miniature 9-contact socket. Type 4HS8 is identical with type 6HS8 except for heater ratings.



9FG

	4HS8	6HS8	
Heater Voltage (ac/dc)	4.2	6.3	volts
Heater Current	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:			
Grid No.3 to Plate (Each Unit)		2	pF
Grid No.1 to All Other Electrodes		6	pF
Grid No.3 (Each Unit) to All Other Electrodes		3.6	pF
Plate (Each Unit) to All Other Electrodes		3	pF
Grid No.3 (Unit No.1) to Grid No.3 (Unit No.2)		0.015 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage (Each Unit)	300	volts
Grid-No.3 (Suppressor-Grid) Voltage (Each Unit):		
Peak positive value	50	volts
DC negative value	50	volts
DC positive value	3	volts
Grid-No.2 (Screen-Grid) Voltage	150	volts
Grid-No.1 (Control-Grid) Voltage, Negative-bias value	50	volts
Cathode Current	12	mA
Plate Dissipation (Each Unit)	1.1	watts
Grid-No.2 Input	0.75	watt

CHARACTERISTICS — **With One Unit Operating•**

Plate Voltage	100	100	volts
Grid-No.3 Voltage	0	0	volts
Grid-No.2 Voltage	67.5	67.5	volts
Grid-No.1 Voltage	0	▪	volts
Transconductance, Grid No.3 to Plate	—	450	µmhos
Transconductance, Grid No.1 to Plate	1100	—	µmhos
Plate Current	—	2	mA
Grid-No.3 Voltage (Approx.) for plate current of 100 µA	—	—3.5	volts
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—	—2.3	volts

With Both Units Operating

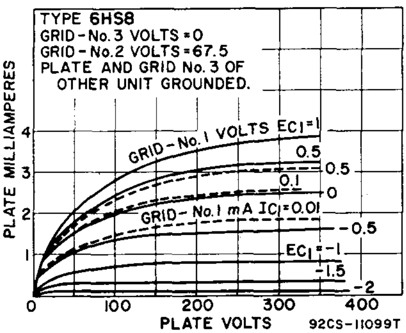
Plate Voltage (Each Unit)	100	100	volts
Grid-No.3 Voltage (Each Unit)	—10	0	volts
Grid-No.2 Voltage	67.5	67.5	volts
Grid-No.1 Voltage	▪	▪	volts
Plate Current (Each Unit)	—	2	mA
Grid-No.2 Current	7	4.4	mA
Cathode Current	7.1	8.5	mA

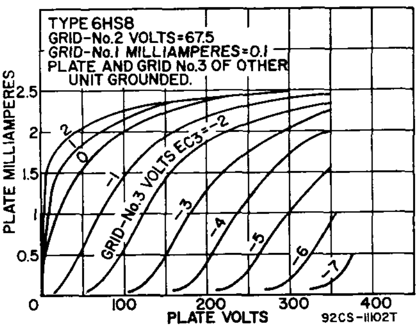
MAXIMUM CIRCUIT VALUES

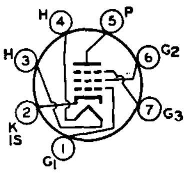
Grid-No.3-Circuit Resistance (Each Unit)	0.5	megohm
Grid-No.1-Circuit Resistance	0.5	megohm

• With plate and grid No.3 of other unit connected to ground.

▪ Adjusted to give grid-No.1 current of 0.1 milliampere.







7EN

SHARP-CUTOFF PENTODE **6HZ6**

5HZ6

Miniature type used as sound-detector tube in FM and color and black-and-white television receivers. Tube has two independent control grids. **Outlines section, 5C;** requires miniature 7-contact socket. Type 5HZ6 is *identical with type 6HZ6 except for heater ratings.*

	5HZ6	6HZ6	
Heater Voltage (ac/dc)	4.75	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.023	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	8.2	pF
Grid No.1 to Grid No.3	0.09	pF
Grid No.3 to Plate	1.6	pF
Grid No.3 to Cathode, Heater, Grid No.1, Grid No.2, Plate, and Internal Shield	7.2	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Supply Voltage	150	volts
Grid-No.3 Supply Voltage	0	volts
Grid-No.2 Supply Voltage	100	volts
Grid-No.1 Supply Voltage	0	volts
Cathode-Bias Resistor	180	ohms
Plate Resistance (Approx.)	0.11	megohm
Transconductance, Grid No.1 to Plate	3400	μmhos
Transconductance, Grid No.3 to Plate	600	μmhos
Plate Current	3.2	mA
Grid-No.2 Current	3.2	mA
Grid-No.3 Supply Voltage (Approx.) for plate current of 20 μA	—7	volts
Grid-No.1 Supply Voltage (Approx.) for plate current of 20 μA	—4.5	volts

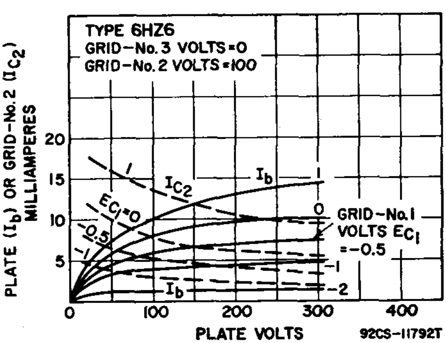
FM Sound Detector

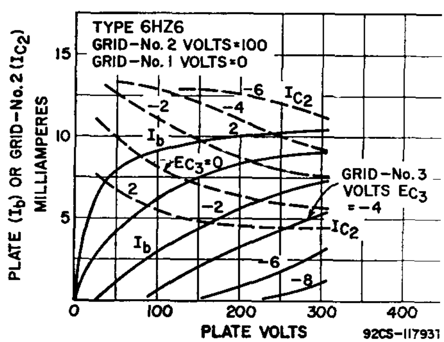
MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	300	volts
Grid-No.3 (Control-Grid) Voltage:		
Negative value (dc and peak ac)	100	volts
Positive value (dc and peak ac)	25	volts
Grid-No.2 (Screen-Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate Dissipation	1.7	watts
Grid-No.3 Input	0.1	watt
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	1	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

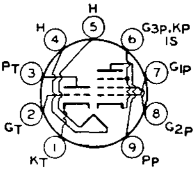
MAXIMUM CIRCUIT VALUES

Grid-No.3-Circuit Resistance	0.68	megohm
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.22	megohm
For cathode-bias operation	0.47	megohm





HIGH-MU TRIODE— SHARP-CUTOFF PENTODE 6HZ8



9DX

Miniature type used in television receiver applications. The triode unit is used as a voltage amplifier or sync separator, and the pentode unit as a video amplifier. **Outlines section,** 8E; requires miniature 9-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 1.125; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive bias value	0	0	volts
Plate Dissipation	1	8	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	2	watts
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	

CHARACTERISTICS				
Plate Voltage	200	60	250	volts
Grid-No.2 Supply Voltage	—	170	170	volts
Grid-No.1 Voltage	−2	0	—	volts
Cathode-Bias Resistor	—	—	100	ohms
Amplification Factor	70	—	—	
Plate Resistance (Approx.)	—	—	0.14	megohm
Transconductance	4000	—	12600	μmhos
Plate Current	3.5	90▪	29	mA
Grid-No.2 Current	—	22.5▪	6	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	−5	—	−11.5	volts

MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

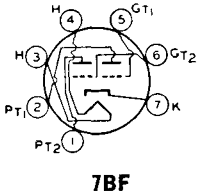
▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Refer to chart at end of section. **6J5 6J5GT**

Refer to chart at end of section. **6J6**

MEDIUM-MU TWIN TRIODE 6J6A

5J6



7BF

Miniature type used as combined rf power amplifier and oscillator or as twin af amplifier. With push-pull arrangement of the grids and the plates in parallel, this type can also be used as a mixer at frequencies as high as 600 MHz. **Outlines section,** 5C; requires miniature 7-contact socket. Type 5J6 is identical with type 6J6A except for heater ratings.

	5J6	6J6A	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	volts
Direct Interelectrode Capacitances (Each Unit, Approx.):	Unshielded	Shielded	
Grid to Plate	1.6	1.6	pF
Grid to Cathode and Heater	2.2	2.6	pF
Plate to Cathode and Heater (Unit No.1)	0.4	1.6	pF
Plate to Cathode and Heater (Unit No.2)	0.4	1	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	300	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation	1.5	watts

CHARACTERISTICS

Plate Voltage	100	volts
Cathode-Bias Resistor	50†	ohms
Amplification Factor	38	
Plate Resistance (Approx.)	7100	ohms
Transconductance	5300	μmhos
Plate Current	8.5	mA

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance:		
For fixed-bias operation	Not recommended	
For cathode-bias operation	0.5	megohm

† Value is for both units operating at the specified conditions.

RF Power Amplifier and Oscillator—Class C Telegraphy

Key-down conditions per tube without modulation

MAXIMUM RATINGS (Design-Center Values, Each Unit)

Plate Voltage	300	volts
Grid Voltage:		
Negative-bias value	40	volts
Positive-bias value	0	volts
Plate Current	15	mA
Grid Current	8	mA
Plate Input	4.5	watts
Plate Dissipation	1.5	watts

TYPICAL PUSH-PULL OPERATION (Both Units)

Plate Voltage	150	volts
Grid Voltage°	—10	volts
Plate Current	30	mA
Grid Current (Approx.)	16	mA
Driving Power (Approx.)	0.35	watt
Power Output (Approx.)	3.5	watts

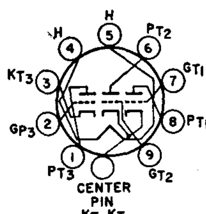
° Obtained by grid resistor (625 ohms), cathode-bias resistor (220 ohms), or fixed supply.

6J7
6J7G
6J7GT

Refer to chart at end of section.

6J8G

Refer to chart at end of section.



10G

6J9 HIGH-MU TRIPLE TRIODE

Miniature type used as rf-amplifier, oscillator, and mixer into the vhf range. **Outlines section, 6B**, except center pin is added to base; requires miniature 10-contact socket.

	Series	Parallel	
Heater Arrangement	Series	Parallel	
Heater Voltage (ac/dc)	6.3	6.3	volts
Heater Current	0.45	0.45	mA
Heater Warm-up Time (Average)	11	—	seconds
Peak Heater-Cathode Voltage	±100 max	±100 max	volts

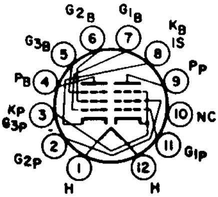
Class A_1 Amplifier (Each Unit Unless Otherwise Specified)

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate Dissipation	2	watts
Total Plate Dissipation (All plates)	5	watts

CHARACTERISTICS

Plate Voltage	125	volts
Grid Voltage	—1	volt
Amplification Factor	57	
Plate Resistance	11000	ohms
Transconductance	5200	µmhos
Plate Current	6	mA
Grid Voltage (Approx.) for plate current of 20 µA	—5.4	volts



12BT

PENTODE— BEAM POWER TUBE 6J10

Duodecar type used in FM and color and black-and-white television receivers. The pentode unit is used as a gated-beam discriminator and the beam power unit is used in audio power-output stages in FM and television limiter and discriminator applications. Outlines section, 8B; requires duodecar 12-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.95	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	275	volts
Grid-No.2 (Screen-Grid) Voltage	275	volts
Plate Dissipation	10	watts
Grid-No.2 Input	2	watts

CHARACTERISTICS AND TYPICAL OPERATION

Plate Voltage	250	volts
Grid-No.2 Voltage	250	volts
Grid-No.1 Voltage	—8	volts
Peak AF Grid-No.1 Voltage	8	volts
Plate Resistance (Approx.)	0.1	megohm
Transconductance	6500	µmhos
Zero-Signal Plate Current	35	mA
Maximum-Signal Plate Current	39	mA
Zero-Signal Grid-No.2 Current	2.5	mA
Maximum-Signal Grid-No.2 Current	7	mA
Load Resistance	5000	ohms
Total Harmonic Distortion (Approx.)	10	per cent
Maximum-Signal Power Output	4.2	watts

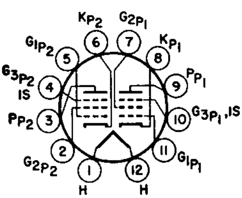
MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	0.5	megohm

Beam Power Unit as Gated-Beam Discriminator

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	330	volts
Grid-No.2 (Accelerator-Grid) Voltage	110	volts
Peak Positive Grid-No.1 Voltage	60	volts
Average Cathode Current	13	mA



12BW

SHARP-CUTOFF TWIN PENTODE 6J11

Duodecar type used as if-amplifier tube in television receivers. Outlines section, 8A; requires duodecar 12-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.8	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:**		
Unit No. 1:		
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, Grid No.3 of Unit No.2, and Internal Shield	11	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, Grid No.3 of Unit No.2, and Internal Shield	2.8	pF
Unit No. 2:		
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, Grid No.3 of Unit No.1, and Internal Shield	11	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, Grid No.3 of Unit No.1, and Internal Shield	3.2	pF
Grid No.1 to Plate (Each Unit)	0.04 max	pF
Cathode of Unit No.1 to Cathode of Unit No.2	0.02 max	pF
Grid No.1 of Unit No.1 to Plate of Unit No.2	0.003 max	pF
Grid No.1 of Unit No.2 to Plate of Unit No.1	0.003 max	pF
Plate of Unit No.1 to Plate of Unit No.2	0.03 max	pF

** With external shield connected to cathode.

Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	3.1	watts
Grid-No.2 Input:		
For grid No.2 voltages up to 165 volts	0.65	watt
For grid No.2 voltages between 165 and 300 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	125	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket	
Grid-No.2 Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.2	megohm
Transconductance	13000	μmhos
Plate Current	11	mA
Grid-No.2 Current	3.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—3	volts

MAXIMUM CIRCUIT VALUE

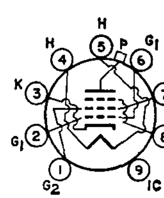
Grid-No.1-Circuit Resistance, for cathode-bias operation	0.25	megohm

6JB6 Refer to chart at end of section.

6JB6A

BEAM POWER TUBE

12JB6A, 17JB6A



9QL

Novar types used as high-efficiency horizontal-deflection amplifiers in television receivers. **Outlines section, 32A**; requires novar 9-contact socket. Types 12JB6A and 17JB6A are identical with type 6JB6A except for heater ratings.

	6JB6A	12JB6A	17JB6A	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.2	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	15	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	6	pF

Class A_1 Amplifier

CHARACTERISTICS	Triode Connection▲	Pentode Connection		
Plate Voltage	150	60	150	volts
Grid No.3 (Suppressor Grid)		Connected to cathode at socket		
Grid-No.2 (Screen-Grid) Voltage	—	150	150	volts
Grid-No.1 (Control-Grid) Voltage	—22.5	0	—22.5	volts
Mu-Factor, Grid No.2 to Grid No.1	4.4	—	—	
Plate Resistance (Approx.)	—	—	15000	ohms
Transconductance	—	—	7100	μmhos
Plate Current	—	390■	70	mA
Grid-No.2 Current	—	32■	2.1	mA
Grid-No.1 Voltage for plate current of 1 mA	—	—	—42	volts

▲ Grid No.2 connected to plate.

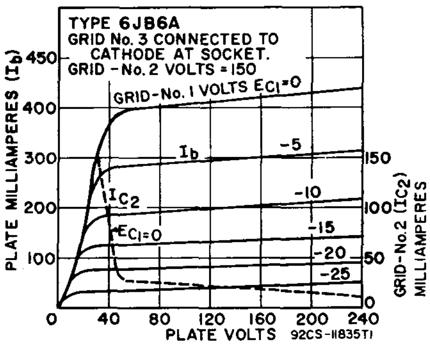
■ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

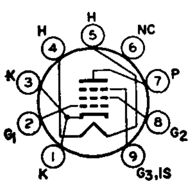
MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage†	70	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage	—55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	550	mA
Average Cathode Current	175	mA
Plate Dissipation•	17.5	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	240	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance, for grid-resistor-bias operation	1	megohm

\# Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
† For horizontal-deflection service, a positive voltage may be applied to grid No.3 to minimize "snivets" interference in both vhf and uhf television receivers. A typical value is 30 volts.
• A bias resistor or other means is required to protect the tube in absence of excitation.



6JC6
6JC6A

SHARP-CUTOFF PENTODE



9PM

3JC6A, 4JC6, 4JC6A

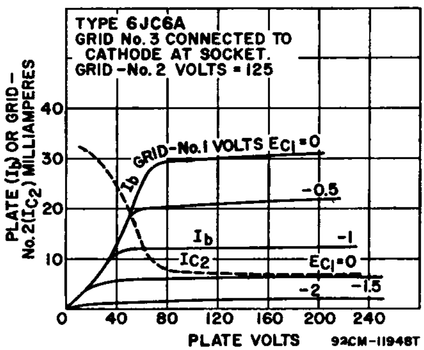
Miniature type with frame grid used in if-amplifier stages of color and black-and-white television receivers utilizing intermediate frequencies in the order of 40 MHz. Outlines section, 6B; requires miniature 9-contact socket. Type 4JC6 is identical with type 6JC6 except for heater ratings. Types 3JC6A and 4JC6A are identical with type 6JC6A except for heater ratings.

	3JC6A	4JC6 4JC6A	6JC6 6JC6A	
Heater Voltage (ac/dc)	3.5	4.5	6.3	volts
Heater Current	0.6	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:	6JC6	6JC6A	
Grid No.1 to Plate	0.019 max	0.019 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	8.2	8.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3	3	pF

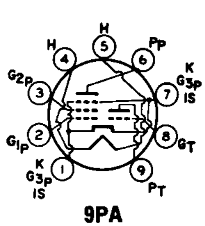
Class A_1 Amplifier

	6JC6	6JC6A	
MAXIMUM RATINGS (Design-Maximum Values)			
Plate Voltage	330	330	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	330	volts
Grid-No.2 Voltage			See curve page 98
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	3.1	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	0.6	0.7	watt
For grid-No.2 voltages between 165 and 330 volts			See curve page 98
CHARACTERISTICS			
Plate Supply Voltage	125	125	volts
Grid No.3	Connected to cathode at socket		
Grid-No.2 Supply Voltage	125	125	volts
Cathode-Bias Resistor	56	56	ohms
Plate Resistance (Approx.)	0.18	0.18	megohm
Transconductance	15000	16000	µmhos
Plate Current	13	14	mA
Grid-No.2 Current	3.2	3.4	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—3	—3	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.25	0.25	megohm
For cathode-bias operation	1	1	megohm



6JC8

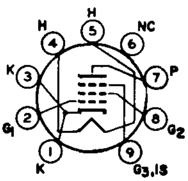
MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE



Miniature type used as combined vhf oscillator and mixer tube in television receivers. **Outlines section,** 6B; requires miniature 9-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.45; warm-up time (average), 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A₁ Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit		
Plate Voltage	275	275		volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	275		volts
Grid-No.2 Voltage		See curve page 98		
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0		volts
Plate Dissipation	1.7	2.3		watts
Grid-No.2 Input:				
For grid-No.2 voltages up to 137.5 volts	—	0.45		watt
For grid-No.2 voltages between 137.5 and 275 volts		See curve page 98		
CHARACTERISTICS				
Plate Voltage	125	100	125	volts
Grid-No.2 Voltage	—	70	125	volts
Grid-No.1 Voltage	—1	0	—1	volt
Amplification Factor	40	—	—	
Plate Resistance (Approx.)	6000	—	300000	ohms
Transconductance	6500	5700	5500	μmhos
Plate Current	12	—	9	mA
Grid-No.2 Current	—	—	2.2	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—7	—	6.5	volts
MAXIMUM CIRCUIT VALUES				
Grid-No.1-Circuit Resistance:				
For fixed-bias operation	—	0.1		megohm
For cathode-bias operation	—	0.5		megohm



9PM

SHARP-CUTOFF PENTODE 6JD6

3JD6, 4JD6

Miniature type used as if-amplifier tube in color and black-and-white television receivers utilizing an intermediate frequency in the order of 40 MHz. Outlines section, 6B; requires miniature 9-contact socket. Types 3JD6 and 4JD6 are identical with type 6JD6 except for heater ratings.

	3JD6	4JD6	6JD6	
Heater Voltage (ac/dc)	3.5	4.5	6.3	volts
Heater Current	0.6	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

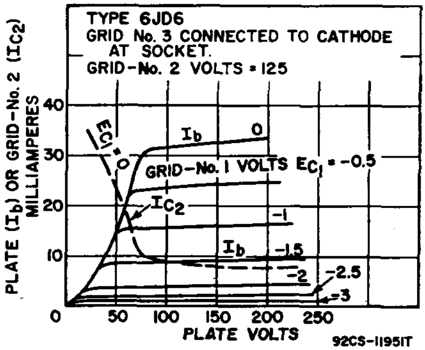
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.019 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	8.2	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3	pF

Class A₁ Amplifier

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	330	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	2.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.6	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	
CHARACTERISTICS		
Plate Supply Voltage	125	volts
Grid-No.3 Voltage	0	volts
Grid-No.2 Supply Voltage	125	volts
Grid-No.1 Supply Voltage	0	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	160000	ohms
Transconductance	14000	μmhos
Plate Current	15	mA
Grid-No.2 Current	4	mA
Grid-No.1 Voltage (Approx.) for transconductance of 600 μmhos	—4.5	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

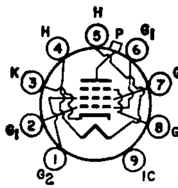


6JE6 Refer to chart at end of section.

6JE6A

24JE6A

BEAM POWER TUBE



9QL

Novar type used as horizontal-deflection amplifier in color television receivers. **Outlines section, 32B;** requires novar 9-contact socket. Type 24JE6A is identical to type 6JE6A except for heater ratings.

	6JE6A	24JE6A	
Heater Voltage (ac/dc)	6.3	24	volts
Heater Current	2.5	0.6	amperes
Heater Warm-up Time	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.56	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	22	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	11	pF

Class A_1 Amplifier

CHARACTERISTICS	Triode Connection▲	Pentode Connection		
Plate Voltage	125	70	175	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket			
Grid-No.2 (Screen-Grid) Voltage	—	125	125	volts
Grid-No.1 (Control-Grid) Voltage	−25	0	−25	volts
Amplification Factor	3	—	—	
Plate Resistance (Approx.)	—	—	—	ohms
Transconductance	—	—	—	µmhos
Plate Current	—	600†	130	mA
Grid-No.2 Current	—	36†	2.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—	−54	volts

† This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

▲ Grid No.2 connected to plate.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	990	volts
Peak Positive-Pulse Plate Voltage#	7500	volts

Peak Negative-Pulse Plate Voltage	1100	volts
DC Grid-No.3 Voltage•	75	volts
DC Grid-No.2 Voltage	220	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	1200	mA
Average Cathode Current	350	mA
Grid-No.2 Input	5	watts
Plate Dissipation▪	30	watts
Bulb Temperature (At hottest point)	250	°C

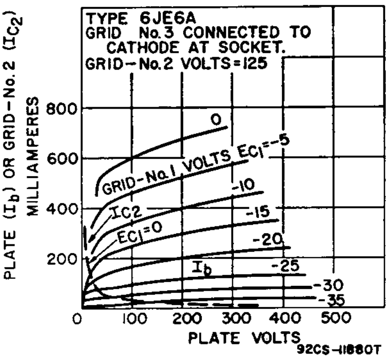
MAXIMUM CIRCUIT VALUES

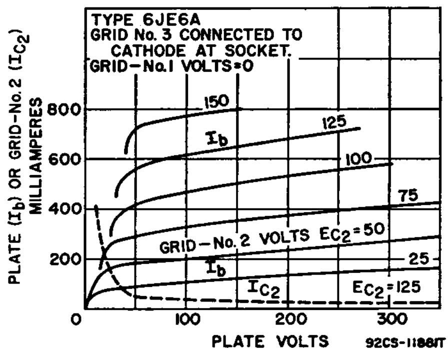
Grid-No.1-Circuit Resistance:		
For grid-resistor-bias operation▪	0.47	megohm
For plate-pulsed operation (horizontal-deflection circuits only)	10	megohms

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

• In this service, a positive voltage may be applied to grid No.3 to minimize "snivets" interference; a typical value for this voltage is 30 volts.

▪ A bias resistor or other means is required to protect the tube in absence of excitation.



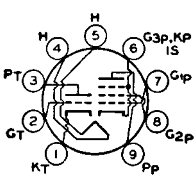


Refer to type 6LQ6. **6JE6B**

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE

6JE8

11JE8



9DX

Miniature type used in television receiver applications. The triode unit is used as a voltage amplifier or sync separator, and the pentode unit as a video amplifier. Outlines section, 6E; requires miniature 9-contact socket. Type 11JE8 is identical with type 6JE8 except for heater ratings.

	6JE8	11JE8	
Heater Voltage (ac/dc)	6.3	10.9	volts
Heater Current	0.78	0.45	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A₁ Amplifier

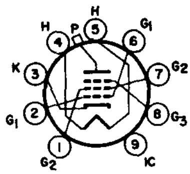
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit		
Plate Voltage	300	330		volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330		volts
Grid-No.2 Voltage	—	See curve page 98		
Grid-No.1 (Control-Grid) Voltage Positive-bias value	0	0		volts
Plate Dissipation	1	5		watts
Grid-No.2 Input:				
For plate voltages up to 165 volts	—	1.5*		watts
For plate voltages between 165 and 330 volts	—	See curve page 98		
CHARACTERISTICS				
Plate Voltage	200	60	250	volts
Grid-No.2 Voltage	—	170	170	volts
Grid-No.1 Voltage	—2	0	—	volts

Cathode-Bias Resistor	—	—	82	ohms
Amplification Factor	70	—	—	
Plate Resistance (Approx.)	—	—	0.14	megohm
Transconductance	4200	—	12000	µmhos
Plate Current	4.5	48■	22	mA
Grid-No.2 Current	—	12■	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 µA	—5	—	—10	volts
MAXIMUM CIRCUIT VALUES				
Grid-No.1-Circuit Resistance:				
For fixed-bias operation	0.5	0.25		megohm
For cathode-bias operation	1	1		megohm

* Grid-No.2 input may reach 2 watts for plate-dissipation values of 4 watts or less.
■ This value may be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

6JF6
22JF6

BEAM POWER TUBE



9QL

Novar type used as horizontal-deflection amplifier in black-and-white television receivers. Outlines section, 18A; requires novar 9-contact socket. Type 22JF6 is identical with type 6JF6 except for heater ratings.

	6JF6	22JF6	
Heater Voltage (ac/dc)	6.3	22	volts
Heater Current	1.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	*100 max*	*100 max*	volts
Direct Interelectrode Capacitances (Approx.):			
Grid No.1 to Plate		1.2	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3		22	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3		9	pF

Class A_1 Amplifier

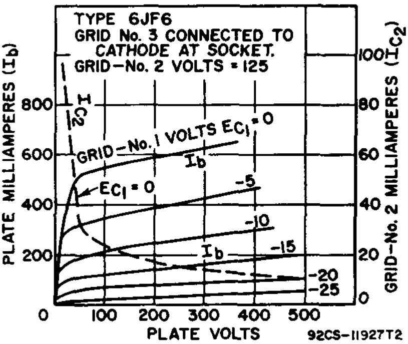
CHARACTERISTICS	Triode■ Connection	Pentode Connection			
Plate Voltage	125	—	50	130	volts
Peak Positive-Pulse Plate Voltage#	—	6500	—	—	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket				
Grid-No.2 (Screen-Grid) Voltage	—	125	125	125	volts
Grid-No.1 (Control-Grid) Voltage	—20	—	0	—20	volts
Triode Amplification Factor	4.1	—	—	—	
Plate Resistance (Approx.)	—	—	—	12000	ohms
Transconductance	—	—	—	10000	µmhos
Plate Current	—	—	525†	80	mA
Grid-No.2 Current	—	—	32†	2.5	mA
Grid-No.1 Voltage for plate current of 1 mA	—	—125	—	—40	volts

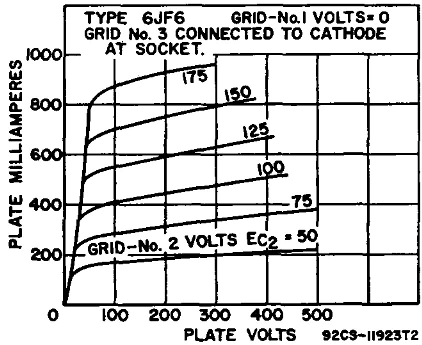
Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

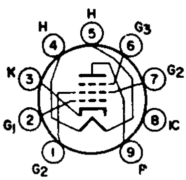
MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage*	100	volts
DC Grid-No.2 Voltage	220	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	950	mA
Average Cathode Current	275	mA
Grid-No.2 Input	3.5	watts
Plate Dissipation†	17	watts
Bulb Temeprature (At hottest point)	240	°C
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance:		
For grid-resistor-bias operation‡	0.47	megohm
For plate-pulsed operation (horizontal-deflection circuits only)	10	megohms

▪ Grid-No.2 connected to plate at socket.
† This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.
# Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
* In this service, a positive value may be applied to grid No.3 to minimize "snivets" interference; a typical value for this voltage is 50 volts.
‡ A bias resistor or other means is required to protect the tube in absence of excitation.





Refer to chart at end of section. **6JG6**



9QU

6JG6A

17JG6A, 22JG6A

BEAM POWER TUBE

Novar type used as horizontal-deflection amplifier in low-B+, black-and-white television receivers. Outlines section, 31B; requires novar 9-contact socket. For curves of average plate characteristics, refer to type 6JF6. Types 17JG6A and 22JG6A are identical with type 6JG6A except for heater ratings.

	6JG6A	17JG6A	22JG6A	
Heater Voltage (ac/dc)	6.3	16.8	22	volts
Heater Current	1.6	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.7	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No. 3	22	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	9	pF

Class A₁ Amplifier

CHARACTERISTICS	Triode▪ Connection	Pentode Connection		
Plate Voltage	125	50	130	volts
Grid No.3 (Suppressor Grid)		Connected to cathode at socket		
Grid-No.2 (Screen-Grid) Voltage	—	125	125	volts
Grid-No.1 (Control-Grid) Voltage	—20	0	—20	volts
Amplification Factor	4.1	—	—	
Plate Resistance (Approx.)	—	—	12000	ohms
Transconductance	—	—	10000	μmhos
Plate Current	—	525•	80	mA
Grid-No.2 Current	—	32•	2.5	mA
Grid-No.1 Voltage (Approx.), for plate current of 1 mA	—	—	—40	volts

▪ With grid No.2 connected to plate at socket.
• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage*	75	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage, Negative-bias value	—55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	950	mA
Average Cathode Current	275	mA
Plate Dissipation†	17	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	240	°C

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, for grid-No.1-resistor-bias operation	2.2	megohms

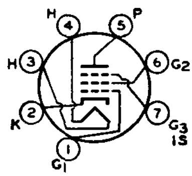
Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

* In a horizontal-deflection-amplifier service, a positive voltage (typical value, 30 volts) may be applied to grid No.3 to reduce "snivets" interference, which may occur in both vhf and uhf television receivers.

† A bias resistor or other means is required to protect the tube in absence of excitation.

6JH6 SEMIREMOTE-CUTOFF PENTODE

4JH6



7CM

Miniature type used in the gain-controlled picture if-amplifier stages of color and black-and-white television receivers. Outlines section, 5C; requires miniature 7-contact socket. For curves of average plate characteristics, refer to type 6BZ6. Type 4JH6 is identical with type 6JH6 except for heater ratings.

	4JH6	6JH6	
Heater Arrangement	Series	Parallel	
Heater Voltage (ac/dc)	4.2	6.3	volts
Heater Current	0.45	0.3	ampere
Heater Warm-up Time	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:	Unshielded	Shielded▪	
Grid No.1 to Plate	0.025 max	0.015 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	7	7	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2	3	pF

▪ With external shield connected to cathode.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

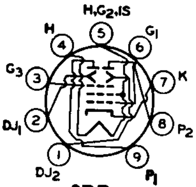
Plate Voltage	300	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	0.55	watt
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	125	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	0.26	megohm
Transconductance	8000	μmhos
Transconductance Range for grid-No.1 voltage of —4.5 volts and cathode-bias resistor of 56 ohms	400-900	μmhos
Plate Current	14	mA
Grid-No.2 Current	3.6	mA
Grid-No.1 Voltage (Approx.) for transconductance of 50 μmhos	—19	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm



9DP

BEAM-DEFLECTION TUBE 6JH8

Miniature type used in color-demodulator and burst-gate circuits in color television receivers. This type has two plates and two deflecting electrodes; the control grid varies beam deflection. **Outlines section, 6E;** requires miniature 9-contact socket. Pin 5 should be connected to cathode at socket. The 6JH8 should be so located in the equipment that it is not subjected to stray magnetic fields.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.3	amperes
Direct Interelectrode Capacitances:		
Grid No.1 to All Other Electrodes, Except Both Plates	7.5	pF
Grid No.1 to Deflecting Electrode No.1	0.04 max	pF
Grid No.1 to Deflecting Electrode No.2	0.07 max	pF
Plate No.1 to All Other Electrodes	5.0	pF
Plate No.2 to All Other Electrodes	5.0	pF
Plate No.1 to Plate No.2	0.4	pF
Deflecting Electrode No.1 to All Other Electrodes	4.8	pF
Deflecting Electrode No.2 to All Other Elecetrodes	4.8	pF
Deflecting Electrode No. 1 to Deflecting Electrode No.2	0.38	pF

Color TV Demodulator

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage (Each Plate)	330	volts
Peak Deflecting-Electrode Voltage (Each Electrode):		
Negative value	165	volts
Positive value	165	volts
Grid-No.3 (Accelerating-Grid) Voltage	330	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Cathode Current	33	mA
Plate Dissipation (Each Plate)	3	watts
Grid-No.3 Input	1	watt

MAXIMUM CIRCUIT VALUES

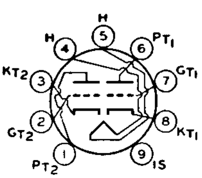
Grid-No.1 Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.25	megohm

Class A_1 Amplifier

With both plates connected together and with both deflecting electrodes connected to cathode at socket

CHARACTERISTICS

Plate-No.1 Supply Voltage	250	volts
Plate-No.2 Supply Voltage	250	volts
Grid-No.3 Voltage	250	volts
Cathode-Bias Resistor	220	ohms
Transconductance	4400	μmhos
Total Plate Current	14	mA
Grid-No.3 Current	1.5	mA
Grid-No.1 Voltage (Approx.) for total plate current of 10 μA	—13	volts



9AJ

DUAL TRIODE 6JK8

Miniature type used as combined rf-amplifier and mixer-oscillator tube in FM tuners. Unit No.1 is used as an oscillator-mixer, and unit No.2 is used as an rf amplifier. **Outlines section, 6B;** requires miniature 9-contact socket and may be mounted in any position.

Heater Voltage (ac/dc)		6.3	volts
Heater Current		0.4	ampere
Peak Heater-Cathode Voltage		±100 max	volts
Direct Interelectrode Capacitances:	Unit No.1	Unit No.2	
Grid to Plate	1.4	0.6	pF
Grid to Cathode, Heater, and Internal Shield	3	5	pF
Plate to Cathode, Heater, and Internal Shield	1	4	pF
Heater to Cathode	2.8	2.8	pF
Grid of Unit No.1 to Grid of Unit No.2		0.003 max	pF
Plate of Unit No.1 to Plate of Unit No.2		0.009 max	pF

Class A_1 Amplifier

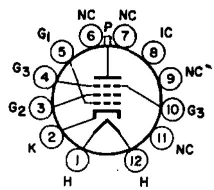
MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1 Oscillator	Unit No.2 RF Amplifier	
Plate Voltage	165	200	volts
Negative Grid Voltage	50	50	volts
Average Cathode Current	22	22	mA
Plate Dissipation	1	2	watts
CHARACTERISTICS			
Plate Voltage	100	135	volts
Grid Voltage	—1	—1.2	volts
Amplification Factor	55	70	
Plate Resistance (Approx.)	8000	5400	ohms
Transconductance	6800	13000	µmhos
Plate Current	5.3	10	mA
Grid Voltage (Approx.):			
For plate current of 20 µA	—4.4	—	volts
For transconductance of 150 µmhos	—	—5.5	volts
For transconductance of 1500 µmhos	—	—2.8	volts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance, for cathode-bias operation	1	1	megohm

6JM6 Refer to chart at end of section.

6JM6A

17JM6A

BEAM POWER TUBE



12FJ

Duodecar types used as horizontal-amplifier tubes in color and black-and-white television receivers. Outlines section, 16A; requires duodecar 12-contact socket. Type 17JM6A is identical with type 6JM6A except for heater ratings.

	6JM6A	17JM6A	
Heater Voltage (ac/dc)	6.3	16.8	volts
Heater Current	1.2	0.45	amperes
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:			
Grid No.1 to Plate		0.6	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3		16	pF
Plate to Cathode, Heater, Grid No. 2, and Grid No. 3		7	pF

Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection			Triode** Connection	
Plate Voltage	5000	55	250	150	volts
Grid-No.3 (Suppressor-Grid)	Connected to cathode at socket				
Grid-No.2 (Screen-Grid) Voltage	150	150	150	150	volts
Grid-No.1 (Control-Grid) Voltage	—	0	—22.5	—22.5	volts
Plate Resistance (Approx.)	—	—	15000	—	ohms
Transconductance	—	—	7300	—	µmhos
Plate Current	—	345*	65	—	mA
Grid-No.2 Current	—	30*	1.8	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 µA	—100	—	—42	—	volts
Amplification Factor	—	—	—	4.4	

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	1	megohm

* This value can be measured by a method utilizing a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

** Grid No.2 tied to plate.

Horizontal-Deflection Amplifier

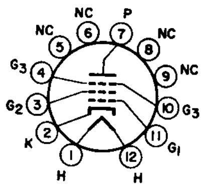
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage	70	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage, Negative-bias value	55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Average Cathode Current	175	mA
Peak Cathode Current	550	mA
Plate Dissipation##	17.5	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	220	°C

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

A bias resistor or other means is required to protect the tube in absence of excitation.



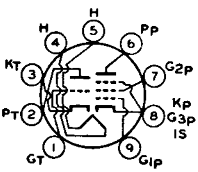
12FK

BEAM POWER TUBE — 6JN6

12JN6, 17JN6

Duodecar type used as horizontal-amplifier tube in color and black-and-white television receivers. Outlines section, 15A; requires duodecar 12-contact socket. This type is electrically identical with type 6JM6 except that it has a slightly lower grid-No.1-to-plate capacitance. Types 12JN6 and 17JN6 are identical with type 6JN6 except for heater ratings.

	6JN6	12JN6	17JN6	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Direct Interelectrode Capacitances:				
Grid No.1 to Plate			0.34	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3			16	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3			7	pF



9FA

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE — 6JN8

19JN8

Miniature type used as FM converter and rf amplifier in radio receivers. Outlines section, 6B; requires miniature 9-contact socket. Type 19JN8 is identical with type 6JN8 except for heater ratings.

	6JN8	19JN8	
Heater Voltage (ac/dc)	6.3	18.9	volts
Heater Current	0.45	0.15	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:*			
Pentode Unit:			
Grid No.1 to Plate		0.01	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield		5.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield		3.4	pF

Triode Unit:		
Grid to Plate	1.7	pF
Grid to Cathode, Heater, Pentode Cathode, Grid No.3, and Internal Shield	3.2	pF
Plate to Cathode, Heater, Pentode Cathode, Grid No.3, and Internal Shield	2.2	pF

[3] With external shield connected to cathode of unit under test.

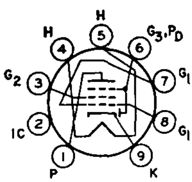
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	300	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	2.5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 150 volts	—	0.55	watt
For grid-No.2 voltages between 150 and 300 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Voltage	125	125	volts
Grid-No.2 Voltage	—	125	volts
Grid-No.1 Voltage	—1	—1	volt
Amplification Factor	46	—	
Plate Resistance (Approx.)	5400	200000	ohms
Transconductance	8500	7500	μmhos
Plate Current	13.5	12	mA
Grid-No.2 Current	—	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—8	—8	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	2.2	2.2	megohms
For cathode-bias operation	2.2	2.2	megohms

6JQ6

12JQ6, 17JQ6, 25JQ6

BEAM POWER TUBE with integral diode



9RA

Miniature type featuring integral diode, internally connected to grid No.3, used in feedback-stabilized vertical-deflection-amplifier applications in color and black-and-white television receivers. **Outlines section, 6G**; requires miniature 9-contact socket. Types 12JQ6, 17JQ6, and 25JQ6 are identical with type 6JQ6 except for heater ratings.

	6JQ6	12JQ6	17JQ6	25JQ6	
Heater Voltage (ac/dc)	6.3	12.6	16.8	25.2	volts
Heater Current	1.2	0.6	0.45	0.3	amperes
Heater Warm-up Time (Average)	—	11	11	11	seconds
Heater-Cathode Voltage:					
Peak value	±200 max	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.32	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Diode Plate	13	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Diode Plate	6	pF

Class A_1 Amplifier

CHARACTERISTICS			
Plate Voltage	40	140	volts
Grid-No.3 (Suppressor-Grid) Voltage	0	0	volts
Grid-No.2 (Screen-Grid) Voltage	120	140	volts
Grid-No.1 (Control-Grid) Voltage	0	—18	volts
Triode Amplification Factor•	—	6.5	
Plate Resistance (Approx.)	—	10500	ohms
Transconductance	—	4200	μmhos
Plate Current	150#	35	mA
Grid-No.2 Current	20#	2.5	mA
Grid-No.1 Voltage for plate current of 1 mA	—	—37	volts
Instantaneous Diode-Plate-to-Cathode Voltage Drop for Instantaneous Diode-Plate Current of 2 mA	—	5	volts

Vertical-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Voltage	425	volts
Peak Positive-Pulse Plate Voltage (Absolute-Maximum Value)*	2000	volts
DC Grid-No.3 and Diode-Plate Voltage	+10 / −150	volts
DC Grid-No.2 Voltage	330	volts
Peak Negative-Pulse Grid-No.1 Voltage	150	volts
Average Cathode Current	70	mA
Peak Cathode Current	250	mA
Average Diode-Plate (and Grid-No.3) Current	1	mA
Plate Dissipation	10	watts
Grid-No.2 Input	2	watts
Bulb Temperature (At hottest point)	240	°C

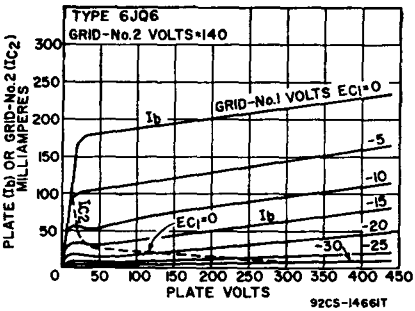
MAXIMUM CIRCUIT VALUES

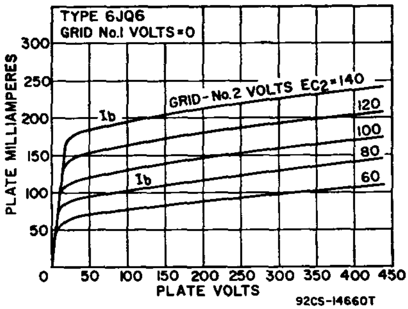
Grid-No.1—Circuit Resistance:		
For grid-No.1-resistor-bias operation	2.2	megohms
For cathode-bias operation	2.2	megohms

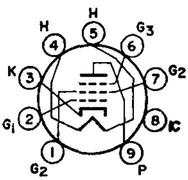
• Grid No.3 and diode plate connected to cathode, and grid-No.2 connected to plate at socket.

This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

* Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).







9QU

6JR6

17JR6, 22JR6

BEAM POWER TUBE

Novar type used for horizontal-deflection amplifier service in low B+, black-and-white television receivers. Outlines section, 31B; requires novar 9-contact socket. Types 17JR6 and 22JR6 are identical with type 6JR6 except for heater ratings.

	6JR6	17JR6	22JR6	
Heater Voltage (ac/dc)	6.3	16.8	22	volts
Heater Current	1.6	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):

Grid No.1 to Plate	0.7	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	22	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	9	pF

Class A₁ Amplifier

CHARACTERISTICS	Triode* Connection	Pentode Connection			
Plate Voltage	125	—	50	130	volts
Peak Positive-Pulse Plate Voltage#	—	6500	—	—	volts
Grid No.3 (Suppressor Grid)		Connected to cathode at socket			
Grid-No.2 (Screen-Grid) Voltage	125	125	125	125	volts
Grid-No.1 (Control-Grid) Voltage	—20	—	0	—20	volts
Plate Resistance (Approx.):	—	—	—	18000	ohms
Transconductance	—	—	—	7000	μmhos
Plate Current	—	—	470‡	45	mA
Grid-No.2 Current	—	—	32‡	1.5	mA
Grid-No.1 Voltage for plate current of 1 mA	—	—75	—	—32	volts
Amplification Factor	4.7	—	—	—	

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Ratings)

Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
Grid-No.3 Voltage▪	75	volts
Grid-No.2 Voltage	220	volts
Grid-No.1 Voltage, Negative-bias value	55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	950	mA
Average Cathode Current	275	mA
Grid-No.2 Input	3.5	watts
Plate Dissipation•	17	watts
Bulb Temperature (At hottest point)	240	°C

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance		
For grid-resistor-bias operation	0.47	megohm
For plate-pulsed operation (horizontal-deflection circuits only)	10	megohms

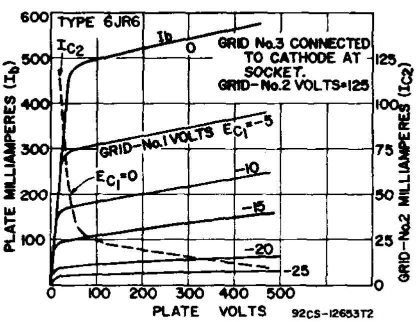
* Grid No. 2 connected to plate at socket.

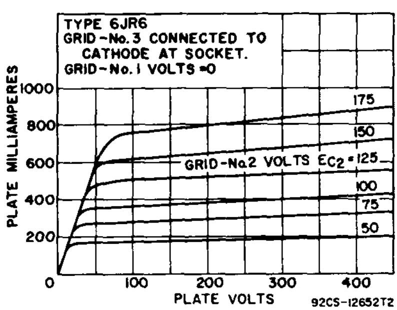
Pulse duration must not exceed 15% of one horizontal scanning cycle (10 microseconds).

‡ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

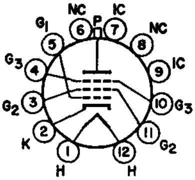
▪ In this service, a positive value may be applied to grid No.3 to minimize "snivets" interference; a typical value for this voltage is 30 volts.

• A bias resistor or other means is required to protect the tube in absence of excitation.





Refer to chart at end of section. **6JS6**



12FY

BEAM POWER TUBE

6JS6A

31JS6A

Duodecar types used as horizontal-deflection amplifiers in color and black-and-white television receivers. Outlines section, 16B; requires duodecar 12-contact socket. Type 31JS6A is identical with type 6JS6A except for heater ratings.

	6JS6A	31JS6A	
Heater Voltage (ac/dc)	6.3	31.5	volts
Heater Current	2.25	0.45	amperes
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

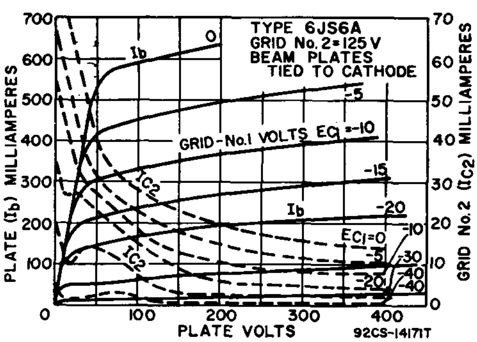
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.7	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	24	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	10	pF

Class A_1 Amplifier

CHARACTERISTICS	Triode†† Connection	Pentode Connection			
Plate Voltage	125	5000	70	175	volts
Grid No.3 (Suppressor Grid)	—	Connected to cathode at socket			
Grid-No.2 (Screen-Grid) Voltage	125	5000	70	175	volts
Grid-No.1 (Control-Grid) Voltage	−25	—	0	−25	volts
Plate Resistance (Approx.)	—	—	—	5600	ohms
Transconductance	—	—	—	11300	μmhos
Plate Current	—	—	570†	125	mA
Grid-No.2 Current	—	—	34†	4.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	−140	—	−54	volts
Triode Amplification Factor	3	—	—	—	

† This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

†† Grid No.2 connected to plate.



Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	990	volts
Peak Positive-Pulse Plate Voltage#	7500	volts
Peak Negative-Pulse Plate Voltage	1100	volts

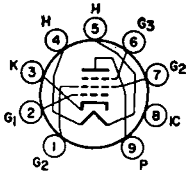
DC Grid-No.3 Voltage	70	volts
DC Grid-No.2 Voltage	190	volts
Peak Negative-Pulse Grid-No.1 Voltage	250	volts
Average Cathode Current	315	mA
Peak Cathode Current	1100	mA
Plate Dissipation**	28	watts
Grid-No.2 Input	5.5	watts
Bulb Temperature (At hottest point)	225	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	1	megohm

Pulse duration must not exceed 15% of one horizontal scanning cycle (10 microseconds).
** A bias resistor or other means is required to protect the tube in absence of excitation.

6JT6 Refer to chart at end of section.

6JT6A BEAM POWER TUBE

12JT6A, 17JT6A



9QU

Novar types used as horizontal-deflection amplifiers in high-efficiency deflection circuits of black-and-white television receivers employing wide-angle or high-voltage picture tubes. **Outlines section, 31A;** requires novar 9-contact socket. Types 12JT6A and 17JT6A are identical with type 6JT6A except for heater ratings.

	6JT6A	12JT6A	17JT6A	
Heater Voltage (ac/dc)	6.3	12.6	16.8	volts
Heater Current	1.2	0.6	0.45	amperes
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts
Direct Interelectrode Capacitances:				
Grid No.1 to Plate			0.26	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3			15	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3			6.5	pF

Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection		Triode* Connection	
Plate Voltage	60	250	150	volts
Grid No.3 (Suppressor Grid)		Connected to cathode at socket		
Grid-No.2 (Screen-Grid) Voltage	150	150	150	volts
Grid-No.1 (Control-Grid) Voltage	0	—22.5	—22.5	volts
Triode Amplification Factor	—	—	4.4	
Plate Resistance (Approx.)	—	15000	—	ohms
Transconductance	—	7100	—	µmhos
Plate Current	390▪	70	—	mA
Grid-No.2 Current	32▪	2.1	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—42	—	volts

* Grid No.2 connected to plate.

▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

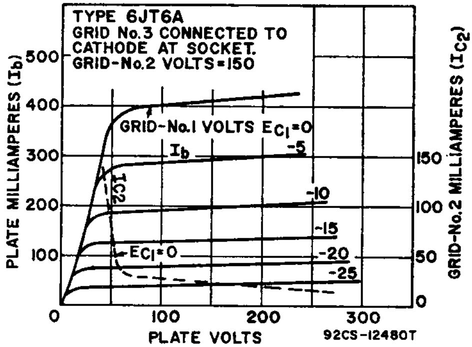
MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage▲	70	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage, Negative-bias value	55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	550	mA

Average Cathode Current	175	mA
Plate Dissipation†	17.5	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	240	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance, for grid-resistor-bias operation	1	megohm

\# Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

▲ A positive voltage may be applied to grid No.3 to reduce interference from "snivets" which may occur in television receivers. A typical value for this voltage is 30 volts.

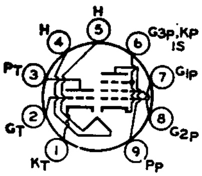
† A bias resistor or other means is required to protect the tube in absence of excitation.



HIGH-MU TRIODE— SHARP-CUTOFF PENTODE

6JT8

10JT8



9DX

Neonoval type with frame-grid pentode unit used in color and black-and-white television receivers. The triode unit is used as a voltage-amplifier or sync-separator tube, and the pentode unit is used as a video-amplified tube. **Outlines section,** 10A, except base is small-button miniature 9-pin; requires miniature 9-contact socket. Type 10JT8 is identical with type 6JT8 except for heater ratings.

	6JT8	10JT8	
Heater Voltage (ac/dc)	6.3	10.2	volts
Heater Current	0.725	0.45	ampere
Heater Warm-up Time (Average	—	11	seconds
Heater-Cathode Voltage:	±200 max	±200 max	volts
Peak value	100 max	100 max	volts
Average value			

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	1	4	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	

CHARACTERISTICS				
Plate Supply Voltage	250	50	200	volts
Grid-No.2 Supply Voltage	—	100	100	volts
Grid-No.1 Voltage	—2	0	—	volts
Cathode-Bias Resistor	—	—	82	ohms
Amplification Factor	100	—	—	
Plate Resistance (Approx.)	37000	—	50000	ohms

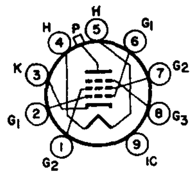
	Triode Unit	Pentode Unit		
Transconductance	2700	—	20000	µmhos
Plate Current	1.5	55•	17	mA
Grid-No.2 Current	—	18•	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—	—	—5	volts
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—5.3	—	—	volts
MAXIMUM CIRCUIT VALUES				
Grid-No.1-Circuit Resistance:				
For fixed-bias operation	0.5	0.25		megohm
For cathode-bias operation	1	1		megohm

• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

6JU6

22JU6

BEAM POWER TUBE



9QL

Novar type used as horizontal-deflection amplifier in color television receivers. Outlines section, 18A; requires novar 9-contact socket. Type 22JU6 is identical with type 6JU6 except for heater ratings.

	6JU6	22JU6	
Heater Voltage (ac/dc)	6.3	20	volts
Heater Current	1.6	0.45	amperes
Heater Warm-up Time	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:			
Grid No.1 to Plate		1.2	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3		22	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3		9	pF

Class A₁ Amplifier

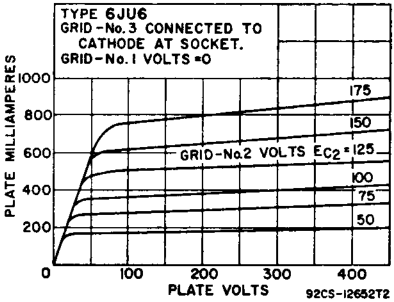
CHARACTERISTICS	Triode† Connection	Pentode Connection			
Plate Voltage	125	—	50	130	volts
Peak Positive-Pulse Plate Voltage#	—	6500	—	—	volts
Grid No.3 (Suppressor Grid)		Connected to cathode at socket			
Grid-No.2 (Screen-Grid) Voltage	125	125	125	125	volts
Grid-No.1 (Control-Grid) Voltage	—20	—	0	—20	volts
Amplification Factor	4.7	—	—	—	
Plate Resistance (Approx.)	—	—	—	18000	ohms
Transconductance	—	—	—	7000	µmhos
Plate Current	—	—	470††	45	mA
Grid-No.2 Current	—	—	32††	1.5	mA
Grid-No.1 Voltage for plate current of 1 mA	—	—75	—	—32	volts

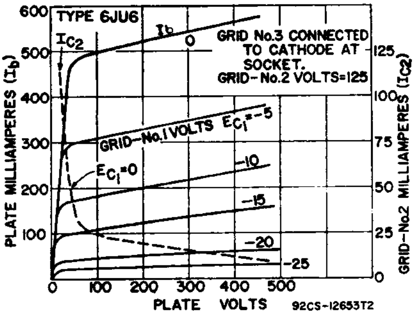
Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage■	75	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage, Negative-bias value	55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	950	mA
Average Cathode Current	275	mA
Grid-No.2 Input	3.5	watts
Plate Dissipation■■	17	watts
Bulb Temperature (At hottest point)	240	°C
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance:		
For grid-resistor-bias operation	0.47	megohm
For plate-pulsed operation	10	megohms

Pulse duration must not exceed 15% of one horizontal scanning cycle (10 microseconds).
† Grid No.2 connected to plate.
†† This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.
▪ In this service, a positive value may be applied to grid No.3 to minimize "snivets" interference; a typical value for this voltage is 30 volts.
▪▪ A bias resistor or other means is required to protect the tube in absence of excitation.



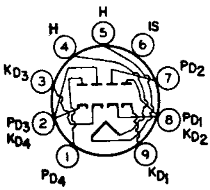


Refer to chart at end of section. **6JU8**

QUADRUPLE DIODE

6JU8A

8JU8A



9PQ

Miniature type used in phase-detector and noise-immune color-killer circuits of color television receivers, and in bridge-matrixing circuits in FM stereo multiplex equipment. **Outlines section, 6B**; requires miniature 9-contact socket. Units 1 and 2 are shielded from units 3 and 4 to minimize coupling between the series-connected pairs of diodes. Type 8JU8A is identical with type 6JU6A except for heater ratings.

	6JU8A	8JU8A	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time	—	11	seconds
Peak Heater-Cathode Voltage	±300 max	±300 max	volts
Direct Interelectrode Capacitances (Approx.):			
Plate of Unit No.1 and Cathode of Unit No.2 to Cathode of Unit No.1		1.8	pF
Plate of Unit No.1 and Cathode of Unit No.2 to Plate of Unit No.2		2.2	pF
Plate of Unit No.2 to Heater and Internal Shield		0.62	pF
Plate of Unit No.3 and Cathode of Unit No.4 to Cathode of Unit No.3		1.9	pF
Plate of Unit No.3 and Cathode of Unit No.4 to Plate of Unit No.4		2.2	pF
Plate of Unit No.4 to Heater and Internal Shield		0.94	pF
Cathode of Unit No.1 to Heater and Internal Shield		1.8	pF
Cathode of Unit No.3 to Heater and Internal Shield		1.9	pF

MAXIMUM RATINGS (Design-Center Values, Each Diode Unit)

Peak Inverse Plate Voltage	300	volts
Peak Plate Current	54	mA
Average Output Current	9	mA

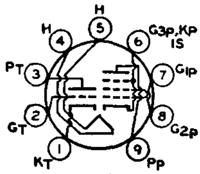
CHARACTERISTICS, Instantaneous Value (Each Unit)

Plate Current for plate voltage of 10 volts	60	mA

6JV8
8JV8

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE

Miniature type used in television receiver applications, particularly those having low-voltage "B" supplies. The triode unit is used in sound-if, keyed-agc, sync-separator, sync-amplifier, and noise-suppression circuits. The pentode unit is especially useful as a video amplifier tube. **Outlines section, 6E**; requires miniature 9-contact socket. Type 8JV8 is identical with type 6JV8 except for heater ratings.



9DX

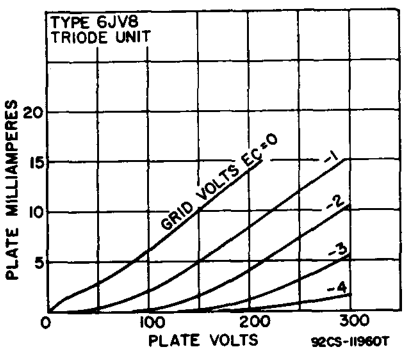
	6JV8	8JV8	
Heater Voltage (ac/dc)	6.3	8.5	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Triode Unit:		
Grid to Plate	2.2	pF
Grid to Cathode and Heater	3	pF
Plate to Cathode and Heater	2	pF
Pentode Unit:		
Grid No.1 to Plate	0.08 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	8	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3.2	pF
Pentode Grid No.1 to Triode Plate	0.012 max	pF
Pentode Plate to Triode Plate	0.24 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Voltage	—	330	volts
Grid-No.1 (Control-Grid) Voltage:			
Positive-bias value	0	0	volts
Negative-bias value	50	50	volts
Plate Dissipation	1.1	4	watts
Grid-No.2 Input	—	1.7	watts

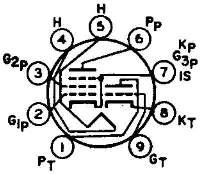
CHARACTERISTICS	Triode Unit	Pentode Unit			
Plate Voltage	200	60	125	200	volts
Grid-No.2 Voltage	—	200	125	200	volts
Grid-No.1 Voltage	−2	0	−1	−2.9	volts
Amplification Factor	70	—	—	—	
Plate Resistance (Approx.)	0.0175	—	0.1	0.15	megohm
Transconductance	4000	—	11500	10700	µmhos
Plate Current	4	51•	22	22	mA
Grid-No.2 Current	—	14•	4	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	−5	—	−5.5	−9	volts



MAXIMUM CIRCUIT VALUES

Grid-No.1-Current Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.



9AE

6JW8/ ECF802

6LX8/LCF802

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

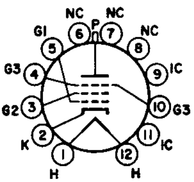
Miniature type used as horizontal-oscillator and frequency-control tube in color and black-and-white television receivers. **Outlines section, 6B**; requires miniature 9-contact socket. Type 6LX8/LCF802 is identical with type 6JW8/ECF802 except for heater ratings.

	6JW8/ ECF802	6LX8/ LCF802	
Heater Voltage (ac/dc)	6.3	6	volts
Heater Current	0.43	0.45	ampere
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	550	volts
Plate Voltage	250	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	550	volts
Grid-No.2 Voltage	—	250	volts
Peak Cathode Current•	—	50	mA
Cathode Current	10	15	mA
Plate Dissipation	1.4	1.2	watts
Grid-No.2 Input	—	0.8	watts
Input Impedance at 60 Hz	50	300	kohms
CHARACTERISTICS			
Plate Voltage	200	100	volts
Grid-No.2 Voltage	—	100	volts
Grid-No.1 (Control-Grid) Voltage	—2	—1	volts
Mu Factor, Grid-No.1 to Grid-No.2	—	47	
Amplification Factor	70	—	
Input Resistance	0.2	0.4	megohm
Transconductance	3500	5500	μmhos
Plate Current	3.5	6	mA
Grid-No.2 Current	—	1.7	mA
Plate Current:			
For grid-No.1 voltage of 0 volts	—	12.5	mA
For grid current of 10 μA	10	—	mA
Grid-No.2 Current for grid-No.1 voltage of 0 volts	—	8.5	mA
Grid-No.1 Voltage:			
For grid-No.1 current of +0.3 μA	—1.3	—1.3	volts
For plate and grid-No.2 voltage of 200 volts and plate current of 10 μA	—	—16	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	—	0.56	megohm
For cathode-bias operation	3	1	megohms

• With a maximum duty factor of 0.30 and maximum pulse duration of 30 microseconds.



12GD

6JZ6

BEAM POWER TUBE

Duodecar type used as horizontal-deflection amplifier in black-and-white television receivers. **Outlines section, 16A**; requires duodecar 12-contact socket. Heater: volts (ac/dc), 6.3; amperes, 1.5; maximum heater-cathode volts, ±200 peak, 100 average.

Class A₁ Amplifier

CHARACTERISTICS	Triode* Connection	Pentode Connection			
Plate Voltage	130	5000	50	130	volts
Grid No.3 (Suppressor Grid)		Connected to cathode at socket			
Grid-No.2 (Screen-Grid) Voltage	130	130	130	130	volts
Grid-No.1 (Control-Grid) Voltage	—20	—	0▲	—20	volts
Plate Resistance (Approx.)	—	—	—	9900	ohms
Transconductance	—	—	—	9000	μmhos
Plate Current	—	—	450°	46	mA
Grid-No.2 Current	—	—	29°	1.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—64	—	—	—32	volts
Triode Amplification Factor	4.8	—	—	—	

* Grid No.2 connected to plate.
▲ Applied for short interval (2 seconds max.) so as not to damage tube.
° This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

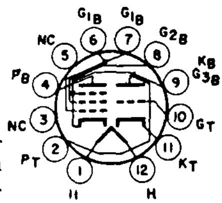
MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate-Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage, Positive-bias value	70	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage, Negative-bias value	55	volts
Peak Negative Grid-No.1 Voltage	330	volts
Peak Cathode Current	800	mA
Average Cathode Current	230	mA
Plate Dissipation**	18	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	220	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	1	megohm

\# Pulse duration must not exceed 15% of one horizontal scanning cycle (10 microseconds).
** A bias resistor or other means is required to protect the tube in absence of excitation.

6JZ8

17JZ8

MEDIUM-MU TRIODE— BEAM POWER TUBE



12DZ

Duodecar type used in combined vertical-deflection-oscillator and vertical-deflection-amplifier applications in television receivers. Outlines section, 8C; requires duodecar 12-contact socket.

	6JZ8	17JZ8	
Heater Voltage (ac/dc)	6.3	16.8	volts
Heater Current	1.2	0.45	amperes
Heater Warm-up Time	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A₁ Amplifier

CHARACTERISTICS	Triode Unit	Beam Power Unit		
Plate Voltage	150	45	120	volts
Grid-No.2 (Screen-Grid) Voltage	—	110	110	volts
Grid-No.1 (Control-Grid) Voltage	—5	0	—8	volts
Amplification Factor	20	—	—	
Plate Resistance (Approx.)	8500	—	11700	ohms
Transconductance	2350	—	7100	μmhos
Plate Current	5.5	122■	46	mA
Grid-No.2 Current	—	16.5■	3.5	mA

	Triode Unit	Beam Power Unit		
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—10	—	—	volts
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—	—25	volts

• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Vertical-Deflection Oscillator and Amplifier

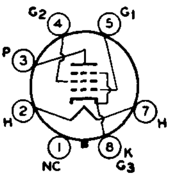
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit Oscillator	Beam Power Unit Amplifier	
DC Plate Voltage	250	250	volts
Peak Positive-Pulse Plate Voltage#	—	2000	volts
DC Grid-No.2 Voltage	—	200	volts
Peak Negative-Pulse Grid-No.1 Voltage	400	150	volts
Peak Cathode Current	70	245	mA
Average Cathode Current	20	70	mA
Plate Dissipation•	1	7	watts
Grid-No.2 Input	—	1.8	watts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	1	1	megohm
For cathode-bias operation	2.2	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

• A bias resistor or other means is required to protect the tube in absence of excitation.

Refer to chart at end of section. **6K5GT**



7S

POWER PENTODE 6K6GT

Glass octal type used in output stage of radio receivers and, triode-connected, as a vertical-deflection amplifier in television receivers. This type may be supplied with pin No.1 omitted. **Outlines section, 13D;** requires octal socket. This tube, like other power-handling tubes, should be adequately ventilated.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.4	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.5	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	5.5	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	6	pF

Class A_1 Amplifier

MAXIMUM RATING (Design-Center Values)		
Plate Voltage	315	volts
Grid-No.2 (Screen-Grid) Voltage	285	volts
Plate Dissipation	8.5	watts
Grid-No.2 Input	2.8	watts

TYPICAL OPERATION				
Plate Voltage	100	250	315	volts
Grid-No.2 Voltage	100	250	250	volts
Grid-No.1 (Control-Grid) Voltage	—7	—18	—21	volts
Peak AF Grid-No.1 Voltage	7	18	21	volts
Zero-Signal Plate Current	9	32	25.5	mA
Maximum-Signal Plate Current	9.5	33	28	mA
Zero-Signal Grid-No.2 Current	1.6	5.5	4.0	mA
Maximum-Signal Grid-No.2 Current	3	10	9	mA
Plate Resistance (Approx.)	104000	90000	110000	ohms
Transconductance	1500	2300	2100	μmhos
Load Resistance	12000	7600	9000	ohms
Total Harmonic Distortion	11	11	15	per cent
Maximum-Signal Power Output	0.35	3.4	4.5	watts

TYPICAL PUSH-PULL OPERATION (Values are for two tubes)

	Fixed Bias	Cathode Bias	
Plate Supply Voltage	285	285	volts
Grid-No.2 Supply Voltage	285	285	volts
Grid-No.1 Voltage	—25.5	—	volts
Cathode-Bias Resistor	—	400	ohms
Peak AF Grid-No.1-to-Grid-No.1 Voltage	51	51	volts
Zero-Signal Plate Current	55	55	mA
Maximum-Signal Plate Current	72	61	mA
Zero-Signal Grid-No.2 Current	9	9	mA
Maximum-Signal Grid-No.2 Current	17	13	mA
Effective Load Resistance (Plate-to-plate)	12000	12000	ohms
Total Harmonic Distortion	6	4	per cent
Maximum-Signal Power Output	10.5	9.8	watts

CHARACTERISTICS (Triode Connection)*

Plate Voltage	250	volts
Grid-No.1 Voltage	—18	volts
Plate Current	37.5	mA
Transconductance	2700	μmhos
Amplification Factor	6.8	
Plate Resistance (Approx.)	2500	ohms
Grid-No.1 Voltage (Approx.) for plate current of 0.5 mA	—48	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:

For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

* Grid-No.2 connected to plate.

Vertical Deflection Amplifier (Triode Connection)*

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS

DC Plate Voltage	315	volts
Peak Positive-Pulse Plate Voltage# (Absolute maximum)	1200°	volts
Peak Negative-Pulse Grid-No.1 Voltage	250	volts
Peak Cathode Current	75	mA
Average Cathode Current	25	mA
Plate Dissipation	7	watts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, for cathode-bias operation	2.2	megohms

* Grid No.2 connected to plate.

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

° Under no circumstances should this absolute value be exceeded.

6K7
6K7G
6K7GT

Refer to chart at end of section.

6K8
6K8G
6K8GT

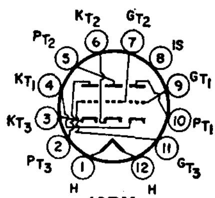
Refer to chart at end of section.

6K11

Refer to chart at end of section.

6K11/ 6Q11

THREE-UNIT TRIODE

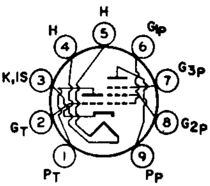


12BY

Duodecar type used as combined agc, sync, and noise-inverter tube in television receivers. **Outlines section,** 8A; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 6.3 amperes, 0.6; warm-up time (average), 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1	Units Nos.2 and 3	
Plate Voltage	330	330	volts
Grid Voltage:			
Negative-bias value	50	50	volts
Positive-bias value	0	0	volts
Cathode Current	20	—	mA
Plate Dissipation	2.75	0.3	watts
CHARACTERISTICS			
Plate Voltage	250	250	volts
Grid Voltage	—8.5	—2	volts
Amplification Factor	17	100	
Plate Resistance (Approx.)	7700	62500	ohms
Transconductance	2200	1600	µmhos
Plate Current	10.5	1.2	mA
Grid Voltage (Approx.) for plate current of 10 µA	—24	—	volts



9PV

HIGH-MU TRIODE—SHARP-CUTOFF PENTODE

6KA8

8KA8

Miniature type used in color and black-and-white television receivers. The triode unit is used in sync-separator circuits; the pentode unit has two independent control grids and is used in gated-agc-amplifier and noise-inverter circuits. Outlines section, 6E; requires miniature 9-contact socket. For curves of average plate characteristics for triode unit, refer to type 6AW8A. Type 8KA8 is identical with type 6KA8 except for heater ratings.

	6KA8	8KA8	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	2.2	pF
Grid to Cathode, Heater, and Internal Shield	2.8	pF
Plate to Cathode, Heater, and Internal Shield	2.2	pF
Pentode Unit:		
Grid-No.1 to Plate	0.1 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	9.5	pF
Grid No.1 to Grid No.3	0.5	pF
Grid No.3 to Plate	2.2	pF
Grid No.3 to All Other Electrodes, Heater, and Internal Shield	7	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	
Plate Voltage	300	volts
Grid Voltage:		
Positive-bias value	0	volts
Negative-bias value	50	volts
Plate Dissipation	1.1	watts

CHARACTERISTICS	Triode Unit	Pentode Unit	
Plate Supply Voltage	200	150	volts
Grid-No.3 Supply Voltage	—	0	volts
Grid-No.2 Supply Voltage	—	100	volts
Grid-No.1 Supply Voltage	—2	0	volts
Cathode-Bias Resistor	—	180	ohms
Amplification Factor	70	—	
Plate Resistance (Approx.)	17500	100000	ohms
Transconductance, Grid No.1 to Plate	4000	4400	µmhos
Transconductance, Grid No.3 to Plate	—	600	µmhos
Plate Current	4	4	mA

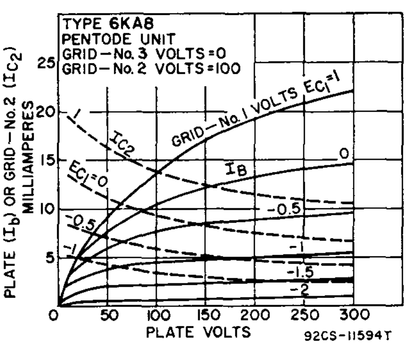
	Triode Unit	Pentode Unit	
Grid-No.2 Current	—	2.8	mA
Grid-No.1 Supply Voltage (Approx.):			
For plate current of 10 µA	—5	—	volts
For plate current of 20 µA	—	—4	volts
Grid No.3 Supply Voltage (Approx.) for plate current of 20 µA	—	—7	volts

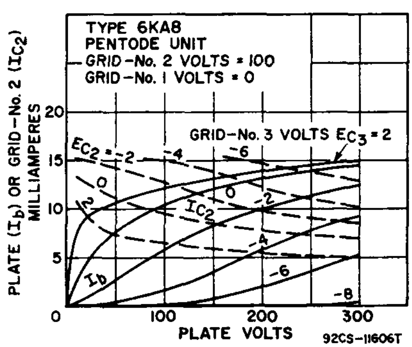
MAXIMUM CIRCUIT VALUES	Triode Unit	
Grid-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

Gated AGC Amplifier and Noise Inverter

MAXIMUM RATINGS (Design-Maximum Values)	Pentode Unit	
DC Plate Voltage	300	volts
Peak Positive-Pulse Plate Voltage#	600	volts
Grid-No.3 (Control-Grid) Voltage:		
Positive-bias value	0	volts
Negative-bias value	—100	volts
Grid-No.2 (Screen-Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:		
Positive-bias value	0	volts
Negative-bias value	—50	volts
Plate Dissipation	2	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	1.1	watts
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	
MAXIMUM CIRCUIT VALUES		
Grid-No.3-Circuit Resistance	0.68	megohm
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).



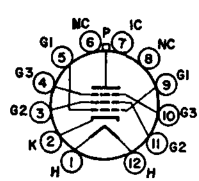


6KD6

40KD6

BEAM POWER TUBE

Duodecar type used as horizontal-deflection amplifier in television receivers. Outlines section, 16C; requires duodecar 12-contact socket. Type 40KD6 is identical with type 6KD6 except for heater ratings.



12GW

	6KD6	40KD6	
Heater Voltage	6.3	40	volts
Heater Current	2.85	0.45	amperes
Heater Warm-up Time	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	990	volts
Peak Positive-Pulse Plate Voltage#	7000	volts
Positive DC Grid-No.3 Voltage	70	volts
Grid-No.2 Voltage	200	volts
Peak Negative-Pulse Grid-No.1 Voltage	250	volts
Peak Cathode Current	1400	mA
Average Cathode Current	400	mA
Plate Dissipation▪	33	watts
Grid-No.2 Input	5	watts
Bulb Temperature (At hottest point)	225	°C

Class A_1 Amplifier

CHARACTERISTICS	Triode† Connection	Pentode Connection		
Plate Voltage	150	60	150	volts
Grid No.3 (Suppressor Grid)		Connected to cathode at socket		
Grid-No.2 (Screen-Grid) Voltage	150	110	110	volts
Grid-No.1 (Control-Grid) Voltage	—22.5	0	—22.5	volts
Amplification Factor	4	—	—	
Plate Resistance (Approx.)	—	—	6000	ohms
Transconductance	—	—	14000	μmhos
Plate Current	—	750▪▪	120	mA
Grid-No.2 Current	—	42▪▪	1.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 1.0 μA	—	—	—40	volts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	2.2	megohms

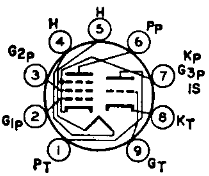
▪ A bias resistor or other means is required to protect the tube in absence of excitation.

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

†Grid-No.2 connected to plate at socket.

▪▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Refer to type 6U8A/6KD8. **6KD8**



9DC

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

6KE8

4KE8, 5KE8

Miniature type with frame-grid pentode unit used as combined oscillator-mixer tube in television receivers using an intermediate frequency in the order of 40 MHz. Outlines section, 6B; requires miniature 9-contact socket. Types 4KE8 and 5KE8 are identical with type 6KE8 except for heater ratings.

	4KE8	5KE8	6KE8	
Heater Voltage (ac/dc)	4.5	5.6	6.3	volts
Heater Current	0.6	0.45	0.4	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:†

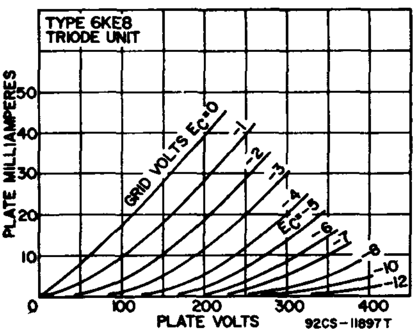
Triode Unit:		
Grid to Plate	1.3	pF
Grid to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	2.4	pF
Plate to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	2	pF
Pentode Unit:		
Grid No.1 to Plate	0.015 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3.4	pF
Heater to Triode Cathode and Pentode Cathode	5.5▪	pF

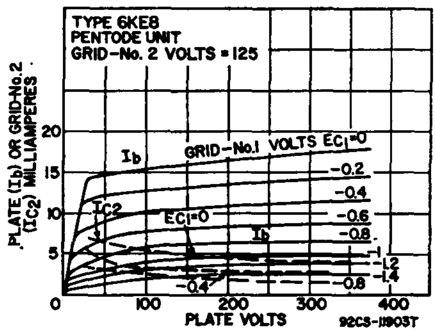
† With external shield connected to cathode of unit under test, except as noted.

▪ With external shield connected to ground.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	280	280	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	280	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Cathode Current	20	20	mA
Plate Dissipation	2	2	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 140 volts	—	0.5	watt
For grid-No.2 voltages between 140 and 280 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	125	125	volts
Grid-No.2 Supply Voltage	—	125	volts
Grid-No.1 Supply Voltage	0	0	volts
Cathode-Bias Resistor	68	33	ohms
Amplification Factor	40	—	
Plate Resistance (Approx.)	5000	125000	ohms
Transconductance	8000	12000	µmhos
Plate Current	13	10	mA
Grid-No.2 Current	—	2.8	mA
Grid-No.1 Voltage (Approx.):			
For plate current of 100 µA	5	—	volts
For plate current of 50 µA	—	—3	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	0.5	megohm



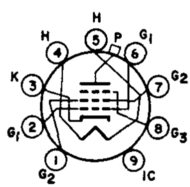


6KL8 Refer to chart at end of section.

6KM6

22KM6

BEAM POWER TUBE



9QL

Novar type used as horizontal-deflection amplifier in color and black-and-white television receivers. Outlines section, 18A; requires novar 9-contact socket. Type 22KM6 is identical with type 6KM6 except for heater ratings.

	6KM6	22KM6	
Heater Voltage (ac/dc)	6.3	22	volts
Heater Current	1.6	0.45	amperes
Heater Warm-up Time	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:			
Grid No.1 to Plate		1.2	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3		22	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3		9	pF

Class A_1 Amplifier

CHARACTERISTICS	Triode Connection	Pentode Connection			
Plate Voltage	140	—	60	140	volts
Peak Positive-Pulse Plate Voltage**	—	6500	—	—	volts
Grid-No.3 (Suppressed-Grid) Voltage	0	30	30	30	volts
Grid-No.2 (Screen-Grid) Voltage	140	140	140	140	volts
Grid-No.1 (Control-Grid) Voltage	—24.5	—	0	—24.5	volts
Amplification Factor†	4	—	—	—	
Plate Resistance (Approx.)	—	—	—	6000	ohms
Transconductance	—	—	—	9500	μmhos
Plate Current	—	—	560††	80	mA
Grid-No.2 Current	—	—	31††	2.4	mA
Grid-No.1 Voltage for plate current of 1 mA	—	—110	—	—42	volts

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage▪	75	volts
DC Grid-No.2 Voltage	220	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	950	mA
Average Cathode Current	275	mA
Grid-No.2 Input	3.5	watts
Plate Dissipation▪▪	20	watts
Bulb Temperature (At hottest point)	240	°C
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance:		
For grid-resistor-bias operation	0.47	megohm
For plate-pulsed operation	10	megohms

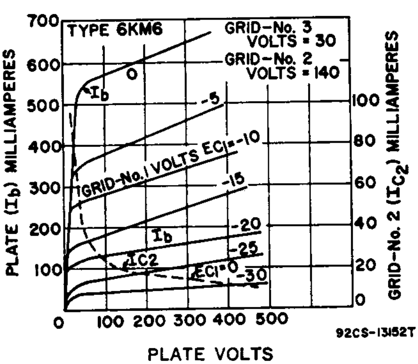
\# Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

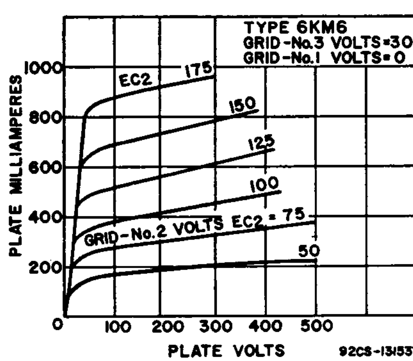
† With grid No.3 and grid No.2 connected, respectively, to cathode and plate at socket.

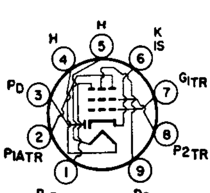
†† This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

▪ In this service, a positive value may be applied to grid No.3 to minimize "snivets" interference; a typical value for this voltage is 30 volts.

▪▪ A bias resistor or other means is required to protect the tube in absence of excitation.







9QG

DIODE—SHARP-CUTOFF THREE-PLATE TETRODE 6KM8

Miniature type used in frequency-divider and complex-wave generator circuits of electronic musical instruments. In such circuits the tetrode unit can provide three independent output-signal voltages; the diode unit can be used as a key in a vibrato circuit. Outlines section, 6E; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.3	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Tetrode Unit:		
Grid No.1 to Plate No.1A	0.02 max	pF
Grid No.1 to Plate No.1B	0.02 max	pF
Grid No.1 to Plate No.2	0.06 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Internal Shield	5.5	pF
Plate No.1A to Cathode, Heater, Grid No.2, and Internal Shield	1.2	pF
Plate No.1B to Cathode, Heater, Grid No.2, and Internal Shield	1.3	pF
Plate No.2 to Cathode, Heater, Grid No.2, and Internal Shield	1.8	pF
Tetrode Grid No.1 to Diode Plate	0.024 max	pF
Tetrode Plate No.1A to Diode Plate	0.18	pF
Tetrode Plate No.1B to Diode Plate	0.024	pF
Tetrode Plate No.2 to Diode Plate	0.013	pF

Tetrode Unit as Class A_1 Amplifier

Plates No. 1A, 1B, and 2 connected together

CHARACTERISTICS

Plate Voltage	100	volts
Grid-No.2 Voltage	100	volts
Grid-No.1 Supply Voltage	0	volts
Grid-No.1 Resistor (Bypassed)	2.2	megohms
Plate Resistance (Approx.)	30000	ohms
Transconductance	3400	μmhos
Plate Current	4.2	mA
Grid-No.2 Current	1.7	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—4	volts

Triode Connection—Plates No.1A, 1B, and 2 connected to grid No.2

Plate Voltage	100	volts
Grid-No.1 Supply Voltage	0	volts
Grid-No.1 Resistor (Bypassed)	2.2	megohms
Transconductance	4500	μmhos
Amplification Factor	45	
Plate Current	5.5	mA

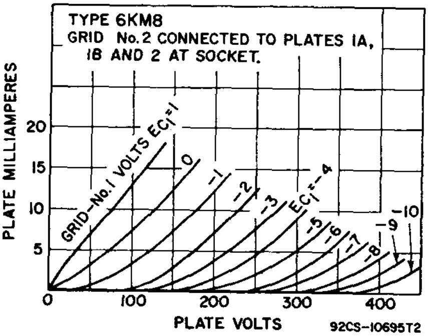
Separate-plate operation; plates not under test grounded

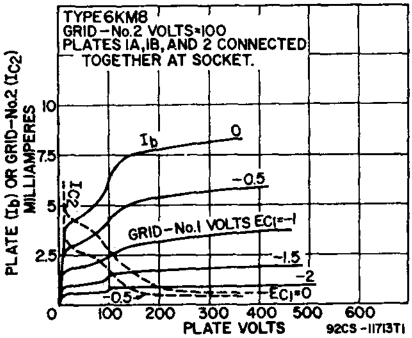
Plate	1A	1B	2	
Plate Voltage	100	100	100	volts
Grid-No.2 Voltage	100	100	100	volts
Grid-No.1 Supply Voltage	0	0	0	volts
Grid-No.1 Resistor (Bypassed)	2.2	2.2	2.2	megohms
Transconductance	2000	2000	1800	μmhos
Plate Resistance (Approx.)	0.1	0.1	0.12	megohm
Plate Current	2.3	2.3	2.1	mA
Grid-No.2 Current	3.8	3.8	3.3	mA

Tetrode Unit as Frequency Divider and Complex-Wave Generator

MAXIMUM RATINGS *(Design-Maximum Values)*

Plate Voltage (Each plate)	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid(Voltage:		
Positive-bias value	0	volts
Negative-bias value	50	volts
Plate Dissipation (Each plate)	1	watt

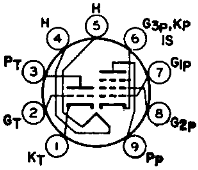




Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.65	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance, for grid-No.1-resistor-bias operation	2.2	megohms

Diode Unit

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Current	1	mA
CHARACTERISTICS, Instantaneous Value		
Tube Voltage Drop for plate current of 2 mA	10	volts



9DX

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE 6KR8

10KR8

Miniature type used in television receiver applications. The triode unit is used as a general-purpose amplifier; the pentode unit is used as a video amplifier. **Outlines section,** 6E; requires miniature 9-contact socket. Type 10KR8 is identical with type 6KR8 except for heater ratings.

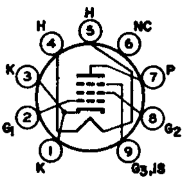
	6KR8	10KR8	
Heater Voltage (ac/dc)	6.3	10.5	volts
Heater Current	0.75	0.45	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	—	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2	5	watts
Grid-No.2 Input:			
For voltages up to 165 volts	—	1.1	watts
For voltages between 165 and 330 volts	—	See curve page 98	

CHARACTERISTICS	Triode Unit	Pentode Unit		
Plate Supply Voltage	125	35	200	volts
Grid-No.2 Supply Voltage	—	100	100	volts
Grid-No.1 Voltage	—	0	—	volts
Cathode-Bias Resistor	68	—	82	ohms
Amplification Factor	46	—	—	
Plate Resistance (Approx.)	4400	—	60000	ohms
Transconductance	10400	—	20000	μmhos
Plate Current	15	54	19.5	mA
Grid-No.2 Current	—	13.5	3	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—8	—	—	volts
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—	—6.3	volts

MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:	Triode Unit	Pentode Unit	
For fixed-bias operation	0.5	0.5	megohm
For cathode-bias operation	1	1	megohm



9PM

SEMIREMOTE-CUTOFF PENTODE 6KT6

3KT6, 4KT6

Miniature type used as if-amplifier tube in television receivers utilizing an intermediate frequency in the order of 40 MHZ. **Outlines section,** 6B; requires miniature 9-contact socket. Types 3KT6 and 4KT6 are identical with type 6KT6 except for heater ratings.

	3KT6	4KT6	6KT6	
Heater Voltage (ac/dc)	3.5	4.5	6.3	volts
Heater Current	0.6	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.019 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	9.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

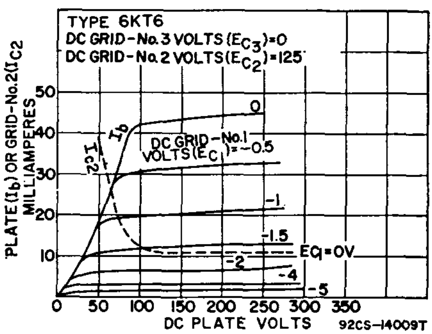
Plate Voltage	330	volts
Grid-No.3 (Suppressor-Grid) Voltage	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage	0	volts
Plate Dissipation	3.1	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	0.6	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	125	170	volts
Grid-No.3 Voltage	0	0	volts
Grid-No.2 Supply Voltage	125	170	volts
Cathode-Bias Resistor	56	56	ohms
Plate Resistor	160000	—	ohms
Transconductance	18000	—	µmhos
Plate Current	17	—	mA
Grid-No.2 Current	4.2	—	mA
Grid-No.1 Voltage (Approx.) for transconductance *of 10 µmhos*	—	—22	volts

MAXIMUM CIRCUIT VALUES

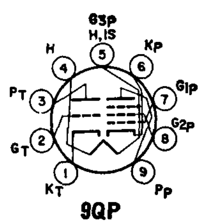
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	—	0.25	megohm
For cathode-bias operation	—	1	megohm



6KT8

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE

Miniature type used in color and black-and-white television receiver applications. The pentode unit is used as an if-amplifier tube, and the triode unit as a sync-separator or voltage-amplifier tube. Outlines section, 6B; requires miniature 9-contact socket.

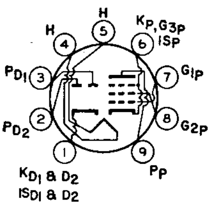


9QP

Heater Voltage (ac/dc)	6.3		volts
Heater Current	0.6		ampere
Heater-Cathode Voltage:			
Peak value	±200 max		volts
Average value	100 max		volts
Direct Interelectrode Capacitances:			
Triode Unit:	Unshielded	Shielded	
Grid to Plate	3	3	pF
Grid to Cathode, Heater, Grid No.3 of Pentode Unit, and Internal Shield	3.2	3.2	pF
Plate to Cathode, Heater, Grid No.3 of Pentode Unit, and Internal Shield	1.6	2.4	pF
Pentode Unit:			
Grid No.1 to Plate	0.046 max	0.030 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	7.5	7.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.2	2.8	pF
Grid of Triode Unit to Plate of Pentode Unit	0.018 max	0.003 max	pF
Grid No.1 of Pentode Unit to Plate of Triode Unit	0.006 max	0.002 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	1	2.5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Voltage	250	125	volts
Grid-No.2 Voltage	—	125	volts
Grid-No.1 Voltage	—2	—1	volts
Amplification Factor	100	—	
Plate Resistance (Approx.)	31500	150000	ohms
Transconductance	3200	10000	µmhos
Plate Current	1.8	12	mA
Grid-No.2 Current	—	4.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—3.5	—7	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.5	megohm
For cathode-bias operation	1	1	megohm



9LT

TWIN DIODE— SHARP-CUTOFF PENTODE

6KU8

10KU8

Neonoval type with frame-grid pentode used in television receiver applications. Diode units are used as horizontal phase detectors and the pentode unit is used as a video amplifier. **Outlines section, 10A;** requires neonoval 9-contact socket. Type 10KU8 is identical with type 6KU8 except for heater ratings.

	6KU8	10KU8	
Heater Voltage (ac/dc)	6.3	10.2	volts
Heater Current	0.725	0.45	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Diode Units:		
Plate of Diode Unit No.1 to All Other Electrodes	1.1	pF
Plate of Diode Unit No.2 to All Other Electrodes	1.1	pF
Diode Cathode to Plate of Diode Unit No.1	5.5	pF
Diode Cathode to Plate of Diode Unit No.2	5.5	pF
Pentode Unit:		
Grid No.1 to Plate	0.1 max	pF
Grid No.1 to Pentode Cathode, Diode Cathode, Heater, Grid No. 2, Grid No.3, and Internal Shields	12	pF

Plate to Pentode Cathode, Diode Cathode, Heater, Grid No.2, Grid No.3, and Internal Shields	3	pF
Pentode Grid No.1 to Plate of Diode Unit No.1	0.003 max	pF
Pentode Grid No.1 to Plate of Diode Unit No.2	0.003 max	pF
Pentode Plate to Plate of Diode Unit No.1	0.008 max	pF
Pentode Plate to Plate of Diode Unit No.2	0.008 max	pF

Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	4	watts
Grid-No.2 Input:		
For voltages up to 165 volts	1.1	watts
For voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	50	200	volts
Grid-No.2 Supply Voltage	100	100	volts
Grid-No.1 Voltage	0	0	volts
Cathode-Bias Resistor	—	82	ohms
Transconductance	—	20000	μmhos
Plate Resistance (Approx.)	—	50000	ohms
Plate Current	55*	17	mA
Grid-No.2 Current	18*	3.5	mA
Grid-No.1 Voltage for plate current of 100 μA	—	—5	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

* This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Diode Units (Each Unit)

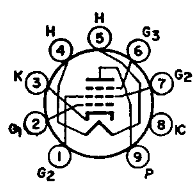
CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 2 mA	10	volts

6KV6

17KV6

BEAM POWER TUBE



9QU

Novar type used for high-voltage pulse- or shunt-regulator applications in color television receivers. Outlines section, 31D; requires novar 9-contact socket. Type 17KV6 is identical with type 6KV6 except for heating ratings.

	6KV6	17KV6	
Heater Voltage (ac/dc)	6.3	16.8	volts
Heater Current	1.6	0.6	amperes
Heater Warm-up Time	—	11	seconds
Heater-Cathode Voltage:			
Peak value	+200 max —500 max	+200 max —500 max	volts volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances (Approx.):			
Grid No.1 to Plate		1.2	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3		22	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3		9	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	100	140	volts
Grid-No.3 (Suppressor-Grid) Voltage	0	0	volts
Grid-No.2 (Screen-Grid) Voltage	140	140	volts
Grid-No.1 (Control-Grid) Voltage	0	—24.5	volts
Triode Amplification Factor#	—	4	
Plate Resistance (Approx.)	—	6000	ohms
Transconductance	—	9500	μmhos
Plate Current	580▪	80	mA
Grid-No.2 Current	24▪	2.4	mA
Grid-No.1 Voltage for plate current of 1 mA	—	—42	volts

High-Voltage-Pulse Shunt Regulator

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage (I_b = 0 mA)	770	volts
Peak Positive-Pulse Plate Voltage▲	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage	75	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage, Negative-bias value	75	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	950	mA
Average Cathode Current	275	mA
Plate Dissipation‡	20*	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	240	°C

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance:		
For grid-No.1-resistor-bias operation	1	megohm

Grid-No.3 and grid-No.2 connected, respectively, to cathode and plate at socket.

■ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

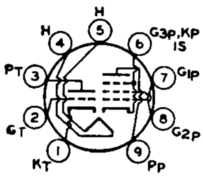
▲ Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

‡ Adequate circuit precautions must be taken to protect the tube in the absence of grid-No.1 bias.

* Plate dissipations up to 24 watts maximum are permissible for short periods of time (up to 10 seconds maximum) provided the maximum envelope-temperature rating is not exceeded.

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE 6KV8

11KV8



9DX

Miniature type with frame-grid pentode unit used in black-and-white television receivers. The triode unit is used in general-purpose voltage-amplifier, sync-separator, and sound-if-amplifier applications. The pentode unit is used as a video-output tube. **Outlines section,** 6E; requires miniature 9-contact socket. For curves of average plate characteristics for triode unit, refer to type 6AW8A. Type 11KV8 is identical with type 6KV8 except for heater ratings.

	6KV8	11KV8	
Heater Voltage (ac/dc)	6.3	10.9	volts
Heater Current	0.775	0.45	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Triode Unit:		
Grid to Plate	3.7	pF
Grid to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	2.5	pF
Plate to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	2.4	pF
Triode Grid to Pentode Plate	0.015 max	
Pentode Unit:		
Grid No.1 to Plate	0.12 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	13	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4.8	pF
Pentode Plate to Triode Plate	0.17 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	300	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts

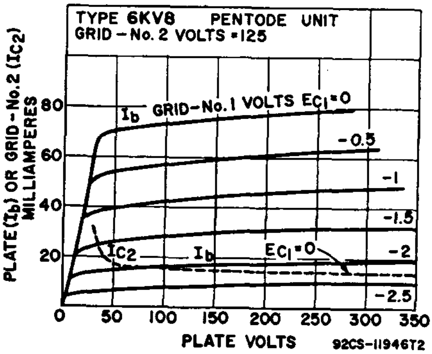
	Triode Unit	Pentode Unit	
Plate Dissipation	1	5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 150 volts	—	1	watt
For grid-No.2 voltages between 150 and 300 volts	—	See curve page 98	

CHARACTERISTICS

	Triode Unit	Pentode Unit		
Plate Supply Voltage	200	125	200	volts
Grid-No.2 Supply Voltage	—	125	125	volts
Grid-No.1 Supply Voltage	—2	0	0	volts
Cathode-Bias Resistor	—	82	68	ohms
Amplification Factor	70	—	—	
Plate Resistance (Approx.)	17500	55000	75000	ohms
Transconductance	4000	21000	23000	μmhos
Plate Current	4	16.5	20	mA
Grid-No.2 Current	—	3.1	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—4.5	—4.2	—4.2	volts

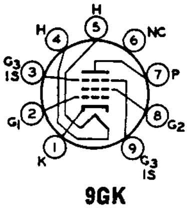
MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:	Triode Unit	Pentode Unit	
For fixed-bias operation	0.5	0.1	megohm
For cathode-bias operation	1	0.25	megohm



6KY6
7KY6

SHARP-CUTOFF PENTODE



Miniature type with frame grid used as video output amplifier in color and black-and-white television receivers. Outlines secton, 6E; requires miniature 9-contact socket. Type 7KY6 is identical with type 6KY6 except for heater ratings.

	6KY6	7KY6	
Heater Voltage (ac/dc)	6.3	7.3	volts
Heater Current	0.52	0.45	ampere
Heater Warm-up Time	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.16 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	14	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	6	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts

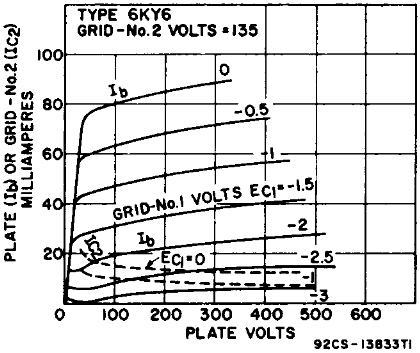
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	9	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	1	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

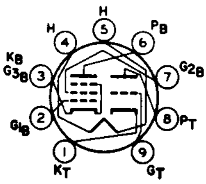
Plate Supply Voltage	200	volts
Grid-No.3 Voltage	Connected to cathode at socket	
Grid-No.2 Supply Voltage	135	volts
Grid-No.1 Supply Voltage	0	volts
Cathode-Bias Resistor	47	ohms
Plate Resistance (Approx.)	40000	ohms
Transconductance	30000	μmhos
Plate Current	30	mA
Grid-No.2 Current	5.2	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—4.5	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.25	megohm



Refer to chart at end of section. **6KY8**



9QT

HIGH-MU TRIODE—BEAM POWER TUBE **6KY8A**

15KY8A

Novar type used in combined vertical-deflection-oscillator and vertical-deflection-amplifier applications in black-and-white television receivers having low-voltage "B" supplies. **Outlines section,** 30A; requires novar 9-contact socket. Type 15KY8A is identical with type 6KY8A except for heater ratings.

	6KY8A	15KY8A	
Heater Voltage (ac/dc)	6.3	15	volts
Heater Current	1.1	0.45	amperes
Heater Warm-up Time (Average)	—	11	seconds
Heater Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Triode Unit:		
Grid to Plate	0.44	pF
Grid to Cathode and Heater	15	pF
Plate to Cathode and Heater	7	pF

Pentode Unit:

Grid No.1 to Plate	0.048	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	2.6	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	0.28	pF

Class A_1 Amplifier

CHARACTERISTICS	Triode Unit	Beam Power Unit			
Plate Voltage	250	50	135	120	volts
Grid-No.2 (Screen-Grid) Voltage	—	120	120	*	volts
Grid-No.1 (Control-Grid) Voltage	—3	0	—10	—10	volts
Amplification Factor	64	—	—	7	
Plate Resistance (Approx.)	40000	—	18000	—	ohms
Transconductance	1600	—	8400	—	µmhos
Plate Current	1.4	170•	39	—	mA
Grid-No.2 Current	—	20•	3	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—	—24	—	volts

* Triode connection, grid No.2 connected to plate at socket.

• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

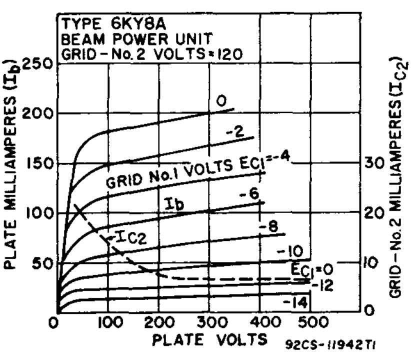
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit Oscillator	Beam Power Unit Amplifier	
DC Plate Voltage	330	300	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	—	2200†	volts
DC Grid-No.2 Voltage	—	150	volts
Peak Negative-Pulse Grid-No.1 Voltage	400	250	volts
Peak Cathode Current	77	200	mA
Average Cathode Current	22	60	mA
Plate Dissipation	1.5	12	watts
Grid-No.2 Input	—	1.9	watts

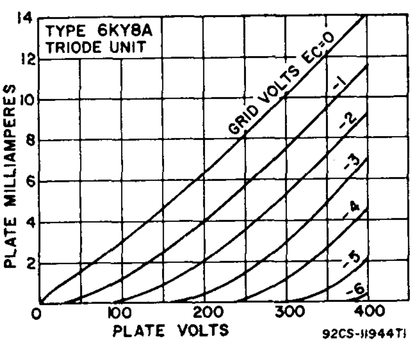
MAXIMUM CIRCUIT VALUES

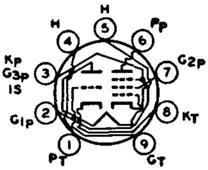
Grid-No.1-Circuit Resistance: For grid-resistor-bias operation	2.2	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

† Under no conditions should this maximum value be exceeded.







9FZ

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

6KZ8

5KZ8, 9KZ8

Miniature type used as combined oscillator and mixer in vhf color and black-and-white television receivers. Outlines section, 6B; requires miniature 9-contact socket. Types 5KZ8 and 9KZ8 are identical with type 6KZ8 except for heater ratings.

	5KZ8	6KZ8	9KZ8	
Heater Voltage (ac/dc)	4.7	6.3	9.45	volts
Heater Current	0.6	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:*		
Triode Unit:		
Grid to Plate	1.6	pF
Grid to Triode Cathode, Pentode Cathode, Heater, Pentode Grid No.3, and Heater	3.2	pF
Plate to Triode Cathode, Pentode Cathode, Heater, Pentode Grid No.3, and Heater	1.8	pF
Pentode Unit:		
Grid No.1 to Plate	0.01 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3.4	pF
Heater to Cathode (Each Unit)	3.2#	pF

* With external shield connected to cathode.

\# With external shield connected to ground.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	2.5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	

CHARACTERISTICS

Plate Voltage	125	125	volts
Grid-No.2 Voltage	—	125	volts
Grid-No.1 Voltage	—1	—1	volt
Amplification Factor	46	—	
Plate Resistance (Approx.)	5400	200000	ohms
Transconductance	8500	7500	µmhos
Plate Current	13.5	12	mA
Grid-No.2 Current	—	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 µA	—8	—8	volts

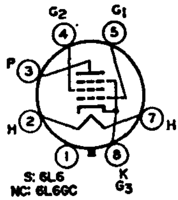
MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.25	0.25	megohm
For cathode-bias operation	0.5	0.5	megohm

Refer to chart at end of section. **6L5G**

6L6
6L6GC

BEAM POWER TUBE



7AC

Metal type 6L6 and glass octal type 6L6GC are used in the output stage of audio amplifying equipment, especially units designed to have ample reserve of power-delivering ability. Outlines section, 4 and 19D, respectively; require octal socket. These tubes, like other power-handling tubes, should be adequately ventilated. Type 6L6GC can be used in place of type 6L6 and may be supplied with pin 1 omitted.

	6L6	6L6GC	
Heater Voltage (ac/dc)		6.3	volts
Heater Current		0.9	ampere
Heater-Cathode Voltage:	6L6	6L6GC	
Peak value	±180 max	±200 max	volts
Average value	—	100 max	volts
Direct Interelectrode Capacitances (Approx.):			
Grid No.1 to Plate	0.4*	0.6	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	10*	10	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	12*	6.5	pF

* With pin 1 connected to pin 8.

Class A_1 Amplifier

MAXIMUM RATINGS	6L6 Design-Center Values	6L6GC Design Maximum Values	
Plate Voltage	360	500	volts
Grid-No.2 (Screen-Grid) Voltage	270	450▲	volts
Plate Dissipation	19	30	watts
Grid-No.2 Input	2.5	5	watts

TYPICAL OPERATION				
Plate Voltage	250	300	350	volts
Grid-No.2 Voltage	250	200	250	volts
Grid-No.1 (Control-Grid) Voltage	—14	—12.5	—18	volts
Peak AF Grid-No.1 Voltage	14	12.5	18	volts
Zero-Signal Plate Current	72	48	54	mA
Maximum-Signal Plate Current	79	55	66	mA
Zero-Signal Grid-No.2 Current	5	2.5	2.5	mA
Maximum-Signal Grid-No.2 Current	7.3	4.7	7	mA
Plate Resistance (Approx.)	22500	35000	33000	ohms
Transconductance	6000	5300	5200	μmhos
Load Resistance	2500	4500	4200	ohms
Total Harmonic Distortion	10	11	15	per cent
Maximum-Signal Power Output	6.5	6.5	10.8	watts

▲ In push-pull circuits where grid No.2 of each tube is connected to a tap on the plate winding of the output transformer, this maximum rating is 500 volts.

Class A_1 Amplifier (Triode Connection)†

MAXIMUM RATINGS	6L6 Design-Center Values	6L6GC Design-Maximum Values	
Plate Voltage	275	450	volts
Plate Dissipation (Total)	19	30	watts

TYPICAL OPERATION		
Plate Voltage	250	volts
Grid-No.1 Voltage	—20	volts
Peak AF Grid-No.1 Voltage	20	volts
Zero-Signal Plate Current	40	mA
Maximum-Signal Plate Current	44	mA
Plate Resistance (Approx.)	1700	ohms
Amplification Factor	8	
Transconductance	4700	μmhos
Load Resistance	5000	ohms
Total Harmonic Distortion	5	per cent
Maximum-Signal Power Output	1.4	watts

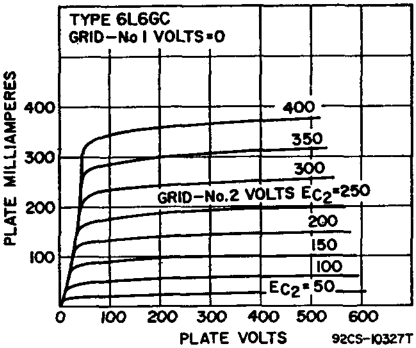
† Grid No.2 connected to plate.

Push-Pull Class A_1 Amplifier

MAXIMUM RATINGS (Same as for Class A₁ Amplifier)

TYPICAL OPERATION (Values are for two tubes)

Plate Voltage	250	270	volts
Grid-No.2 Voltage	250	270	volts
Grid-No.1 Voltage	—16	—17.5	volts
Peak AF Grid-No.1-to-Grid-No.1 Voltage	32	35	volts
Zero-Signal Plate Current	120	134	mA
Maximum-Signal Plate Current	140	155	mA
Zero-Signal Grid-No.2 Current	10	11	mA
Maximum-Signal Grid-No.2 Current	16	17	mA
Effective Load Resistance (Plate-to-plate)	5000	5000	ohms
Total Harmonic Distortion	2	2	per cent
Maximum-Signal Power Output	14.5	17.5	watts



MAXIMUM RATINGS (Same as for Class A₁ Amplifier)

TYPICAL OPERATION (Values are for two tubes)

	6L6		6L6GC	
Plate Voltage	360	360	450	volts
Grid-No.2 Voltage	270	270	400	volts
Grid-No.1 Voltage	—22.5	—22.5	—37	volts
Peak AF Grid-No.1-to-Grid-No.1 Voltage	45	45	70	volts
Zero-Signal Plate Current	88	88	116	mA
Maximum-Signal Plate Current	132	140	210	mA
Zero-Signal Grid-No.2 Current	5	5	5.6	mA
Maximum-Signal Grid-No.2 Current	15	11	22	mA
Effective Load Resistance (Plate-to-plate)	6600	3800	5600	ohms
Total Harmonic Distortion	2	2	1.8	per cent
Maximum-Signal Power Output	26.5	18	55	watts

Push-Pull Class AB_2 Amplifier

MAXIMUM RATINGS (Same as for Class A₁ Amplifier)

TYPICAL OPERATION (Values are for two tubes)

Plate Voltage	360	360	volts
Grid-No.2 Voltage	225	270	volts
Grid-No.1 Voltage	—18	—22.5	volts
Peak AF Grid-No.1-to-Grid-No.1 Voltage	52	72	volts
Zero-Signal Plate Current	78	88	mA
Maximum-Signal Plate Current	142	205	mA
Zero-Signal Grid-No.2 Current	3.5	5	mA
Maximum-Signal Grid-No.2 Current	11	16	mA
Effective Load Resistance (Plate-to-plate)	6000	3800	ohms
Total Harmonic Distortion	2	2	per cent
Maximum-Signal Power Output	31	47	watts

MAXIMUM CIRCUIT VALUES

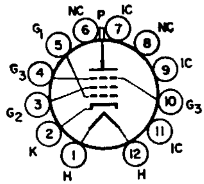
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

Refer to chart at end of section. **6L6G**

Refer to chart at end of section. **6L6GB**

6L7 Refer to chart at end of section.

6L7G Refer to chart at end of section.



12GJ

6LB6 BEAM POWER TUBE

Duodecar type used as horizontal-deflection amplifier in color and black-and-white television receivers. Outlines section, 16E; requires duodecar 12-contact socket. Heater: volts (ac/dc), 6.3; amperes, 2.25; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

CHARACTERISTICS	Triode* Connection	Pentode Connection			
Plate Voltage	125	5000	50	150	volts
Grid No.3 (Suppressor Grid)		Connected to cathode at socket			
Grid-No.2 Voltage	125	110	110	110	volts
Grid-No.1 Voltage	—25	—	—	—20	volts
Plate Resistance (Approx.)	—	—	—	6600	ohms
Transconductance	—	—	—	13400	μmhos
Plate Current	—	—	560‡	105	mA
Grid-No.2 Current	—	—	46‡	2	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—125	—	—40	volts
Amplification Factor	4	—	—	—	

* Grid No.2 tied to plate.

‡ This value may be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

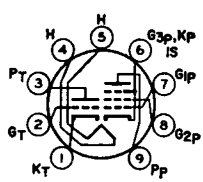
MAXIMUM RATINGS (Design-Maximum Values)		
DC Plate Supply Voltage	990	volts
Peak Positive Pulse Plate Voltage# (Absolute Maximum Value)	7000	volts
Peak Negative-Pulse Plate Voltage	100	volts
Grid-No.3 Voltage, Positive-bias value	0	volts
Grid-No.2 Voltage	200	volts
Peak Negative Grid-No.1 Voltage	300	volts
Peak Cathode Current	1100	mA
Average Cathode Current	315	mA
Plate Dissipation▪ (Absolute Maximum Value)	30	watts
Grid-No.2 Input	5	watts
Bulb Temperature (At hottest point)	250	°C
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance:		
With feedback-type high voltage regulation	1.2	megohms
With shunt-type high voltage regulation (switching mode)	10	megohms
Grid-No.3-Circuit Resistance	0	ohms

\# Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

▪ A bias resistor or other means is required to protect the tube in absence of excitation.

6LB8 MEDIUM-MU TRIODE—SHARP-CUTOFF PENTODE

10LB8



9DX

Neonoval type with frame-grid pentode used in television receivers. The triode unit is used as a voltage amplifier; the pentode unit is used as a video amplifier. Outlines section, 10A; requires neonoval 9-contact socket. Type 10LB8 is identical with type 6LB8 except for heater ratings.

	6LB8	10LB8	
Heater Arrangement	Parallel	Series	
Heater Voltage	6.3	10.2	volts
Heater Current	0.725	0.45	ampere
Heater Warm-up Time	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2	4	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	

CHARACTERISTICS				
Plate Supply Voltage	125	50	200	volts
Grid-No.2 Supply Voltage	—	100	100	volts
Grid-No.1 Voltage	0	0	0	volts
Cathode-Bias Resistor	68	—	82	ohms
Amplification Factor	30	—	—	
Plate Resistance (Approx.)	6000	—	50000	ohms
Transconductance	5000	—	20000	μmhos
Plate Current	13	55*	17	mA
Grid-No.2 Current	—	18*	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—10	—	—	volts
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—	—5	volts

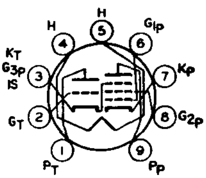
MAXIMUM CIRCUIT VALUES			
Grid-No.1 Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

* This value can be measured by a method involving a recurrent waveform such that the *maximum range of the tube will not be exceeded.*

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE

6LC8

8LC8



9QY

Miniature type used in color and black-and-white television receiver applications. Pentode unit is used in noise-immune gated-agc-amplifier circuits, and the triode unit in sync-separator circuits. Outlines section, 6E; requires miniature 9-contact socket. Type 8LC8 is identical with type 6LC8 except for heater ratings. For curves of average plate characteristics, refer to type 6KA8.

	6LC8	8LC8	
Heater Voltage (ac/dc)	6.3	8.4	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	2.2	pF
Grid to Cathode, Heater, Pentode Grid No.3, and Internal Shield	2.8	pF
Plate to Cathode, Heater, Pentode Grid No.3, and Internal Shield	2.2	pF
Pentode Unit:		
Grid No.1 to Plate	0.10 max	pF
Grid No.1 to Cathode, Heater, Grid No.3, Triode Cathode, and Internal Shield	10	pF
Grid No.3, Triode Cathode, and Internal Shield to Plate	3.4	pF
Grid No.1 to Grid No.3, Triode Cathode, and Internal Shield	0.36	pF
Grid No.3, Triode Cathode, and Internal Shield to Plate, Cathode, Heater, Grid No.1, and Grid No.2	12.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	
Plate Voltage	300	volts
Grid Voltage:		
Positive-bias value	0	volts
Negative-bias value	50	volts
Plate Dissipation	1.1	watts

CHARACTERISTICS	Triode Unit	Pentode Unit	
Plate Supply Voltage	200	150	volts
Grid-No.2 Supply Voltage	—	100	volts
Grid-No.1 Voltage	—2	—	volts
Cathode-Bias Resistor	—	180	ohms
Amplification Factor	70	—	
Plate Resistance (Approx.)	17500	100000	ohms
Transconductance, Grid No.1 to Plate	4000	4400	μmhos
Transconductance, Grid No.3 to Plate	—	600	μmhos
Plate Current	4	4	mA
Grid-No.2 Current	—	2.8	mA
Grid-No.1 Voltage (Approx.):			
For plate current of 10 μA	—5	—	volts
For plate current of 20 μA	—	—4	volts
Grid-No.3 Voltage (Approx.) for plate current of 20 μA	—	—7*	volts

MAXIMUM CIRCUIT VALUES	Triode Unit	
Grid-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

* With no external connection to triode plate and triode grid.

Gated AGC Amplifier and Noise Inverter

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Pentode Unit	
DC Plate Voltage	300	volts
Peak Positive-Pulse Plate Voltage#	600	volts
Grid-No.3 (Control-Grid) Voltage:		
Positive-bias value	0	volts
Negative-bias value	100	volts
Grid-No.2 (Screen-Grid) Supply Voltage	300	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:		
Positive-bias value	0	volts
Negative-bias value	50	volts
Plate Dissipation	2	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 150 volts	1.1	watts
For grid-No.2 voltages between 150 and 300 volts	See curve page 98	

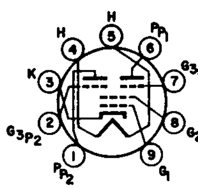
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6LE8

10LE8, 15LE8

TWIN PENTODE



9QZ

Miniature type used as combined color demodulator and matrix amplifier in color television receivers utilizing high-level demodulation systems. Outlines section, 6G; requires miniature 9-contact socket. Types 10LE8 and 15LE8 are identical with type 6LE8 except for heater ratings.

	6LE8	10LE8	15LE8	
Heater Voltage (ac/dc)	6.3	10.0	15.0	volts
Heater Current	0.76	0.45	0.30	ampere
Heater Warm-up Time (Average)	—	11	11	seconds
Heater-Cathode Voltage:				
Peak value	+200, —300 max			volts
Average value	100 max			volts

Direct Interelectrode Capacitances:		
Plate (Each Unit) to All Other Electrodes	3.7	pF
Grid No.1 to All Other Electrodes	15.5	pF
Grid No.3 (Each Unit) to All Other Electrodes	6	pF
Grid No.3 to Plate (Each Unit)	2.7	pF
Grid No.3 (Unit No.1) to Grid No.3 (Unit No.2)	0.1	pF

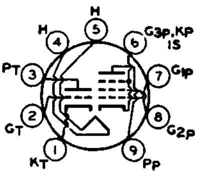
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage (Each Unit)	300	volts
Grid-No.2 (Screen-Grid) Voltage	150	volts
Plate Dissipation (Each Unit)	*2*	watts
Grid-No.2 Input	2	watts

CHARACTERISTICS

	G_1 Control	G_3 Control	
Plate Voltage	100	100	volts
Grid-No.3 (Suppressor-Grid) Voltage	0	0	volts
Grid-No.2 Voltage	100	100	volts
Grid-No.1 (Control-Grid) Voltage, Negative-bias value	2.5	2.5	volts
Transconductance (Approx.)	5800	350	µmhos
Plate Resistance (Approx.)	50000	50000	ohms
Plate Current	8	7.6	mA
Grid-No.2 Current	15	14.5	mA
Grid-No.1 Voltage for plate current of 20 µA	—7.2	—	volts
Grid-No.1 Voltage for plate current of 100 µA	—6.3	—	volts
Grid-No.3 Voltage for plate current of 20 µA	—	—17.4	volts
Grid-No.3 Voltage for plate current of 100 µA	—	—16.5	volts



9DX

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE 6LF8

Miniature type used in video-amplifier stages of color and black-and-white television receivers. Outlines section, 6E; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.6	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	*100 max*	volts
Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	2.2	pF
Grid to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	3.2	pF
Plate to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	1.8	pF
Pentode Unit:		
Grid No.1 to Plate	0.06 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	10	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3.6	pF
Pentode Grid No.1 to Triode Plate	0.008 max	pF
Pentode Plate to Triode Plate	0.15 max	pF

Class A Amplifier

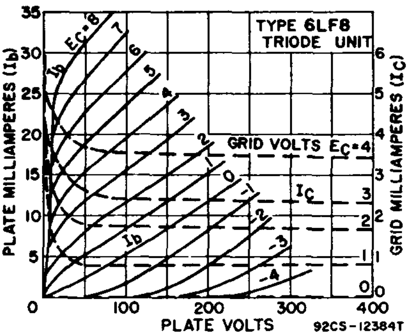
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage:			
Positive-bias value	4	0	volts
Negative-bias value	—55	—55	volts
Grid-No.1 Current	8	0	mA
Plate Dissipation	1.1	3.75	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	

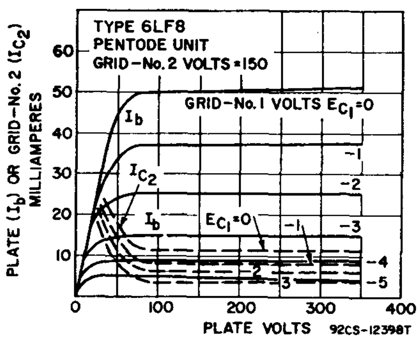
CHARACTERISTICS	Triode Unit		Pentode Unit		
Plate Voltage	200	40	75	100	volts
Grid-No.2 Voltage	—	—	150	150	volts
Grid-No.1 Voltage	—2	3	0	—2.5	volts
Amplification Factor	70	40	—	—	
Plate Resistance (Approx.)	17500	10000	—	200000	ohms
Transconductance	4000	4000	—	11000	μmhos
Plate Current	4	11	50▪	20	mA
Grid-No.2 Current	—	—	12▪	5	mA
Grid-No.1 Current	0	2.7	0	0	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—5	—	—	—8	volts

MAXIMUM CIRCUIT VALUES

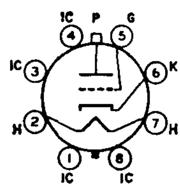
Grid-No.1-Circuit Resistance:	Triode Unit	Pentode Unit	
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.





6LJ6 BEAM TRIODE



8MQ

Glass octal type used for the shunt regulation of high-voltage, low-current power supplies in color and black-and-white television receivers. Outlines section, 21D; requires octal socket. For high-voltage and X-ray safety considerations, refer to type 1G3GT/1B3GT.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.2	amperes
Heater Cathode Voltage	+0, —450*	volts
Direct Interelectrode Capacitances:		
Grid to Plate	1	pF
Grid to Cathode and Heater	2.6	pF
Plate to Cathode and Heater	1	pF

* Series impedance should be used with the cathode to limit the cathode current under prolonged short-circuit conditions to 450 mA.

Shunt Voltage-Regulator Service

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	27000	volts
Unregulated Plate Supply Voltage	60000	volts
Negative Grid Voltage	135	volts
Peak Negative Grid Voltage▪	440	volts
Plate Dissipation	40	watts
Average Plate Current	1.6	mA

TYPICAL OPERATION

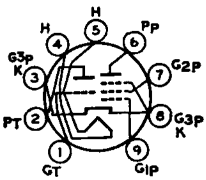
Unregulated DC Supply Voltage	36000	volts
Equivalent Resistance of Unregulated Supply	11	megohms
DC Reference Voltage	200	volts

Equivalent Resistance of Reference Supply	1000	ohms
Effective Grid-Plate Transconductance	200	µmhos
DC Plate Current for Load Current of 0 mA	1000	µA
DC Plate Current for Load Current of 1 mA	45	µA
Regulated DC Output Voltage for Load Current of 0 mA	25000	volts
Regulated DC Ouput Voltage for Load Current of 1 mA	24500	volts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance:		
For use with "Flyback Transformer" high voltage supply	3	megohms

▪ For interval of 20 seconds maximum during equipment warm-up period.



9GF

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

6LJ8

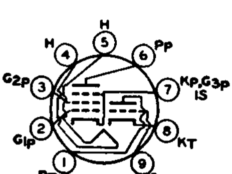
4LJ8, 5LJ8

Miniature type used as a combined oscillator and mixer in vhf television receivers. Outlines section, 6B; requires 9-contact socket. Types 4LJ8 and 5LJ8 are identical with type 6LJ8 except for heater ratings.

	4LJ8	5LJ8	6LJ8	
Heater Voltage (ac/dc)	4.3	5.6	6.3	volts
Heater Current	0.6	0.45	0.4	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	280	280	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	280	volts
Grid-No.2 Voltage	—	See curve page 98	
Cathode Current	20	20	mA
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2	2	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 140 volts	—	0.5	watts
For grid-No.2 voltages between 140 and 280 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Voltage	125	125	volts
Grid-No.2 Voltage	—	125	volts
Cathode-Bias Resistor	68	33	ohms
Amplification Factor	40	—	
Plate Resistance (Approx.)	5000	125000	ohms
Transconductance	8000	13000	µmhos
Plate Current	13	12	mA
Grid-No.2 Current	—	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 30 µA	—6.5	—4	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	1	0.5	megohm
For cathode-bias operation	0.5	0.25	megohm



9AE

MEDIUM-MU TRIODE—SEMI-REMOTE-CUTOFF PENTODE

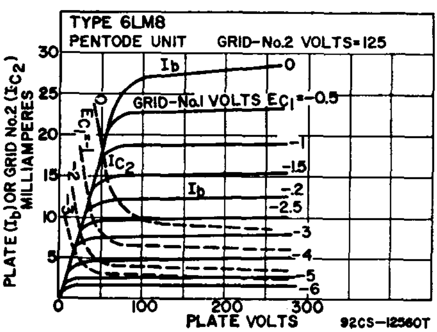
6LM8

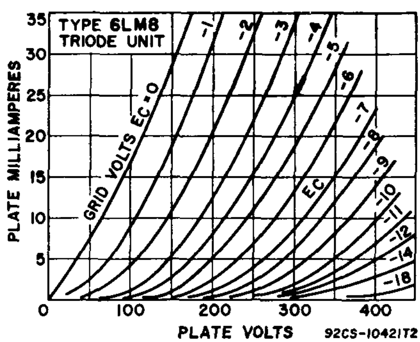
Miniature type used in color and black-and-white television receiver applications. The pentode unit is used in burst-amplifier circuits, and the triode unit as a general-purpose amplifier tube. Outlines section, 6B; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	1.8	pF
Grid to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	3.2	pF
Plate to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	1.9	pF
Pentode Unit:		
Grid No.1 to Plate	0.015 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3.8	pF
Heater to Cathode (Each Unit)	3.2	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	350	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	2.5	volts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watts
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Voltage	125	125	volts
Grid-No.2 Voltage	—	125	volts
Grid No.1 Voltage	—1	—2	volts
Amplification Factor	46	—	
Plate Resistance (Approx.)	5400	150000	ohms
Transconductance	8500	6000	μmhos
Plate Current	13.5	12	mA
Grid-No.2 Current	—	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—8	—14	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	0.5	megohm

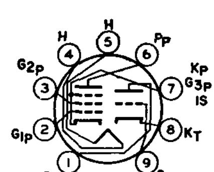




6LN8/ LCF80

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

Miniature type used in frequency-changer service in television receivers. Outlines section, 6B; requires miniature 9-contact socket.



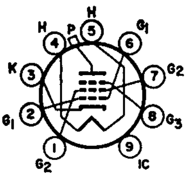
9DC

Heater Voltage (ac/dc)	6	volts
Heater Current	0.45	ampere
Peak Heater-Cathode Voltage	±100 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	550	volts
Plate Voltage	250	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	550	volts
Grid-No.2 Voltage:			
With cathode current of 14 mA	—	175	volts
With cathode current less than 10 mA	—	200	volts
Cathode Current	14	14	mA
Plate Dissipation	1.5	1.7	watts
Grid-No.2 Input:			
With plate dissipation greater than 1.2 watts	—	0.5	watt
With plate dissipation less than 1.2 watts	—	0.75	watt
CHARACTERISTICS			
Plate Voltage	100	170	volts
Grid-No.2 Voltage	—	170	volts
Grid-No.1 Voltage	—2	—2	volts
Amplification Factor	20	—	
Mu-Factor, Grid No.2 to Grid No.1	—	47	
Plate Resistance (Approx.)	—	0.4	megohm
Transconductance	5000	6200	μmhos
Plate Current	14	10	mA
Grid-No.2 Current	—	2.8	mA
Input Resistance at frequency of 50 MHz	—	0.01	megohm
Equivalent Noise Resistance	—	1500	ohms
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.5	megohm
For cathode-bias operation	0.5	1	megohm

6LQ6 6LQ6/ 6JE6B



9QL

BEAM POWER TUBE

24LQ6, 24LQ6/24JE6C

Novar types used as horizontal-deflection amplifier in color and black-and-white television receivers. Outlines section, 32C; require novar 9-contact socket. Types 24LQ6 and 24LQ6/24JE6C are identical with type 6LQ6 except for heater ratings.

	6LQ6 6LQ6/6JE6B	24LQ6 24LQ6/24JE6C	
Heater Voltage (ac/dc)	6.3	24	volts
Heater Current	2.5	0.6	amperes
Heater Warm-up Time	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances:			
Grid No.1 to Plate		0.56	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3		22	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3		11	pF

Class A_1 Amplifier

CHARACTERISTICS	Triode* Connection	Pentode Connection			
Peak Positive-Pulse Plate Voltage#	—	5000	—	—	volts
Plate Voltage	145	—	60	175	volts
Grid-No.3 (Suppressor-Grid) Voltage	—	30	30	30	volts
Grid-No.2 (Screen-Grid) Voltage	145	145	145	145	volts
Grid-No.1 (Control-Grid) Voltage	—35	—	0	—35	volts
Plate Resistance (Approx.)	—	—	—	7000	ohms
Transconductance	—	—	—	7500	μmhos
Plate Current	—	—	710‡	95	mA
Grid-No.2 Current	—	—	55‡	2.4	mA

	Triode* Connection		Pentode Connection		
Grid-No.1 Voltage for plate current of 1 mA	—	—125	—	—60	volts
Amplification Factor	2.8	—	—	—	

* Grid No.3 and grid No.2 connected, respectively, to cathode and plate at socket.

‡ This value may be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS *(Design-Maximum Values)*

Plate Supply Voltage	990	volts
Peak Positive-Pulse Plate Voltage#	7500	volts
Peak Negative-Pulse Plate Voltage	1100	volts
Grid-No.3 Voltage■	75	volts
Grid-No.2 Voltage	220	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	1200	mA
Average Cathode Current	350	mA
Plate Dissipation°	30	watts
Plate Dissipation (Temporary overload)▲	200	watts
Grid-No.2 Input	5	watts
Envelope Temperature (At hottest point)	250	°C

MAXIMUM CIRCUIT VALUES

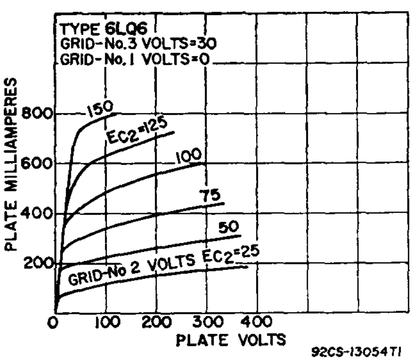
Grid-No.1-Circuit Resistance:		
For grid-No.1-resistor-bias operation	0.47	megohm
For plate-pulsed operation (horizontal-deflection circuits only)	10	megohms

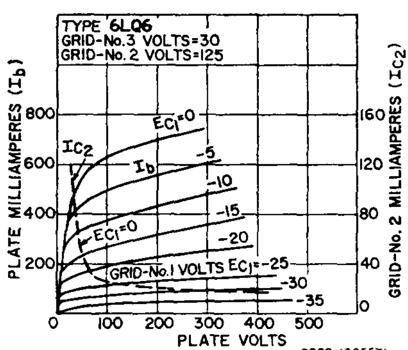
Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

■ For horizontal-deflection service, a positive voltage may be applied to grid-No.3 to minimize "snivets" interference in both vhf and uhf television receivers. A typical value is 30 volts.

° A bias resistor or other means is required to protect the tube in absence of excitation.

▲ Total continuous or accumulated time not to exceed 40 seconds.

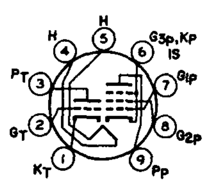




6LQ8
11LQ8

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

Miniature type used in color and black-and-white television receiver applications. The pentode unit is used as a video output tube. The triode unit is used in sync separator and sound-if circuits. **Outlines section,** 6E; requires miniature 9-contact socket. Type 11LQ8 is identical with type 6LQ8 except for heater ratings.



9DX

	6LQ8	11LQ8	
Heater Voltage (ac/dc)	6.3	10.9	volts
Heater Current	0.7	0.45	ampere
Heater Warm-up Time	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

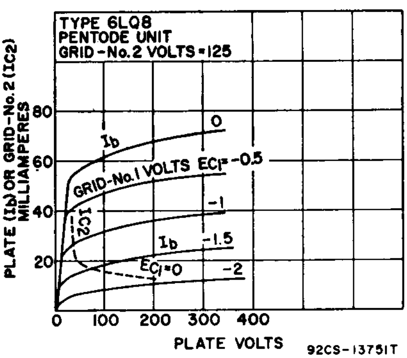
Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	2.8	pF
Grid to Triode Cathode, Pentode Cathode, Heater, Pentode Grid No.3, and Internal Shield	4.2	pF
Plate to Triode Cathode, Pentode Cathode, Heater, Pentode Grid No.3, and Internal Shield	2.4	pF
Pentode Unit:		
Grid No.1 to Plate	0.12 max	pF
Grid No.1 to Cathode Heater, Grid No.2, Grid No.3, and Internal Shield	14	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4.8	pF
Triode Grid to Pentode Plate	0.015 max	pF
Pentode Plate to Triode Plate	0.17 max	pF

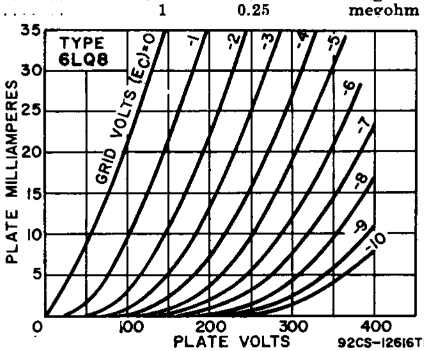
Class A_1 Amplifier

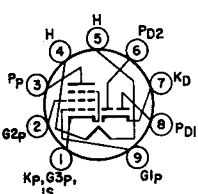
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	300	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	300	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2	5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 150 volts	—	1	watts
For grid-No.2 voltages between 150 and 300 volts	—	See curve page 98	

CHARACTERISTICS	Triode Unit	Pentode Unit		
Plate Supply Voltage	125	125	200	volts
Grid-No.2 Supply Voltage	—	125	125	volts
Cathode-Bias Resistor	68	82	68	ohms
Amplification Factor	46	—	—	
Plate Resistance (Approx.)	4400	55000	75000	ohms
Transconductance	10400	21000	23000	µmhos
Plate Current	15	16.5	20	mA
Grid-No.2 Current	—	3.1	3.5	mA
Grid-No.1 Voltage(Approx.) for plate current of 100 µA	—6	—4.2	—4.2	volts

MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:	Triode Unit	Pentode Unit	
For fixed-bias operation	0.5	0.1	megohm
For cathode-bias operation	1	0.25	megohm







9RL

TWIN DIODE— SHARP-CUTOFF PENTODE

6LT8

8LT8

Miniature type used in television receiver applications. The pentode unit is used in low-frequency horizontal-oscillator applications. The diode units are used in horizontal afc discriminator circuits. **Outlines section, 6B**; requires miniature 9-contact socket. Type 8LT8 is identical with type 6LT8 except for heater ratings.

	6LT8	8LT8	
Heater Voltage	6.3	8.1	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Pentode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	3.1	watts
Grid-No.2 Input:		
For grid-No 2 voltages up to 165 volts	0.65	watt
For grid-No.2 voltages between 165 and 33 volts	See curve page 98	

CHARACTERISTICS

Plate Voltage	125	volts
Grid No.3 (Suppressor Grid)	Connected to ground	
Grid-No.2 Voltage	125	volts
Cathode-Bias Resistor	56	ohms
Plate Resistance (Approx.)	200000	ohms
Transconductance	13000	μmhos
Plate Current	10	mA
Grid-No.2 Current	3.4	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—3.5	volts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, for cathode-bias operation	1	megohm

Diode Unit (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

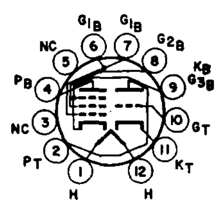
Plate Current (Continuous Operation)	5	mA

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 20 mA	5	volts

6LU8
21LU8

MEDIUM-MU TRIODE— BEAM POWER TUBE



12DZ

Duodecar type used as a combined vertical-deflection oscillator and vertical-deflection amplifier in color television receivers. **Outlines section,** 15D; requires duodecar 12-contact socket. Type 21LU8 is identical with type 6LU8 except for heater ratings.

	6LU8	21LU8	
Heater Arrangement	Parallel	Series	
Heater Voltage	6.3	21	volts
Heater Current	1.5	0.45	amperes
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS	Triode Unit	Beam Power Unit			
Plate Voltage	250	45	135	120	volts
Grid-No.2 (Screen-Grid) Voltage	—	125	120	120•	volts
Grid-No.1 (Control-Grid) Voltage	—4	0	—10	—10	volts
Amplification Factor	58	—	—	6.5	
Plate Resistance (Approx.)	16000	—	12000	—	ohms
Transconductance	3600	—	9300	—	μmhos
Plate Current	2.3	200••	56	—	mA
Grid-No.2 Current	—	20••	3	—	mA
Grid-No.1 Voltage (Approx.):					
For plate current of 10 μA	—6.6	—	—	—	volts
For plate current of 100 μA	—	—	—30	—	volts
For plate current of 1 mA	—	—	—26	—	volts

• Triode connection, Grid No.2 connected to plate at socket.

•• *This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.*

Vertical-Deflection Oscillator and Amplifier

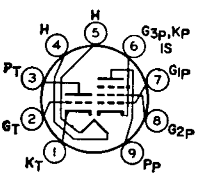
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit Oscillator	Beam Power Unit Amplifier	
Plate Voltage	400	400	volts
Grid-No.2 Voltage	—	300	volts
Peak Positive-Pulse Plate Voltage#	—	2500	volts
Peak Negative-Pulse Grid-No.1 Voltage	400	250	volts
Plate Dissipation▪	2.5	14	watts
Peak Cathode Current	105	260	mA
Average Cathode Current	30	75	mA
Grid-No.2 Input	—	2.75	watts
Bulb Temperature (At hottest point)	—	210	°C
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For fixed-bias operation	—	1	megohm
For cathode-bias operation	2.2	2.2	megohms

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

▪ A bias resistor or other means is required to protect the tube in absence of excitation.

Refer to type 6JW8/ECF802. **6LX8/LCF802**



9DX

HIGH-MU TRIODE—SHARP-CUTOFF PENTODE 6LY8

Miniature type used in color and black-and-white television receiver applications. The pentode unit is used as a video amplifier, and the triode unit for general-purpose use. **Outlines section,** 6E; requires 9-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.75; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

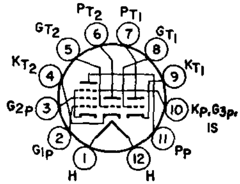
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	1	5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	

CHARACTERISTICS				
Plate Voltage	250	35	200	volts
Grid-No.2 Voltage	—	100	100	volts
Grid-No.1 Voltage	—2.0	0	—	volts
Cathode-Bias Resistor	—	—	82	ohms
Amplification Factor	100	—	—	
Plate Resistance (Approx.)	59000	—	60000	ohms
Transconductance	1700	—	20000	μmhos
Plate Current	1.0	54	19.5	mA
Grid-No.2 Current	—	13.5	3	mA
Grid Voltage (Approx.) for plate current of 10 μA	—5	—	—	volts
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—	—6.3	volts

MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.5	megohm
For cathode-bias operation	1	1	megohm

6M11

HIGH-MU TWIN TRIODE— SHARP-CUTOFF PENTODE



12CA

Duodecar type used in television receiver applications. The triode units are used in sync-separator and agc-amplifier circuits; the pentode unit is used in if-amplifier circuits. **Outlines section,** 8B; requires duodecar 12-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.77	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:**		
Triode Units:		
Grid to Plate	1.8	pF
Grid to Triode Cathode, Pentode Cathode, Heater, Pentode Grid No.3, and Internal Shield	3.4	pF
Plate to Triode Cathode, Pentode Cathode, Heater, Pentode Grid No.3, and Internal Shield	0.8	pF
Pentode:		
Grid No.1 to Plate	0.03	pF
Grid No.1 to Cathode, Grid No.2, Grid No.3, and Internal Shield	12	pF
Plate to Cathode, Grid No.2, Grid No.3, and Internal Shield	2.8	pF

** With external shield connected to pentode cathode, grid No.3, and internal shield.

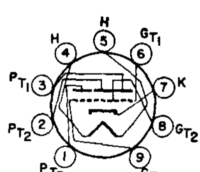
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Each Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.25	3.1	watts
Grid-No.2 Input:			
For voltages up to 165 volts	—	0.65	watt
For voltages between 165 and 330 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Supply Voltage	125	125	volts
Grid-No.2 Supply Voltage	—	125	volts
Cathode-Bias Resistor	125	56	ohms
Amplification Factor	58	—	
Plate Resistance (Approx.)	7250	200000	ohms
Transconductance	8000	13000	µmhos
Plate Current	8	11	mA
Grid-No.2 Current	—	3.4	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—	—3.5	volts
Grid-No.1 Voltage (Approx.) for plate current of 50 µA	—4.5	—	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance, for cathode-bias operation	0.68	1	megohm

6MD8

12MD8

MEDIUM-MU TRIPLE TRIODE



9RQ

Novar type used in matrixing circuits of color and black-and-white television receivers. **Outlines section,** 11E; requires novar 9-contact socket. Type 12MD8 is identical with type 6MD8 except for heater ratings.

	6MD8	12MD8	
Heater Arrangement	Parallel	Series	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	0.9	0.45	ampere
Heater Warm-up Time (Average)	—	11	seconds

Heater-Cathode Voltage:				
Peak value		±200 max	±200 max	volts
Average value		100 max	100 max	volts
	Unit No.1	Unit No.2	Unit No.3	
Direct Interelectrode Capacitances (Approx.):				
Grid to Plate	3	3	3	pF
Grid to Cathode and Heater	3.6	3.6	3.4	pF
Plate to Cathode and Heater	0.48	0.48	0.36	pF

Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

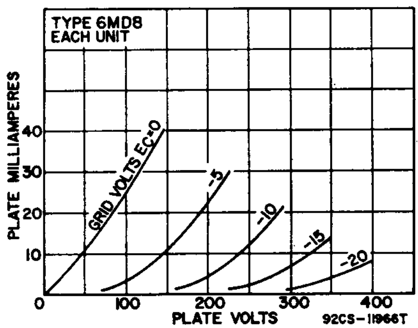
Plate Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation	3	watts

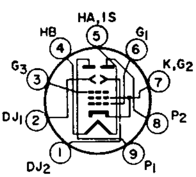
CHARACTERISTICS

Plate Voltage	250	volts
Grid Voltage	—10.5	volts
Amplification Factor	17	
Plate Resistance (Approx.)	5500	ohms
Transconductance	3100	μmhos
Plate Current	11.5	mA
Plate Current for grid voltage of —14 volts	4	mA
Grid Voltage (Approx.) for plate current of 50 μA	—23	volts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance, for fixed-bias operation	1	megohm





9RU

TWO-PLATE BEAM-DEFLECTION TUBE 6ME8

Miniature type used for color-demodulator applications in color television receivers and a variety of other switching and gate applications. **Outlines section, 6E;** requires miniature 9-contact socket. Pin 5 should be connected directly to ground. The 6ME8 should be so located in the equipment that it is not subjected to stray magnetic fields.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.3	ampere
Direct Interelectrode Capacitances:		
Grid No.1 to All Other Electrodes Except Plates	7.5	pF
Either Plate to All Other Electrodes	6	pF
Either Deflecting Electrode to All Other Electrodes	6	pF
Plate No.1 to Plate No.2	0.4	pF
Deflecting Electrode No.1 to Deflecting Electrode No.2	0.4	pF
Grid No.1 to Deflecting Electrode No.1	0.07 max	pF
Grid No.1 to Deflecting Electrode No.2	0.1 max	pF

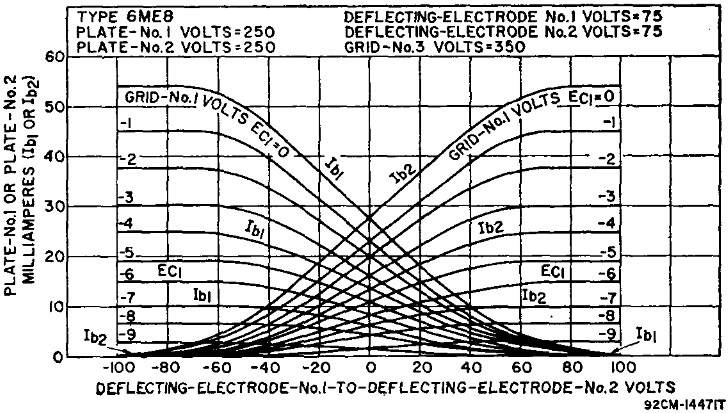
Color TV Demodulator

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage (Each Plate)	400	volts
Peak Deflecting-Electrode Voltage (Each Electrode)	±200	volts
Deflecting-Electrode Voltage (Each Electrode)	100	volts
Grid-No.3 (Accelerating-Grid) Voltage	400	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Cathode Current	30	mA
Plate Dissipation (Each Plate)	2	watts
Grid-No.3 Input	2	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.25	megohm

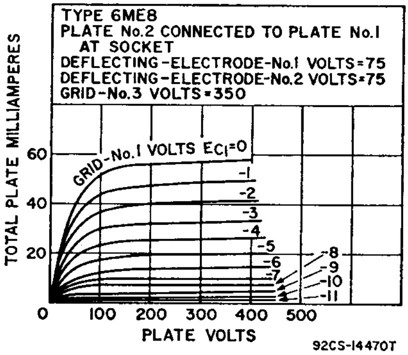


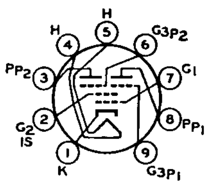
Class A₁ Amplifier

CHARACTERISTICS

Plate-No.2 Supply Voltage	250	volts
Plate No.2	Connected to Plate No.1	
Plate-No.1 Supply Voltage	250	volts
Grid-No.3 Supply Voltage	350	volts
Grid-No.1 Supply Voltage	0	volts
Deflecting-Electrode-No.2 Supply Voltage	75	volts
Deflecting-Electrode-No.1 Supply Voltage	75	volts
Cathode-Bias Resistor	390	ohms
Transconductance, Grid No.1 to both plates	4400	μmhos
Total Plate Current	14.5	mA
Grid-No.3 Current	0.7	mA
Grid-No.1 Voltage for total plate current of 10 μA	—16	volts
Deflecting-Electrode Switching Voltage*	30 max	volts
Voltage Difference between Deflecting Electrodes for equal plate currents	0	volts
Plate-No.1 Current with Deflecting-Electrode-No.1 Voltage = 55V and Deflecting-Electrode-No.2 Voltage = 95V	1.3 max	mA
Plate-No.2 Current with Deflecting-Electrode-No.1 Voltage = 95V and Deflecting-Electrode-No.2 Voltage = 55V	1.3 max	mA
Deflecting-Electrode-No.1 Current with Deflecting-Electrode-No.1 Voltage = 125V and Deflecting-Electrode-No.2 Voltage = 25V	0.04 max	mA
Deflecting-Electrode-No.2 Current with Deflecting-Electrode-No.1 Voltage = 25V and Deflecting-Electrode-No.2 Voltage = 125V	0.04 max	mA

* Defined as the total voltage change from 75 volts on either deflecting electrode with an equal and opposite change on the other deflecting electrode required to switch the plate current from one plate to the other.





9FG

SHARP-CUTOFF TWIN PENTODE — 6MK8

Miniature type used in sync-separator, clipper, agc, and low-level color-demodulator circuits in television receivers. Outlines section, 6E; requires miniature 9-contact socket.

Heater Voltage	6.3	volts
Heater Current	0.3	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid No.3 to Plate (Each Section)	2	pF
Grid No.1 to All Electrodes	6	pF
Grid No.3 (Each Section) to All Electrodes	3.6	pF
Plate (Each Section) to All Electrodes	3	pF
Grid No.3 (Section 1) to Grid No.3 (Section 2)	0.015 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage (Each Unit)	300	volts
Grid-No.3 (Suppressor-Grid) Voltage (Each Unit)		
Peak positive value	50	volts
DC negative value	50	volts
DC positive value	3	volts
Grid-No.2 (Screen-Grid) Voltage	150	volts
Grid-No.1 (Control-Grid) Voltage, Negative-bias value	50	volts
Cathode Current	12	mA
Plate Dissipation (Each Section)	1.1	watts
Grid-No.2 Input	0.75	watt

MAXIMUM PLATE CURRENT RATIO (Balance)—1.3 to 1

Plate Voltage	100	volts
Grid-No.2 Voltage	67.5	volts
Grid-No.1 Voltage	67.5	volts
Grid-No.3 Voltage	0	volts
Grid-No.1 Resistance	0.68	megohm

CHARACTERISTICS **With One Unit Operating▪**

Plate Voltage	100	100	volts
Grid-No.3 Voltage	0	0	volts
Grid-No.2 Voltage	67.5	67.5	volts
Grid-No.1 Voltage	0	*	volts
Transconductance, Grid No.3 to Plate	—	450	μmhos
Transconductance, Grid No.1 to Plate	1100	—	μmhos

Plate Current	—	2	mA
Grid-No.3 Voltage (Approx.) for plate current of 100 μA	—	—3.5	volts
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—2.3	volts
With Both Units Operating			
Plate Voltage (Each Unit)	100	100	volts
Grid-No.3 Voltage (Each Unit)	—10	0	volts
Grid-No.2 Voltage	67.5	67.5	volts
Grid-No.1 Voltage	*	*	
Plate Current (Each Section)	—	2	mA
Cathode Current	7.1	8.5	mA
Grid-No.2 Current	7	4.4	mA

MAXIMUM CIRCUIT VALUES

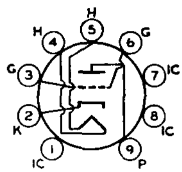
Grid-No.3-Circuit Resistance (Each Unit)	0.5	megohm
Grid-No.1-Circuit Resistance	0.5	megohm

▪ With plate and grid No.3 of other unit grounded.

* Grid current adjusted for 100 μA dc.

6N6G	Refer to chart at end of section.
6N7 **6N7GT**	Refer to chart at end of section.
6P5GT	Refer to chart at end of section.
6P7G	Refer to chart at end of section.
6Q7 **6Q7G** **6Q7GT**	Refer to chart at end of section.
6Q11	Refer to chart at end of section.
6R7 **6R7G** **6R7GT**	Refer to chart at end of section.
6S4	Refer to chart at end of section.

6S4A MEDIUM-MU TRIODE



9AC

Miniature type used as vertical-deflection amplifier in color and black-and-white television receivers. Outlines section, 6E; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.6	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid to Plate	2.4	pF
Grid to Cathode and Heater	4.2	pF
Plate to Cathode and Heater	0.6	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	250	volts

Grid Voltage	—8	volts
Amplification Factor	16.5	
Plate Resistance (Approx.)	3700	ohms
Transconductance	4500	μmhos
Plate Current	24	mA
Plate Current for grid voltage of —15 volts	4	mA
Grid Voltage (Approx.) for plate current of 50 μA	—22	volts

Vertical-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Voltage	550	volts
Peak Positive-Pulse Plate Voltage#	2200	volts
Peak Negative-Pulse Grid Voltage	250	volts
Peak Cathode Current	105	mA
Average Cathode Current	30	mA
Plate Dissipation	8.5	watts
MAXIMUM CIRCUIT VALUE		
Grid-Circuit Resistance, for cathode-bias operation	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

Refer to chart at end of section. **6S7 6S7G**

Refer to chart at end of section. **6S8GT**

Refer to chart at end of section. **6SA7 6SA7GT**

Refer to chart at end of section. **6SB7Y**

Refer to chart at end of section. **6SC7**

Refer to chart at end of section. **6SF5 6SF5GT**

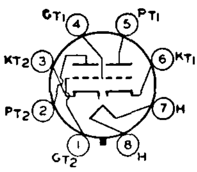
Refer to chart at end of section. **6SF7**

Refer to chart at end of section. **6SG7**

Refer to chart at end of section. **6SH7**

Refer to chart at end of section. **6SJ7 6SJ7GT**

Refer to chart at end of section. **6SK7 6SK7GT**



8BD

6SL7GT
HIGH-MU TWIN TRIODE
12SL7GT

Glass octal type used as phase inverter in radio equipment. Each unit may also be used in resistance-coupled amplifier circuits. **Outlines section,** 13D; requires octal socket. Except for the common heater, each triode unit is independent of the other. For typical operation as phase inverter or resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section. Type 12SL7GT is identical with type 6SL7GT except for heater ratings.

	6SL7GT	12SL7GT	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	0.3	0.15	ampere
Peak Heater-Cathode Voltage	±90 max	±90 max	volts

Direct Interelectrode Capacitances (Approx.):°	Unit No.1	Unit No.2	
Grid to Plate	2.8	2.8	pF
Grid to Cathode and Heater	3	3.4	pF
Plate to Cathode and Heater	3.8	3.2	pF

° With external shield connected to cathode.

Class A_1 Amplifier

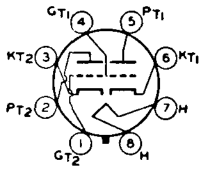
MAXIMUM RATINGS (Design-Center Values)		
Plate Voltage	300	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation	1	watt
CHARACTERISTICS		
Plate Voltage	250	volts
Grid Voltage	—2	volts
Amplification Factor	70	
Plate Resistance (Approx.)	44000	ohms
Transconductance	1600	µmhos
Plate Current	2.3	mA

6SN7GT
6SN7GTA

Refer to chart at end of section.

6SN7GTB MEDIUM-MU TWIN TRIODE

12SN7GTA



8BD

Glass octal type used as combined vertical oscillator and vertical-deflection amplifier, and as horizontal-deflection oscillator, in color and black-and-white television receivers. Each unit may also be used in multivibrator or resistance-coupled amplifier circuits in radio equipment. **Outlines section,** 13D; requires octal socket. Except for the common heater, each triode unit is independent of the other. For typical operation as resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section. Type 12SN7GTA is identical with type 6SN7GTB except for heater ratings.

	6SN7GTB	12SN7GTA	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	0.6	0.3	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	4.0	3.8	pF
Grid to Cathode and Heater	2.2	2.6	pF
Plate to Cathode and Heater	0.7	0.7	pF

Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Center Values)		
Plate Voltage	450	volts
Cathode Current	20	mA
Plate Dissipation:		
For either plate	5	watts
For both plates with both units operating	7.5	watts

CHARACTERISTICS			
Plate Voltage	90	250	volts
Grid Voltage	0	—8	volts
Amplification Factor	20	20	
Plate Resistance (Approx.)	6700	7700	ohms
Transconductance	3000	2600	µmhos
Plate Current	10	9	mA
Plate Current for grid voltage of —12.5 volts	—	1.3	mA
Grid Voltage (Approx.) for plate current of 10 µA	—7	—18	volts

MAXIMUM CIRCUIT VALUE		
Grid-Circuit Resistance, for fixed-bias operation	1	megohm

Oscillator (Each Unit)

For operation in a 525-line, 30-frame system

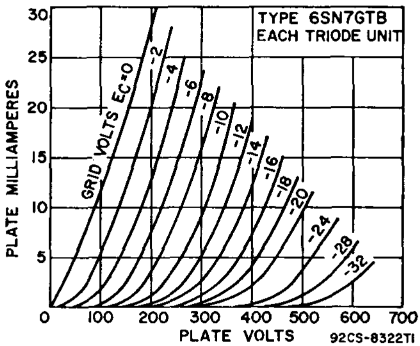
MAXIMUM RATINGS (Design-Center Values)	Vertical-Deflection Oscillator	Horizontal-Deflection Oscillator	
DC Plate Voltage	450	450	volts
Peak Negative-Pulse Grid Voltage	400	600	volts
Peak Cathode Current	70	300	mA
Average Cathode Current	20	20	mA
Plate Dissipation:			
For either plate	5	5	watts
For both plates with both units operating	7.5	7.5	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance	2.2	2.2	megohms

Vertical Deflection Amplifier (Each Unit)

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)		
DC Plate Voltage	450	volts
Peak Positive-Pulse Plate Voltage# (Absolute maximum)	1500▪	volts
Peak Negative-Pulse Grid Voltage	250	volts
Peak Cathode Current	70	mA
Average Cathode Current	20	mA
Plate Dissipation:		
For either plate	5	watts
For both plates with both units operating	7.5	watts
MAXIMUM CIRCUIT VALUE		
Grid-Circuit Resistance, for cathode-bias operation	2.2	megohms

Pulse duration must not exceed 15% of a vertical cycle (2.5 milliseconds).
▪ Under no circumstances should this absolute value be exceeded.



Refer to chart at end of section. **6SQ7 6SQ7GT**

Refer to chart at end of section. **6SR7**

Refer to chart at end of section. **6SS7**

Refer to chart at end of section. **6ST7**

Refer to chart at end of section. **6SZ7**

Refer to chart at end of section. **6T4**

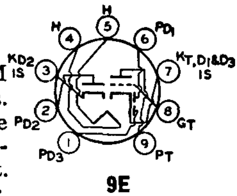
Refer to chart at end of section. **6T7G**

Refer to chart at end of section. **6T8**

6T8A

5T8

TRIPLE DIODE— HIGH-MU TRIODE



9E

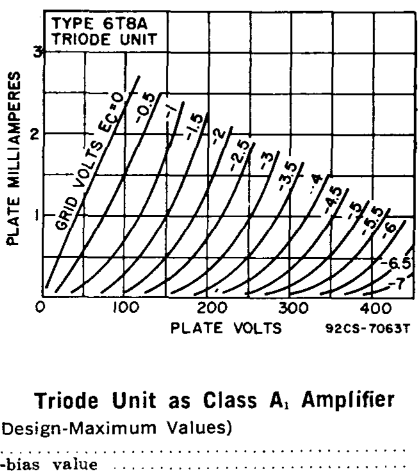
Miniature type used as combined audio amplifier, AM detector, and FM detector in AM/FM radio receivers. Diode unit No.1 is used for AM detection, and diode units No.2 and No.3 are used for FM detection. **Outlines section**, 6B; requires miniature 9-contact socket. For typical operation as resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section. Type 5T8 is identical with type 6T8A except for heater ratings.

	5T8	6T8A	
Heater Voltage (ac/dc)	4.7	6.3	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±100 max	volts
Average value	100 max	—	volts
Direct Interelectrode Capacitances:	Unshielded	Shielded	
Triode Unit:			
Grid to Plate	1.7	1.7	pF
Grid to Cathode, Internal Shield (pin 7), and Heater	1.6	1.7	pF
Plate to Cathode, Internal Shield (pin 7), and Heater	1.2	2.4	pF
Diode Units:			
Diode-No.1 Plate to Cathode, Internal Shield (pin 7), and Heater	3.8	3.8	pF
Diode-No.2 Plate to Cathode, Internal Shield (pin 3), and Heater	3.8	3.8•	pF
Diode-No.3 Plate to Cathode, Internal Shield (pin 7), and Heater	3.4	3.6	pF
Diode-No.2 Cathode, Internal Shield (pin 3) to All Other Electrodes, and Heater	7.5	8.5▪	pF
Triode Grid to any Diode Plate	0.034 max	0.034 max	pF

* With external shield connected to pin 7 except as noted.

• With external shield connected to pin 3.

▪ With external shield connected to pins 4 and 5.



Triode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation	1.1	watts

CHARACTERISTICS

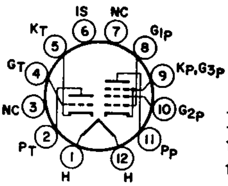
Plate Voltage	100	250	volts
Grid Voltage	—1	—3	volts
Amplification Factor	70	70	

Plate Resistance (Approx.)	54000	58000	ohms
Transconductance	1300	1200	µmhos
Plate Current	0.8	1	mA

Diode Units

MAXIMUM RATINGS (Design-Maximum Values)

Plate Current (Each Unit)	5.5	mA



12FM

HIGH-MU TRIODE—POWER PENTODE 6T9

Duodecar type used in audio-frequency circuits. The triode unit is used as a voltage amplifier; the pentode unit is used as a power amplifier. Outlines section, 8B; requires duodecar 12-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.93	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Pentode Section:		
Grid No.1 to Plate	0.2	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	11	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	11	pF
Triode Unit:		
Grid to Plate	2.6	pF
Grid to Cathode, Heater, and Internal Shield	3.4	pF
Plate to Cathode, Heater, and Internal Shield	1.1	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	275	volts
Grid-No.2 (Screen-Grid) Voltage	—	275	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	1.5	12	watts
Grid-No.2 Input	—	2	watts

CHARACTERISTICS (Triode Unit)

Plate Voltage	250	volts
Grid Voltage	—2	volts
Amplification Factor	95	
Plate Resistance (Approx.)	45000	ohms
Transconductance	2100	µmhos
Plate Current	1.5	mA

TYPICAL OPERATION (Pentode Unit)

Plate Voltage	250	volts
Grid-No.2 Voltage	250	volts
Grid-No.1 Voltage	—8	volts
Peak AF Grid-No.1 Voltage	8	volts
Zero-Signal Plate Current	35	mA
Maximum-Signal Plate Current	39	mA
Zero-Signal Grid-No.2 Current	2.5	mA
Maximum-Signal Grid-No.2 Current	7	mA
Plate Resistance (Approx.)	0.1	megohm
Transconductance	6500	µmhos
Load Resistance	5000	ohms
Total Harmonic Distortion (Approx.)	10	per cent
Maximum-Signal Power Output	4.2	watts

MAXIMUM CIRCUIT VALUES

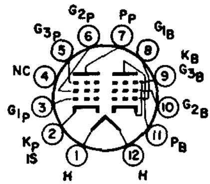
Grid-No.1-Circuit Resistance:	Triode Unit	Pentode Unit	
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1*	0.5	megohm

* For cathode-bias operation of the triode unit, a maximum resistance of 10 megohms can be used provided the plate dissipation never exceeds 0.25 watt.

6T10
12T10

BEAM POWER TUBE— SHARP-CUTOFF PENTODE

Duodecar type used as combined FM detector and audio-frequency output amplifier in color and black-and-white television receivers. The beam power unit is used in af output stages, and the sharp-cutoff, dual-control pentode unit is used as an FM detector. Outlines section, 8C; requires duodecar 12-contact socket. For maximum ratings and characteristics, refer to type 6AL11. Type 12T10 is identical with type 6T10 except for heater ratings.



12EZ

	6T10	12T10	
Heater Voltage (ac/dc)	6.3	12.6	volts
Heater Current	0.95	0.45	amperes
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Unit No.1:		
Grid No.1 to Plate	0.22	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	11	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	10	pF
Unit No.2:		
Grid No.1 to Plate	0.032	pF
Grid No.3 to Plate	3	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3 and Internal Shield	6.5	pF
Grid No.3 to Cathode, Heater, Grid No.1, Grid No.2, Plate, and Internal Shield	7.5	pF
Grid No.1 to Grid No.3	0.12	pF
Plate of Unit No.1 to Plate of Unit No.2	0.13	pF

6U5 Refer to chart at end of section.

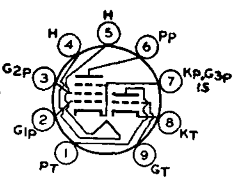
6U7G Refer to chart at end of section.

6U8 Refer to chart at end of section.

6U8A
6U8A/6KD8
5U8, 9U8A

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

Miniature types used as combined oscillator and mixer tube in color and black-and-white television receivers utilizing an intermediate frequency in the order of 40 MHz. Outlines section, 6B; require miniature 9-contact socket. Types 5U8 and 9U8A are identical with type 6U8A except for heater ratings.



9AE

	5U8	6U8A 6U8A/6KD8	9U8A	
Heater Voltage (ac/dc)	4.7	6.3	9.45	volts
Heater Current	0.6	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	11	11	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:	Unshielded	Shielded▲	
Triode Unit:			
Grid to Plate	1.8	1.8	pF
Grid to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	2.8	2.8	pF
Plate to Cathode, Heater, Pentode Cathode, Pentode Grid No.3, and Internal Shield	1.5	2	pF

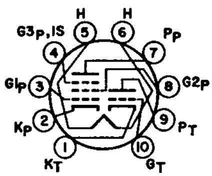
Pentode Unit:			
Grid No.1 to Plate	0.010 max	0.006 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5	5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.6	3.5	pF
Triode Cathode to Heater	3	3•	pF
Pentode Cathode, Pentode Grid No.3, and Internal Shield	3	3•	pF
Pentode Grid No.1 to Triode Plate	0.2 max	0.2 max	pF
Pentode Plate to Triode Plate	0.1 max	0.02 max	pF

▲ With external shield connected to pin 4 except as noted.

• With external shield connected to pin 6.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Voltage	125	125	volts
Grid-No2 Voltage	—	110	volts
Grid-No.1 Voltage	—1	—1	volts
Amplification Factor	40	—	
Plate Resistance (Approx.)	—	0.2	megohm
Transconductance	7500	5000	µmhos
Plate Current	13.5	9.5	mA
Grid-No.2 Current	—	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—9	—8	volts



10K

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

6U9/ ECF201

5U9/LCF201

Miniature type used in if-amplifier and sound and sync applications in television receivers. Outlines section, 6B, except has 10-pin base; requires miniature 10-contact socket. Type 5U9/LCF201 is identical with type 6U9/ECF201 except for heater ratings.

	5U9/LCF201	6U9/ECF201	
Heater Voltage	5.9	6.3	volts
Heater Current	0.45	0.41	ampere
Peak Heater-Cathode Voltage	±100 max	±150 max	volts

Direct Interelectrode Capacitance:		
Pentode Unit:		
Plate to All Other Elements (except grid No.1)	3.5	pF
Grid No.1 to All Other Elements (except plate)	6.5	pF
Grid No.1 to Cathode	4	pF
Plate to Grid No.1	<6.5	fF
Grid No.1 to Grid No.2	1.8	pF
Triode Unit:		
Plate to All Other Elements (except grid)	3	pF
Grid to All Other Elements (except plate)	2.5	pF
Plate to Grid	2	pF
Pentode Plate to Triode Plate	<15	fF
Pentode Grid No.1 to Triode Plate	<1.2	fF
Pentode Grid No.1 to Triode Grid	<1.5	fF

Class A_1 Amplifier

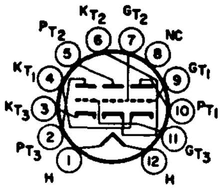
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	550	volts
Plate Voltage	250	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	550	volts
Grid-No.2 Voltage	—	250	volts
Peak Cathode Current•	50	—	mA

	Triode Unit	Pentode Unit	
Cathode Current	18	18	mA
Plate Dissipation	1.5	2.1	watts
Grid-No.2 Input	—	0.7	watt
CHARACTERISTICS			
Plate Voltage	100	160	volts
Grid-No.3 (Suppressor-Grid) Voltage	—	0	volts
Grid-No.2 Voltage	—	110	volts
Grid-No.1 (Control-Grid) Voltage	—2	—1.4	volts
Mu Factor, Grid No.1 to Grid No.2	—	45	
Amplification Factor	17	—	
Transconductance	5000	12000	µmhos
Plate Current	14	13	mA
Grid-No.2 Current	—	5	mA
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance	1	1	megohm

• With a maximum duty cycle of 0.10 and maximum pulse duration of 10 microseconds.

6U10 THREE-UNIT TRIODE

Duodecar type used in amplifier applications. Units No.1 and No.3 are medium-mu triode units, and unit No.2 is a high-mu triode unit. **Outlines section,** 8A; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 6.3; amperes, 0.6; warm-up time (average), 11 seconds); maximum heater-cathode volts, ±275 (peak) for units 1 and 3; ±200 (peak) for unit 2; 100 (average) for each unit.



12FE

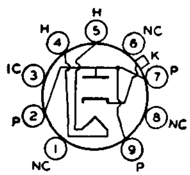
Class A₁ Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Units Nos. 1 and 3	Unit No.2	
Plate Voltage	330	330	volts
DC Grid Voltage:			
Positive-bias value	0	0	volts
Negative-bias value	50	50	volts
Average Cathode Current	20	—	mA
Plate Dissipation	2	1	watts
CHARACTERISTICS			
Plate Voltage	200	200	volts
Grid Voltage	—6	—1.5	volts
Amplification Factor	17.5	90	
Plate Resistance (Approx.)	7700	61000	ohms
Transconductance	2300	1600	µmhos
Plate Current	9.6	1.2	mA
Grid Voltage (Approx.):			
For plate current of 100 µA	—15	—	volts
For plate current of 35 µA	—	—3	volts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For fixed-bias operation	1	0.5	megohm
For cathode-bias operation	2.2	1*	megohms

* This value may reach 10 megohms provided the plate-supply voltage and load resistance are such that the plate dissipation can never exceed 0.5 watt.

6V3A HALF-WAVE VACUUM RECTIFIER

Miniature type used as a damper tube in horizontal-deflection circuits of television receivers. **Outlines section,** 7B; requires miniature 9-contact socket. This tube, like other power-handling tubes, should be adequately ventilated. **Heater:** volts (ac/dc), 6.3; amperes, 1.75.



9BD

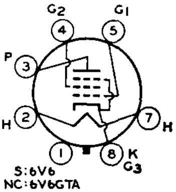
Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)

Peak Inverse Plate Voltage# (Absolute Maximum)		6000†	volts
Peak Plate Current		800	mA
Average Plate Current		135	mA
Heater-Cathode Voltage:			
Peak value	+300	−6750†	volts
Average value	+100	−750†	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
† Under no circumstances should this absolute value be exceeded.



7AC

6V6
6V6GTA

BEAM POWER TUBE

5V6GT, 12V6GT

Metal type 6V6 and glass octal type 6V6GTA are used as output amplifiers in automobile, battery-operated, and other receivers in which reduced plate-current drain is desirable. **Outlines section,** 2B and 13D, respectively; require octal socket. These tubes are equivalent in performance to type 6AQ5A. Refer to type 6AQ5A for average plate characteristic curves. Types 5V6GT and 12V6GT are identical with type 6V6GTA except for heater ratings.

	5V6GT	6V6	6V6GTA	12V6GT	
Heater Voltage (ac/dc)	4.7	6.3	6.3	12.6	volts
Heater Current	0.6	0.45	0.45	0.225	ampere
Heater Warm-up Time (Average)	11	—	11	—	seconds
Heater-Cathode Voltage:					
Peak value	±200 max	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	100 max	volts

	6V6°	6V6GTA	
Direct Interelectrode Capacitances (Approx.):			
Grid No.1 to Plate	0.3	0.7	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	10	9	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	11	7.5	pF

° With shell connected to cathode.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	350	volts
Grid-No.2 (Screen-Grid) Voltage	315	volts
Plate Dissipation	14	watts
Grid-No.2 Input	2.2	watts

TYPICAL OPERATION

Plate Voltage	180	250	315	volts
Grid-No.2 Voltage	180	250	225	volts
Grid-No.1 (Control-Grid) Voltage	−8.5	−12.5	−13	volts
Peak AF Grid-No.1 Voltage	8.5	12.5	13	volts
Zero-Signal Plate Current	29	45	34	mA
Maximum-Signal Plate Current	30	47	35	mA
Zero-Signal Grid-No.2 Current	3	4.5	2.2	mA
Maximum-Signal Grid-No.2 Current	4	7	6	mA
Plate Resistance (Approx.)	50000	50000	80000	ohms
Transconductance	3700	4100	3750	μmhos
Load Resistance	5500	5000	8500	ohms
Total Harmonic Distortion	8	8	12	per cent
Maximum-Signal Power Output	2	4.5	5.5	watts

CHARACTERISTICS (Triode Connection)▲

Plate Voltage	250	volts
Grid-No.1 (Control-Grid) Voltage	−12.5	volts
Amplification Factor	9.8	
Plate Resistance (Approx.)	1960	ohms
Transconductance	5000	μmhos
Plate Current	49.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 0.5 mA	−36	volts

▲ Grid No.2 connected to plate.

Push-Pull Class A_1 Amplifier

MAXIMUM RATINGS (Same as for class A_1 amplifier)

TYPICAL OPERATION (Values are for two tubes)

Plate Voltage	250	285	volts
Grid-No.2 Voltage	250	285	volts
Grid-No.1 (Control-Grid) Voltage	—15	—19	volts
Peak AF Grid-No.1-to-Grid-No.1 Voltage	30	38	volts
Zero-Signal Plate Current	70	70	mA
Maximum-Signal Plate Current	79	92	mA
Zero-Signal Grid-No.2 Current	5	4	mA
Maximum-Signal Grid-No.2 Current	13	13.5	mA
Effective Load Resistance (Plate-to-Plate)	10000	8000	ohms
Total Harmonic Distortion	5	3.5	per cent
Maximum-Signal Power Output	10	14	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

Vertical-Deflection Amplifier (Triode Connection)▲

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Voltage	350	volts
Peak Positive-Pulse Plate Voltage#	1200	volts
Peak Negative-Pulse Grid-No.1 (Control-Grid) Voltage	275	volts
Peak Cathode Current	115	mA
Average Cathode Current	40	mA
Plate Dissipation	10	watts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, for cathode-bias operation	2.2	megohms

▲ Grid No.2 connected to plate.

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

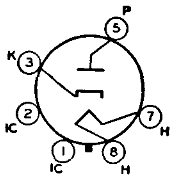
6V6GT Refer to chart at end of section.

6V7G Refer to chart at end of section.

6W4GT

HALF-WAVE VACUUM RECTIFIER

Glass octal type used as damper tube in television receivers. **Outlines section**, 13D; requires octal socket. This type may be supplied with pin No.1 omitted. Socket terminals 1, 2, 4, and 6 should not be used as tie points. This tube, like other power-handling tubes, should be adequately ventilated. Power-rectifier operation of this type is not recommended.



4CG

Heater Voltage (ac)	6.3	volts
Heater Current	1.2	amperes
Direct Interelectrode Capacitances (Approx.):		
Plate to Cathode and Heater	6	pF
Cathode to Plate and Heater	13	pF
Heater to Cathode	7	pF

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)

Peak Inverse Plate Voltage (*Absolute Maximum*)#		3850	volts
Peak Plate Current		750	mA
Average Plate Current		125	mA
Plate Dissipation		3.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—2300	volts
Average value	+100	—500	volts

CHARACTERISTICS, Instantaneous Value

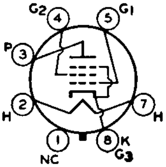
Tube Voltage Drop for plate current of 250 mA	21	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

6W6GT

12W6GT, 25W6GT

BEAM POWER TUBE



7AC

Glass octal type used in the audio output stage of radio and color and black-and-white television receivers. Triode-connected, it is used as a vertical-deflection amplifier in television receivers. Outlines section, 13D; requires octal socket. This type may be supplied with pin No.1 omitted. Types 12W6GT and 25W6GT are identical with type 6W6GT except for heater ratings.

	6W6GT	12W6GT	25W6GT	
Heater Voltage (ac/dc)	6.3	12.6	25	volts
Heater Current	1.2	0.6	0.3	ampere
Heater Warm-up Time (Average)	—	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	+200 max —300 max	±200 max	volts
Average value	100 max	+100 max —200 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.8	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	15	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	9	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.2 (Screen-Grid) Voltage	165	volts
Plate Dissipation	12	watts
Grid-No.2 Input	1.35	watts

TYPICAL OPERATION

Plate Supply Voltage	110	200	volts
Grid-No.2 Supply Voltage	110	125	volts
Grid-No.1 (Control-Grid) Voltage	—7.5	—	volts
Cathode-Bias Resistor	—	180	ohms
Peak AF Grid-No.1 Voltage	7.5	8.5	volts
Zero-Signal Plate Current	49	46	mA
Maximum-Signal Plate Current	50	47	mA
Zero-Signal Grid-No.2 Current	4	2.2	mA
Maximum-Signal Grid-No.2 Current	10	8.5	mA
Plate Resistance (Approx.)	13000	28000	ohms
Transconductance	8000	8000	μmhos
Load Resistance	2000	4000	ohms
Total Harmonic Distortion (Approx.)	10	10	per cent
Maximum-Signal Power Output	2.1	3.8	watts

CHARACTERISTICS (Triode Connection)*

Plate Voltage	225	volts
Grid-No.1 Voltage	—30	volts
Amplification Factor	6.2	
Plate Resistance (Approx.)	1600	ohms
Transconductance	3800	μmhos
Plate Current	22	mA
Grid No.1 Voltage (Approx.) for plate current of 0.5 mA	—42	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1 Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

* Grid No.2 connected to plate.

Vertical Deflection Amplifier

For operation in a 525-line, 30-frame system

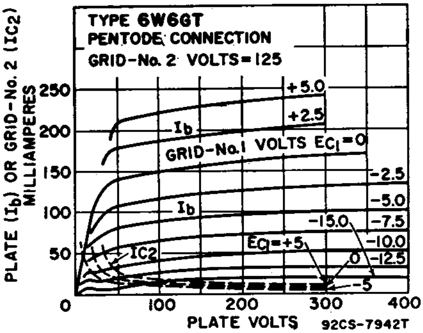
MAXIMUM RATINGS (Design-Maximum Values)	Triode Connection*	Pentode Connection	
DC Plate Voltage	330	330	volts
Peak Positive-Pulse Plate Voltage#	1200	1500	volts
DC Grid No.2 (Screen-Grid) Voltage	—	165	volts
Peak Negative-Pulse Grid-No.1 Voltage	275	275	volts
Peak Cathode Current	195	195	mA
Average Cathode Current	65	65	mA
Plate Dissipation	8.5	8	watts
Grid-No.2 Input	—	1.2	watts

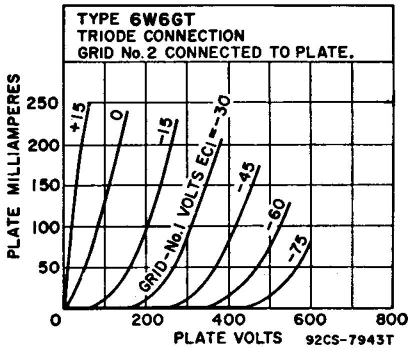
MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, for cathode-bias operation	2.2	2.2	megohms

* Grid No.2 connected to plate.

\# Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

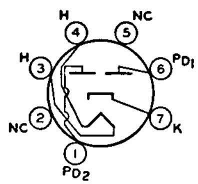




6W7G Refer to chart at end of section.

6X4
12X4

FULL-WAVE VACUUM RECTIFIER



5BS

Miniature type used in power supply of automobile and ac-operated radio receivers. Equivalent in performance to larger type 6X5GT. **Outlines section,** 5D; requires miniature 7-contact socket. This tube, like other power-handling tubes, should be adequately ventilated. For discussion of Rating Chart and Operation Characteristics, refer to **Interpretation of Tube Data.** Type 12X4 is identical with type 6X4 except for heater ratings.

	6X4	12X4	
Heater Voltage (ac/dc)	6.3▲	12.6	volts
Heater Current	0.6	0.3	ampere
Heater-Cathode Voltage:			
Peak value	+200, —450 max		volts
Average value	100 max		volts

▲ When the heater is operated from a 3-cell (nominal-6-volt) storage-battery source, the permissible heater-voltage range is from 5 to 8 volts.

Full-Wave Rectifier

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage	1250	volts
Steady-State Peak Plate Current (Per Plate)	245	mA
AC Plate Supply Voltage (Per Plate, rms)	See Rating Chart	
DC Output Voltage (At filter input)†	350	volts
Average Output Current (Each plate)†	45	mA
Hot-Switching Transient Plate Current	#	

† This rating applies when the 6X4 is used in vibrator operation with a minimum duty cycle of 75 per cent.

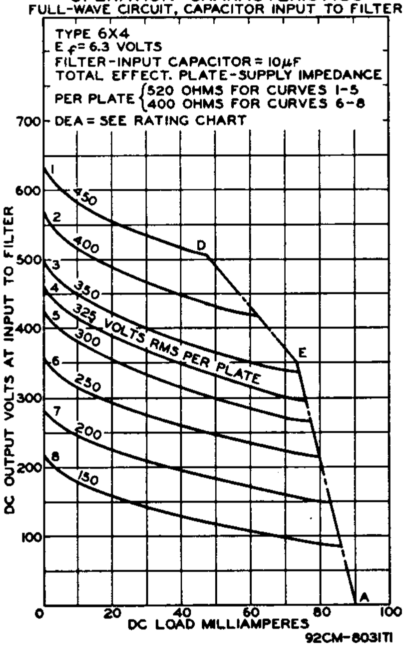
\# If hot-switching is regularly required in operation, the use of choke-input circuits is recommended. Such circuits limit the hot-switching current to a value no higher than that of the peak plate current. When capacitor-input circuits are used, a maximum peak current value per plate of 1.1 amperes during the initial cycles of the hot-switching transient should not be exceeded.

TYPICAL OPERATION

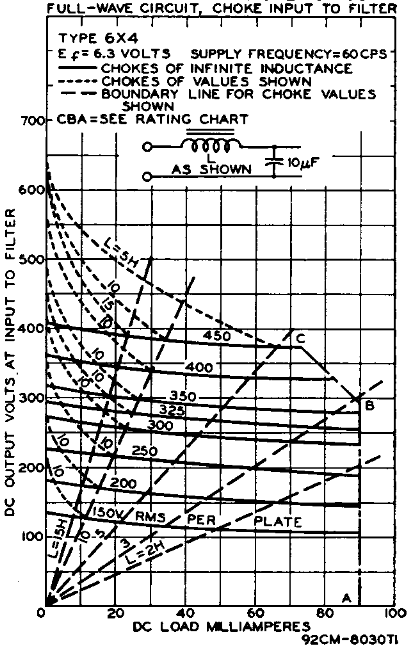
Filter Input	Sine Wave Operation Capacitor	Sine Wave Operation Choke	Vibrator Operation Capacitor	
AC Plate Supply Voltage (Each plate, rms)• ..	325	400	—	volts
Filter Input Capacitor	10	—	10	μF
Effective Plate Supply Impedance (Each plate) ..	525	—	—	ohms
Filter Input Choke	—	10	—	henries
Average Output Current	70	70	70	mA
DC Output Voltage at Input to Filter (Approx.)	310	340	240	volts

• AC plate supply voltage is measured without load.

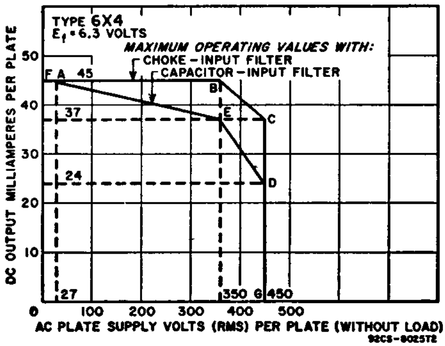
OPERATION CHARACTERISTICS
FULL-WAVE CIRCUIT, CAPACITOR INPUT TO FILTER



OPERATION CHARACTERISTICS
FULL-WAVE CIRCUIT, CHOKE INPUT TO FILTER



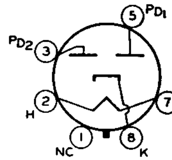
RATING CHART



Refer to chart at end of section. **6X5**

6X5GT

FULL-WAVE VACUUM RECTIFIER



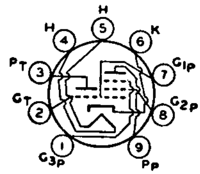
6S

Glass octal type used in power supply of automobile and ac-operated receivers. **Outlines section, 13D;** requires octal socket. This type may be supplied with pin No.1 omitted. For maximum ratings, and typical operation, refer to type 6X4.

6X8A

5X8, 19X8

MEDIUM-MU TRIODE—SHARP-CUTOFF PENTODE



9AK

Miniature type used as combined oscillator and mixer tube in television receivers utilizing an intermediate frequency in the order of 40 MHz and in AM/FM receivers. **Outlines section, 6B;** requires miniature 9-contact socket. Types 5X8 and 19X8 are identical with type 6X8A except for heater ratings.

	5X8	6X8A	19X8	
Heater Voltage (ac/dc)	4.7	6.3	18.4	volts
Heater Current	0.6	0.45	0.15	ampere
Heater Warm-up Time (Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances:	Unshielded	Shielded▲	
Triode Unit:			
Grid to Plate	1.5	1.5	pF
Grid to Cathode and Heater	2	2.4	pF
Plate to Cathode and Heater	0.5	1	pF
Pentode Unit:			
Grid No.1 to Plate	0.09 max	0.06 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	4.6	4.8	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	0.9	1.6	pF
Pentode Grid No.1 to Triode Plate	0.05 max	0.04 max	pF
Pentode Plate to Triode Plate	0.05 max	0.008 max	pF
Heater to Cathode	6.5	6.5•	pF

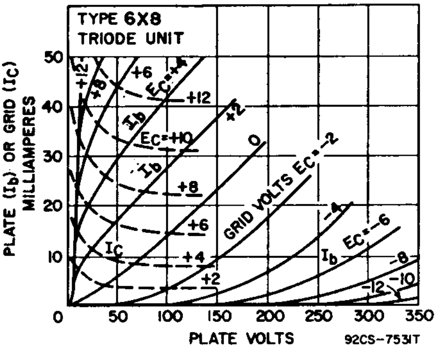
▲ With external shield connected to cathode except as noted.

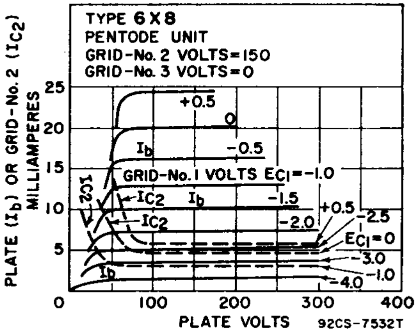
• Wilth external shield connected to pentode plate.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	275	275	volts
Grid No.2 (Screen-Grid) Supply Voltage	—	275	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	1.7	2.3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 137.5 volts	—	0.45	watt
For grid-No.2 voltages between 137.5 and 275 volts	—	See curve page 98	

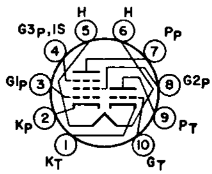
CHARACTERISTICS			
Plate Voltage	125	125	volts
Grid No.3	Connected to cathode at socket		
Grid-No.2 Voltage	—	125	volts
Grid-No.1 Voltage	−1	−1	volt
Amplification Factor	40	—	
Plate Resistance (Approx.)	6000	300000	ohms
Transconductance	6500	5500	μmhos
Plate Current	12	9	mA
Grid-No.2 Current	—	2.2	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	−7	−6.5	volts





Refer to chart at end of section.

6X8



10K

6X9/ ECF200

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE

Miniature type used as if-amplifier tube in television receivers. **Outlines section 6B**, except has 10-pin base; requires miniature 10-contact socket.

Heater Voltage	6.3	volts
Heater Current	0.41	ampere
Peak Heater-Cathode Voltage	±150 max	volts
Direct Interelectrode Capacitances:		
Triode Unit:		
Plate to All Other Elements (except grid)	3	pF
Grid to All Other Elements (except plate)	2.5	pF
Plate to Grid	2	pF
Pentode Unit:		
Plate to All Other Elements (except grid No.1)	3.5	pF
Grid No.1 to All Other Elements (except plate)	6.5	pF
Grid No.1 to Cathode	4	pF
Plate to Grid No.1	<6.5	fF
Grid No.1 to Grid No.2	1.8	pF
Pentode Grid No.1 to Triode Plate	15	fF
Pentode Grid No.1 to Triode Grid	<1.2	fF
Pentode Plate to Triode Plate	<1.5	fF

Class A_1 Amplifier

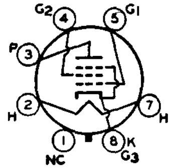
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	550	volts
Plate Voltage	250	250	volts
Peak Plate Voltage•	600	—	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	550	volts
Grid-No.2 Voltage	—	250	volts
Cathode Current	18	18	mA
Plate Dissipation	1.5	2.1	watts
Grid-No.2 Input	—	0.7	watt
CHARACTERISTICS			
Plate Voltage	170	160	volts
Grid-No.3 (Suppressor-Grid) Voltage	—	0	volts
Grid-No.2 Voltage	—	135	volts
Grid-No.1 (Control-Grid) Voltage	—1	—1.7	volts
Mu Factor, Grid-No.1 to Grid-No.2	—	55	
Amplification Factor	55	—	
Transconductance	4800	14000	μmhos
Plate Current	8.5	13	mA
Grid-No.2 Current	—	5	mA
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance	1	1	megohm

• With a maximum duty factor of 0.18 and maximum pulse duration of 18 microseconds.

6Y5 Refer to chart at end of section.

6Y6GA/ 6Y6G

BEAM POWER TUBE



7AC

Glass octal type used as output amplifier in radio receivers and in rf-operated, high-voltage power supplies in television equipment. Outlines section, 19B; requires octal socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	1.25	amperes
Peak Heater-Cathode Voltage	±180 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.7	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	12	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	7.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	200	volts
Grid-No.2 (Screen-Grid) Supply Voltage	200	volts
Grid-No.2 Voltage	See curve page 98	
Plate Dissipation	12.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 100 volts	1.75	watts
For grid-No.2 voltages between 100 and 200 volts	See curve page 98	

TYPICAL OPERATION

Plate Voltage	135	200	volts
Grid-No.2 Voltage	135	135	volts
Grid-No.1 (Control-Grid) Voltage	—13.5	—14	volts
Peak AF Grid-No.1 Voltage	13.5	14	volts
Zero-Signal Plate Current	58	61	mA
Maximum-Signal Plate Current	60	66	mA
Zero-Signal Grid-No.2 Current	3.5	2.2	mA
Maximum-Signal Grid-No.2 Current	11.5	9	mA
Plate Resistance (Approx.)	9300	18300	ohms
Transconductance	7000	7100	µmhos
Load Resistance	2000	2600	ohms
Total Harmonic Distortion	10	10	per cent
Maximum-Signal Power Output	3.6	6	watts

MAXIMUM CIRCUIT VALUES

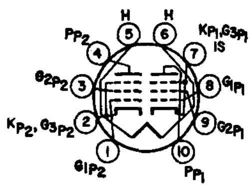
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

6Y7G Refer to chart at end of section.

6Y9

11Y9/LFL200

DUAL PENTODE



10L

Miniature type for use in color and black-and-white television receiver applications. Unit No. 1 is used as a video output pentode, and unit No. 2 as a sound if amplifier, agc amplifier, or sync separator. Outlines section, 6L, except has 10-pin base; requires miniature 10-contact socket. Type 11Y9/LFL200 is identical with type 6Y9 except for heater ratings.

	6Y9	11Y9/LFL200	
Heater Voltage	6.3	11	volts
Heater Current	0.8	0.45	ampere
Peak Heater-Cathode Voltage	±200 max	±200 max	volts
Direct Interelectrode Capacitances:			
Unit No.1:			
Plate to All Other Elements (except grid No.1)		7	pF
Grid No.1 to All Other Elements (except plate)		12	pF
Plate to Grid No.1		95	fF

Unit No.2:		
Plate to All Other Elements (except grid No.1)	11	pF
Grid No.1 to All Other Elements (except plate)	10	pF
Plate to Grid No.1	140	fF
Grid No.1 to Heater	<100	fF
Plate to Plate	<150	fF
Grid to Grid	<10	fF
Plate (Unit No.1) to Grid No.1 (Unit No.2)	<100	fF
Plate (Unit No.2) to Grid No.1 (Unit No.2)	<5	fF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1	Unit No.2	
Plate Supply Voltage	550	550	volts
Plate Voltage	250	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	550	550	volts
Grid-No.2 Voltage	250	250	volts
Cathode Current	60	15	mA
Plate Dissipation	5	1.5	watts
Grid-No.2 Input	2.5	0.5	watts
CHARACTERISTICS			
Plate Voltage	170	150	volts
Grid-No.2 Voltage	170	150	volts
Grid-No.1 (Control-Grid) Voltage	—2.6	—2.3	volts
Mu Factor, Grid-No.1 to Grid-No.2	38	35	
Internal Resistance	40	160	kohms
Transconductance	21000	8500	µmhos
Plate Current	30	10	mA
Grid-No.2 Current	6.5	3	mA
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance	1	1	megohm

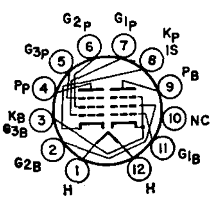
Refer to chart at end of section. **6Z5**

Refer to chart at end of section. **6Z7G**

6Z10
6Z10/6J10

PENTODE— BEAM POWER TUBE

13Z10, 13Z10/13J10



12BT

Duodecar types used as a combined limiter, discriminator, and audio power-output tube in FM radio and television receivers. **Outlines section,** 8C; require duodecar 12-contact socket. Types 13Z10 and 13Z10/13J10 are identical with type 6Z10 except for heater ratings.

	6Z10 6Z10/6J10	13Z10 13Z10/13J10	
Heater Voltage (ac/dc)	6.3	13.2	volts
Heater Current	0.95	0.45	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances:		
Pentode Unit:		
Grid No.1 to Grid No.3	0.009	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, Plate, and Internal Shield	4.4	pF
Grid No.3 to Cathode, Heater, Grid No.1, Grid No.2, Plate, and Internal Shield	3.2	pF
Beam Power Unit:		
Grid No.1 to Plate	0.22	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	11	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	7.5	pF

Pentode Unit As Class A_1 Amplifier

CHARACTERISTICS				
Plate Voltage	135	135	135	volts
Grid-No.3 (Suppressor-Grid) Voltage	4	4	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	280	280	volts

Grid-No.2 Voltage	75	—	—	volts
Grid No. 1 (Control-Grid) Voltage	0	0	0	volts
Grid-No.2 Resistor	—	33	33	kohms
Transconductance, Grid No.1 to Plate	—	—	360	μmhos
Transconductance, Grid No.3 to Plate	—	—	700	μmhos
Average Plate Current	—	5	—	mA
Grid-No.2 Current	4.5	—	—	mA
Grid No.1 Voltage (Approx.) for plate current of 20 μA	—	—	—4	volts
Grid No.3 Voltage (Approx.) for plate current of 20 μA	—	—	—4	volts

Beam Power Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	275	volts
Grid-No.2 (Screen-Grid) Voltage	275	volts
Plate Dissipation	10	watts
Grid-No.2 Input	2	watts

TYPICAL OPERATION

Plate Voltage	250	volts
Grid-No.2 Voltage	250	volts
Grid-No.1 (Control-Grid) Voltage	—8	volts
Peak AF Grid-No.1 Voltage	8	volts
Zero-Signal Plate Current	35	mA
Maximum-Signal Plate Current	39	mA
Zero-Signal Grid-No.2 Current	3	mA
Maximum-Signal Grid-No.2 Current	13	mA
Plate Resistance (Approx.)	0.1	megohm
Transconductance	6500	μmhos
Load Resistance	5000	ohms
Total Harmonic Distortion (Approx.)	8.5	per cent
Maximum-Signal Power Output	4.2	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	0.5	megohm

Pentode Unit as Limiter and Discriminator

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	330	volts
Grid-No.2 Voltage	330	volts
Grid-No.1 Voltage, Peak positive value	60	volts
Average Cathode Current	13	mA

6ZY5G Refer to chart at end of section.

7A4 Refer to chart at end of section.

7A5 Refer to chart at end of section.

7A6 Refer to chart at end of section.

7A7 Refer to chart at end of section.

7A8 Refer to chart at end of section.

7AD7 Refer to chart at end of section.

7AF7 Refer to chart at end of section.

7AG7 Refer to chart at end of section.

7AH7 Refer to chart at end of section.

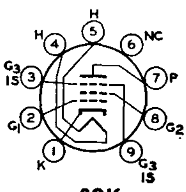
7AU7 Refer to type 12AU7A.

7B4 Refer to chart at end of section.

7B5 Refer to chart at end of section.

7B6 Refer to chart at end of section.

Refer to chart at end of section.	**7B7**
Refer to chart at end of section.	**7B8**
Refer to chart at end of section.	**7C5**
Refer to chart at end of section.	**7C6**
Refer to chart at end of section.	**7C7**
Refer to chart at end of section.	**7E6**
Refer to chart at end of section.	**7E7**
Refer to chart at end of section.	**7EY6**
Refer to chart at end of section.	**7F7**
Refer to chart at end of section.	**7F8**
Refer to chart at end of section.	**7G7**
Refer to chart at end of section.	**7H7**
Refer to chart at end of section.	**7HG8**
Refer to type 6HG8/ECF86.	**7HG8/PCF86**
Refer to chart at end of section.	**7J7**
Refer to chart at end of section.	**7K7**
Refer to type 6KY6.	**7KY6**



9GK

SHARP-CUTOFF PENTODE 7KZ6

Miniature type with frame grid used for video-output-amplifier service in color and black-and-white television receivers. Outlines section, 6E; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	7.3	volts
Heater Current	0.45	ampere
Heater Warm-up Time	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.16 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	13	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	6	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

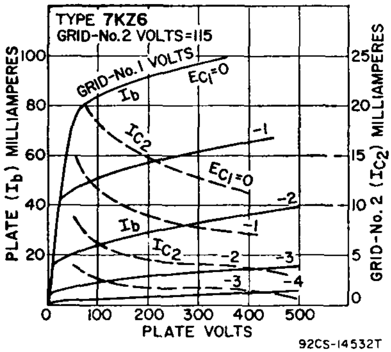
Plate Voltage	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	9	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	1	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

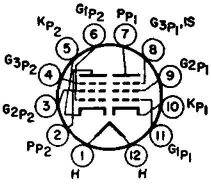
Plate Supply Voltage	250	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket	
Grid-No.2 Supply Voltage	115	volts
Grid-No.1 Supply Voltage	0	volts
Cathode-Bias Resistor	75	ohms
Plate Resistance (Approx.)	45000	ohms
Transconductance	24000	µmhos
Average Plate Current	25	mA
Grid-No.2 Current	3.6	mA
Grid-No.1 Voltage for plate current of 100 µA	—5.2	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.25	megohm



7L7	Refer to chart at end of section.
7N7	Refer to chart at end of section.
7Q7	Refer to chart at end of section.
7R7	Refer to chart at end of section.
7S7	Refer to chart at end of section.
7V7	Refer to chart at end of section.
7W7	Refer to chart at end of section.
7X7	Refer to chart at end of section.
7Y4	Refer to chart at end of section.
7Z4	Refer to chart at end of section.
8AR11	Refer to type 6AR11.
8AU8	Refer to type 6AU8A.
8AW8A	Refer to type 6AW8A.
8B10	Refer to type 6B10.
8BA8A	Refer to type 6BA8A.
8BA11	Refer to type 6BA11.
8BH8	Refer to type 6BH8.



12FU

DUAL PENTODE **8BM11**

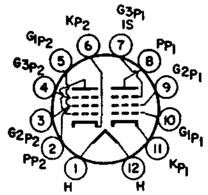
Duodecar type used as if amplifier in television receivers. Unit No.1 is a semiremote-cutoff pentode, and unit No. 2 is a sharp-cutoff pentode. **Outlines section,** 8B; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 8.4; amperes, 0.45; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1	Unit No.2	
Plate Voltage	160	160	volts
Grid-No.3 (Suppressor-Grid) Voltage	0	0	volts
Grid-No.2 (Screen-Grid) Voltage	160	160	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.2	2.2	watts
Grid-No.2 Input	0.55	0.55	watt
CHARACTERISTICS			
Plate Supply Voltage	125	125	volts
Grid No.3	Connected to cathode at socket		
Grid-No.2 Voltage	125	125	volts
Cathode-Bias Resistor	56	120	ohms
Plate Resistance (Approx.)	220000	300000	ohms
Transconductance	8800	8500	μmhos
Plate Current	14	9	mA
Grid-No.2 Current	3.6	2.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—	—5.5	volts
Grid-No.1 Voltage (Approx.) for transconductance of 50 μmho	—16.5	—	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance, for cathode-bias operation	1	0.25	megohm

Refer to type 6BN8. **8BN8**

Refer to type 6BQ5. **8BQ5**



12DM

SEMIREMOTE-CUTOFF DUAL PENTODE **8BQ11**

11BQ11

Duodecar type used as intermediate-frequency amplifier in television receivers. Outlines section, 8B; requires duodecar 12-contact socket. Type 11BQ11 is identical with type 8BQ11 except for heater ratings.

	8BQ11	11BQ11	
Heater Voltage (ac/dc)	8.4	11.2	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
	Unit No.1	**Unit No.2**	
Direct Interelectrode Capacitances:			
Grid No.1 to Plate	0.022	0.024	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	10	—	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.8	—	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, Grid No.3 of Unit No.1, and Internal Shield	—	11	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, Grid No.3 of Unit No.1, and Internal Shield	—	2.8	pF

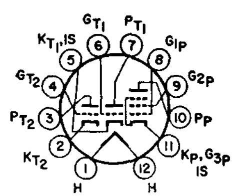
Plate of Unit No.1 to Plate of Unit No.2	0.015	pF
Grid No.1 of Unit No.1 to Plate of Unit No.2	0.002	pF
Grid No.1 of Unit No.2 to Plate of Unit No.1	0.008	pF
Grid No.1 of Unit No.1 to Grid No.1 of Unit No.2	0.002	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Unit No.1	Unit No.2	
Plate Voltage			
Grid-No.3 (Suppressor-Grid) Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	0	0	volts
Grid-No.2 Voltage	330	330	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	See curve page 98		
Plate Dissipation	0	0	volts
Grid-No.2 Input:	3.1	3.1	watts
For grid-No.2 voltages up to 165 volts	0.65	0.65	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98		
CHARACTERISTICS			
Plate Supply Voltage	125	125	volts
Grid No.3	Connected to cathode at socket		
Grid-No.2 Voltage	125	125	volts
Cathode-Bias Resistor	56	56	ohms
Plate Resistance (Approx.)	0.2	0.2	megohm
Transconductance	10500	13000	μmhos
Plate Current	11	11	mA
Grid-No.2 Current	3.5	3.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—	—3	volts
Grid-No.1 Voltage (Approx.) for transconductance of 50 μmho	—15	—	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance, for cathode-bias operation	1	0.25	megohm

8BU11

MEDIUM-MU TWIN TRIODE— SHARP-CUTOFF PENTODE



12FP

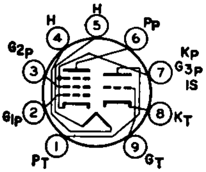
Duodecar type used in television receiver applications. **Outlines** section, 8C; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 7.8; amperes, 0.6; warm-up time, 11 seconds, maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Pentode Unit	Each Triode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	—	volts
Grid-No.2 Voltage	See curve page 98	—	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	1.8	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	0.55	—	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	—	
CHARACTERISTICS			
Plate Supply Voltage	125	125	volts
Grid-No.2 Voltage	125	—	volts
Grid-No.1 Voltage	—1	—	volts
Cathode-Bias Resistor	—	68	ohms
Amplification Factor	—	43	
Plate Resistance (Approx.)	200000	50000	ohms
Transconductance	7500	8600	μmhos
Plate Current	12	13.5	mA
Grid-No.2 Current	4	—	mA
Grid Voltage (Approx.) for plate current of 100 μA	—	—8	volts
Grid-No.1 Voltage (Approx.) for plate current of 30 μA	—8	—	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.5	megohm
For cathode-bias operation	1	1	megohm

8CM7 Refer to type 6CM7.

Refer to type 6CN7.	**8CN7**
Refer to type 6CS7.	**8CS7**
Refer to type 6CW5/EL86.	**8CW5/XL86**
Refer to type 6CW5.	**8CW5**
Refer to type 6CX8.	**8CX8**
Refer to type 6EB8.	**8EB8**
Refer to type 6EM5.	**8EM5**
Refer to type 6ET7.	**8ET7**
Refer to chart at end of section.	**8FQ7**
Refer to type 6FQ7/6CG7.	**8FQ7/8CG7**
Refer to chart at end of section.	**8GJ7**
Refer to type 6GJ7/ECF801.	**8GJ7/PCP801**
Refer to type 6GN8.	**8GN8**
Refer to type 6GU7.	**8GU7**
Refer to type 6JU8A.	**8JU8A**
Refer to type 6JV8.	**8JV8**
Refer to type 6KA8.	**8KA8**
Refer to type 6LC8.	**8LC8**
Refer to type 6LT8.	**8LT8**
Refer to chart at end of section.	**9A8**



9DC

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

9A8/ PCF80

Miniature type used as combined oscillator and mixer tubes in vhf color and black-and-white television receivers. **Outlines** section, 6B; requires miniature 9-contact socket. **Heater:** volts (ac/dc), 9; amperes, 0.3; maximum heater-cathode volts, +100, —200 peak; —120 average.

Class A_1 Amplifier

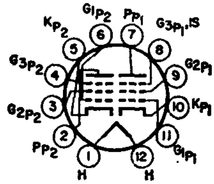
MAXIMUM RATINGS (Design-Center Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	550	volts
Plate Voltage	250	250	volts
Grid-No.2 (Screen-Grid) Voltage	—	175	volts
Cathode Current	14	14	mA
Plate Dissipation	1.5	1.7	watts
Grid-No.2 Input	—	0.5	watt
CHARACTERISTICS			
Plate Voltage	100	170	volts
Grid-No.2 Voltage	—	170	volts
Grid-No.1 Voltage	—2	—2	volts
Amplification Factor	20	47*	
Plate Resistance (Approx.)	—	0.4	megohm
Transconductance	5000	6200	μmhos
Plate Current	14	10	mA
Grid-No.2 Current	—	2.8	mA

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.5	megohm
For cathode-bias operation	0.5	1	megohm

* Grid No.2 to Grid No.1.

9AU7 Refer to type 12AU7A.



12FU

9BJ11 BEAM POWER TUBE— SHARP-CUTOFF PENTODE

Duodecar type used in two-stage video-if-amplifier in television receivers. Pentode unit is used as the input stage and beam power unit as the output stage. Outlines section, 8B; requires duodecar 12-contact socket.

Heater Voltage (ac/dc)	9.6	volts
Heater Current	0.45	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Pentode Unit:		
Grid No.1 to Plate	0.008	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, Grid No.3 of Beam Power Unit, and Internal Shield	9.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, Grid No.3 of Beam Power Unit, and Internal Shield	3.4	pF
Beam Power Unit:		
Grid No.1 to Plate	0.016	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	8.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	3	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Pentode Unit	Beam Power Unit	
Plate Voltage	160	160	volts
Grid-No.3 (Suppressor-Grid) Voltage:			
Positive-bias value	10	0	volts
Negative-bias value	50	—	volts
Grid-No.2 (Screen-Grid) Voltage	160	160	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.8	2.2	watts
Grid-No.2 Input	1.25	0.55	watts
CHARACTERISTICS			
Plate Supply Voltage	110	125	volts
Grid No.3	Connected to cathode at socket		
Grid-No.2 Voltage	110	125	volts
Grid-No.1 Resistor	0.1	—	megohm
Cathode-Bias Resistor	—	120	ohms
Plate Resistance (Approx.)	40000	40000	ohms
Transconductance	7500	9600	μmhos
Plate Current	5.8	8.5	mA
Grid-No.2 Input	6.8	2.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—3	—4.5	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance	0.1	—	megohm
Grid-No.3-Circuit Resistance	1.5	—	megohm

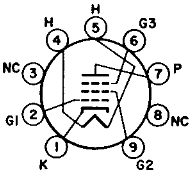
9BR7 Refer to chart at end of section.

9CL8 Refer to chart at end of section.

9EA8 Refer to type 6EA8.

9GV8 Refer to chart at end of section.

9GV8/XCL85 Refer to type 6GV8/EL500.



9RF

SHARP-CUTOFF PENTODE 9KC6

Miniature type used as chroma bandpass amplifier, color demodulator, or video amplifier in color television receivers. **Outlines section, 6E;** requires miniature 9-contact socket. **Heater:** volts, 8.7; amperes, 0.45; warm-up time, 11 seconds; maximum heater-cathode voltage, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

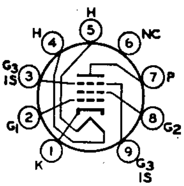
Plate Voltage	400	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.3 (Suppressor-Grid) Voltage:		
Positive-bias value	0	volts
Negative-bias value	100	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	7	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	1.5	watts
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	250	250	50	volts
Grid-No.2 Voltage	150	100	100	volts
Grid-No.1 Voltage	0	—1	0	volts
Grid-No.3 Voltage (referred to negative end of cathode)	0	—25	—	volts
Cathode-Bias Resistor	56	0	—	ohms
Plate Current	18	1	25	mA
Grid-No.2 Current	9	13	25	mA
Transconductance:				
Grid No.1 to plate	24000	—	—	µmhos
Grid No.3 to plate	500	—	—	µmhos
Plate Resistance (Approx.)	55000	—	—	ohms
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—4.1	—	—	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	0.5	megohm
Grid-No.3-Circuit Resistance	1	megohm



9GK

SHARP-CUTOFF PENTODE 9KX6

Miniature type with frame grid used as video output amplifier in color and black-and-white television receivers. **Outlines section, 6E;** requires miniature 9-contact socket. **Heater:** volts, 8.7; amperes, 0.45; warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	400	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive value	0	volts
Plate Dissipation	11.5	watts
Grid-No.2 Input	1.5	watts

CHARACTERISTICS

Plate Voltage	250	50	volts
Grid-No.3 Voltage	Connected to cathode at socket		
Grid-No.2 Supply Voltage	150	125	volts
Grid-No.1 Voltage	0	0	volts

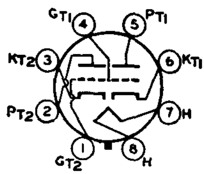
Cathode-Bias Resistor, Bypassed	56	—	ohms
Plate Resistance (Approx.)	50000	—	ohms
Transconductance (Grid No.1 to Plate)	36000	—	µmhos
Plate Current	28	70	mA
Grid-No.2 Current	6.5	24	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—5.7	—	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.25	megohm

9KZ8	Refer to type 6KZ8.
9U8A	Refer to type 6U8A.
10	Refer to chart at end of section.
10AL11	Refer to type 6AL11.
10BQ5	Refer to type 6BQ5.
10C8	Refer to chart at end of section.
10CW5	Refer to chart at end of section.
10CW5/LL86	Refer to type 6CW5/EL86.
10DE7	Refer to type 6DE7.
10DR7	Refer to type 6DR7.
10DX8	Refer to chart at end of section.
10DX8/LCL84	Refer to type 6DX8/ECL84.

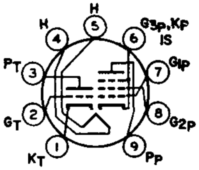
10EG7 DUAL TRIODE



8BD

Glass octal type used as combined vertical-deflection oscillator and vertical-deflection amplifier in television receivers. Outlines section, 13B; requires octal socket. Heater: volts (ac/dc), 9.7; amperes, 0.6; warm-up time (average), 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average. For maximum ratings and characteristics, refer to type 6EW7.

10EM7	Refer to type 6EM7.
10EW7	Refer to type 6EW7.
10GF7	Refer to chart at end of section.
10GF7A	Refer to type 6GF7A.
10GK6	Refer to type 6GK6.
10GN8	Refer to type 6GN8.
10HF8	Refer to type 6HF8.



9DX

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE **10JA8**

Miniature type used in color and black-and-white television receiver applications. The triode unit is used as a sync separator, sync clipper, and phase inverter; the pentode unit is used as a video amplifier. **Outlines section**, 6E; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	10.5	volts
Heater Current	0.45	ampere
Heater Warm-up Time	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	4	pF
Grid to Cathode, Pentode Cathode, Heater, Pentode Grid No.3, and Internal Shield	2.6	pF
Plate to Cathode, Pentode Cathode, Heater, Pentode Grid No.3, and Internal Shield	2.6	pF
Pentode Unit:		
Grid No.1 to Plate	0.1 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	11	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4.4	pF
Grid No.1 to Triode Plate	0.005 max	pF
Plate to Triode Grid	0.018 max	pF
Plate to Triode Plate	0.17 max	pF

Class A_1 Amplifier

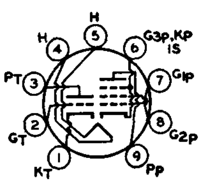
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	1	5	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	1.5	watts
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	

CHARACTERISTICS	Triode Unit		Pentode Unit			
Plate Voltage	135	200	30	135	200	volts
Grid-No.2 Voltage	—	—	135	135	135	volts
Grid-No.1 Voltage	−2	−2	0	−1.5	−1.5	volts
Amplification Factor	60	70	—	—	—	
Plate Resistance	39000	19000	—	66000	70000	ohms
Transconductance	1550	3700	—	12600	14000	µmhos
Plate Current	1	3.5	32•	17	18	mA
Grid-No.2 Current	—	—	14•	4.2	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 µA	−4.8	−7	—	−5	−5	volts

MAXIMUM CIRCUIT VALUES	Triode Unit	Pentode Unit	
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Refer to type 6JT8. **10JT8**



9DX

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE **10JY8**

Miniature type used in television receiver applications. The pentode unit is used as a video amplifier, and the triode unit as a sync separator. **Outlines section**, 6E; requires miniature 9-contact socket. **Heater**: volts (ac/dc), 10.5; amperes, 0.45; warm-up time (average), 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average (—300 peak, —200 average for triode unit).

Class A₁ Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit		
Plate Voltage	330	330		volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330		volts
Grid-No.2 Voltage	—	See curve page 98		
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0		volts
Plate Dissipation	2	5		watts
Grid-No.2 Input:				
For grid-No.2 voltages up to 165 volts	—	1.1		watts
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98		
CHARACTERISTICS				
Plate Voltage	125	50	200	volts
Grid-No.2 Voltage	—	150	150	volts
Grid-No.1 Voltage	—	0	—	volts
Cathode-Bias Resistor	68	—	100	ohms
Amplification Factor	46	—	—	
Plate Resistance (Approx.)	4400	—	55000	ohms
Transconductance	10400	—	11000	μmhos
Plate Current	15	60■	24	mA
Grid-No.2 Current	—	18■	4.8	mA
Grid Voltage (Approx.) for plate current of 10 μA	—8	—	—10	volts
MAXIMUM CIRCUIT VALUES				
Grid-No.1-Circuit Resistance:				
For fixed-bias operation	0.5	0.25		megohm
For cathode-bias operation	1	1		megohm

■ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

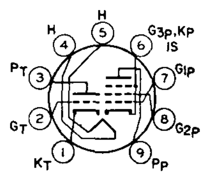
10KR8 Refer to type 6KR8.

10KU8 Refer to type 6KU8.

10LB8 Refer to type 6LB8.

10LE8 Refer to type 6LE8.

10LW8 HIGH-MU TRIODE— SHARP-CUTOFF PENTODE



9DX

Miniature type used in television receivers applications. The pentode unit is used as a video amplifier, and the triode unit for general-purpose use. Outlines section, 6E; requires miniature 9-contact socket. Heater: volts (ac/dc), 10.5; amperes, 0.45; average warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

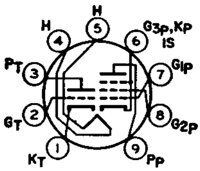
Class A₁ Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit		
Plate Voltage	330	330		volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330		volts
Grid-No.2 Voltage		See curve page 98		
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0		volts
Plate Dissipation	1.5	4		watts
Grid-No.2 Input:				
For grid-No.2 voltages up to 165 volts	—	1.5		watts
For grid-No.2 voltages between 165 and 330 volts		See curve page 98		
CHARACTERISTICS				
Plate Supply Voltage	200	35	200	volts
Grid-No.2 Supply Voltage	—	100	100	volts
Grid-No.1 Voltage	—2	0	—	volts
Cathode-Bias Resistor	—	—	82	ohms
Amplification Factor	75	—	—	
Plate Resistance (Approx.)	18700	—	60000	ohms
Transconductance	4000	—	19000	μmhos
Plate Current	2.6	48	16.5	mA
Grid-No.2 Current	—	12.5	2.8	mA
Grid Voltage (Approx.) for plate current of 30 μA	—4	—	—	volts
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—	—5.5	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.5	megohm
For cathode-bias operation	1	1	megohm

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE 10LZ8



9DX

Miniature type used in television receiver applications. The pentode unit is used as a video amplifier, and the triode unit for general-purpose use. **Outlines section,** 6E; requires miniature 9-contact socket. **Heater:** volts (ac/dc), 10.5; amperes, 0.45; warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

	Triode Unit	Pentode Unit	
Plate Voltage	300	225	volts
Grid-No.2 (Screen-Grid) Voltage	—	160	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	1	4.5	watts
Grid-No.2 Input	—	2	watts

CHARACTERISTICS

Plate Supply Voltage	250	30	30	200	volts
Grid-No.2 Supply Voltage	—	140	140	140	volts
Grid-No.1 Voltage	—2	0	—1	—2	volts
Amplification Factor	110	—	—	—	
Plate Resistance (Approx.)	52000	—	—	150000	ohms
Transconductance	2100	—	11000	9500	μmhos
Plate Current	1.1	30	16	12	mA
Grid-No.2 Current	—	13.5	9.5	2.5	mA
Grid Voltage (Approx.) for plate current of 10 μA	—3.6	—	—	—	volts
Grid-No.1 Voltage (Approx.) for plate current of 500 μA	—	—	—4	—4.2	volts

MAXIMUM CIRCUIT VALUES

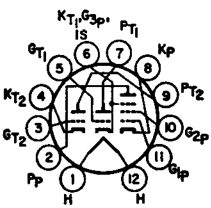
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.5	megohm
For cathode-bias operation	1	1	megohm

Refer to chart at end of section. **11**

Refer to type 6AR11. **11AR11**

Refer to type 8BQ11. **11BQ11**

DUAL TRIODE— SHARP-CUTOFF PENTODE 11BT11



12GS

Duodecar type used in television receiver applications. The triode units are used for general-purpose applications; the pentode unit is used in video-amplifier service. **Outlines section,** 8B; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 10.7; amperes, 0.6; warm-up time (average), 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

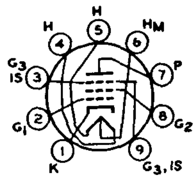
	Triode Unit No. 1	Triode Unit No. 2	Pentode Unit	
Plate Voltage	330	330	165	volts
Grid-No.2 (Screen-Grid) Voltage	—	—	165	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	0	volts

	Triode Unit No. 1	Triode Unit No. 2	Pentode Unit		
Plate Dissipation	1.5	2	3.5		watts
Grid-No.2 Input	—	—	1.5		watts
CHARACTERISTICS					
Plate Voltage	200	200	35	150	volts
Grid-No.2 Voltage	—	—	100	100	volts
Grid-No.1 Voltage	—	—	0	—	volts
Cathode-Bias Resistor	270	470	—	82	ohms
Amplification Factor	69	40	—	—	
Plate Resistance (Approx.)	12500	7600	—	51000	ohms
Transconductance	5500	5300	—	19000	µmhos
Plate Current	7.1	7.2	54	17.4	mA
Grid-No.2 Current	—	—	13.5	3.2	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—	—8	—	—6.6	volts
Grid-No.1 Voltage (Approx.) for plate current of 50 µA	—5.5	—	—	—	volts
MAXIMUM CIRCUIT VALUES					
Grid-No.1-Circuit Resistance:					
For fixed-bias operation	0.5	0.5	0.05		megohm
For cathode-bias operation	1	1	0.1		megohm

11CY7 Refer to type 6CY7.

11FY7 Refer to type 6FY7.

11HM7 SHARP-CUTOFF PENTODE



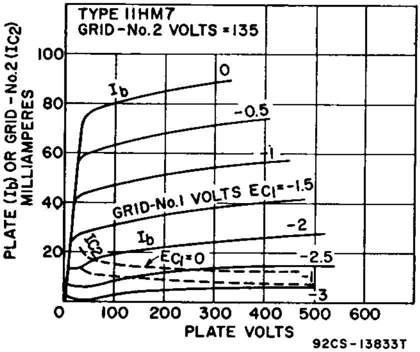
9BF

Miniature type with frame grid used as video output amplifier in color television receivers. Outlines section, 6E; requires miniature 9-contact socket.

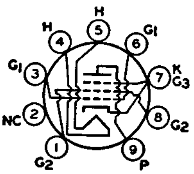
	Series	Parallel	
Heater Arrangement	Series	Parallel	
Heater Voltage (ac/dc)	11	5.5	volts
Heater Current	0.3	0.6	ampere
Heater-Cathode Voltage:			
Peak value		±200 max	volts
Average value		100 max	volts
Direct Interelectrode Capacitances:			
Grid No.1 to Plate		0.15 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield		14	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield		5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	7	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	1	watt
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	
CHARACTERISTICS		
Plate Supply Voltage	200	volts
Grid-No.3 Voltage	0	volts
Grid-No.2 Voltage	135	volts
Cathode-Bias Resistor	47	ohms
Plate Resistance (Approx.)	40000	ohms
Transconductance	30000	µmhos
Plate Current	30	mA
Grid-No.2 Current	5.2	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 µA	—4.5	volts
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.25	megohm



Refer to type 6JE8.	**11JE8**
Refer to type 6KV8.	**11KV8**
Refer to type 6LQ8.	**11LQ8**
Refer to chart at end of section.	**11Y9**
Refer to type 6Y9.	**11Y9/LFL200**
Refer to chart at end of section.	**12A5**
Refer to chart at end of section.	**12A7**
Refer to chart at end of section.	**12A8GT**



9EU

BEAM POWER TUBE **12AB5**

Miniature type used in the output stage of automobile radio receivers operating from a 12-volt storage battery. **Outlines section, 6E;** requires miniature 9-contact socket.

Heater-Voltage Range (ac/dc)•	10 to 15.9	volts
Heater Current (Approx.) at 12.6 volts	0.2	ampere
Peak Heater-Cathode Voltage	±90 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.7 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	8	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	8.5	pF

• For longest life, it is recommended that the heater be operated within the voltage range of 11 to 14 volts.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	315	volts
Grid-No.2 (Screen-Grid) Voltage	285	volts
Plate Dissipation	12	watts
Grid-No.2 Input	2	watts
Bulb Temperature (At hottest point)	250	°C

TYPICAL OPERATION WITH 12.6 VOLTS ON HEATER

Plate Supply Voltage	250	250	volts
Grid-No.2 Supply Voltage	200	250	volts
Grid-No.1 (Control-Grid) Voltage	—	—12.5	volts
Cathode-Bias Resistor	270	—	ohms
Peak AF Grid-No.1 Voltage	10.5	12.5	volts

Zero-Signal Plate Current	33.5	45	mA
Maximum-Signal Plate Current	36	47	mA
Zero-Signal Grid-No.2 Current	1.6	4.5	mA
Maximum-Signal Grid-No.2 Current	3.2	7	mA
Plate Resistance (Approx.)	75000	50000	ohms
Transconductance	4000	4100	μmhos
Load Resistance	6000	5000	ohms
Total Harmonic Distortion	8	8	per cent
Maximum-Signal Power Output	3.3	4.5	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

Push-Pull Class AB_1 Amplifier

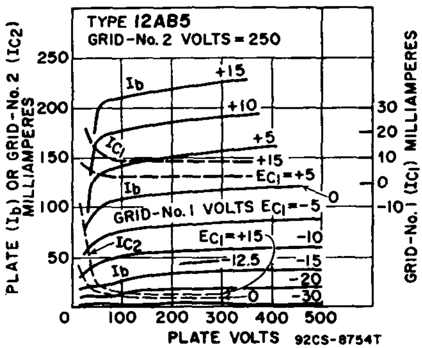
MAXIMUM RATINGS (Same as for Single-Tube Class A_1 Amplifier)

TYPICAL OPERATION WITH 12.6 VOLTS ON HEATER (Values are for two tubes)

Plate Voltage	250	volts
Grid-No.2 Voltage	250	volts
Grid-No.1 Voltage	—15	volts
Peak AF Grid-No.1-to-Grid-No.1 Voltage	30	volts
Zero-Signal Plate Current	70	mA
Maximum-Signal Plate Current	79	mA
Zero-Signal Grid-No.2 Current	5	mA
Maximum-Signal Grid-No.2 Current	13	mA
Effective Load Resistance (Plate-to-Plate)	10000	ohms
Total Harmonic Distortion	5	per cent
Maximum-Signal Power Output	10	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm



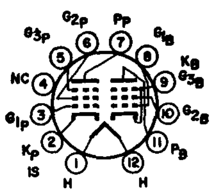
12AC6 Refer to chart at end of section.

12AD6 Refer to chart at end of section.

12AE6
12AE6A Refer to chart at end of section.

12AE7 Refer to chart at end of section.

12AE10 BEAM POWER TUBE— SHARP-CUTOFF PENTODE



12EZ

Duodecar type used as combined FM detector and audio-frequency output amplifier in television receivers. The beam power unit is used in af output stages and the pentode unit as an FM detector. Outlines section, 8C; requires duodecar 12-contact socket. Heater: volts (ac/dc), 12.6; amperes, 0.45; warm-up time (average), 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Beam Power Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	165	volts
Grid-No.2 (Screen-Grid) Voltage	150	volts
Cathode Current	60	mA
Plate Dissipation	6	watts
Grid-No.2 Input	1.25	watts

TYPICAL OPERATION

Plate Voltage	145	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 (Control-Grid) Voltage	—7	volts
Peak AF Grid-No.1 Voltage	7	volts
Zero-Signal Plate Current	34	mA
Maximum-Signal Plate Current	39	mA
Zero-Signal Grid-No.2 Current	6.5	mA
Maximum-Signal Grid-No.2 Current	9.3	mA
Plate Resistance (Approx.)	33000	ohms
Transconductance	5600	μmhos
Load Resistance	2500	ohms
Total Harmonic Distortion (Approx.)	12	per cent
Maximum-Signal Power Output	1.45	watts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance:		
For cathode-bias operation	1	megohm

Pentode Unit as Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	150	volts
Grid-No.3 (Suppressor-Grid) Voltage	0	volts
Grid-No.2 Voltage	100	volts
Cathode-Bias Resistor	560	ohms
Plate Resistance (Approx.)	0.15	megohm
Transconductance, Grid No.1	1000	μmhos
Transconductance, Grid No.3	400	μmhos
Plate Current	1.3	mA
Grid-No.2 Current	2	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—4.5	volts
Grid-No.3 Voltage (Approx.) for plate current of 10 μA	—4.5	volts

Pentode Unit as FM Detector

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid-No.3 Voltage	28	volts
Grid-No.2 Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 Voltage, Positive-bias value	0	volts
Plate Dissipation	1.7	watts
Grid-No.2 Input	1.1	watts

Refer to type 6AF3. **12AF3**

Refer to chart at end of section. **12AF6**

Refer to chart at end of section. **12AH7GT**

Refer to chart at end of section. **12AJ6**

Refer to type 6AL5. **12AL5**

Refer to chart at end of section. **12AL8**

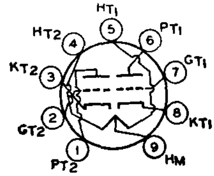
Refer to type 6AL11. **12AL11**

Refer to type 6AQ5A. **12AQ5**

Refer to type 6AT6. **12AT6**

12AT7
12AT7/ ECC81

HIGH-MU TWIN TRIODE



9A

Miniature types used as push-pull cathode-drive amplifiers or frequency converters in the FM and television broadcast bands. **Outlines section,** 6B; require miniature 9-contact socket. Each triode unit is independent of the other except for the common heater. For typical operation as a resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section.

	Series	Parallel	
Heater Arrangement:			
Heater Voltage (ac/dc)	12.6	6.3	volts
Heater Current	0.15	0.3	ampere
Peak Heater-Cathode Voltage		±90 max	volts
Direct Interelectrode Capacitances:			
Grid-Drive Operation:			
Grid to Plate (Each unit)		1.5	pF
Grid to Cathode and Heater (Each unit)		2.2	pF
Plate to Cathode and Heater:			
Unit No.1		0.5	pF
Unit No.2		0.4	pF
Cathode-Drive Operation:			
Cathode to Plate (Each unit)		0.2	pF
Cathode to Grid and Heater (Each unit)		4.6	pF
Plate to Grid and Heater (Each unit)		1.8	pF
Heater to Cathode (Each Unit)		2.4	pF

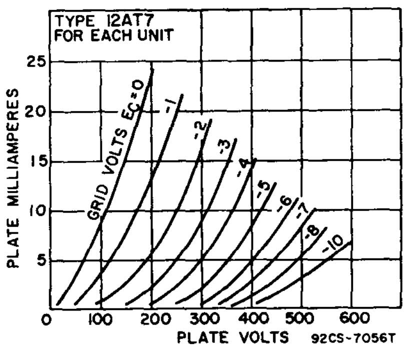
Class A_1 Amplifier (Each Unit)

MAXIMUM AND MINIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts
Grid Voltage, Negative-bias value	50	volts
Plate Dissipation	2.5	watts

CHARACTERISTICS

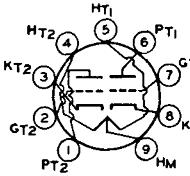
Plate Supply Voltage	100	250	volts
Cathode-Bias Resistor	270	200	ohms
Amplification Factor	60	60	
Plate Resistance (Approx.)	15000	10900	ohms
Transconductance	4000	5500	µmhos
Grid Voltage (Approx.) for plate current of 10 µA	—5	—12	volts
Plate Current	3.7	10	mA



12AU6 Refer to type 6AU6A.

12AU7 Refer to chart at end of section.

12AU7A
12AU7A/ ECC82
7AU7, 9AU7



9A

MEDIUM-MU TWIN TRIODE

Miniature types used as phase inverters or push-pull amplifiers in ac/dc radio equipment and as multivibrators or oscillators in industrial control devices. Also used as combined vertical oscillators and vertical-deflection amplifiers, and as horizontal-deflection oscillators, in color and black-and-white television receivers. **Outlines section, 6B**; require miniature 9-contact socket. Each triode unit is independent of the other except for the common heater. For typical operation as a resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section. Types 7AU7 and 9AU7 are identical with type 12AU7 and 12AU7A/ECC82 except for heater ratings.

	7AU7	9AU7	12AU7A 12AU7A/ ECC82	
Heater Voltage (ac/dc):				
Series	7	9.4	12.6	volts
Parallel	3.5	4.7	6.3	volts
Heater Current:				
Series	0.3	0.225	0.15	ampere
Parallel	0.6	0.45	0.3	ampere
Heater Warm-up Time (Parallel, Average)	11	11	—	seconds
Heater-Cathode Voltage:				
Peak value	±200 max	±200 max	±200 max	volts
Average value	100 max	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	1.5	1.5	pF
Grid to Cathode and Heater	1.6	1.6	pF
Plate to Cathode and Heater	0.5	0.35	pF

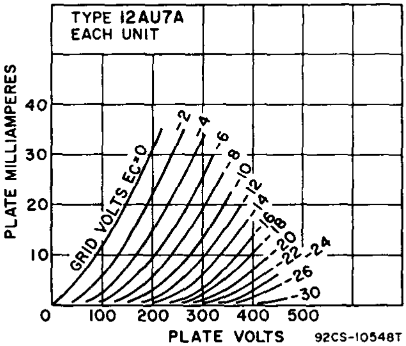
Class A₁ Amplifier (Each Unit Unless Otherwise Specified)

MAXIMUM RATINGS (Design-Maximum Values)			
Plate Voltage		330	volts
Cathode Current		22	mA
Plate Dissipation:			
Each Plate		2.75	watts
Both Plates (Both units operating)		5.5	watts
CHARACTERISTICS			
Plate Voltage	100	250	volts
Grid Voltage	0	—8.5	volts
Amplification Factor	19.5	17	
Plate Resistance (Approx.)	6250	7700	ohms
Transconductance	3100	2200	μmhos
Plate Current	11.8	10.5	mA
Grid Voltage (Approx.) for plate current of 10 μA	—	—24	volts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance:			
For fixed-bias operation		0.25	megohm
For cathode-bias operation		1	megohm

Oscillator (Each Unit Unless Otherwise Specified)

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Vertical-Deflection Oscillator	Horizontal-Deflection Oscillator	
DC Plate Voltage	330	330	volts
Peak Negative-Pulse Grid Voltage	440	660	volts
Peak Cathode Current	66	330	mA
Average Cathode Current	22	22	mA
Plate Dissipation:			
Each Plate	2.75	2.75	watts
Both Plates (Both units operating)	5.5	5.5	watts
MAXIMUM CIRCUIT VALUES			
Grid-Circuit Resistance	2.2	2.2	megohms



Vertical-Deflection Amplifier (Each Unit Unless Otherwise Specified)

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Voltage	330	volts
Peak Positive-Pulse Plate Voltage#	1200	volts
Peak Negative-Pulse Grid Voltage	275	volts
Peak Cathode Current	66	mA
Average Cathode Current	22	mA
Plate Dissipation:		
Each Plate	275	volts
Both Plates *(Both units operating)*	5.5	watts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance, for cathode-bias operation	2.2	megohms

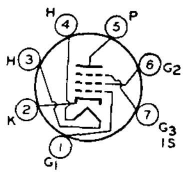
Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).

12AV5GA Refer to type 6AV5GA.

12AV6 Refer to type 6AV6.

12AV7 Refer to chart at end of section.

12AW6 SHARP-CUTOFF PENTODE



7CM

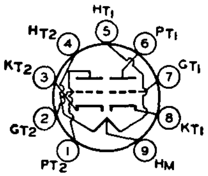
Miniature type used as an rf or if amplifier up to 400 MHz in compact ac/dc FM receivers. **Outlines section,** 5C; requires miniature 7-contact socket. **Heater:** volts (ac/dc), 12.6; amperes, 0.15. Except for heater ratings and terminal connections, this type is identical with miniature type 6AG5.

12AX3 Refer to type 6AX3.

12AX4GT
12AX4GTA Refer to chart at end of section.

12AX4GTB Refer to type 6AX4GTB.

12AX7 Refer to chart at end of section.



9A

HIGH-MU TWIN TRIODE 12AX7A 12AX7A/ ECC83

Miniature types used as phase inverters or twin resistance-coupled amplifiers in radio equipment. **Outlines section**, 6B; require miniature 9-contact socket. Each triode unit is independent of the other except for common heater. For characteristics and curves, refer to type 6AV6. For typical operation as a resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section.

Heater Arrangement:	**Series**	**Parallel**	
Heater Voltage (ac/dc)	12.6	6.3	volts
Heater Current	0.15	0.3	ampere
Heater-Catnode-Voltage:			
Peak value		±200 max	volts
Average value		100 max	volts
Direct Interelectrode Capacitances (Approx.):	**Unit No.1**	**Unit No.2**	
Grid to Plate	1.7	1.7	pF
Grid to Cathode and Heater	1.6	1.6	pF
Plate to Cathode and Heater	0.46	0.34	pF

Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid Voltage:		
Negative-bias value	55	volts
Positive-bias value	0	volts
Plate Dissipation	1.2	watts

EQUIVALENT-NOISE AND HUM VOLTAGE (References To Grid, Each Unit)•

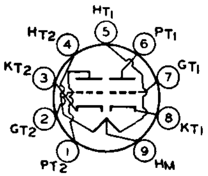
Average Value	1.8	μV rms

• Measured in "true rms" units under the following conditions: Heater voltage (parallel connection), 6.3 volts ac; center tap of heater transformer grounded; plate supply voltage, 250 volts dc; plate load resistor, 100000 ohms; cathode resistor, 2700 ohms bypassed by 100-μF capacitor; grid resistor, 0 ohms; and amplifier covering frequency range between 25 and 10000 Hz.

Refer to chart at end of section. **12AY3**

Refer to type 6AY3B. **12AY3A**

MEDIUM-MU TWIN TRIODE 12AY7



9A

Miniature type used in the first stages of high-gain audio-frequency amplifiers. **Outlines section**, 6B; requires miniature 9-contact socket. Each triode unit is independent of the other except for the common heater. Use of the 12.6-volt connection with an ac heater supply is not recommended for applications involving low hum. For typical operation as a resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section.

Heater Arrangement:	**Series**	**Parallel**	
Heater Voltage (ac/dc)	12.6	6.3	volts
Heater Current	0.15	0.3	ampere
Peak Heater-Cathode Voltage		±90 max	volts
Direct Interelectrode Capacitances (Approx., Each Unit)			
Grid tc Plate		1.3	pF
Grid to Cathode and Heater		1.3	pF
Plate to Cathode and Heater		0.6	pF

Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	300	volts

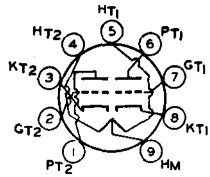
Grid Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Cathode Current	10	mA
Plate Dissipation	1.5	watts

CHARACTERISTICS

Plate Voltage	250	volts
Grid Voltage	—4	volts
Amplification Factor	40	
Plate Resistance	22800	ohms
Transconductance	1750	μmhos
Plate Current	3	mA
Grid Voltage (Approx.) *for plate current of 10 mA*	—11	volts

12AZ7 Refer to chart at end of section.

12AZ7A HIGH-MU TWIN TRIODE



9A

Miniature type used in direct-coupled cathode-drive rf amplifier circuits of vhf color and black-and-white television tuners. **Outlines section, 6B**; requires miniature 9-contact socket. For characteristics as class A_1 amplifier, refer to miniature type 12AT7.

Heater Voltage (ac/dc):			
Series		12.6	volts
Parallel		6.3	volts
Heater Current:			
Series		0.225	ampere
Parallel		0.45	ampere
Heater Warm-up Time (Average)		11	seconds
Heater-Cathode Voltage:			
Peak value		±200 max	volts
Average value		100 max	volts
Direct Interelectrode Capacitance (Approx.):	Unshielded	Shielded▲	
Grid to Plate (Each unit)	2	1.9	pF
Grid to Cathode and Heater (Each unit)	2.6	2.8	pF
Plate to Cathode and Heater:			
Unit No.1	0.44	1.4	pF
Unit No.2	0.36	1.6	pF

▲ With external shield connected to cathode of unit under test.

Class A_1 Amplifier (Each Unit)

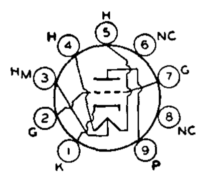
MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid Voltage, Negative-bias value	55	volts
Plate Dissipation	2.5	watts

MAXIMUM CIRCUIT VALUES (Each Unit)

Grid-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

12B4A LOW-MU TRIODE



9AG

Miniature type used as vertical-deflection amplifier in television receivers. **Outlines section, 6E**; requires miniature 9-contact socket.

	Series	Parallel	
Heater Voltage	12.6	6.3	volts
Heater Current	0.3	0.6	ampere
Heater Warm-up Time	—	11	seconds
Heater-Cathode Voltage:			
Peak value		±200 max	volts
Average value		100 max	volts

Direct Interelectrode Capacitances:

Grid to Plate	4.8	pF
Grid to Cathode and Heater	5	pF
Plate to Cathode and Heater	1.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

Plate Voltage	550	volts
Grid Voltage, Negative-bias value	50	volts
Plate Dissipation	5.5	watts

CHARACTERISTICS

Plate Voltage	150	volts
Grid Voltage	—17.5	volts
Amplification Factor	6.5	
Plate Resistance (Approx.)	1030	ohms
Transconductance	6300	µmhos
Plate Current	34	mA
Plate Current for grid voltage of —23 volts	9.6	mA
Grid Voltage (Approx.) for plate current of 200 µA	—32	volts

MAXIMUM CIRCUIT VALUES

Grid-Circuit Resistance:

For fixed-bias operation	0.47	megohm
For cathode-bias operation	2.2	megohms

Vertical-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)

DC Plate Voltage	550	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	1000†	volts
Peak Negative-Pulse Grid Voltage	250	volts
Peak Cathode Current	105	mA
Average Cathode Current	30	mA
Plate Dissipation	5.5	watts

MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance, for cathode-bias operation	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).
† Under no circumstances should this absolute value be exceeded.

Refer to chart at end of section. **12B8GT**

Refer to type 6BA6. **12BA6**

Refer to chart at end of section. **12BA7**

Refer to chart at end of section. **12BD6**

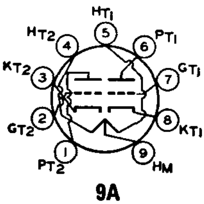
Refer to type 6BE3. **12BE3**

Refer to type 6BE6. **12BE6**

Refer to chart at end of section. **12BF6**

Refer to type 6BF11. **12BF11**

Refer to chart at end of section. **12BH7**



9A

MEDIUM-MU TWIN TRIODE 12BH7A

Miniature type used as combined vertical-deflection amplifier and vertical oscillator, and as horizontal-deflection oscillator, in television receivers, and in phase-inverter and multivibrator circuits. Outlines section, 6E; requires miniature 9-contact socket. Each triode unit is independent of the other except for the common heater.

Heater Arrangement:	Series	Parallel	
Heater Voltage (ac/dc)	12.6	6.3	volts
Heater Current	0.3	0.6	ampere
Heater Warm-up Time (Average)	—	11	seconds

Heater-Cathode Voltage:

Peak value	±200 max	volts
Average value	100 max	volts

Direct Interelectrode Capacitances (Approx.):	Unit No.1	Unit No.2	
Grid to Plate	2.6	2.6	pF
Grid to Cathode and Heater	3.2	3.2	pF
Plate to Cathode and Heater	0.5	0.4	pF
Plate of Unit No.1 to Plate of Unit No.2	0.8		pF

Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Center Values)		
Plate Voltage	300	volts
Grid Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Cathode Current	20	mA
Plate Dissipation:		
Each Plate	3.5	watts
Both plates (Both units operating)	7	watts
CHARACTERISTICS		
Plate Voltage	250	volts
Grid Voltage	—10.5	volts
Amplification Factor	16.5	
Plate Resistance (Approx.)	5300	ohms
Transconductance	3100	μmhos
Plate Current	11.5	mA
Plate Current for grid voltage of —14 volts	4	mA
Grid Voltage (Approx.) for plate current of 50 μA	—23	volts
MAXIMUM CIRCUIT VALUES		
Grid-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

Oscillator (Each Unit)

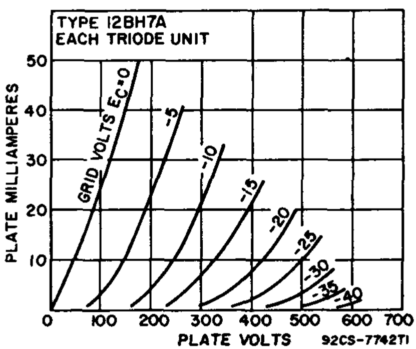
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)	Vertical-Deflection Oscillator	Horizontal-Deflection Oscillator	
DC Plate Voltage	450	450	volts
Peak Negative-Pulse Grid Voltage	400	600	volts
Peak Cathode Current	70	300	mA
Average Cathode Current	20	20	mA
Plate Dissipation:			
Each Plate	3.5	3.5	watts
Both Plates (Both units operating)	7	7	watts
MAXIMUM CIRCUIT VALUE			
Grid-Circuit Resistance	2.2	2.2	megohms

Vertical-Deflection Amplifier (Each Unit)

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)		
DC Plate Voltage	450	volts
Peak Positive-Pulse Plate Voltage# (Absolute maximum)	1500*	volts
Peak Negative-Pulse Grid Voltage	250	volts
Peak Cathode Current	70	mA
Average Cathode Current	20	mA
Plate Dissipation:		
Each Plate	3.5	watts
Both Plates (Both units operating)	7	watts



MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance for cathode-bias operation	2.2	megohms

Pulse duration must not exceed 15% of a vertical scanning cycle (2.5 milliseconds).
* Under no circumstances should this absolute value be exceeded.

Refer to chart at end of section. **12BK5**

Refer to chart at end of section. **12BL6**

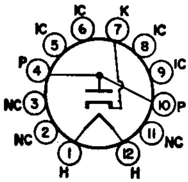
Refer to type 6BN6. **12BN6**

Refer to type 6BQ6GTB/6CU6. **12BQ6GTB/12CU6**

Refer to chart at end of section. **12BR7**

Refer to chart at end of section. **12BS3**

Refer to type 6BS3A. **12BS3A**



12BL

HALF-WAVE VACUUM RECTIFIER 12BT3

Duodecar type used as damper tube in horizontal-deflection circuits of television receivers. Outlines section, 8C; requires duodecar 12-contact socket. Heater: volts (ac/dc), 12.6; amperes, 0.45.

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		3300	volts
Peak Plate Current		1000	mA
Average Plate Current		165	mA
Plate Dissipation		5.3	watts
Heater-Cathode Voltage:			
Peak value	+300	—3300	volts
Average value	+100	—600	volts

CHARACTERISTICS, Instantaneous Value

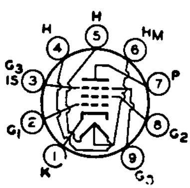
Tube Voltage Drop for plate current of 250 mA max	21	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

Refer to chart at end of section. **12BV7**

Refer to chart at end of section. **12BW4**

Refer to chart at end of section. **12BY7**



9BF

SHARP-CUTOFF PENTODE 12BY7A

Miniature type used as video amplifier in television receivers. Outlines section, 6E; requires miniature 9-contact socket.

Heater Arrangement:	Series	Parallel	
Heater Voltage (ac/dc)	12.6	6.3	volts
Heater Current	0.3	0.6	ampere
Heater Warm-up Time (Average)	—	11	seconds
Heater-Cathode Voltage:			
Peak value		±200 max	volts
Average value		100 max	volts

Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.063	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	10.2	pF
Plate to Cathode, Heater, Grid No.2, and Internal Shield	3.5	pF

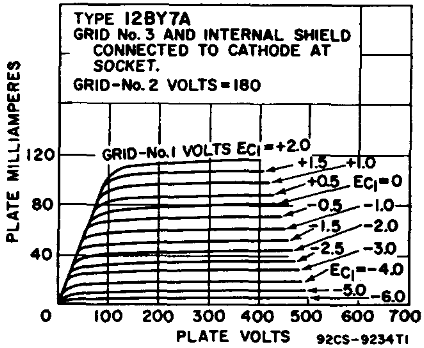
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	330	volts
Grid-No.3 (Suppressor-Grid) Voltage, Positive value	0	volts
Grid-No.2 (Screen-Grid) Voltage	190	volts
Grid-No.1 (Control-Grid) Voltage		
Negative-bias value	55	volts
Positive-bias value	0	volts
Plate Dissipation	6.5	watts
Grid-No.2 Input	1.2	watts

CHARACTERISTICS

Plate Supply Voltage	250	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	180	volts
Cathode-Bias Resistor	100	ohms
Plate Resistance (Approx.)	93000	ohms
Transconductance	11000	µmhos
Plate Current	26	mA
Grid-No.2 Current	5.75	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 µA	—11.6	volts

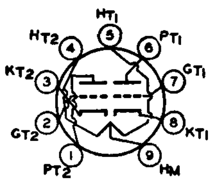


MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	1	megohm

12BZ6 Refer to type 6BZ6.

12BZ7 HIGH-MU TWIN TRIODE



9A

Miniature type used in sync-separator and sync-amplifier circuits of television receivers, and in clipping circuits and audio-amplifier applications. Outlines section, 6E; requires miniature 9-contact socket.

	Series	Parallel	
Heater Voltage	12.6	6.3	volts
Heater Current	0.3	0.6	ampere
Peak Heater-Cathode Voltage	±180 max	±180 max	volts
Direct Interelectrode Capacitances:	Unit No.1	Unit No.2	
Grid to Plate	2.5	2.5	pF
Grid to Cathode, and Heater	6.5	6.5	pF
Plate to Cathode, and Heater	0.7	0.55	pF
Plate of Unit No.1 to Plate of Unit No.2	1.3		pF

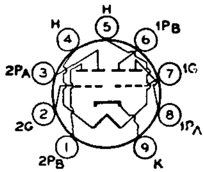
Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Center Values)		
Plate Voltage	300	volts
Grid Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate Dissipation	1.5	watts
CHARACTERISTICS		
Plate Voltage	250	volts
Grid Voltage	—2	volts
Amplification Factor	100	
Plate Resistance (Approx.)	31800	ohms
Transconductance	3200	μmhos
Plate Current	2.5	mA
MAXIMUM CIRCUIT VALUE		
Grid-Circuit Resistance for contact-potential-bias operation	5	megohms

Refer to type 6CU5.	**12C5**
Refer to chart at end of section.	**12C8**
Refer to type 6CA5.	**12CA5**
Refer to type 6CK3.	**12CK3**
Refer to type 6CL3.	**12CL3**
Refer to chart at end of section.	**12CN5**
Refer to type 6CR6.	**12CR6**
Refer to type 6CS6.	**12CS6**
Refer to chart at end of section.	**12CT8**
Refer to type 6CU5.	**12CU5/12C5**
Refer to chart at end of section.	**12CX6**
Refer to type 6DA4.	**12D4**
Refer to type 6DB5.	**12DB5**
Refer to chart at end of section.	**12DE8**
Refer to type 6DK6.	**12DK6**
Refer to chart at end of section.	**12DK7**
Refer to chart at end of section.	**12DL8**
Refer to chart at end of section.	**12DM4** **12DM4A**
Refer to chart at end of section.	**12DQ6A**
Refer to chart at end of section.	**12DQ6B**
Refer to chart at end of section.	**12DQ7**
Refer to chart at end of section.	**12DS7** **12DS7A**
Refer to type 6DT5.	**12DT5**
Refer to type 6DT8.	**12DT8**
Refer to chart at end of section.	**12DU7**
Refer to chart at end of section.	**12DV8**
Refer to chart at end of section.	**12DW7**
Refer to chart at end of section.	**12DY8**
Refer to chart at end of section.	**12DZ6**

12EA6	Refer to chart at end of section.
12EC8	Refer to chart at end of section.
12ED5	Refer to chart at end of section.
12EG6	Refer to chart at end of section.
12EH5	Refer to chart at end of section.
12EK6	Refer to chart at end of section.
12EL6	Refer to chart at end of section.
12EM6	Refer to chart at end of section.
12EN6	Refer to chart at end of section.
12EQ7	Refer to type 6EQ7.
12F5GT	Refer to chart at end of section.
12F8	Refer to chart at end of section.
12FK6	Refer to chart at end of section.
12FM6	Refer to chart at end of section.

12FQ8 HIGH-MU TWIN DOUBLE-PLATE TRIODE



9KT

Miniature type used in frequency-divider and complex-wave-generator circuits of electronic musical instruments. Outlines section, 6B; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	12.6	volts
Heater Current	0.15	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid to Either Plate (Each Unit)	0.9	pF
Grid to Cathode, and Heater (Each Unit)	1.8	pF
Plate A of Unit No.1 to Cathode, and Heater	0.34	pF
Plate B of Unit No.1 to Cathode, and Heater	0.24	pF
Plate A of Unit No.2 to Cathode, and Heater	0.3	pF
Plate B of Unit No.2 to Cathode, and Heater	0.18	pF
Plate A to Plate B (Each Unit)	0.7	pF
Plate A of Unit No.1 to Plate A of Unit No.2	0.4	pF

Class A_1 Amplifier (Each Unit)

CHARACTERISTICS▪

Plate Voltage	250	volts
Grid Voltage	−1.5	volts
Amplification Factor	95	
Plate Resistance (Approx.)	76000	ohms
Transconductance	1250	µmhos
Plate Current	1.5	mA

▪ Using either plate A or plate B, with plate not in use connected to ground.

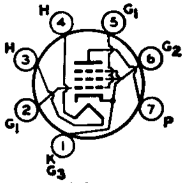
Frequency-Divider and Complex-Wave Generator (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

Plate A Voltage	330	volts
Plate B Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Plate A Dissipation	0.5	watt
Plate B Dissipation	0.5	watt

12FR8	Refer to chart at end of section.

Refer to chart at end of section. **12FV7**



7CV

POWER PENTODE 12FX5

60FX5

Miniature type used in output stages of audio amplifiers. Outlines section, 5D; requires miniature 7-contact socket. Type 60FX5 is identical with type 12FX5 except for heater ratings.

	12FX5	60FX5	
Heater Voltage (ac/dc)	12.6	60	volts
Heater Current	0.45	0.1	ampere
Heater Warm-up Time (Average)	11	—	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.65	pF
Grid No 1 to Cathode, Heater, Grid No.2, and Grid No.3	17	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	9	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Plate Dissipation	5.5	watts
Grid-No.2 Input	2	watts
Bulb Temperature (At hottest point)	225	°C

TYPICAL OPERATION

Plate Supply Voltage	110	volts
Grid-No.2 Supply Voltage	115	volts
Cathode-Bias Resistor	62	ohms
Peak AF Grid-No.1 Voltage	3	volts
Zero-Signal Plate Current	36	mA
Maximum-Signal Plate Current	35	mA
Zero-Signal Grid No.2 Current	10	mA
Maximum-Signal Grid No.2 Current	12	mA
Plate Resistance	17500	ohms
Transconductance	13500	µmhos
Load Resistance	3000	ohms
Total Harmonic Distortion	8	per cent
Maximum-Signal Power Output	1.3	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

Refer to chart at end of section. **12FX8 12FX8A**

Refer to chart at end of section. **12GA6**

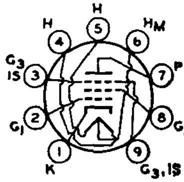
Refer to chart at end of section. **12GC6**

Refer to type 6GE5. **12GE5**

Refer to chart at end of section. **12GJ5**

Refer to type 6GJ5A. **12GJ5A**

Refer to chart at end of section. **12GN7**



12GN7A SHARP-CUTOFF PENTODE

9BF

Miniature type with frame grid used as video-amplifier tube in color and black-and-white television receivers. **Outlines section,** 6E; requires miniature 9-contact socket. **Heater:** volts, 6.3 (parallel), 12.6 (series); amperes, 0.6 (parallel), 0.3 (series); warm-up time (average), 11 seconds, maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	400	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	11.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	1.5	watts
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	50	250	volts
Grid-No.2 Supply Voltage	125	150	volts
Grid-No.1 Voltage	0	0	volts
Cathode-Bias Resistor	—	56	ohms
Plate Resistance (Approx.)	—	0.05	megohm
Transconductance	—	36000	μmhos
Plate Current	70•	28	mA
Grid-No.2 Current	24•	6.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—5.7	volts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	0.25	megohm

• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

12GT5 Refer to chart at end of section.

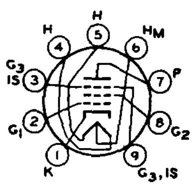
12GT5A Refer to chart at end of section.

12GW6/12DQ6B Refer to type 6GW6/6DQ6B.

12H6 Refer to chart at end of section.

12HE7 Refer to type 38HE7.

12HG7 12HG7/ 12GN7A SHARP-CUTOFF PENTODE



9BF

Neonoval types with frame grid used as video amplifier in color and black-and-white television receivers. **Outlines section,** 10C; require 9-contact neonoval socket.

Heater Arrangement:	Series	Parallel	
Heater Voltage (ac/dc)	12.6	6.3	volts
Heater Current	0.26	0.52	ampere

Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.15 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	14 max	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	4.4 max	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

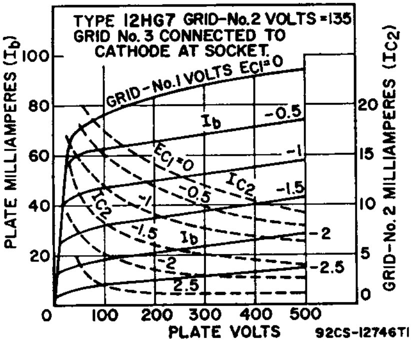
Plate Voltage	400	volts
Grid-No.2 (Screen-Grid) Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	10	watts
Grid-No.2 Input:		
For Grid-No.2 voltages up to 165 volts	1	watt
For Grid-No.2 voltages between 165 and 330 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	300	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket	
Grid-No.2 Supply Voltage	135	volts
Grid No.1	Connected to negative end of cathode resistor	
Cathode Resistor	47	ohms
Plate Resistance (Approx.)	60000	ohms
Transconductance	32000	μmhos
Plate Current	31	mA
Grid-No.2 Current	4.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—4.5	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.25	megohm



Refer to chart at end of section. **12J5GT**

Refer to chart at end of section. **12J7GT**

Refer to chart at end of section. **12J8**

Refer to chart at end of section. **12JB6**

Refer to type 6JB6A. **12JB6A**

Refer to type 6JN6. **12JN6**

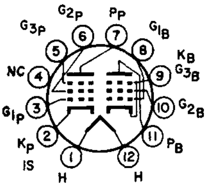
Refer to chart at end of section. **12JN8**

Refer to type 6JQ6. **12JQ6**

12JT6	Refer to chart at end of section.
12JT6A	Refer to type 6JT6A.
12K5	Refer to chart at end of section.
12K7GT	Refer to chart at end of section.
12K8	Refer to chart at end of section.
12KL8	Refer to chart at end of section.
12L6GT	Refer to chart at end of section.
12MD8	Refer to type 6MD8.
12Q7GT	Refer to chart at end of section.
12R5	Refer to chart at end of section.
12S8GT	Refer to chart at end of section.
12SA7 **12SA7GT**	Refer to chart at end of section.
12SC7	Refer to chart at end of section.
12SF5 **12SF5GT**	Refer to chart at end of section.
12SF7	Refer to chart at end of section.
12SG7	Refer to chart at end of section.
12SH7	Refer to chart at end of section.
12SJ7 **12SJ7GT**	Refer to chart at end of section.
12SK7 **12SK7GT**	Refer to chart at end of section.
12SL7GT	Refer to type 6SL7GT.
12SN7GT	Refer to chart at end of section.
12SN7GTA	Refer to type 6SN7GTB.
12SQ7 **12SQ7GT**	Refer to chart at end of section.
12SR7 **12SR7GT**	Refer to chart at end of section.
12T10	Refer to type 6T10.
12U7	Refer to chart at end of section.
12V6GT	Refer to type 6V6.
12W6GT	Refer to type 6W6GT.
12X4	Refer to type 6X4.
12Z3	Refer to chart at end of section.
13CW4	Refer to type 6CW4.
13DE7	Refer to type 6DE7.

Refer to type 6DR7.	**13DR7**
Refer to type 6EM7.	**13EM7/15EAY**
Refer to type 6FD7.	**13FD7**
Refer to type 6FM7.	**13FM7**
Refer to chart at end of section.	**13GB5**
Refer to type 6GB5/EL500.	**13GB5/XL500**
Refer to type 6GF7A.	**13GF7A**
Refer to chart at end of section.	**13J10**

BEAM POWER TUBE— SHARP-CUTOFF PENTODE 13V10



12EZ

Duodecar type used as combined FM detector and audio-frequency output amplifier in television receivers. The beam power unit is used in af output stages and the pentode unit as an FM detector. Outlines section, 8C; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 13.2; amperes, 0.45; average warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Beam Power Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	165	volts
Grid-No.2 (Screen-Grid) Voltage	150	volts
Cathode Current	65	mA
Plate Dissipation	6.5	watts
Grid-No.2 Input	1.8	watts

TYPICAL OPERATION

Plate Voltage	145	volts
Grid-No.2 Voltage	125	volts
Grid-No.1 (Control-Grid) Voltage	—6	volts
Peak AF Grid-No.1 Voltage	6	volts
Zero-Signal Plate Current	34	mA
Maximum-Signal Plate Current	36	mA
Zero-Signal Grid-No.2 Current	2.2	mA
Maximum-Signal Grid-No.2 Current	5.5	mA
Plate Resistance (Approx.)	0.058	megohm
Transconductance	6400	μmhos
Load Resistance	3000	ohms
Total Harmonic Distortion (Approx.)	7	per cent
Maximum-Signal Power Output	1.5	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	0.5	megohm

Pentode Unit as Class A_1 Amplifier

CHARACTERISTICS

Plate Supply Voltage	150	volts
Grid-No.3 (Suppressor-Grid) Voltage	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	100	volts
Cathode-Bias Resistor	560	ohms
Plate Resistance (Approx.)	0.15	megohm
Transconductance, Grid No.1 to Plate	1000	μmhos
Transconductance, Grid No.3 to Plate	400	μmhos
Plate Current	1.3	mA
Grid-No.2 Current	2	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—4.5	volts
Grid-No.3 Voltage (Approx.) for plate current of 10 μA	—4.5	volts

Pentode Unit as FM Detector

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	330	volts
Grid-No.3 Voltage	28	volts
Grid-No.2 Supply Voltage	330	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	1.7	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 165 volts	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	See curve page 98	

13Z10 Refer to type 6Z10.

13Z10/13J10 Refer to type 6Z10.

14A4 Refer to chart at end of section.

14A5 Refer to chart at end of section.

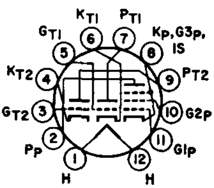
14A7 Refer to chart at end of section.

14AF7 Refer to chart at end of section.

14B6 Refer to chart at end of section.

14B8 Refer to chart at end of section.

14BL11 DUAL TRIODE— SHARP-CUTOFF PENTODE



12GC

Duodecar type used in television receiver applications. The pentode unit is used for video amplifier service, and the triode units for general-purpose use. Outlines section, 8B; requires duodecar 12-contact socket. Heater: volts (ac/dc), 14.2; amperes, 0.45; average warm-up time 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

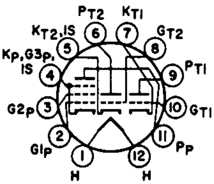
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit No.1	Triode Unit No.2	Pentode Unit	
Plate Voltage	330	330	250	volts
Grid-No.2 (Screen-Grid) Voltage	—	—	125	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	0	volts
Plate Dissipation	1.5	2	2.5	watts
Grid-No.2 Input	—	—	1.25	watts

CHARACTERISTICS	Triode Unit No.1	Triode Unit No.2	Pentode Unit		
Plate Voltage	200	200	35	200	volts
Grid-No.2 Voltage	—	—	100	100	volts
Grid-No.1 Voltage	—	—	0	—	volts
Cathode-Bias Resistor	470	270	—	82	ohms
Amplification Factor	40	69	—	—	
Plate Resistance (Approx.)	7600	12500	—	70000	ohms
Transconductance	5300	5500	—	19000	μmhos
Plate Current	7.2	7.1	40	16	mA
Grid-No.2 Current	—	—	13	3	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	−8	−5.5	—	−5.5	volts

MAXIMUM CIRCUIT VALUES	Triode Unit No.1	Triode Unit No.2	Pentode Unit	
Grid-No.1-Circuit Resistance:				
For fixed-bias operation	0.5	0.5	0.1	megohm
For cathode-bias operation	1	1	0.25	megohm
	0	0	—	volts

DUAL TRIODE— SHARP-CUTOFF PENTODE 14BR11



12GL

Duodecar type used in television receiver applications. The high-mu triode unit No. 1 is used for general-purpose use, the medium-mu triode unit No. 2 for sync separator service, and the pentode unit for video amplifier service. Outlines section, 8C; requires duodecar 12-contact socket. Heater: volts (ac/dc), 14.2; amperes, 0.45; warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit No.1	Triode Unit No.2	Pentode Unit	
Plate Voltage	330	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	—	330	volts
Grid-No.2 Voltage	—	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	0	volts
Plate Dissipation	1.5	2	4	watts
Grid-No.2 Input:				
For grid-No.2 voltages up to 165 volts	—	—	1.1	watts
For grid-No.2 voltages between 165 and 330 volts	—	—	See curve page 98	

CHARACTERISTICS	Triode Unit No.1	Triode Unit No.2	Pentode Unit	Pentode Unit	
Plate Voltage	200	200	35	135	volts
Grid-No.2 Voltage	—	—	135	135	volts
Grid-No.1 Voltage	—2	—	0	—	volts
Cathode-Bias Resistor	—	220	—	100	ohms
Amplification Factor	68	41	—	—	
Plate-Resistance (Approx.)	12400	9400	—	45000	ohms
Transconductance	5500	4400	—	10400	μmhos
Plate Current	7	9.2	34	17	mA
Grid-No.2 Current	—	—	13	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—5.5	—6.5	—	—6	volts

MAXIMUM CIRCUIT VALUES	Triode Unit No.1	Triode Unit No.2	Pentode Unit	
Grid-No.1-Circuit Resistance:				
For fixed-bias operation	0.5	0.5	1	megohm
For cathode-bias operation	1	1	1	megohm

14C5
Refer to chart at end of section.

14C7
Refer to chart at end of section.

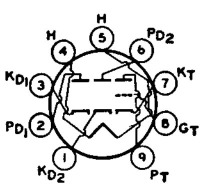
14E6
Refer to chart at end of section.

14E7
Refer to chart at end of section.

14F7
Refer to chart at end of section.

14F8
Refer to chart at end of section.

TWIN DIODE— HIGH-MU TRIODE 14GT8



9KR

Miniature type used as combined detector and af voltage amplifier in radio receivers. Outlines section, 6B; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	14	volts
Heater Current	0.15	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Direct Interelectrode Capacitances:

Triode Unit:		
Grid to Plate	1.8	pF
Grid to Cathode and Heater	1.6	pF
Plate to Cathode and Heater	0.24	pF
Diode Units:		
Diode No.1 Plate to Triode Grid	0.09 max	pF
Diode No.2 Plate to Triode Grid	0.07 max	pF
Either Diode Cathode to All Other Tube Electrodes	6.5	pF
Diode Plate to Cathode and Heater (Each Unit)	2.4	pF

Triode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation	1.1	watts

CHARACTERISTICS, Instantaneous Value

Plate Voltage	250	volts
Grid Voltage	—3	volts
Amplification Factor	72	
Plate Resistance (Approx.)	72000	ohms
Transconductance	1000	µmhos
Plate Current	0.7	mA

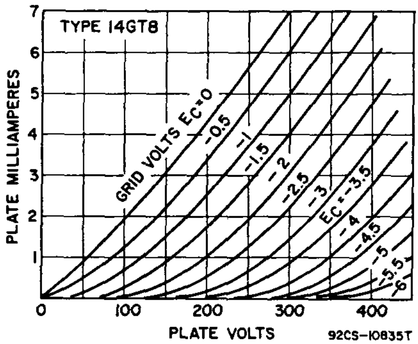
Diode Units (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

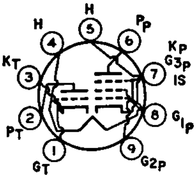
Plate Current	5	mA

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 18 mA	5	volts



14H7	Refer to chart at end of section.
14J7	Refer to chart at end of section.
14JG8	Refer to chart at end of section.
14N7	Refer to chart at end of section.
14Q7	Refer to chart at end of section.
14R7	Refer to chart at end of section.
15	Refer to chart at end of section.
15AF11	Refer to type 6AF11.
15BD11	Refer to type 6BD11.
15CW5	Refer to chart at end of section.
15CW5/PL84	Refer to type 6CW5/EL86.



9HX

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE

15DQ8

Miniature type used in color and black-and-white television receiver applications. The triode unit is used as a sync-separator, sync-amplifier, keyed-agc, or noise-suppressor tube. The pentode unit is used as a video-output tube. Outlines section, 6E; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	15	volts
Heater Current	0.3	ampere
Peak Heater-Cathode Voltage	±200 max	volts

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	550	volts
Peak Plate Voltage, with maximum plate current of 0.1 mA•	600	—	volts
Plate Voltage	250	250	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	550	volts
Grid-No.2 Voltage	—	250	volts
Cathode Current	12	40	mA
Plate Dissipation	1	4	watts
Grid-No.2 Input	—	1.7	watts

CHARACTERISTICS	Triode Unit	Pentode Unit			
Plate Voltage	200	170	200	200	volts
Grid-No.2 Voltage	—	170	200	220	volts
Grid-No.1 Voltage	—1.7	—2.1	—2.9	—3.4	volts
Amplification Factor	65	—	—	—	
Mu-Factor, Grid-No.2 to Grid-No.1	—	36	36	36	
Plate Resistance (Approx.)	—	0.1	0.13	0.15	megohm
Transconductance	4000	11000	10400	10000	μmhos
Plate Current	3	18	18	18	mA
Grid-No.2 Current	—	3	3	3	mA

TYPICAL OPERATION OF PENTODE UNIT AS VIDEO OUTPUT TUBE				
Plate Supply Voltage	170	200	220	volts
Series Plate Resistor	3000	3000	3000	ohms
Grid-No.2 Voltage	170	200	220	volts
Grid-No.1 Voltage	—2	—2.8	—3.3	volts
Transconductance	10400	10000	9700	μmhos
Plate Current	18	18	18	mA
Grid-No.2 Current	3.2	3.1	3.1	mA

MAXIMUM CIRCUIT VALUES	Triode Unit	Pentode Unit	
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	1	1	megohm
For cathode-bias operation	3	2	megohms

• With maximum duty factor of 0.18 and maximum pulse duration of 18 microseconds.

Refer to type 6EW7. **15EW7**

Refer to type 6FM7. **15FM7**

Refer to type 6FY7. **15FY7**

Refer to chart at end of section. **15HB6**

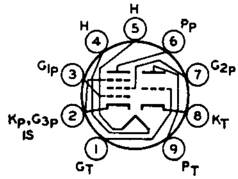
Refer to chart at end of section. **15KY8**

Refer to type 6KY8A. **15KY8A**

Refer to type 6LE8. **15LE8**

16A8

HIGH-MU TRIODE—POWER PENTODE



9EX

Miniature type used in television receiver applications. The triode unit is used as a vertical oscillator or as an af amplifier, and the pentode unit is used as a vertical output tube or as an audio output tube. Outlines section, 6G; requires 9-contact socket. Heater: volts (ac/dc), 16; amperes, 0.3; maximum heater-cathode volts, ±200.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Supply Voltage	550	550	volts
Peak Plate Voltage•	600	2500	volts
Plate Voltage	250	250	volts
Peak Inverse Plate Voltage	—	500	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	550	volts
Grid-No.2 Voltage	—	250	volts
Cathode Current	15	50	mA
Plate Dissipation (Frame Output)	—	5	watts
Plate Dissipation (Audio Output)	—	7	watts
Grid-No.2 Input	—	1.8	watts
Peak Grid-No.2 Input	—	3.2	watts

CHARACTERISTICS	Triode Unit	Pentode Unit				
Plate Voltage	100	100	170	200	200	volts
Grid-No.2 Voltage	—	100	170	200	200	volts
Grid-No.1 Voltage	0	—6	—11.5	—12.5	—16	volts
Amplification Factor	70	—	—	—	—	
Mu Factor, Grid No.2 to Grid No.1	—	10	9.5	9.5	9.5	
Plate Resistance	—	15000	16000	20500	20000	ohms
Transconductance	2500	6800	7500	6800	6400	μmhos
Plate Current	3.5	26	41	35	35	mA
Grid-No.2 Current	—	5	8	6.5	7	mA

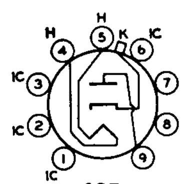
MAXIMUM CIRCUIT VALUES	Triode Unit	Pentode Unit	
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	1	1	megohm
For cathode-bias operation	3	2	megohms

• With a maximum duty factor of 0.04 and maximum pulse duration of 0.8 milliseconds.

16AQ3 Refer to chart at end of section.

16AQ3/ XY88

DIODE



9CB

Miniature type used as booster diodes in line-time-base circuits of transformerless television receivers. Outlines section, 7D; requires miniature 9-contact socket. Heater: volts (ac/dc), 16.4; amperes, 0.6; maximum heater-cathode volts, 6600 peak.

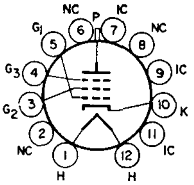
MAXIMUM RATINGS (Design-Center Values)		
Supply Voltage at zero current	550	volts
Supply Voltage	250	volts
Peak Plate Current	550	mA
Average Plate Current	220	mA
Plate Dissipation	5	watts
Peak Negative-Pulse Plate Voltage*	6000#	volts

* Under no conditions should an absolute maximum value of 7500 volts be exceeded.

The pulse duration must not exceed 22 per cent of a cycle, or a maximum of 18 microseconds.

16GK6 Refer to type 6GK6.

16GY5 Refer to type 6GY5.



12GH

16KA6
21KA6

BEAM POWER TUBE

Duodecar type used as horizontal-deflection amplifier in television receivers. Outlines section, 16A; requires duodecar 12-contact socket. A separate connection is provided for grid No.3 to minimize "snivets." Type 21KA6 is identical with type 16KA6 except for heater ratings.

	16KA6	21KA6	
Heater Voltage	15.8	21	volts
Heater Current	0.6	0.45	ampere
Heater Warm-up Time	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	5000	60	60	130	volts
Grid-No.3 (Suppressor-Grid) Voltage	0	0	25	0	volts
Grid-No.2 (Screen-Grid) Voltage	130	130	130	130	volts
Grid-No.1 (Control-Grid) Voltage	—	0	0	—20	volts
Plate Resistance (Approx.)	—	—	—	11000	ohms
Transconductance	—	—	—	9100	μmhos
Plate Current	—	410*	410*	50	mA
Grid-No.3 Current	—	—	2	—	mA
Grid-No.2 Current	—	24*	23*	1.75	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—66	—	—	—33	volts
Triode Amplification Factor	—	—	—	4.7	

* This value may be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

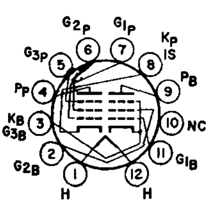
MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
Grid-No.3 Voltage, Positive-bias value	70	volts
Grid-No.2 Voltage	220	volts
Grid-No.1 Voltage, Negative-bias value	55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Average Cathode Current	230	mA
Peak Cathode Current	800	mA
Plate Dissipation	18	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	220	°C

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).



12BT

17AB10
17AB10/
17AX10

PENTODE—
BEAM POWER TUBE

Duodecar type used as a combined limiter, discriminator, and audio power-output tube in FM radio and television receivers. Outlines section, 8C; requires duodecar 12-contact socket.

Heater Volts (ac/dc)	16.8	volts
Heater Current	0.45	ampere
Heater Warm-up Time	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Pentode Unit:		
Grid No.1 to Grid No.3	0.01	pF
Grid No.1 to All Other Electrodes	4.4	pF
Grid No.3 to All Other Electrodes	3.2	pF
Beam Power Unit:		
Grid No.1 to Plate	0.22	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	12	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	7.5	pF

Pentode Unit as Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	135	135	135	volts
Grid-No.3 (Suppressor-Grid) Voltage	4	4	0	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	280	280	volts
Grid-No.2 Voltage	75	—	—	volts
Grid-No.1 (Control-Grid) Voltage	0	0	0	volts
Grid-No.2 Resistor	—	33	33	kohms
Transconductance, Grid-No.1 to Plate	—	—	360	μmhos
Transconductance, Grid-No.3 to Plate	—	—	700	μmhos
Plate Current	—	5	—	mA
Grid-No.2 Current	4.5	—	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—	—	—4	volts
Grid-No.3 Voltage (Approx.) for plate current of 20 μA	—	—	—4	volts

Beam Power Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	165	volts
Grid-No.2 Voltage	150	volts
Plate Dissipation	6.5	watts
Grid-No.2 Input	1.8	watts
Cathode Current	65	mA

CHARACTERISTICS

Plate Voltage	145	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 Voltage	—6	volts
Peak AF Grid-No.1 Voltage	6	volts
Plate Resistance (Approx.)	30000	ohms
Transconductance	8600	μmhos
Zero-Signal Plate Current	36	mA
Maximum-Signal Plate Current	40	mA
Zero-Signal Grid-No.2 Current	3	mA
Maximum-Signal Grid-No.2 Current	9	mA
Load Resistance	3000	ohms
Total Harmonic Distortion (Approx.)	10	per cent
Maximum-Signal Power Output	2.4	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.25	megohm
For cathode-bias operation	0.5	megohm

Pentode Unit as Limiter and Discriminator

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	330	volts
Grid-No.2 Supply Voltage	330	volts
Grid-No.1 Voltage, Peak positive value	60	volts
Average Cathode Current	13	mA

17AX3 Refer to type 6AX3.

17AX4GT Refer to chart at end of section.

17AX4GTA Refer to type 6AX4GTB.

17AX10 Refer to type 17AB10.

Refer to chart at end of section. **17AY3**

Refer to type 6AY3B. **17AY3A**

Refer to type 6BE3. **17BE3**

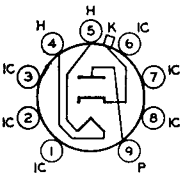
Refer to type 6BE3. **17BE3/17BZ3**

Refer to type 6BF11. **17BF11**

Refer to chart at end of section. **17BH3**

Refer to type 6BH3A. **17BH3A**

Refer to type 6BQ6GTB/6CU6. **17BQ6GTB**



9CB

HALF-WAVE VACUUM RECTIFIER **17BR3**

Miniature type used as damper tube in horizontal-deflection circuits of television receivers. Outlines section, 7D; requires miniature 9-contact socket. Heater: volts (ac/dc), 16.8; amperes, 0.45; warm-up time (average), 11 seconds.

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		5500	volts
Peak Plate Current		1200	mA
Average Plate Current		200	mA
Plate Dissipation		6.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—5500	volts
Average value	+100	—900	volts
Bulb Temperature (At hottest point)		180	°C

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 250 mA	19	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

Refer to chart at end of section. **17BS3**

Refer to type 6BS3A. **17BS3A**

Refer to chart at end of section. **17BZ3**

Refer to chart at end of section. **17C5**

Refer to type 6C9. **17C9**

Refer to type 6CK3. **17CK3**

Refer to chart at end of section. **17CL3**

Refer to type 6CT3. **17CT3**

Refer to type 6CU5. **17CU5**

Refer to type 6CU5. **17CU5/17C5**

Refer to type 6DA4. **17D4**

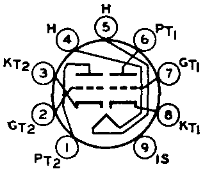
Refer to type 6DE4. **17DE4**

Refer to chart at end of section. **17DM4**

17DM4A	Refer to type 6DM4A.
17DQ6A	Refer to chart at end of section.
17EW8	Refer to chart at end of section.

17EW8/ HCC85

HIGH-MU TWIN TRIODE



9AJ

Miniature type used in rf-amplifier and oscillator-mixer circuits in FM and AM radio receivers. Outlines section, 6B; requires miniature 9-contact socket.

Heater Voltage	17.5	volts
Heater Current	0.15	ampere
Peak Heater-Cathode Voltage	±90 max	volts
Direct Interelectrode Capacitances:		
Plate to Grid (Each Unit)	1.5	pF
Plate to Cathode (Each Unit)	0.18	pF
Plate to Cathode, Heater, and Internal Shield (Each Unit)	1.2	pF
Grid to Cathode, Heater, and Internal Shield (Each Unit)	3	pF
Plate of Unit No.1 to Plate of Unit No.2	0.04 max	pF
Grid of Unit No.1 to Grid of Unit No.2	0.003 max	pF
Plate of Unit No.1 to Grid of Unit No.2	0.008 max	pF
Plate of Unit No.2 to Grid of Unit No.1	0.008 max	pF
Plate of Unit No.1 to Cathode of Unit No.2	0.008 max	pF
Plate of Unit No.2 to Cathode of Unit No.1	0.008 max	pF
Grid of Unit No.1 to Triode of Unit No.2	0.003 max	pF
Grid of Unit No.2 to Triode of Unit No.1	0.003 max	pF

Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	250	volts
Grid-Voltage, Negative-bias Value	100	volts
Cathode Current	15	mA
Plate Dissipation	2.5	watts

CHARACTERISTICS

Plate Voltage	100	170	200	volts
Grid Voltage	—1.1*	—1.5	—2.1	volts
Amplification Factor	50	50	48	
Transconductance	4600	6200	5800	µmhos
Plate Current	4.5	10	10	mA

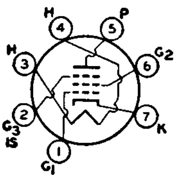
MAXIMUM CIRCUIT VALUE

Grid-Circuit Resistance	1	megohm

* Should not be used if grid current is not permissible.

17GE5	Refer to type 6GE5.
17GJ5	Refer to chart at end of section.
17GJ5A	Refer to type 6GJ5A.
17GT5	Refer to chart at end of section.
17GT5A	Refer to type 6GT5A.
17GV5	Refer to type 6GV5.
17GW6/17DQ6B	Refer to chart at end of section.
17H3	Refer to chart at end of section.
17JB6	Refer to chart at end of section.
17JB6A	Refer to type 6JB6A.

Refer to chart at end of section.	**17JG6**
Refer to type 6JG6A.	**17JG6A**
Refer to chart at end of section.	**17JM6**
Refer to type 6JM6A.	**17JM6A**
Refer to type 6JN6.	**17JN6**
Refer to type 6JQ6.	**17JQ6**
Refer to type 6JR6.	**17JR6**
Refer to chart at end of section.	**17JT6**
Refer to type 6JT6A.	**17JT6A**
Refer to type 6JZ8.	**17JZ8**
Refer to type 6KV6.	**17KV6**
Refer to chart at end of section.	**17LD8**
Refer to chart at end of section.	**17X10**
Refer to chart at end of section.	**18A5**
Refer to chart at end of section.	**18FW6**



7CC

REMOTE-CUTOFF PENTODE 18FW6A

Miniature type used as rf- and if-amplifier tube in ac/dc radio receivers. Outlines section, 5C; requires miniature 7-contact socket.

Heater Voltage	18	volts
Heater Current	0.1	ampere
Heater Warm-up Time (Average)	20	seconds
Peak Heater-Cathode Voltage	±100 max	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.0035 max	
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	5.5	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

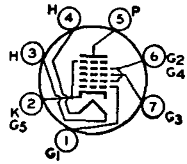
Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Supply Voltage	150	volts
Grid-No.2 Voltage	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	2.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 75 volts	0.6	watt
For grid-No.2 voltages between 75 and 150 volts	See curve page 98	

CHARACTERISTICS

Plate Supply Voltage	100	volts
Grid No.3	Connected to cathode at socket	
Grid-No.2 Supply Voltage	100	volts
Cathode-Bias Resistor	68	ohms
Plate Resistance (Approx.)	0.25	megohm
Transconductance	4400	μmhos
Plate Current	11	mA
Grid-No.2 Current	4.4	mA
Grid-No.1 Voltage (Approx.) for transconductance of 25 μmhos	—20	volts

18FX6 Refer to chart at end of section.

18FX6A PENTAGRID CONVERTER



7CH

Miniature type used for converter applications in ac/dc radio receivers. Outlines section, 5C; requires miniature 7-contact socket. Heater: volts (ac/dc), 18; amperes, 0.1; warm-up time (average), 20 seconds; maximum heater-cathode volts, ±100 peak.

Converter

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	150	volts
Grids-No.2-and-No.4 (Screen-Grid) Supply Voltage	150	volts
Grids-No.2-and-No.4 Voltage	110	volts
Grids-No.2-and-No.4 Input	1.2	watts
Plate Dissipation	1	watt

TPYICAL OPERATION (Separate Excitation)*

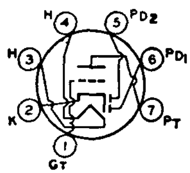
Plate Voltage	100	volts
Grids-No.2-and-No.4 (Screen-Grid) Voltage	100	volts
Grid-No.3 (Control-Grid) Voltage	—1.5	volts
Grid-No.1 (Oscillator-Grid) Resistor	20000	ohms
Plate Resistance (Approx.)	0.4	megohm
Conversion Transconductance	480	μmhos
Plate Current	2.3	mA
Grids-No.2-and-No.4 Current	6.2	mA
Grid-No.1 Current	0.5	mA
Total Cathode Current	9	mA
Grid-No.3 Voltage (Approx.) for conversion transconductance of 10 μmhos	—21	volts

NOTE: The transconductance between grid No.1 and grids No.2 and No.4 connected to plate (not oscillating) is approximately 7000 μmhos under the following conditions: grids No.1 and No.3 at 0 volts; grids No.2 and No.4 and plate at 100 volts. Under the same conditions, the plate current is 24 μA, and the amplification factor is 22.

* The characteristics shown with separate excitation correspond very closely with those obtained in a self-excited oscillator circuit operating with zero bias.

18FY6 Refer to chart at end of section.

18FY6A TWIN DIODE— HIGH-MU TRIODE



7BT

Miniature type used as combined detector, amplifier, and avc tube in compact ac/dc radio receivers. Outlines section, 5C; requires miniature 7-contact socket. Heater: volts (ac/dc), 18; amperes, 0.1; warm-up time (average), 20 seconds; maximum heater-cathode volts, ±100 peak.

Triode Unit as Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	150	volts
Grid Voltage, Positive-bias value	0	volts
Plate Dissipation	0.5	watt

CHARACTERISTICS

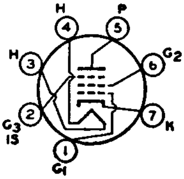
Plate Voltage	100	volts
Grid Voltage	—1	volt
Amplification Factor	100	
Plate Resistance (Approx.)	77000	ohms
Transconductance	1300	μmhos
Plate Current	0.6	mA

Diode Units (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

Plate Current	1	mA

Refer to type 6GB5/EL500. **18GB5**



7BK

SHARP-CUTOFF PENTODE **18GD6A**

Miniature type used in the if, rf, and converter stages of ac/dc AM radio receivers. Outlines section, 5C; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	18	volts
Heater Current	0.1	ampere
Heater Warm-up Time (Average)	20	seconds
Peak Heater-Cathode Voltage	±100 max	volts
Direct Interelectrode Capacitances:•		
Grid No.1 to Plate	0.0035	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	6	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3 and Internal Shield	5	pF

• Values are same without external shield, or with external shield connected to cathode.

Class A_1 Amplifier

CHARACTERISTICS

Plate Supply Voltage	100	volts
Grid No.3 (Suppressor Grid)	Connected to cathode at socket	
Grid-No.2 (Screen-Grid) Voltage	100	volts
Cathode-Bias Resistor	150	ohms
Plate Resistance (Approx.)	0.5	megohm
Transconductance	4300	μmhos
Plate Current	5	mA
Grid-No.2 Current	2	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—4.7	volts

RF Amplifier and Converter

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	150	volts
Grid-No.2 Supply Voltage	150	volts
Grid-No.2 Voltage	See curve page 98	
Plate Dissipation	2.5	watts
Grid-No.2 Input:		
For grid-No.2 voltages up to 75 volts	0.6	watt
For grid-No.2 voltages between 75 and 150 volts	See curve page 98	

Refer to chart at end of section. **19**

Refer to chart at end of section. **19AU4**

Refer to chart at end of section. **19AU4GTA**

Refer to chart at end of section. **19BG6G** **19BG6GA**

Refer to type 6CG3. **19CG3**

Refer to type 6CL8A. **19CL8A**

Refer to type 6EA8. **19EA8**

Refer to type 6EZ8. **19EZ8**

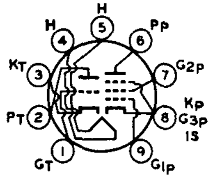
Refer to type 6GQ7. **19GQ7**

Refer to type 6HR6. **19HR6**

Refer to chart at end of section. **19HS6**

19HV8

HIGH-MU TRIODE— SHARP-CUTOFF PENTODE



9FA

Miniature type used as if-amplifier and af voltage-amplifier tube in radio receivers. Outlines section, 6B; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	18.9	volts
Heater Current	0.15	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Direct Interelectrode Capacitances:	Unshielded	Shielded	
Pentode Unit:			
Grid No.1 to Plate	0.016	0.007	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5.5	5.5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2.4	3.4	pF
Heater to Cathode	2.8	2.8	pF
Triode Unit:			
Grid to Plate	0.9	0.9	pF
Grid to Cathode, Cathode of Pentode Unit, Heater, Grid No.3, and Internal Shield	1.7	1.9	pF
Plate to Cathode, Cathode of Pentode Unit, Heater, Grid No.3, and Internal Shield	1.7	2.6	pF
Heater to Cathode	2.8	2.8	pF

Class A_1 Amplifier

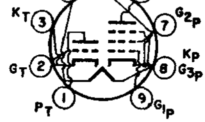
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—	330	volts
Grid-No.2 Voltage	—	See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	0.55	3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.55	watt
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Voltage	100	125	volts
Grid-No.2 Voltage	—	125	volts
Grid-No.1 Voltage	—1	—1	volt
Amplification Factor	70	—	
Plate Resistance (Approx.)	54000	200000	ohms
Transconductance	1300	6500	μmhos
Plate Current	0.8	12	mA
Grid-No.2 Current	—	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 50 μA	—1.5	—	volts
Grid-No.1 Voltage (Approx.) for plate current of 20 μA	—	—9	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

19J6 Refer to chart at end of section.

19JN8 Refer to type 6JN8.

19KG8

MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

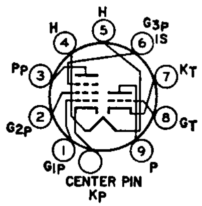


9LY

Miniature type used as combined oscillator and mixer. Outlines section, 6B; requires miniature 9-contact socket. Heater: volts (ac/dc), 18.9; amperes, 0.15; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	300	300	volts
Grid No.2 (Screen-Grid) Supply Voltage	—	300	volts
Grid No.2 Voltage	—	See curve page 98	
Grid No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.5	2.5	watts
Grid No.2 Input:			
For grid-No.2 voltages up to 150 volts	—	0.55	watt
For grid-No.2 voltages between 150 and 300 volts	—	See curve page 98	
CHARACTERISTICS			
Plate Voltage	125	125	volts
Grid-No.2 Voltage	—	125	volts
Grid-No.1 Voltage	—1	—1	volts
Amplification Factor	46	—	
Plate Resistance (Approx.)	5400	200000	ohms
Transconductance	8500	7500	µmhos
Plate Current	13.5	12	mA
Grid-No.2 Current	—	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 µA	—8	—8	volts
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	2.2	2.2	megohms
For cathode-bias operation	2.2	2.2	megohms



10H

MEDIUM-MU TRIODE— SEMIREMOTE-CUTOFF PENTODE 19Q9

Miniature type used as FM rf amplifier and autodyne mixer. **Outlines section**, 6B, except center pin is added to base; requires miniature 10-contact socket. **Heater:** volts, 18.9; amperes, 0.15; warm-up time, 17 seconds; maximum heater-cathode voltage, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Section	Pentode Section		
Plate Voltage	330		330	volts
Grid-No.2 (Screen-Grid) Supply Voltage	—		330	volts
Grid-No.2 Voltage	—		See curve page 98	
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0		0	volts
Plate Dissipation	2.5		3	watts
Grid-No.2 Input	—		0.55	watt
CHARACTERISTICS				
Plate Voltage	125	100	125	volts
Grid-No.2 Voltage	—	70	125	volts
Grid-No.1 Voltage	—1		—1	volt
Amplification Factor	40	—	—	
Plate Resistance	5000	—	200000	ohms
Transconductance	8000	7000	6500	µmhos
Plate Current	14	—	12	mA
Grid-No.2 Current	—	—	4	mA
Grid-No.1 (Approx.) for plate current of 20 µA	—9	—	—9	volts
MAXIMUM CIRCUIT VALUES				
Grid-No.1-Circuit Resistance:				
For fixed-bias operation	0.5		0.25	megohm
For cathode-bias operation	1		1	megohm

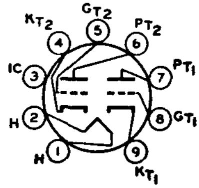
Refer to chart at end of section. **19T8**

Refer to type 6X8A. **19X8**

Refer to chart at end of section. **20**

Refer to chart at end of section. **20EQ7**

20EZ7 HIGH-MU TWIN TRIODE



9PG

Miniature type used in high-gain, resistance-coupled, low-level audio amplifiers such as preamplifiers for stereo phonographs. Outlines section, 6B; requires miniature 9-contact socket. For typical operation as resistance-coupled amplifier, refer to Resistance-Coupled Amplifier section.

Heater Voltage (ac/dc)		20	volts
Heater Current		0.1	ampere
Heater Warm-up Time (Average)		20	seconds
Heater-Cathode Voltage:			
Peak value		±200 max	volts
Average value		100 max	volts
Direct Interelectrode Capacitances:	Unit No.1	Unit No.2	
Grid to Plate	1.5	1.5	pF
Grid to Cathode and Heater	1.6	1.6	pF
Plate to Cathode and Heater	0.2	0.3	pF

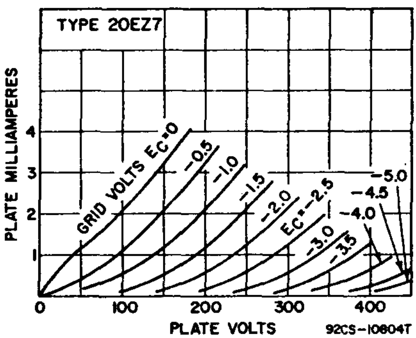
Class A_1 Amplifier (Each Unit)

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	330	volts
Grid Voltage:		
Negative-bias value	55	volts
Positive-bias value	0	volts
Plate Dissipation	1.2	watts

CHARACTERISTICS

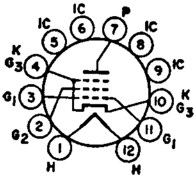
Plate Voltage	100	250	volts
Grid Voltage	—1	—2	volts
Amplification Factor	100	100	
Plate Resistance (Approx.)	80000	62500	ohms
Transconductance	1250	1600	μmhos
Plate Current	0.5	1.2	mA



21EX6 Refer to chart at end of section.

21GY5 Refer to type 6GY5.

21HB5 Refer to chart at end of section.



12BJ

BEAM POWER TUBE 21HB5A

Duodecar type used as horizontal-deflection amplifier in television receivers. Outlines section, 15B; requires duodecar 12-contact socket. For maximum ratings, refer to type 6HB5. Heater: volts (ac/dc), 21; amperes, 0.45; warm-up time (average), 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

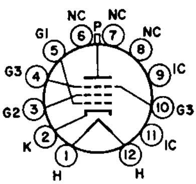
CHARACTERISTICS	Pentode Connection			Triode* Connection	
Plate Voltage	5000	50	130	130	volts
Grid-No.2 (Screen-Grid) Voltage	130	130	130	130	volts
Grid-No.1 (Control-Grid) Voltage	—	0	—20	—20	volts
Amplification Factor	—	—	—	4.8	
Plate Resistance (Approx.)	—	—	9900	—	ohms
Transconductance	—	—	9000	—	μmhos
Plate Current	—	450■	46	—	mA
Grid-No.2 Current	—	29■	1.8	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—64	—	—32	—	volts

* Grid-No.2 tied to plate

■ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Refer to chart at end of section. **21HJ5**

Refer to type 33JV6. **21JV6**



12GD

BEAM POWER TUBE 21JZ6

Duodecar type used as horizontal-deflection amplifier in television receivers. Outlines section, 15C; requires duodecar 12-contact socket. Heater: volts (ac/dc), 21; amperes, 0.45; average warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

CHARACTERISTICS	Triode▲ Connection	Pentode Connection			
Plate Voltage	130	5000	50	130	volts
Grid No.3 (Suppressor Grid)		Connected to cathode at socket			
Grid-No.2 (Screen-Grid) Voltage	—	130	130	130	volts
Grid-No.1 (Control-Grid) Voltage	—20	—	0	—20	volts
Amplification Factor	4.8	—	—	—	
Plate Resistance (Approx.)	—	—	—	9900	ohms
Transconductance	—	—	—	9000	μmhos
Plate Current	—	—	450	46	mA
Grid-No.2 Current	—	—	29	1.8	mA
Grid-No.1 Voltage (Approx.) for plate current of 1.0 mA	—	—64	—	—32	volts

▲ Grid No.2 connected to plate.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage, Positive-bias value	70	volts
Grid-No.2 Voltage	220	volts

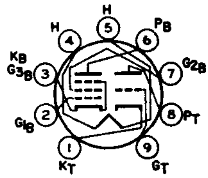
DC Grid-No.1 Voltage, Negative-bias value	55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	800	mA
Average Cathode Current	230	mA
Plate Dissipation•	18	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	220	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	1	megohm

• A bias resistor or other means is required to protect the tube in absence of excitation.
Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

21KA6 Refer to type 16KA6.

21LR8

HIGH-MU TRIODE— BEAM POWER TUBE



9QT

Novar type used in combined vertical-deflection-oscillator and vertical-deflection-amplifier applications in color and black-and-white television receivers. Outlines section, 17E; requires novar 9-contact socket. Heater: volts, 21; amperes, 0.45; average warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

CHARACTERISTICS	Triode Unit	Beam Power Unit			
Plate Voltage	250	45	135	120	volts
Grid-No.2 (Screen-Grid) Voltage	—	125	120	120•	volts
Grid-No.1 (Control-Grid) Voltage	—4	0	—10	—10	volts
Amplification Factor	58	—	—	6.5	
Plate Resistance (Approx.)	14000	—	14000	—	ohms
Transconductance	4100	—	9200	—	µmhos
Plate Current	2.6	200▪	51	—	mA
Grid-No.2 Current	—	200▪	3	—	mA
Grid-No.1 Voltage:					
For plate current of 10 µA	—6.6	—	—	—	volts
For plate current of 100 µA	—	—	—28	—	volts
For plate current of 1 mA	—	—	—24	—	volts

• Triode connection, Grid No.2 connected to plate at socket.
▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Vertical-Deflection Oscillator and Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit Oscillator	Beam Power Unit Amplifier	
Plate Voltage	400	400	volts
Grid-No.2 Voltage	—	300	volts
Peak Positive-Pulse Plate Voltage#	—	2500	volts
Peak Negative-Pulse Grid-No.1 Voltage	400	250	volts
Peak Cathode Current	105	260	mA
Average Cathode Current	30	75	mA
Peak Power Output	2.5	—	watts
Plate Dissipation‡	2.5	14	watts
Grid-No.2 Input‡	—	2.75	watts
Bulb Temperature	—	210	°C
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	—	1	megohm
For cathode-bias operation	2.2	2.2	megohms

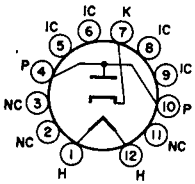
Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
‡ A bias resistor or other means is required to protect the tube in absence of excitation.

21LU8 Refer to type 6LU8.

22 Refer to chart at end of section.

22BH3 Refer to chart at end of section.

22BH3A Refer to type 6BH3A.



12FX

HALF-WAVE VACUUM RECTIFIER 22BW3

Duodecar type used as damper tube in horizontal-deflection circuits of television receivers. Outlines section, 8D; requires duodecar 12-contact socket.

Heater Voltage (ac/dc)	22.4	volts
Heater Current	0.45	ampere
Heater Warm-up Time	11	seconds
Direct Interelectrode Capacitances:		
Cathode to Heater and Plate	8.5	pF
Plate to Cathode and Heater	6	pF
Heater to Cathode	3.8	pF

Damper Service

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		5000	volts
Peak Plate Current		1100	mA
Average Plate Current		175	mA
Plate Dissipation		6.5	watts
Heater-Cathode Voltage:			
Peak value	+300	—5000	volts
Average value	+100	—900	volts

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 350 mA	32	volts

Pulse duration must not exceed 15% of one horizontal scanning cycle (10 microseconds)

Refer to type 6DE4. **22DE4**

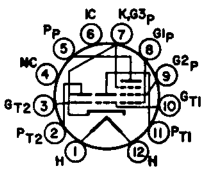
Refer to chart at end of section. **22JG6**

Refer to type 6JG6A. **22JG6A**

Refer to type 6JR6. **22JR6**

Refer to type 6JU6. **22JU6**

Refer to type 6KM6. **22KM6**



12GZ

DUAL TRIODE— BEAM POWER TUBE 23Z9

Duodecar type used in combined vertical-deflection-oscillator and vertical-deflection-amplifier applications in television receivers. Outlines section, 8B; requires duodecar 12-contact socket. Heater: volts (ac/dc), 23; amperes, 0.45; average warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

CHARACTERISTICS	Triode Unit No.1	Triode Unit No.2	Beam Power Unit		
Plate Voltage	150	150	45	120	volts
Grid-No.2 (Screen-Grid) Voltage	—	—	110	110	volts
Grid-No.1 (Control-Grid) Voltage	—2	—5	0	—8	volts
Amplification Factor	43	20	—	—	
Plate Resistance (Approx.)	11000	8500	—	11700	ohms
Transconductance	3900	2350	—	7100	µmhos
Plate Current	5.4	5.5	122	46	mA

	Triode Unit No.1	Triode Unit No.2	Beam Power Unit		
Grid-No.2 Current	—	—	16.5	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—	—	—	—25	volts
Grid Voltage (Approx.) for plate current of 10 μA	—5.7	—11	—	—	volts

Vertical-Deflection Oscillator and Amplifier

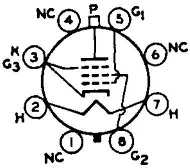
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit No.1 Amplifier	Triode Unit No.2 Oscillator	Beam Power Unit Amplifier	
Plate Voltage	330	250	250	volts
Peak Positive-Pulse Plate Voltage#	—	—	2000	volts
Grid-No.2 Voltage	—	—	200	volts
Peak Negative-Pulse Grid-No.1 Voltage	—	400	150	volts
Grid Voltage, Positive-bias value	0	—	—	volts
Plate Dissipation	125	1	7	watts
Grid-No.2 Input	—	—	1.8	watts
Peak Cathode Current	—	—	245	mA
Average Cathode Current	—	—	70	mA
Peak Plate Current	—	70	—	mA
Average Plate Current	—	20	—	mA
MAXIMUM CIRCUIT VALUES				
Grid-No.1-Circuit Resistance: For fixed-bias operation	0.5	1	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

24A	Refer to chart at end of section.
24JE6A	Refer to type 6JE6A.
24JE6B	Refer to type 6LQ6.
24LQ6	Refer to type 6LQ6.
24LQ6/24JE6B	Refer to type 6LQ6.
25A6 **25A6GT**	Refer to chart at end of section.
25A7GT	Refer to chart at end of section.
25AC5GT	Refer to chart at end of section.
25AV5GA	Refer to type 6AV5GA.
25AX4GT	Refer ty type 6AX4GTB.
25B5	Refer to chart at end of section.
25B6G	Refer to chart at end of section.
25B8GT	Refer to chart at end of section.
25BK5	Refer to chart at end of section.
25BQ6GT	Refer to chart at end of section.
25BQ6GTB/25CU6	Refer to type 6BQ6GTB/6CU6.
25C5	Refer to type 50C5.
25C6G	Refer to chart at end of section.
25CA5	Refer to type 6CA5.
25CD6GA	Refer to chart at end of section.
25CD6GB	Refer to type 6CD6GA.
25CG3	Refer to type 6CG3.
25CM3	Refer to type 6CM3.

Refer to type 6BQ6GTB/6CU6. **25CU6**



5BT

BEAM POWER TUBE **25DN6**

Glass octal type used as horizontal-deflection amplifier in television receivers. Outlines section, 21; requires octal socket. Vertical tube mounting is preferred but horizontal operation is permissible if pins 1 and 3 are in vertical plane.

Heater Voltage (ac/dc)	25	volts
Heater Current	0.6	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS			
Plate Voltage	50	125	volts
Grid-No.2 (Screen-Grid) Voltage	100	125	volts
Grid-No.1 (Control-Grid) Voltage	0	—18	volts
Mu Factor, Grid-No.2 to Grid No.1	—	4.35	
Plate Resistance	—	4000	ohms
Transconductance	—	9000	µmhos
Plate Current	240•	70	mA
Grid-No.2 Current	30•	6.3	mA
Grid-No.1 Voltage (Approx.) for plate current of 0.5 mA	—	—36	volts

• These values can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Horizontal-Deflection Amplifier

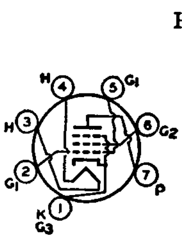
For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Center Values)		
DC Plate Voltage	700	volts
Peak Positive-Pulse Plate Voltage# (Absolute Maximum)	6600□	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.2 (Screen-Grid) Voltage	175	volts
Peak Negative-Pulse Grid-No.1 (Control-Grid) Voltage	200	volts
Peak Cathode Current	700	mA
Average Cathode Current	200	mA
Plate Dissipation†	15	watts
Grid-No.2 Input	3	watts
Bulb Temperature (At hottest point)	225	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	0.47	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
□ Under no circumstances should this absolute value be exceeded.
† A bias resistor or other means is required to protect the tube in absence of excitation.

Refer to chart at end of section. **25EC6**

Refer to type 6EH5. **25EH5**



7CV

BEAM POWER TUBE **25F5A**

Miniature type used in audio-output stage of ac/dc radio receivers employing series-connected heater strings. Outlines section, 5D; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	25	volts
Heater Current	0.15	ampere

Heater Warm-up Time (Average)	17	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.44	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No. 3	12	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	8	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	5.5	watts
Grid-No.2 Input	1.1	watts
Bulb Temperature (At hottest point)	220	°C

TYPICAL OPERATION AND CHARACTERISTICS

Plate Voltage	110	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 Voltage	—7.5	volts
Peak AF Grid-No.1 Voltage	—7.5	volts
Plate Resistance (Approx.)	13000	ohms
Transconductance	6400	µmhos
Zero-Signal Plate Current	43	mA
Maximum-Signal Plate Current	45	mA
Zero-Signal Grid-No.2 Current	3.8	mA
Maximum-Signal Grid-No.2 Current	7.3	mA
Effective Load Resistance	2500	ohms
Total Harmonic Distortion	7	per cent
Maximum-Signal Power Output	1.5	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

Push-Pull Class AB_1 Amplifier

MAXIMUM RATINGS (Same as for class AB_1 amplifier)

TYPICAL OPERATION (Values are for two tubes)

Plate Voltage	110	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 Voltage	—8	volts
Peak AF Grid-No.1-to-Grid-No.1 Voltage	14.4	volts
Zero-Signal Plate Current	82	mA
Maximum-Signal Plate Current	88	mA
Zero-Signal Grid-No.2 Current	7.2	mA
Maximum-Signal Grid-No.2 Current	12.5	mA
Effective Load Resistance (Plate-to-plate)	4500	ohms
Total Harmonic Distortion	2.6	per cent
Maximum-Signal Power Output	2.9	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

25JQ6 Refer to type 6JQ6.

25L6 Refer to chart at end of section.

25L6GT Refer to chart at end of section.

25N6G Refer to chart at end of section.

25W4GT Refer to chart at end of section.

25W6GT Refer to type 6W6GT.

25Y5 Refer to chart at end of section.

25Z5 Refer to chart at end of section.

25Z6
25Z6GT Refer to chart at end of section.

26 Refer to chart at end of section.

Refer to chart at end of section. **27**

Refer to type 6GB5/EL500. **27GB5/PL500**

Refer to chart at end of section. **30**

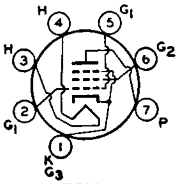
Refer to type 6AG11. **30AG11**

Refer to chart at end of section. **31**

Refer to type 6JS6A. **31JS6A**

Refer to chart at end of section. **32**

Refer to chart at end of section. **32ET5**



7CV

POWER PENTODE 32ET5A

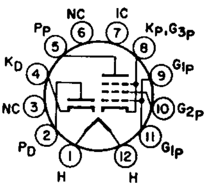
Miniature type used in audio output stage of compact ac/dc radio receivers, **Outlines section**, 5D; requires miniature 7-contact socket. **Heater:** volts (ac/dc), 32; amperes, 0.1; warm-up time (average), 20 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Plate Dissipation	5.4	watts
Grid-No.2 Input	1.2	watts
TYPICAL OPERATION AND CHARACTERISTICS		
Plate Voltage	110	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 (Control-Grid) Voltage	—7.5	volts
Peak AF Grid-No.1 Voltage	7.5	volts
Zero-Signal Plate Current	30	mA
Zero-Signal Grid-No.2 Current	2.8	mA
Plate Resistance (Approx.)	21500	ohms
Transconductance	5500	μmhos
Load Resistance	2800	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	1.2	watts
MAXIMUM CIRCUIT VALUES		
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	0.1	megohm

Refer to chart at end of section. **32L7GT**

Refer to chart at end of section. **33**



12FC

DIODE—BEAM POWER TUBE 33GT7

Duodecar type used in television receiver applications. The diode unit is used for damper service and the beam power unit for horizontal-deflection amplifier service. **Outlines section**, 15A; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 33.6; amperes, 0.45; warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Beam Power Unit as Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection			Triode† Connection	
Plate Voltage	3500	60	130	130	volts
Grid-No.2 (Screen-Grid) Voltage	130	130	130	130	volts
Grid-No.1 (Control-Grid) Voltage	—	0	—22.5	—22.5	volts
Amplification Factor	—	—	—	4	
Plate Resistance (Approx.)	—	—	10000	—	ohms
Transconductance	—	—	6500	—	µmhos
Plate Current	—	320	48	—	mA
Grid-No.2 Current	—	22	2.9	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—60	—	—40	—	volts

† Grid No.2 tied to plate.

Beam Power Unit as Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Ratings)

Plate Voltage	400	volts
Peak Positive-Pulse Plate Voltage#	3500	volts
Peak Negative-Pulse Plate Voltage	0	volts
Grid-No.2 Voltage	150	volts
DC Grid-No.1 DC Voltage, Negative-bias value	55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Average Cathode Current	140	mA
Peak Cathode Current	490	mA
Plate Dissipation▪	9	watts
Grid-No.2 Input	2.5	watts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	1	megohm

▪ A bias resistor or other means is required to protect the tube in absence of excitation.

Damper Service—Diode Unit

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		2500	volts
Peak Plate Current		750	mA
Average Plate Current		125	mA
Plate Dissipation		3.5	watts
Heater-Cathode Voltage:			
Peak value	+200	—2500	volts
Average value	+100	—400	volts
Bulb Temperature (at hottest point)		220	°C

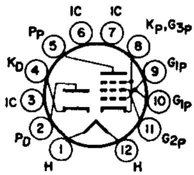
CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 250 mA	21	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

33GY7 Refer to chart at end of section.

33GY7A DIODE—BEAM POWER TUBE



12FN

Duodecar type used as combined damper diode and horizontal-deflection amplifier in television receivers. Socket terminals 1, 3, 6 and 7 should not be used as tie points. *Outlines section,* 15A; requires duodecar 12-contact socket. Heater: volts (ac/dc), 33.6; amperes, 0.45; warm-up time (average), 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Beam Power Unit as Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection			Triode* Connection	
Plate Voltage	5000	60	130	130	volts
Grid-No.2 (Screen-Grid) Voltage	130	130	130	130	volts
Grid-No.1 (Control-Grid) Voltage	—	0	—22.5	—22.5	volts
Amplification Factor	—	—	—	4	
Plate Resistance (Approx.)	—	—	10000	—	ohms
Transconductance	—	—	6500	—	µmhos

	Pentode Connection			Triode* Connection	
Plate Current	—	320▪	48	—	mA
Grid-No.2 Current	—	22▪	2.9	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—80	—	—40	—	volts

* Grid No.2 tied to plate.

▪ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

Beam Power Unit as Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	400	volts
Peak Positive-Pulse Plate Voltage#	5000	volts
Peak Negative-Pulse Plate Voltage	0	volts
DC Grid-No.2 Voltage	150	volts
DC Grid-No.1 Voltage	—55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Peak Cathode Current	540	mA
Average Cathode Current	155	mA
Plate Dissipation†	9	watts
Grid-No.2 Input	3	watts

MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	1	megohm

\# Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).
† A bias resistor or other means is required to protect the tube in absence of excitation.

Damper Service (Diode Unit)

For operation in a 525-line, 30-frame system

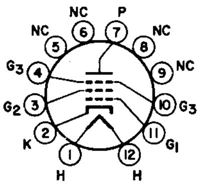
MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		4200	volts
Peak Plate Current		810	mA
Average Plate Current		135	mA
Plate Dissipation		3.8	watts
Heater-Cathode Voltage:			
Peak value	+200	—4200	volts
Average value	+100	—400	volts
Bulb Temperature (At hottest point)		200	°C

CHARACTERISTICS, Instantaneous Value

Tube Voltage Drop for plate current of 250 mA	21	volts

\# Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).



12FK

33JV6

BEAM POWER TUBE

21JV6

Duodecar type used as horizontal-deflection amplifier in television receivers. Outlines section, 15B; requires duodecar 12-contact socket. Type 21JV6 is identical with type 33JV6 except for heater ratings.

	21JV6	33JV6	
Heater Voltage (ac/dc)	21	33	volts
Heater Current	0.45	0.3	ampere
Heater Warm-up Time (Average)	11	11	seconds
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts

Class A_1 Amplifier

CHARACTERISTICS	Triode• Connection	Pentode Connection			
Plate Voltage	130	5000	60	130	volts
Grid No.3 (Suppressor Grid)		Connected to cathode at socket.			
Grid-No.2 (Screen-Grid) Voltage	130	130	130	130	volts
Grid-No.1 (Control-Grid) Voltage	—20	—	0	—20	volts
Plate Resistance (Approx.)	—	—	—	11000	ohms
Transconductance	—	—	—	9100	μmhos
Plate Current	—	—	410	50	mA
Grid-No.1 Current	—	—	24	1.75	mA

	Triode Connection	Pentode Connection			
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—66	—	—33	volts
Amplification Factor	4.7	—	—	—	

• Grid No.2 tied to plate.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

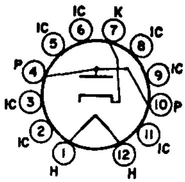
MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6000	volts
Peak Negative-Pulse Plate Voltage	1500	volts
DC Grid-No.3 Voltage	70	volts
DC Grid-No.2 Voltage	220	volts
DC Grid-No.1 Voltage, Negative-bias value	55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Average Cathode Current	230	mA
Peak Cathode Current	800	mA
Plate Dissipation**	18	watts
Grid-No.2 Input	3.5	watts
Bulb Temperature (At hottest point)	220	°C
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	1	megohm

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

** A bias resistor or other means is required to protect the tube in absence of excitation.

34 Refer to chart at end of section.



12GK

34CE3 HALF-WAVE VACUUM RECTIFIER

Duodecar type used as damper tube in television receivers. Outlines section, 8G; requires duodecar 12-contact socket. Heater: volts, 34.5; amperes, 0.45; warm-up time, 11 seconds.

Damper Service

For operation in a 525-line, 30-frame system

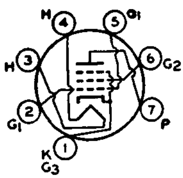
MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		5000	volts
Peak Plate Current		1500	mA
Average Plate Current		350	mA
Plate Dissipation		11	watts
Heater-Cathode Voltage:			
Peak value	+300	—5000	volts
Average value	+100	—900	volts
CHARACTERISTICS, Instantaneous Value			
Tube Voltage Drop for plate current of 680 mA		20	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

34CM3 Refer to type 6CM3.

34GD5 Refer to chart at end of section.



7CV

34GD5A BEAM POWER TUBE

Miniature type used in audio output stages of compact ac/dc radio receivers. Outlines section, 5D; requires miniature 7-contact socket. Heater: volts (ac/dc), 34; amperes 0.1; warm-up time, 20 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

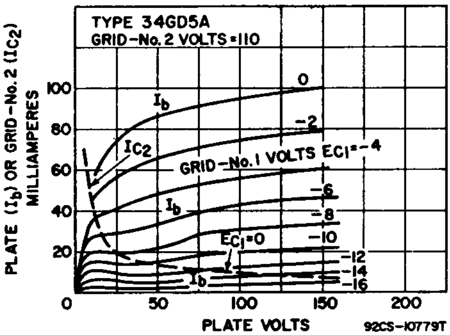
Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Grid-No.1 (Control-Grid) Voltage:		
Negative-bias value	50	volts
Positive-bias value	0	volts
Plate Dissipation	5	watts
Grid-No.2 Input	1.1	watts
Bulb Temperature (At hottest point)	250	°C

TYPICAL OPERATION AND CHARACTERISTICS

Plate Voltage	110	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 Voltage	—7.5	volts
Peak AF Grid-No.1 Voltage	7.5	volts
Zero-Signal Plate Current	35	mA
Zero-Signal Grid-No.2 Current	3	mA
Plate Resistance (Approx.)	13000	ohms
Transconductance	5700	μmhos
Load Resistance	2500	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	1.4	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

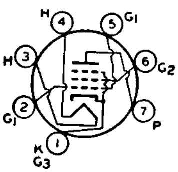


Refer to chart at end of section. **35**

Refer to chart at end of section. **35A5**

Refer to chart at end of section. **35B5**

BEAM POWER TUBE **35C5**



7CV

Miniature type used in output stage of compact, ac/dc radio receivers. Outlines section, 5D; requires miniature 7-contact socket. This tube, like other power-handling tubes, should be adequately ventilated. Except for terminal connections and slightly higher ratings, type 35C5 is equivalent in performance to miniature type 35B5 and, within its maximum ratings, to glass octal type 35L6GT.

Heater Voltage (ac/dc)	35	volts
Heater Current	0.15	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts

Direct Interelectrode Capacitances (Approx.):

Grid No.1 *to* Plate	0.6	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	12	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	9	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Plate Dissipation	5.2	watts
Grid-No.2 Input	1.1	watts
Bulb Temperature (At hottest point)	250	°C

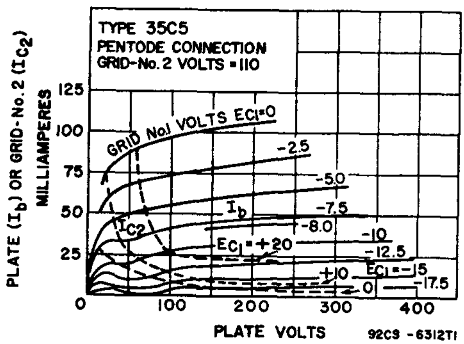
TYPICAL OPERATION

Plate Voltage	110	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 (Control-Grid) Voltage	—7.5	volts
Peak AF Grid-No.1 Voltage	7.5	volts
Zero-Signal Plate Current	40	mA
Maximum-Signal Plate Current	41	mA
Zero-Signal Grid-No.2 Current	3	mA
Maximum-Signal Grid-No.2 Current	7	mA
Plate Resistance (Approx.)	13000	ohms
Transconductance	5800	µmhos
Load Resistance	2500	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	1.5	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:

For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm



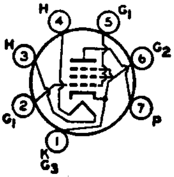
Installation and Application

The 35-volt heater is designed to operate under the normal conditions of line-voltage variation without materially affecting the performance or serviceability of the 35C5. For operation of the 35C5 in series with other types having 0.15-ampere rating, the current in the heater circuit should be adjusted to 0.15 ampere for the normal supply voltage.

In a series-heater circuit of the "dc-power line" type employing several 0.15-ampere types and one or two 35C5s, the heater(s) of the 35C5(s) should be placed on the positive side of the line. Under these conditions, heater-cathode voltage of the 35C5 must not exceed the value given under maximum ratings. In a series-heater circuit of the "universal" type employing rectifier tube 35W4, one or two 35C5s and several 0.15-ampere types, it is recommended that the heater(s) of the 35C5(s) be placed in the circuit so that the higher values of heater-cathode bias will be impressed on the 35C5(s) rather than on the other 0.15-ampere types. This is accomplished by arranging the 35C5(s) on the side of the supply line which is connected to the cathode of the rectifier, i.e., the positive terminal of the rectified voltage supply. Between this side of the line and the 35C5(s), any necessary auxiliary resistance and the heater of the 35W4 are connected in series.

As a power amplifier (class A_1), the 35C5 is recommended for use either singly or in push-pull combination in the power-output stage of ac/dc receivers. The operating values shown under typical operation have been determined on the basis that grid-No.1 current does not flow during any part of the input cycle.

Refer to chart at end of section. **35DZ8**



7CV

POWER PENTODE 35EH5

Miniature type used in the audio output stage of radio and television receivers and in phonographs. Outlines section, 5D; requires miniature 7-contact socket.

Heater Voltage (ac/dc)	35	volts
Heater Current	0.15	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.65	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	17	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	9	pF

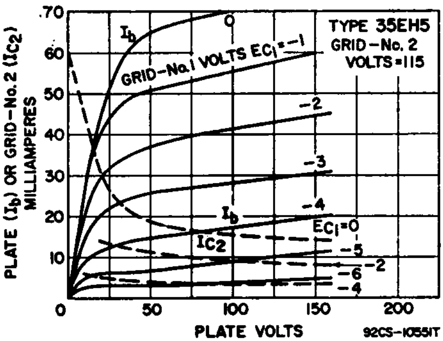
Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	5	watts
Grid-No.2 Input	1.75	watts
Bulb Temperature (At hottest point)	225	°C

TYPICAL OPERATION

Plate Supply Voltage	110	volts
Grid-No.2 Supply Voltage	115	volts
Cathode-Bias Resistor	62	ohms
Peak AF Grid-No.1 Voltage	3	volts
Zero-Signal Plate Current	32	mA
Maximum-Signal Plate Current	32	mA
Zero-Signal Grid-No.2 Current	7.2	mA
Maximum-Signal Grid-No.2 Current	12	mA
Plate Resistance (Approx.)	14000	ohms
Transconductance	3000	μmhos
Load Resistance	3000	ohms
Total Harmonic Distortion	8	per cent
Maximum-Signal Power Output	1.2	watts

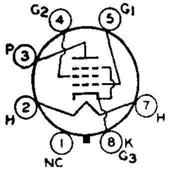


MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

35GL6 Refer to chart at end of section.

35L6GT BEAM POWER TUBE



7AC

Glass octal type used in output stage of ac/dc radio receivers. **Outlines section,** 13D; requires octal socket. This type may be supplied with pin No.1 omitted. Refer to miniature type 35C5 for installation, application information, and curves.

Heater Voltage (ac/dc)	35	volts
Heater Current	0.15	ampere
Peak Heater-Cathode Voltage	±90 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.6	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	13	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	9.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)

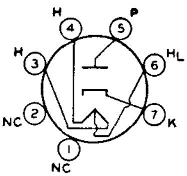
Plate Voltage	200	volts
Grid-No.2 (Screen-Grid) Voltage	125	volts
Plate Dissipation	8.5	watts
Grid-No.2 Input	1	watt

TYPICAL OPERATION	Fixed Bias	Cathode Bias	
Plate Supply Voltage	110	200	volts
Grid-No.2 Supply Voltage	110	125	volts
Grid-No.1 (Control-Grid) Voltage	—7.5	—	volts
Cathode-Bias Resistor	—	180	ohms
Peak AF Grid-No.1 Voltage	7.5	8	volts
Zero-Signal Plate Current	40	43	mA
Maximum-Signal Plate Current	41	43	mA
Zero-Signal Grid-No.2 Current	3	2	mA
Maximum-Signal Grid-No.2 Current	7	5.5	mA
Plate Resistance	14000	34000	ohms
Transconductance	5800	6100	μmhos
Load Resistance	2500	5000	ohms
Total Harmonic Distortion	10	10	per cent
Maximum-Signal Power Output	1.5	3	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

35W4 HALF-WAVE VACUUM RECTIFIER



5BQ

Miniature type used in power supply of ac/dc receivers. **Outlines section,** 5D; requires miniature 7-contact socket. This type is equivalent in performance to glass-octal type 35Z5GT. The heater is provided with a tap for operation of a panel lamp.

	*	**	
Heater Voltage (ac/dc):			
Entire Heater (pins 3 and 4)	35	32	volts
Panel Lamp Section (pins 4 and 6)	7.5	5.5	volts
Heater Current:			
Between Pins 3 and 4	0.15	—	ampere
Between Pins 3 and 6	—	0.15	ampere
Peak Heater-Cathode Voltage		±360 max	volts

* Without panel lamp.

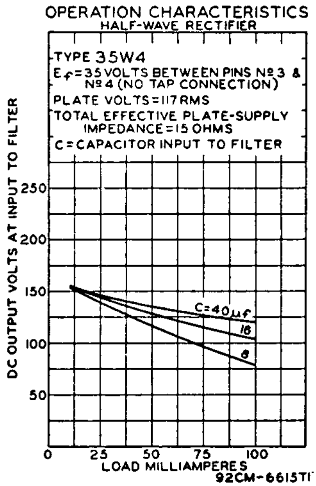
** With No.40 or No.47 panel lamp.

Half-Wave Rectifier

MAXIMUM RATINGS (Design-Maximum Values)

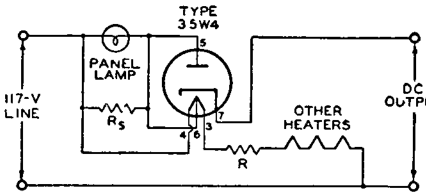
Peak Inverse Plate Voltage	360	volts
Peak Plate Current	660	mA
Average Output Current:		
With Panel Lamp and No Shunting Resistor	66	mA
With Panel Lamp and Shunting Resistor	100	mA
Without Panel Lamp	110	mA
Panel-Lamp-Section Voltage:		
When Panel Lamp Fails	17	volts

Installation and Application



For heater considerations, refer to miniature type 35C5.

With the panel lamp connected as shown in the diagram, the drop across R and all heaters (with panel lamp) should equal 117 volts at 0.15 ampere. The shunting resistor R_s is required when dc output current exceeds 60 milliamperes. Values of R_s for dc output currents greater than 60 milliamperes are given in tabulated data.



AC Plate-Supply Voltage (rms)	117	117	117	117	volts
Filter-Input Capacitor	40	40	40	40	μF
Minimum Total Effective Plate-Supply Impedance	15	15	15	15	ohms
Panel-Lamp Shunting Resistor	—	300	150	100	ohms
Average Output Current	60	70	80	90	mA

† No.40 or No.47 panel lamp used in circuit given below with capacitor-input filter.

TYPICAL OPERATION WITHOUT PANEL LAMP

AC Plate-Supply Voltage (rms)	117	volts
Filter-Input Capacitor	40	μF
Minimum Total Effective Plate-Supply Impedance	15	ohms
Average Output Current	100	mA
DC Output Voltage at Input to Filter (Approx.):		
At half-load current (50 mA)	135	volts
At full-load current (100 mA)	120	volts
Voltage Regulation (Approx.):		
Half-load to full-load current	15	volts

MAXIMUM CIRCUIT VALUES

Panel-Lamp Shunting Resistor:*		
For dc output current of 70 mA	800	ohms
For dc output current of 80 mA	400	ohms
For dc output current of 90 mA	250	ohms

* Required when dc output current is greater than 60 milliamperes.

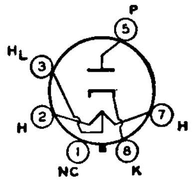
Refer to chart at end of section. **35Y4**

Refer to chart at end of section. **35Z3**

Refer to chart at end of section. **35Z4GT**

35Z5GT

HALF-WAVE VACUUM RECTIFIER



6AD

Glass octal type used in power supply of ac/dc receivers. The heater is provided with a tap for operation of a panel lamp. **Outlines section, 13D**; requires octal socket. This type may be supplied with pin No.1 omitted. For installation and application considerations, refer to miniature type 35W4.

	*	**	
Heater Voltage (ac/dc):			
Entire Heater (pins 2 and 7)	35	32	volts
Panel Lamp Section (pins 2 and 3)	7.5	5.5	volts
Heater Current:			
Between Pins 2 and 7	0.15	—	ampere
Between Pins 3 and 7	—	0.15	ampere
Peak Heater-Cathode Voltage		±350 max	volts

* Without panel lamp.

** With No.40 or No.47 panel lamp.

Half-Wave Rectifier

MAXIMUM RATINGS (Design-Center Values)

Peak Inverse Plate Voltage	700	volts
Peak Plate Current	600	mA
Average Output Current:		
With Panel Lamp and No Shunting Resistor	60	mA
With Panel Lamp and Shunting Resistor	90	mA
Without Panel Lamp	100	mA
Panel-Lamp-Section Voltage (rms):		
When Panel Lamp Fails	15	volts

TYPICAL OPERATION WITH PANEL LAMP†

AC Plate-Supply Voltage (rms)	117	117	117	117	235	volts
Filter-Input Capacitor	40	40	40	40	40	µF
Minimum Total Effective Plate-Supply Impedance	15	15	15	15	100	ohms
Panel-Lamp Shunting Resistor	—	300	150	100	—	ohms
Average Output Current	60	70	80	90	60	mA

† No.40 or No.47 panel lamp used in circuit with capacitor-input filter given under type 35W4.

TYPICAL OPERATION WITHOUT PANEL LAMP†

AC Plate-Supply Voltage (rms)	117	235	volts
Filter-Input Capacitor	40	40	µF
Minimum Total Effective Plate-Supply Impedance	15	100	ohms
Average Output Current	100	100	mA
DC Output Voltage at Input to Filter (Approx.):			
At half-load current (50 mA)	140	280	volts
At full-load current (100 mA)	120	235	volts
Voltage Regulaton (Approx.):			
Half-load to full-load current	20	45	volts

MAXIMUM CIRCUIT VALUES

Panel-Lamp Shunting Resistor•:		
For dc ouptut current of 70 mA	800	ohms
For dc ouptut current of 80 mA	400	ohms
For dc ouptut current of 90 mA	250	ohms

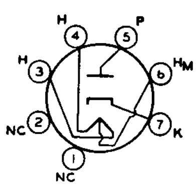
• Required when dc output current is greater than 60 milliamperes.

36 Refer to chart at end of section.

36AM3
36AM3A Refer to chart at end of section.

36AM3B

HALF-WAVE VACUUM RECTIFIER



5BQ

Miniature type used in power supply of ac/dc receivers. This type has a tapped heater so that the heater section between pins 4 and 6 can be used as a limiting resistance in the rectifier plate circuit. This heater section is not to be used as a panel-lamp shunt. **Outlines section, 5D**; requires miniature 7-contact socket.

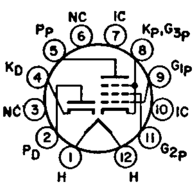
Heater Voltage (ac/dc):			
Entire Heater (Pins 3 and 4)		36	volts
Tap Section (Pins 3 and 6)		32	volts
Heater Current (Pins 3 and 6)		0.1	ampere
Heater Warm-up Time (Average)		20	seconds
Heater-Cathode Voltage:			
Peak value	+200	—350 max	volts
Average value	+100	—350 max	volts

Half-Wave Rectifier

MAXIMUM RATINGS (Design-Maximum Values)			
Peak Inverse Plate Voltage		365	volts
Peak Plate Current		580	mA
Average Output Current		82	mA
TYPICAL OPERATION WITH CAPACITOR INPUT TO FILTER			
AC Plate-Supply Voltage (rms)	120	117	volts
Filter-Input Capacitor	40	40	µF
Total Effective Plate Supply Resistance		See text above	
Average Output Current	75	75	mA
DC Output Voltage	118	105	volts
CHARACTERISTICS			
Tube Voltage Drop for plate current of 150 mA	16	20	volts

Refer to chart at end of section. **37**

Refer to chart at end of section. **38**



12FS

DIODE—BEAM POWER TUBE 38HE7

Duodecar type used in television receiver applications. The diode unit is used for damper service and the beam power unit for horizontal-deflection amplifier service. **Outlines section,** 15D; requires duodecar 12-contact socket. **Heater:** volts (ac/dc), 37.8; amperes, 0.45; warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Beam Power Unit As Class A_1 Amplifier

CHARACTERISTICS	Pentode Connection			Triode** Connection	
Plate Voltage	5000	50	130	130	volts
Grid-No.2 (Screen-Grid) Voltage	130	130	130	130	volts
Grid-No.1 (Control-Grid) Voltage	—	0	—22	—22	volts
Plate Resistance (Approx.)	—	—	6200	—	ohms
Transconductance	—	—	8800	—	µmhos
Plate Current	—	450	60		mA
Grid-No.2 Current	—	40	2.8	—	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—80	—	—39	—	volts
Amplification Factor	—	—	—	4.2	

** Grid No.2 tied to plate.

Beam Power Unit as Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Ratings)		
Plate Voltage	500	volts
Peak Positive-Pulse Plate Voltage#	5000	volts
Peak Negative-Pulse Plate Voltage	0	volts
Grid-No.2 Voltage	150	volts
DC Grid-No.1 Voltage, Negative-bias value	55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Average Cathode Current	230	mA
Peak Cathode Current	800	mA
Plate Dissipation†	10	watts
Grid-No.2 Input	3.5	watts
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	1	megohm

† A bias resistor or other means is required to protect the tube in absence of excitation.

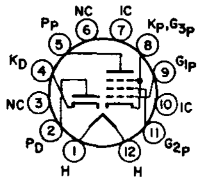
Damper Service—Diode Unit

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)			
Peak Inverse Plate Voltage#		4200	volts
Peak Plate Current		1200	mA
Average Plate Current		200	mA
Heater-Cathode Voltage:			
Peak value	+200	—4200	volts
Average value	+100	—500	volts
Bulb Temperature (at hottest point)		200	°C
CHARACTERISTICS, Instantaneous Value			
Tube Voltage Drop for plate current of 350 mA		21	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

38HK7 DIODE—BEAM POWER TUBE



12FS

Duodecar type used in television receiver applications. The diode unit is used for damper service and the beam power unit for horizontal-deflection amplifier service. Outlines section, 15D; requires duodecar 12-contact socket.

Heater Voltage (ac/dc)	37.8	volts
Heater Current	0.45	ampere
Heater Warm-up Time (Average)	11	seconds
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Diode Unit:		
Plate to Cathode and Heater	10	pF
Cathode to Plate and Heater	9	pF
Heater to Cathode	2	pF
Beam Power Unit:		
Grid No.1 to Plate	0.38	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	19	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	8	pF

Beam Power Unit as Class A_1 Amplifier

CHARACTERISTICS	Triode** Connection	Pentode Connection			
Plate Voltage	130	3500	50	130	volts
Grid-No.2 (Screen-Grid) Voltage	130	130	130	130	volts
Grid-No.1 (Control-Grid) Voltage	—22	—	0	—22	volts
Amplification Factor	4.2	—	—	—	
Plate Resistance	—	—	—	6200	ohms
Transconductance	—	—	—	8800	μmhos
Plate Current	—	—	450	60	mA
Grid-No.2 Current	—	—	40	2.8	mA
Grid-No.1 Voltage (Approx.) *for* plate current of 1 mA	—	—66	—	—39	volts
MAXIMUM CIRCUIT VALUE					
Grid-No.1-Circuit Resistance				1	megohm

** *Grid No.2 tied to plate.*

Beam Power Unit as Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	500	volts
Peak Positive-Pulse Plate Voltage	5000	volts
Peak Negative-Pulse Plate Voltage	0	volts
Grid-No.2 Voltage	150	volts
DC Grid-No.1 Voltage, Negative-bias value	55	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Average Cathode Current	230	mA
Peak Cathode Current	800	mA
Plate Dissipation†	10	watts
Grid-No.2 Input	3.5	watts
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	1	megohm

† A bias resistor or other means is required to protect the tube in absence of excitation.

Damper Service—Diode Unit

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

Peak Inverse Plate Voltage#		4200	volts
Peak Plate Current		1200	mA
Average Plate Current		200	mA
Heater-Cathode Voltage:			
Peak value	+200	—3700	volts
Average value	+100	—500	volts
Bulb Temperature (At hottest point)		200	°C
CHARACTERISTICS, Instantaneous Value			
Tube Voltage Drop for plate current of 350 mA		16	volts

Pulse duration must not exceed 15% of a horizontal scanning cycle (10 microseconds).

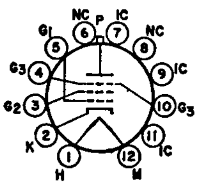
39/44
Refer to chart at end of section.

40
Refer to chart at end of section.

40KD6
Refer to type 6KD6.

41
Refer to chart at end of section.

42
Refer to chart at end of section.



12GU

42KN6
DUAL BEAM PENTODE

Duodecar type used as horizontal-deflection amplifier in color and black-and-white television receivers, with units in parallel. **Outlines section, 16D;** requires duodecar 12-contact socket. Heater: volts, 42; amperes, 0.45; warm-up time, 11 seconds; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

CHARACTERISTICS	Triode* Connection	Pentode Connection			
Plate Voltage	130	5500	60	130	volts
Grid-No.2 (Screen-Grid) Voltage	130	125	125	130	volts
Grid-No.1 (Control-Grid) Voltage	—20	—	0	—20	volts
Plate Resistance	—	—	—	4000	ohms
Transconductance	—	—	—	16000	μmhos
Plate Current	—	—	800▲	100	mA
Grid-No.2 Current	—	—	50▲	4	mA
Grid-No.1 Voltage (Approx.) for plate current of 1 mA	—	—	—	—33	volts
Grid-No.1 Voltage (Approx.) for plate current of 75 μA	—	100	—	—	volts
Amplification Factor	4.5	—	—	—	

▲ This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.

* Grid No.2 connected to plate.

Horizontal-Deflection Amplifier

For operation in a 525-line, 30-frame system

MAXIMUM RATINGS (Design-Maximum Values)

DC Plate Supply Voltage	770	volts
Peak Positive-Pulse Plate Voltage#	6500	volts
Peak Negative-Pulse Plate Voltage	1500	volts
Grid-No.2 Voltage	220	volts
Peak Negative-Pulse Grid-No.1 Voltage	330	volts
Average Cathode Current	400	mA
Peak Cathode Current	1500	mA
Plate Dissipation•	30	watts
Grid-No.2 Input	5	watts
Bulb Temperature (At hottest point)	260	°C

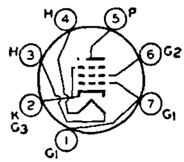
MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance	1	megohm

Pulse duration must not exceed 15% of one horizontal scanning cycle (10 microseconds).
• A bias resistor or other means is required to protect the tube in absence of excitation.

43	Refer to chart at end of section.
45	Refer to chart at end of section.
45Z3	Refer to chart at end of section.
45Z5GT	Refer to chart at end of section.
46	Refer to chart at end of section.
47	Refer to chart at end of section.
48	Refer to chart at end of section.
49	Refer to chart at end of section.
50	Refer to chart at end of section.
50A5	Refer to chart at end of section.

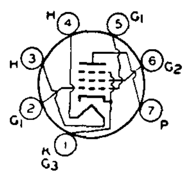
50B5 BEAM POWER TUBE



7BZ

Miniature type used in output stage of compact ac/dc receivers. Outlines section, 5D; requires miniature 7-contact socket. Except for basing arrangement, type 50B5 is identical with miniature type 50C5.

50C5
25C5
BEAM POWER TUBE



7CV

Miniature type used in output stage of compact, ac/dc radio receivers. Outlines section, 5D; requires miniature 7-contact socket. This tube, like other power-handling tubes, should be adequately ventilated. Within its maximum ratings, type 50C5 is equivalent in performance to glass octal type 50L6GT. Type 25C5 is identical with type 50C5 except for heater ratings.

	25C5	50C5	
Heater Voltage (ac/dc)	25	50	volts
Heater Current	0.3	0.15	ampere
Heater-Cathode Voltage:			
Peak value	±200 max	±200 max	volts
Average value	100 max	100 max	volts
Direct Interelectrode Capacitances (Approx.):			
Grid No.1 to Plate		0.6	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3		13	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3		8.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS *(Design-Maximum Values)*

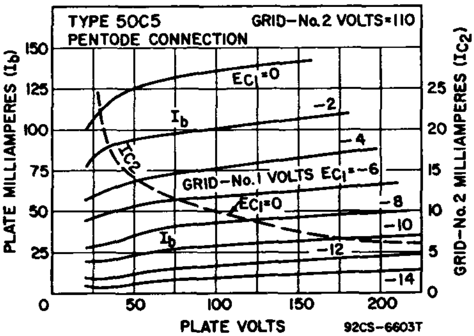
Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Plate Dissipation	7	watts
Grid-No.2 Input	1.4	watts
Bulb Temperature (At hottest point)	220	°C

TYPICAL OPERATION

Plate Voltage	120	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 (Control-Grid) Voltage	—8	volts
Peak AF Grid-No.1 Voltage	8	volts
Zero-Signal Plate Current	49	mA
Maximum-Signal Plate Current	50	mA
Zero-Signal Grid-No.2 Current	4	mA
Maximum-Signal Grid-No.2 Current	8.5	mA
Plate Resistance (Approx.)	10000	ohms
Transconductance	7500	μmhos
Load Resistance	2500	ohms
Total Harmonic Distortion	10	per cent
Maximum-Signal Power Output	2.3	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm



Installation and Application

The 50-volt heater is designed to operate under the normal conditions of line voltage variation without materially affecting the performance or serviceability of the 50C5. For operation of the 50C5 in series with other types having 0.15-ampere rating, the current in the heater circuit should be adjusted to 0.15 ampere for the normal supply voltage.

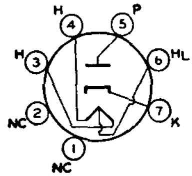
In a series-heater circuit of the "dc power line" type employing several 0.15-ampere types and one or two 50C5s, the heater(s) of the 50C5(s) should be placed on the positive side of the line. Under these conditions, heater-cathode voltage of the 50C5 must not exceed the value given under maximum ratings. In a series-heater circuit of the "universal" type employing rectifier tube 35W4, one or two 50C5s, and several 0.15-ampere types, it is recommended that the heater(s) of the 50C5(s) be placed in the circuit so that the higher values of heater-cathode bias will be impressed on the 50C5(s) rather than on the other 0.15-ampere types. This is accomplished by arranging the 50C5(s) on the side of the supply line which is connected to the cathode of the rectifier, i.e., the positive terminal of the rectified voltage supply. Between this side of the line and the 50C5(s), any necessary auxiliary resistance and the heater of the 35W4 are connected in series.

As a power amplifier (class A_1), the 50C5 is recommended for use either singly or in push-pull combination in the power-output stage of "ac/dc" receivers. The operating values shown under typical operation have been determined on the basis that grid-No. 1 current does not flow during any part of the input cycle.

Refer to chart at end of section. **50C6G**

50DC4

HALF-WAVE VACUUM RECTIFIER



5BQ

Miniature type used in power supply of ac/dc radio receivers. The heater is provided with a tap for operation of a panel lamp. For typical circuit, refer to type 35W4. Outlines section, 5D; requires 7-contact socket.

	*	**	
Heater Voltage (ac/dc):	*	**	volts
Entire Heater (Pins 3 and 4)	50	45	volts
Panel-Lamp Section (Pins 4 and 6)	7.5	5.5	volts
Heater Current:			
Between Pins 3 and 4	0.15	—	ampere
Between Pins 3 and 6	—	0.15	ampere
Peak Heater-Cathode Voltage		±330 max	volts

* Without panel lamp.
** With No.40 or No.47 panel lamp.

Half-Wave Rectifier

MAXIMUM RATINGS (*Design-Maximum Values*)

Peak Inverse Plate Voltage	330	volts
Peak Plate Current	720	mA
Average Output Current:		
With Panel Lamp and No Shunting Resistor	70	mA
With Panel Lamp and Shunting Resistor•	110	mA
Without Panel Lamp	120	mA
Panel-Lamp-Section Voltage (rms):		
When Panel Lamp Fails	16.5	volts

TYPICAL OPERATION WITH PANEL LAMP†

AC Plate-Supply Voltage (rms)	117	117	117	117	volts
Filter-Input Capacitor	40	40	40	40	μF
Minimum Total Effective Plate-Supply Impedance	15	15	15	15	ohms
Panel-Lamp Shunting Resistor	450	200	100	75	ohms
Average Output Current	70	80	90	100	mA

TYPICAL OPERATION WITHOUT PANEL LAMP

AC Plate-Supply Voltage (rms)	117	volts
Filter-Input Capacitor	40	μF
Minimum Total Effective Plate-Supply Impedance	15	ohms
Average Output Current	110	mA
DC Output Voltage at Input to Filter (Approx.):		
At half-load current (55 mA)	130	volts
At full-load current (110 mA)	110	volts
Voltage Regulation (Approx.):		
Half-load to full-load current	20	volts

† No.40 or No.47 panel lamp used in circuit with capacitor-input filter given under type 35W4.
• Required when dc output current is greater than 70 milliamperes.

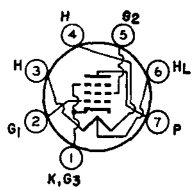
50EH5 Refer to type 6EH5.

50FE5 Refer to chart at end of section.

50FK5 Refer to chart at end of section.

50HC6

POWER PENTODE



7FZ

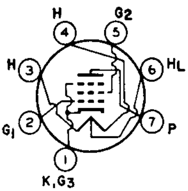
Miniature type used in the audio-frequency power-output stages of radio receivers. Heater is provided with a tap for operation of a panel lamp. Outlines section, 5D; requires miniature 7-contact socket. Heater: volts (ac/dc), 50; amperes, 0.15; tap volts (without panel lamp), 7; maximum heater-cathode volts, ±200 peak, 100 average.

MAXIMUM RATINGS (Design-Maximum Values)

RMS Heater-Tap Voltage, when panel lamp fails	14	volts
Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Plate Dissipation	5.5	watts
Grid-No.2 Input	2	watts

CHARACTERISTICS

Plate Supply Voltage	110	volts
Grid-No.2 Voltage	115	volts
Peak AF Grid-No.1 (Control-Grid) Voltage	3	volts
Cathode-Bias Resistor	62	ohms
Zero-Signal Plate Current	42	mA
Maximum-Signal Plate Current	42	mA
Zero-Signal Grid-No.2 Current	11.5	mA
Maximum-Signal Grid-No.2 Current	14.5	mA
Plate Resistance (Approx.)	11000	ohms
Transconductance	14600	μmhos
Load Resistance	3000	ohms
Total Harmonic Distortion (Approx.)	7	per cent
Maximum-Signal Power Output	1.4	watts



7FZ

POWER PENTODE 50HK6

Miniature type used in audio-frequency power-output stage of radio receivers. **Outlines section,** 5D; requires miniature 7-contact socket. The heater is provided with a tap for operation of a panel lamp. **Heater:** volts (ac/dc), 50; amperes, 0.15; tap volts (without panel lamp), 7; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

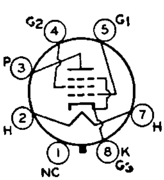
Plate Voltage	150	volts
Grid-No.2 (Screen-Grid) Voltage	130	volts
Plate Dissipation	5.5	watts
Grid-No.2 Input	1.1	watts
RMS Heater-Tap Voltage When Panel Lamp Fails	14	volts

TYPICAL OPERATION AND CHARACTERISTICS

Plate Voltage	110	volts
Grid-No.2 Voltage	110	volts
Grid-No.1 (Control-Grid) Voltage	—7.5	volts
Peak AF Grid-No.1 Voltage	7.5	volts
Zero-Signal Plate Current	49	mA
Maximum-Signal Plate Current	50	mA
Zero-Signal Grid-No.2 Current	4	mA
Maximum-Signal Grid-No.2 Current	8.5	mA
Plate Resistance (Approx.)	10000	ohms
Transconductance	7500	μmhos
Load Resistance	2500	ohms
Total Harmonic Distortion (Approx.)	9	per cent
Maximum-Signal Power Output	1.9	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm



7AC

BEAM POWER TUBE 50L6GT

25L6GT

Glass octal type used in output stage of ac/dc radio receivers. **Outlines section,** 13D; requires octal socket. This type may be supplied with pin No.1 omitted. Refer to miniature type 50C5 for installation and application information. Type 25L6GT is identical with type 50L6GT except for heater ratings.

	25L6GT	50L6GT	
Heater Voltage (ac/dc)	25	50	volts
Heater Current	0.3	0.15	ampere
Peak Heater-Cathode Voltage	±90 max	±90 max	volts
Direct Interelectrode Capacitances (Approx.):			
Grid No.1 to Plate		0.6	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3		15	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3		9.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Center Values)		
Plate Voltage	200	volts
Grid-No.2 (Screen-Grid) Voltage	125	volts
Plate Dissipation	10	watts
Grid-No.2 Input	1.25	watts

TYPICAL OPERATION	Fixed Bias	Cathode Bias	
Plate Supply Voltage	110	200	volts
Grid-No.2 Supply Voltage	110	125	volts
Grid-No.1 (Control-Grid) Voltage	—7.5	—	volts
Peak AF Grid-No.1 Voltage	7.5	8.0	volts
Cathode-Bias Resistor	—	180	ohms
Zero-Signal Plate Current	49	46	mA
Maximum-Signal Plate Current	50	47	mA
Zero-Signal Grid-No.2 Current	4	2.2	mA
Maximum-Signal Grid-No.2 Current	10	8.5	mA
Plate Resistance (Approx.)	13000	28000	ohms
Transconductance	8000	8000	µmhos
Load Resistance	2000	4000	ohms
Total Harmonic Distortion	10	10	per cent
Maximum-Signal Power Output	2.1	3.8	watts

50X6 Refer to chart at end of section.

50Y6GT Refer to chart at end of section.

50Y7GT Refer to chart at end of section.

50Z7G Refer to chart at end of section.

53 Refer to chart at end of section.

60FX5 Refer to type 12FX5.

70L7GT Refer to chart at end of section.

75 Refer to chart at end of section.

78 Refer to chart at end of section.

80 Refer to chart at end of section.

84/6Z4 Refer to chart at end of section.

117L7GT/M7GT Refer to chart at end of section.

117N7GT Refer to chart at end of section.

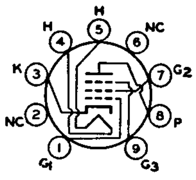
117P7GT Refer to chart at end of section.

117Z3 Refer to chart at end of section.

117Z4GT Refer to chart at end of section.

117Z6GT Refer to chart at end of section.

5879 SHARP-CUTOFF PENTODE



9AD

Miniature type used as audio amplifier in the input stages of medium-gain public-address systems, home sound recorders, and audio systems. **Outlines section,** 6B; requires miniature 9-contact socket. For operation as resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.15	ampere
Peak Heater-Cathode Voltage	±100 max	volts

Direct Interelectrode Capacitances:		
Pentode Connection:		
Grid No.1 to Plate	0.11 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	2.7	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	2.4	pF
Triode Connection*:		
Grid No.1 to Plate	1.4	pF
Grid No.1 to Cathode and Heater	1.4	pF
Plate to Cathode and Heater	0.85	pF

* Grid No.2 and grid No.3 connected to plate.

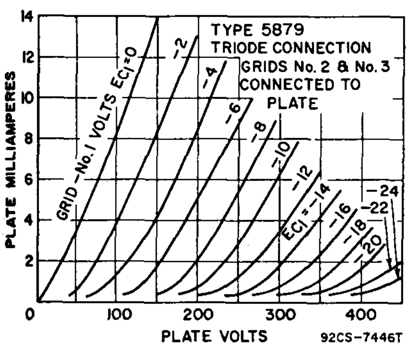
Class A_1 Amplifier

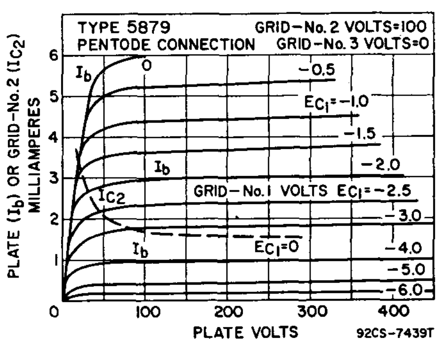
MAXIMUM RATINGS (Design-Maximum Values)	Triode Connection*	Pentode Connection	
Plate Voltage	275	330	volts
Grid-No.2 (Screen-Grid) Voltage	—	See curve page 98	
Grid-No.2 Supply Voltage	—	330	volts
Grid-No.1 (Control-Grid) Voltage:			
Negative-bias value	55	55	volts
Positive-bias value	0	0	volts
Plate Dissipation	1.7	1.25	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.25	watt
For grid-No.2 voltages between 165	—	See curve page 98	

CHARACTERISTICS	Triode	Triode	Pentode	
Plate Voltage	100	250	250	volts
Grid No.3	—	—	Connected to cathode at socket	
Grid-No.2 Voltage	—	—	100	volts
Grid-No.1 Voltage	—3	—8	—3	volts
Amplification Factor	21	21	—	
Plate Resistance (Approx.)	0.017	0.0137	2	megohms
Transconductance	1240	1530	1000	μmhos
Plate Current	2.2	5.5	1.8	mA
Grid-No.2 Current	—	—	0.4	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 μA	—	—	—8	volts

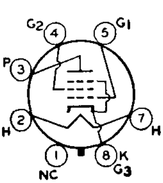
MAXIMUM CIRCUIT VALUE		
Grid-No.1-Circuit Resistance	2.2	megohms

* Grid No.2 and grid No.3 connected to plate.





BEAM POWER TUBE 5881



7AC

Glass octal type used in the output stages of radio receivers and high-fidelity audio amplifiers. **Outlines** section, 29M; requires octal socket. For typical operation as push-pull class A_1, class AB_1 (within maximum ratings), and class AB_2 amplifier, and for curves of average plate characteristics, refer to type 6L6GC. **Heater:** volts (ac/dc), 6.3; amperes, 0.9; maximum heater-cathode volts, ±200 peak.

Class A$_1$ Amplifier

MAXIMUM RATINGS (Design-Center Values)	Triode Connection*	Pentode Connection	
Plate Voltage	400	400	volts
Grid-No.2 (Screen-Grid) Voltage	—	400	volts
Plate Dissipation	26	23	watts
Grid-No.2 Input	—	3	watts

TYPICAL OPERATION AND CHARACTERISTICS	Triode	Triode	Pentode	Pentode	
Plate Voltage	250	300	250	350	volts
Grid-No.2 Voltage	—	—	250	250	volts
Grid-No.1 (Control-Grid) Voltage	—18	—20	—14	—18	volts
Peak AF Grid-No.1 Voltage	18	20	14	18	volts
Zero-Signal Plate Current	52	78	75	53	mA
Maximum-Signal Plate Current	58	85	80	65	mA
Zero-Signal Grid-No.2 Current	—	—	4.3	2.5	mA
Maximum-Signal Grid-No.2 Current	—	—	7.6	8.5	mA
Amplification Factor	8	—	—	—	
Plate Resistance (Approx.)	—	—	30000	48000	ohms
Transconductance	5250	—	6100	5200	μmhos
Load Resistance	4000	4000	2500	4200	ohms
Total Harmonic Distortion	6	5.5	10	13	per cent
Maximum-Signal Power Output	1.4	1.8	6.7	11.3	watts

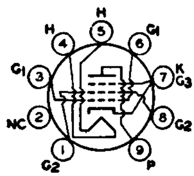
MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

* Grid No.2 connected to plate.

6973

BEAM POWER TUBE



9EU

Miniature type used as power amplifier in compact high-fidelity audio equipment. Outlines section, 6G; requires miniature 9-contact socket.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Grid-No.1 to Plate	0.4 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	9	pF
Plate to Cathode. Heater, Grid No.2, and Grid No.3	6	pF

Class A$_1$ Amplifier

CHARACTERISTICS		
Plate Voltage	250	volts
Grid-No.2 (Screen-Grid) Voltage	250	volts
Grid-No.1 (Control-Grid) Voltage	—15	volts
Plate Resistance (Approx.)	73000	ohms
Transconductance	4800	μmhos
Plate Current	46	mA
Grid-No.2 Current	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 100 μA	—40	volts

Push-Pull Class AB$_1$ Amplifier

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	440	volts
Grid-No.2 Voltage	330	volts
Plate Dissipation	12	watts
Grid-No.2 Input	2	watts
Bulb Temperature (At hottest point)	250	°C

TYPICAL OPERATION (Values are for two tubes)

	Fixed Bias			Cathode Bias		
Plate Supply Voltage	250	350	400	300	310	volts
Grid-No.2 Supply Voltage	250	280	290	300	310	volts
Grid-No.1 Voltage	—15	—22	—25	—	—	volts
Cathode-Bias Resistor	—	—	—	230	270	ohms
Peak AF Grid-No.1-to-Grid-No.1 Voltage	30	44	50	48	55	volts

	Fixed Bias			Cathode Bias		
Zero-Signal Plate Current	92	58	50	80	77	mA
Maximum-Signal Plate Current	105	106	107	96	92	mA
Zero-Signal Grid-No.2 Current	7	3.5	2.5	6	5	mA
Maximum-Signal Grid-No.2 Current	16	14	13.7	14	14	mA
Effective Load Resistance (Plate-to-plate)	8000	7500	8000	5500	6000	ohms
Total Harmonic Distortion	2	1.5	2	2	4	per cent
Maximum-Signal Power Output	12.5	20	24	15	17	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.5	megohm
For cathode-bias operation	1	megohm

Push-Pull Class AB_1 Amplifier

Grid No.2 of Each Tube Connected to Tap on Plate Winding of Output Transformer

MAXIMUM RATINGS (Design-Maximum Values)

Plate and Grid-No.2 Supply Voltage	410	volts
Plate Dissipation	12	watts
Grid-No.2 Input	1.75	watts
Bulb Temperature (At hottest point)	250	°C

TYPICAL OPERATION (Values are for two tubes)	Fixed Bias	Cathode Bias	
Plate Supply Voltage	375	370	volts
Grid-No.2 Supply Voltage	*	#	volts
Grid-No.1 Voltage•	—33.5	–	volts
Cathode-Bias Resistor	—	355	ohms
Peak AF Grid-No.1-to-Grid-No.1 Voltage	67	62	volts
Zero-Signal Cathode Current	62	74	mA
Maximum-Signal Cathode Current	95	84	mA
Effective Load Resistance (Plate-to-plate)	12500	13000	ohms
Total Harmonic Distortion	1.5	1.2	per cent
Maximum-Signal Power Output	18.5	15	watts

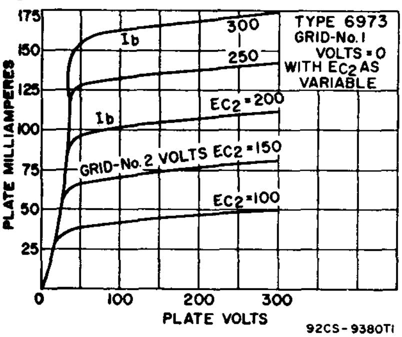
MAXIMUM CIRCUIT VALUES

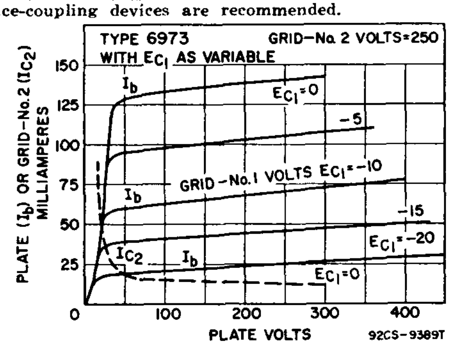
Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.1	megohm
For cathode-bias operation	1	megohm

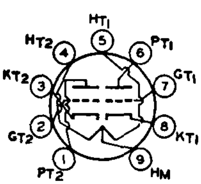
* Obtained from taps on the primary winding of the output transformer. The taps are located on each side of the center tap (B+) so as to apply 50 per cent of the plate signal voltage to grid No.2 of each output tube.

Obtained from taps on the primary winding of the output transformer. The taps are located on each side of the center tap (B+) so as to supply 43 per cent of the plate signal voltage to grid No.2 of each output tube.

• The type of input-coupling network used should not introduce too much resistance in the grid-No.1 circuit. Transformer- or impedance-coupling devices are recommended.







9A

HIGH-MU TWIN TRIODE 7025

Miniature type used as phase inverter or resistance-coupled amplifier in high-quality, high-fidelity audio amplifiers. **Outlines section**, 6B; requires miniature 9-contact socket. This type is identical with miniature type 12AX7A except that it has a controlled equivalent noise and hum characteristic. For operation as resistance-coupled amplifier, refer to **Resistance-Coupled Amplifier** section.

EQUIVALENT-NOISE AND HUM VOLTAGE REFERENCED TO GRID (Each Unit)

Average Value (rms)†	1.8	µV
Maximum Value (rms)•	7	µV

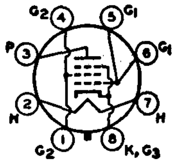
† Measured in "true rms" units under following conditions: heater volts (ac), 6.3 (parallel connection); center tap of heater transformer connected to ground; plate supply volts, 250; plate load resistor, 2700 ohms; cathode-bypass capacitor, 100 µF; grid resistor, 0 ohms; and amplifier covering frequency range between 25 to 10000 cycles per second.

• Same conidtions as for "Average Value" except cathode resistor is unbypassed and grid resistor is 0.05 megohm.

7027

Refer to chart at end of section.

7027A BEAM POWER TUBE



8HY

Glass octal type used in push-pull power amplifier circuits of high-fidelity audio equipment. Outlines section, 9F; requires octal socket. This tube, like other power-handling tubes, should be adequately ventilated.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.9	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	1.5	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	10	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	7.5	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	250	volts
Grid-No.2 (Screen-Grid) Voltage	250	volts
Grid-No.1 (Control-Grid) Voltage	—14	volts
Plate Resistance (Approx.)	22500	ohms
Transconductance	6000	µmhos
Plate Current	72	mA
Grid-No.2 Current	5	mA

Push-Pull Class AB_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	600	volts
Grid-No.2 Voltage	500	volts
Plate Dissipation	35	watts
Grid-No.2 Input	5	watts

TYPICAL OPERATION (Values are for two tubes)

	Fixed Bias			Cathode Bias			
Plate Supply Voltage	400	450	540	400	380	425	volts
Grid-No.2 Supply Voltage	300	350	400	300	380	415	volts
Grid-No.1 Voltage	—25•	—30•	—38•	—	—	—	volts
Cathode-Bias Resistor	—	—	—	200	180	200	ohms
Peak AF Grid-No.1-to-Grid-No.1 Voltage	50	60	76	57	68.5	86	volts
Zero-Signal Plate Current	102	95	100	112	138	150	mA
Maximum-Signal Plate Current	152	194	220	128	170	196	mA
Zero-Signal Grid-No.2 Current	6	3.4	5	7	5.6	8	mA
Maximum-Signal Grid-No.2 Current	17	19.2	21.4	16	20	20	mA
Effective Load Resistance (Plate-to-plate)	6600	6000	6500	6600	4500	3800	ohms
Total Harmonic Distortion	2	1.5	2	2	3.5	4	per cent
Maximum-Signal Power Output	34	50	76	32	36	44	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation•	0.1	megohm
For cathode-bias operation	0.5	megohm

• The type of input coupling network used should not introduce too much resistance in the grid-No.1 circuit. Transformer- or impedance-coupling devices are recommended.

Push-Pull Class AB_1 Amplifier

Grid No.2 of Each Tube Connected to Tap on Plate Winding of Output Transformer

MAXIMUM RATINGS (Design-Maximum Values)

Plate and Grid-No.2 Supply Voltage	600	volts
Plate Dissipation	35	watts
Grid-No.2 Input	4.5	watts

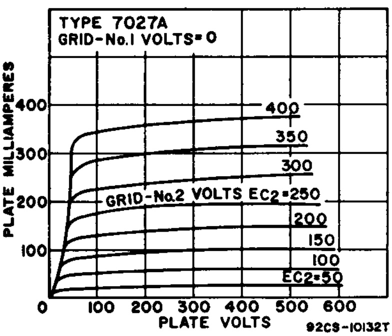
TYPICAL OPERATION (Values are for two tubes)

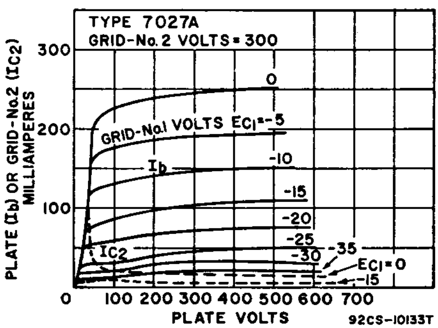
Plate Supply Voltage	410	volts
Grid-No.2 Supply Voltage	*	volts
Cathode-Bias Resistor	220	ohms
Peak AF Grid-No.1-to-Grid-No.1 Voltage	68	volts
Zero-Signal Cathode Current	134	mA
Maximum-Signal Cathode Current	155	mA
Effective Load Resistance (Plate to plate)	8000	ohms
Total Harmonic Distortion	1.6	per cent
Maximum-Signal Power Output	24	watts

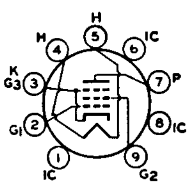
MAXIMUM CIRCUIT VALUE

Grid-No.1-Circuit Resistance, for cathode-bias operation	0.5	megohm

* Obtained from taps on the primary winding of the output transformer. The taps are located on each side of the center tap (B+) so as to apply 43 per cent of the plate signal voltage to grid No.2 of each output tube.







9BL

POWER PENTODE 7189

Miniature type used as power amplifier tube in high-fidelity audio equipment. Outlines section, 6G; requires miniature 9-contact socket.

Heater Voltage	6.3	volts
Heater Current	0.76	ampere
Peak Heater-Cathode Voltage	±100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.5	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	10.8	pF
Plate to Cathode, Heater, Grid-No.2, and Grid No.3	6.5	pF
Grid No.1 to Heater	0.25	pF

Class A_1 Amplifier

CHARACTERISTICS

Plate Voltage	250	volts
Grid-No.2 (Screen-Grid) Voltage	250	volts
Grid-No.1 (Control-Grid) Voltage	—7.3	volts
Mu-Factor, Grid No.2 to Grid No.1	19.5	
Plate Resistance (Approx.)	40000	ohms
Transconductance	11300	µmhos
Plate Current	48	mA
Grid-No.2 Current	5.5	mA

Push-Pull Class AB₁ Amplifier

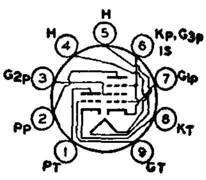
MAXIMUM RATINGS (Design-Center Values)		Grid-No.2 Special Connection•	
Plate Voltage	400	375	volts
Grid-No.2 Voltage	300	•	volts
Cathode Current	65	65	mA
Plate Dissipation	12	12	watts
Zero-Signal Grid-No.2 Input	2	2	watts
Maximum-Signal Grid-No.2 Input	4	4	watts
TYPICAL OPERATION (Values are for two tubes)			
Plate Supply Voltage	—	375	volts
Plate Voltage	400	—	volts
Grid-No.2 Supply Voltage	—	▪	
Grid-No.2 Voltage	300	•	volts
Grid-No.1 Voltage	—15	—	volts
Cathode-Bias Resistor	—	220	ohms
Peak AF Grid-No.1 Voltage	14.8	17.7	volts
Zero-Signal Plate Current	15	70	mA
Maximum-Signal Plate Current	105	81	mA
Zero-Signal Grid-No.2 Current	1.6	•	mA
Maximum-Signal Grid-No.2 Current	25	•	mA
Effective Load Resistance (Plate-to-plate)	8000	11000	ohms
Total Harmonic Distortion	4	3	per cent
Maximum-Signal Power Output	24	16.5	watts
MAXIMUM CIRCUIT VALUES	**Fixed Bias**	**Cathode Bias**	
Grid-No.1-Circuit Resistance	0.3	1	megohm

• Grid No.2 of each tube connected to tap on plate winding of output transformer.

▪ Obtained from taps on primary winding of the output transformer. The taps are located on each side of the center tap (B+) so as to supply 43 per cent of the plate signal voltage to grid No.2 of each output tube.

7199 MEDIUM-MU TRIODE— SHARP-CUTOFF PENTODE

Miniature type used in high-quality, high-fidelity audio equipment, particularly in phase splitters, tone-control amplifiers, and high-gain voltage amplifiers. Outlines section, 6B; requires miniature 9-contact socket. For operation as resistance-coupled amplifier, refer to Resistance-Coupled Amplifier section. In direct-coupled voltage-amplifier phase-splitter circuits, the pentode unit should drive the triode unit.



9JT

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances:		
Triode Unit:		
Grid to Plate	2	pF
Grid to Cathode and Heater	2.3	pF
Plate to Cathode and Heater	0.3	pF
Pentode Unit:		
Grid No.1 to Plate	0.06 max	pF
Grid No.1 to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	5	pF
Plate to Cathode, Heater, Grid No.2, Grid No.3, and Internal Shield	2	pF

EQUIVALENT-NOISE AND HUM VOLTAGE REFERENCED TO GRID

	Triode Unit	Pentode Unit	
Median Value (rms)	10†	35•	µV
Maximum Value (rms)	150†	100•	µV

† Measured in "true rms" units under the following conditions: heater volts (ac), 6.3; center tap of heater transformer connected to ground; plate-supply volts, 250; plate load resistor, 0.1 megohm; cathode resistor, 1500 ohms; grid resistor, 0.05 megohm; and amplifier covering frequency range between 25 and 10000 cycles per second.

• Same conditions as for triode unit except: grid-No.2 supply volts, 250; grid-No.2 resistor, 0.33 megohm; grid-No.2-bypass capacitor, 0.22 µF; cathode resistor, 1200 ohms; and grid-No.1 resistor, 0.05 megohm.

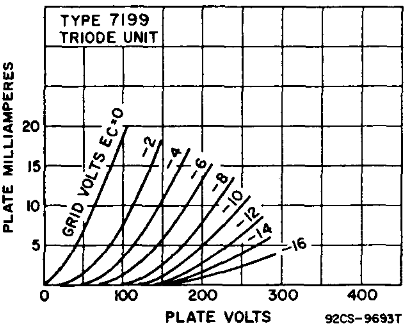
Class A_1 Amplifier

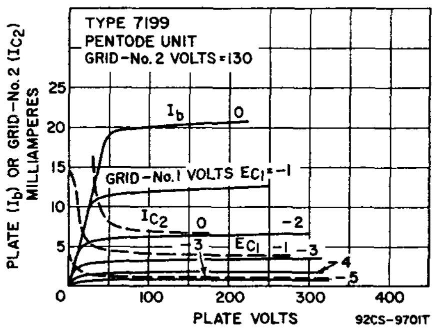
MAXIMUM RATINGS (Design-Maximum Values)	Triode Unit	Pentode Unit	
Plate Voltage	330	330	volts
Grid-No.2 (Screen-Grid) Voltage	—	See curve page 98	
Grid-No.2 Supply Voltage	—	330	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	0	volts
Plate Dissipation	2.4	3	watts
Grid-No.2 Input:			
For grid-No.2 voltages up to 165 volts	—	0.6	watt
For grid-No.2 voltages between 165 and 330 volts	—	See curve page 98	

CHARACTERISTICS	Triode Unit	Pentode Unit		
Plate Supply Voltage	215	100	220	volts
Grid-No.2 Supply Voltage	—	50	130	volts
Grid-No.1 Voltage	—8.5	—	—	volts
Cathode-Bias Resistor	—	1000	62	ohms
Amplification Factor	17	—	—	
Plate Resistance (Approx.)	0.0081	1	0.4	megohm
Transconductance	2100	1500	7000	µmhos
Plate Current	9	1.1	12.5	mA
Grid-No.2 Current	—	0.35	3.5	mA
Grid-No.1 Voltage (Approx.) for plate current of 10 µA	—40	—4	—	volts

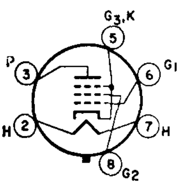
MAXIMUM CIRCUIT VALUES	Triode Unit	Pentode Unit	
Grid-No.1-Circuit Resistance:*			
For fixed-bias operation	0.5	0.25	megohm
For cathode-bias operation	1	1	megohm

* If either unit is operated at maximum rated conditions, grid-No.1-circuit resistance for both units should not exceed the stated value.





Refer to chart at end of section. **7247**



8KN

POWER PENTODE 7355

Glass octal type used in the power-output stage of high-fidelity audio-frequency amplifier systems. **Outlines section**, 13F; requires octal socket. **Heater**: volts (ac/dc), 6.3; amperes, 0.8; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)		
Plate Voltage	500	volts
Grid-No.2 (Screen-Grid) Voltage	400	volts
Grid-No.1 (Control-Grid) Voltage, Positive-bias value	0	volts
Average Cathode Current	100	mA
Plate Dissipation	18	watts
DC Grid-No.2 Input	3.5•	volts

TYPICAL OPERATION AND CHARACTERISTICS

Plate Voltage	250	volts
Grid-No.2 Voltage	225	volts
Grid-No.1 Voltage	—15	volts
Peak AF Grid-No.1 Voltage	15	volts
Plate Resistance (Approx.)	42000	ohms
Transconductance	7600	μmhos
Zero-Signal Plate Current	62	mA
Maximum Signal Plate Current	74	mA
Zero-Signal Grid-No.2 Current	3.2	mA
Maximum-Signal Grid-No.2 Current	16.5	mA
Load Resistance	2500	ohms
Total Harmonic Distortion (Approx.)	15	per cent
Maximum-Signal Power Output	9	watts
Grid-No.1 Voltage (Approx.) for plate current of 500 μA	—35	volts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.3	megohm
For cathode-bias operation	1	megohm

• Grid-No.2 input may reach 7 watts during peak levels of speech and music signals.

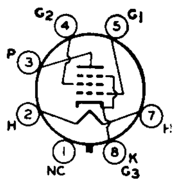
Push-Pull Class AB_1 Amplifier

MAXIMUM RATINGS (Same as for class A_1 amplifier)

TYPICAL OPERATION (Values are for two tubes)

Plate Voltage	300	400	volts
Grid-No.2 Voltage	250	300	volts
Grid-No.1 Voltage	—21	—34	volts
Peak AF Grid-No.1 Voltage	42	60	volts
Zero-Signal Plate Current	100	56	mA
Maximum-Signal Plate Current	185	175	mA
Zero-Signal Grid-No.2 Current	5.5	3.5	mA
Maximum-Signal Grid-No.2 Current	24	24	mA
Effective Load Resistance (Plate-to-plate)	4000	5000	ohms
Total Harmonic Distortion	2	6	per cent
Maximum-Signal Power Output	28.5	40	watts

7408 BEAM POWER TUBE



7AC

Glass octal type used as output amplifier tube in high-quality sound systems. **Outlines section, 13D; requires octal socket.**

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.45	ampere
Heater-Cathode Voltage:		
Peak value	±200	volts
Average value	100	volts
Direct Interelectrode Capacitances:		
Grid No.1 to Plate	0.7	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	9	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	7.5	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	350	volts
Grid-No.2 (Screen-Grid) Voltage	315	volts
Grid-No.2 Input	2.2	watts
Plate Dissipation	14	watts

TYPICAL OPERATION AND CHARACTERISTICS

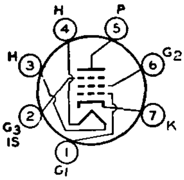
Plate Voltage	60	250	volts
Grid-No.2 Voltage	250	250	volts
Grid-No.1 (Control-Grid) Voltage	0	—12.5	volts
Peak AF Grid-No.1 Voltage	—	12.5	volts
Zero-Signal Plate Current	100•	45	mA
Maximum-Signal Plate Current	—	47	mA
Zero-Signal Grid-No.2 Current	22•	4.5	mA
Maximum-Signal Grid-No.2 Current	—	7	mA
Plate Resistance (Approx.)	—	50000	ohms
Transconductance	—	4100	μmhos
Load Resistance	—	5000	ohms
Total Harmonic Distortion	—	7	per cent
Maximum-Signal Power Output	—	4.5	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:

For fixed-bias operation	0.1	megohm
For cathode-bias operation	0.5	megohm

• This value can be measured by a method involving a recurrent waveform such that the maximum ratings of the tube will not be exceeded.



7BK

SHARP-CUTOFF PENTODE 7543

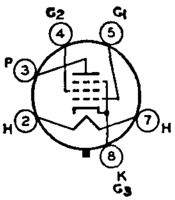
Miniature type used in compact audio equipment. **Outlines section,** 5C; requires miniature 7-contact socket. This type is identical with miniature type 6AU6A except that it has a controlled hum characteristic.

HUM OUTPUT VOLTAGE

Average Value, (rms, cathode bypassed)	1.2†	millivolts
Average Value (rms, cathode unbypassed)	0.9•	millivolt

† Measured in "true rms" units under the following conditions: heater volts (ac), 6.3; center tap of heater transformer connected to ground; plate and grid-No.2 supply volts, 250; plate load resistor, 0.27 megohm; grid No.3 and internal shield connected to cathode at socket; grid-No.2 resistor, 0.68 megohm; grid-No.1 resistor, 0.1 megohm; cathode resistor, 1000 ohms; grid resistor of following stage, 10 megohms; and stage gain, 340.

• Same conditions as above except that cathode resistor is unbypassed and stage gain is 110.



7AC

BEAM POWER TUBE 7581A

Glass octal type used in af power-amplifier applications. **Outlines section,** 19D; requires octal socket. For typical operation as push-pull class A_1, class AB_1, and class AB_2 amplifier, refer to type 6L6GC. This tube, like other power-handling tubes, should be adequately ventilated. **Heater:** volts (ac/dc), 6.3; amperes, 0.9; maximum heater-cathode volts, ±200.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)	Triode Connection*	Pentode Connection	
Plate Voltage	450	500	volts
Grid-No.2 (Screen-Grid) Voltage	—	450#	volts
Plate Dissipation	35	35	watts
Grid-No.2 Input	—	5	watts
CHARACTERISTICS			
Plate Voltage	70	250	volts
Grid-No.2 Voltage	300	250	volts
Grid-No.1 Voltage	0	—14	volts
Plate Resistance (Approx.)	—	22500	ohms
Transconductance	—	6000	μmhos
Plate Current	210	72	mA
Grid-No.2 Current	25	5	mA
MAXIMUM CIRCUIT VALUES			
Grid-No.1-Circuit Resistance:			
For fixed-bias operation	0.1	0.1	megohm
For cathode-bias operation	0.5	0.5	megohm

* Grid No.2 connected to plate.

In push-pull circuits where grid No.2 of each tube is connected to a tap on the plate winding of the output transformer, this maximum rating is 500 volts.

▲ Applied for short interval (2 seconds) so as not to damage tube.

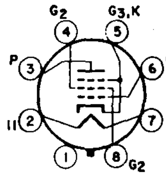
Class A_1 Amplifier (Triode Connection)

MAXIMUM RATINGS (Same as for Class A_1 Amplifier)
TYPICAL OPERATION

Plate Voltage	250	volts
Grid-No.1 Voltage	—20	volts
Peak AF Grid-No.1 Voltage	20	volts
Amplification Factor	8	
Plate Resistance (Approx.)	1700	ohms
Transconductance	4700	μmhos
Zero-Signal Plate Current	40	mA
Maximum-Signal Plate Current	44	mA
Load Resistance	5000	ohms
Total Harmonic Distortion (Approx.)	5	per cent
Maximum-Signal Power Output	1.4	watts

7591 Refer to chart at end of section.

7591A POWER PENTODE



8KQ

Glass octal type used as audio-frequency power-output tube in high-quality audio applications. **Outlines section,** 13D; requires octal socket. Heater: volts (ac/dc), 6.3; amperes, 0.8; maximum heater-cathode volts, ±200 peak, 100 average.

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	550	volts
Grid-No.2 (Screen-Grid) Voltage	440	volts
Cathode Current	90	mA
Plate Dissipation	19	watts
Grid-No.2 Input	3.3•	watts

TYPICAL OPERATION AND CHARACTERISTICS

Plate Voltage	300	volts
Grid-No.2 Voltage	300	volts
Grid-No.1 (Control-Grid) Voltage	—10	volts
Peak AF Grid-No.1 Voltage	10	volts
Zero-Signal Plate Current	60	mA
Maximum-Signal Plate Current	75	mA
Zero-Signal Grid-No.2 Current	8	mA
Maximum-Signal Grid-No.2 Current	15	mA
Triode Amplification Factor*	16.8	
Plate Resistance (Approx.)	29000	ohms
Transconductance	10200	μmhos
Load Resistance	3000	ohms
Total Harmonic Distortion	13	per cent
Maximum-Signal Power Output	11	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.3	megohm
For cathode-bias operation	1	megohm

• Grid-No.2 input may reach 6 watts during peak levels of speech and music signals.

* Triode connection, grid No.2 connected to plate.

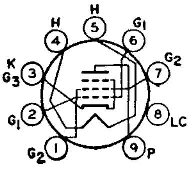
Push-Pull Class AB_1 Amplifier

MAXIMUM RATINGS (Same as for Class A_1 Amplifier)

TYPICAL OPERATION (Values are for two tubes)	Fixed Bias		Cathode Bias	
Plate Supply Voltage	350	450	450	volts
Grid-No.2 Supply Voltage	350	400	400	volts
Grid-No.1 Supply Voltage	—15.5	—21	—	volts
Cathode-Bias Resistor (Common to both cathodes)	—	—	200	ohms
Peak AF Grid-No.1-to-Grid-No.1 Voltage	31	42	28	volts
Zero-Signal Plate Current	92	66	82	mA
Maximum-Signal Plate Current	130	144	94	mA
Zero-Signal Grid-No.2 Current	13	9.4	11.5	mA
Maximum-Signal Grid-No.2 Current	28.6	30	22	mA
Effective Load Resistance (Plate-to-plate)	6600	6600	9000	ohms
Total Harmonic Distortion	2	1.5	2	per cent
Maximum-Signal Power Output	30	45	28	watts

Refer to chart at end of section. **7695**

POWER PENTODE **7868**


9RW

Novar type used in output stages of high-fidelity audio amplifiers and radio receivers. **Outlines section,** 11C or 30D; requires novar 9-contact socket. This tube, like other power-handling tubes, should be adequately ventilated.

Heater Voltage (ac/dc)	6.3	volts
Heater Current	0.8	ampere
Heater-Cathode Voltage:		
Peak value	±200 max	volts
Average value	100 max	volts
Direct Interelectrode Capacitances (Approx.):		
Grid No.1 to Plate	0.15	pF
Grid No.1 to Cathode, Heater, Grid No.2, and Grid No.3	11	pF
Plate to Cathode, Heater, Grid No.2, and Grid No.3	4.4	pF

Class A_1 Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	550▪	volts
Grid-No.2 (Screen-Grid) Voltage	440	volts
Average Cathode Current	90	mA
Plate Dissipation	19	watts
Grid-No.2 Input	3.3•	watts
Bulb Temperature (At hottest point)	240	°C

TYPICAL OPERATION AND CHARACTERISTICS

Plate Supply Voltage	300	volts
Grid-No.2 Voltage	300	volts
Grid-No.1 (Control-Grid) Voltage	—10	volts
Peak AF Grid-No.1 Voltage	10	volts
Zero-Signal Plate Current	60	mA
Maximum-Signal Plate Current	75	mA
Zero-Signal Grid-No.2 Current	8	mA
Maximum-Signal Grid-No.2 Current	15	mA
Plate Resistance (Approx.)	29000	ohms
Transconductance	10200	μmhos
Effective Load Resistance	3000	ohms
Total Harmonic Distortion	13	per cent
Maximum-Signal Power Output	11	watts

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	0.3	megohm
For cathode-bias operation	1	megohm

▪ *In push-pull circuits where the grid No.2 of each tube is connected to a tap on the plate winding of the output transformer, this maximum rating is 440 volts.*

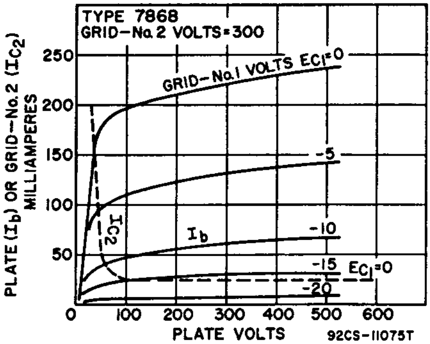
• Grid No.2 input may reach 6 watts during peak levels of speech and music signals.

Push-Pull Class AB_1 Amplifier

MAXIMUM RATINGS (Same as for class A_1 amplifier)

TYPICAL OPERATION (Values are for two tubes)

	Fixed Bias					Cathode Bias	
Plate Supply Voltage	300	350	400	450	450	450	volts
Grid-No.2 Supply Voltage	300	350	350	350	400	400	volts
Grid-No.1 Voltage	—12.5	—15.5	—16	—16.5	—21	—	volts
Cathode-Bias Resistor (Common to both cathodes)	—	—	—	—	—	170	ohms
Peak AF Grid-No.1-to-Grid-No.1 Voltage	25	31	32	33	42	31	volts
Zero-Signal Plate Current	74	72	64	60	40	86	mA
Maximum-Signal Plate Current	116	130	135	142	145	94	mA
Zero-Signal Grid-No.2 Current	10	9.5	8	7.2	5	10	mA
Maximum-Signal Grid-No.2 Current	28	32	28	26	30	20	mA
Effective Load Resistance (Plate-to-plate)	6600	6600	6600	6600	6600	10000	ohms
Total Harmonic Distortion	5	2.5	2	2.5	5	2	per cent
Maximum-Signal Power Output	24	30	34	38	44	28	watts



Push-Pull Class AB$_1$ Amplifier

Grid No.2 of Each Tube Connected to Tap on Plate Winding of Output Transformer*

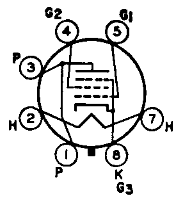
MAXIMUM RATINGS (Same as for class A$_1$ amplifier)

TYPICAL OPERATION (Values are for two tubes)

	Fixed Bias	Cathode Bias	
Plate Supply Voltage	400	425	volts
Grid-No.2 Supply Voltage	*	*	volts
Grid-No.1 Voltage	—20.5	—	volts
Cathode-Bias Resistor (Common to both cathodes)	—	185	ohms
Peak AF Grid-No.1-to-Grid-No.1 Voltage	41	42	volts
Zero-Signal Plate Current	60	88	mA
Maximum-Signal Plate Current	115	100	mA
Zero-Signal Grid-No.2 Current	8	12	mA
Maximum-Signal Grid-No.2 Current	18	16	mA
Effective Load Resistance (Plate-to-plate)	6600	6600	ohms
Total Harmonic Distortion	2.5	3.5	per cent
Maximum-Signal Power Output	23	21	watts

* Grid No.2 supply voltage is obtained from taps on the primary winding of the output transformer. The taps are located on each side of the center tap (B+) so as to apply 50 per cent of the plate signal voltage to the grid No.2 of each output tube.

8417 BEAM POWER TUBE



Glass octal type used as output amplifier in high-fidelity, high-power sound systems. Outlines section, 19J; requires octal socket. This tube, like other power-handling tubes, should be adequately ventilated. **Heater:** volts (ac/dc), 6.3; amperes, 1.6; maximum heater-cathode volts, ±200 peak, 100 average.

Class A$_1$ Amplifier

MAXIMUM RATINGS (Design-Maximum Values)

Plate Voltage	660	volts
Grid-No.2 (Screen-Grid) Voltage	500	volts
Cathode Current	200	mA
Plate Dissipation▪	35	watts
Grid-No.2 Input	5*	watts

CHARACTERISTICS

Plate Voltage	300	volts
Grid-No.2 Voltage	300	volts
Grid-No.1 (Control-Grid) Voltage	—12	volts
Grid-No.1 Voltage for plate current of 1 mA	—37	volts
Plate Resistance	16000	ohms
Transconductance	23000	μmhos
Plate Current	100	mA
Grid-No.2 Current	5.5	mA
Triode Amplification Factor	16.5	

MAXIMUM CIRCUIT VALUES

Grid-No.1-Circuit Resistance:		
For fixed-bias operation	1	megohm
For cathode-bias operation	0.25	megohm

Push-Pull Class AB_1 Amplifier

MAXIMUM RATINGS (Same as for Class A_1 Amplifier)
TYPICAL OPERATION (Values are for two tubes)

Plate Supply Voltage	400	560	volts
Grid-No.2 Supply Voltage	275	300	volts
Grid-No.1 Voltage	—13	—15.5	volts
Peak AF Grid-to-Grid Voltage	24	31	volts
Zero-Signal Plate Current	150	100	mA
Maximum-Signal Plate Current	294	270	mA
Zero-Signal Grid-No.2 Current	4.4	3.4	mA
Maximum-Signal Grid-No.2 Current	34	31	mA
Effective Load (Plate-to-Plate)	2800	4200	ohms
Total Harmonic Distortion	2.5	2	per cent
Maximum Signal Power Output	65	100	watts

▪ A bias resistor or other means is required to protect the tube in absence of excitation.

* Grid-No.2 may reach 8 watts during peak levels of speech and music levels.

Refer to chart at end of section. **EM84/6FG6**

RCA Types for

Key to Chart: Type numbers shown in light face are discontinued types. Outline numbers refer to diagrams shown in the **Outlines section** later in

RCA Type	Name	Out-line	Basing Dia-gram	Heater or Filament (F) Volts	Heater or Filament (F) Amperes	Use Values to right give operating conditions and characteristics for indicated typical use
0Z4	Full-Wave Gas Rectifier	2A	4R	—	—	Rectifier
0Z4G	Full-Wave Gas Rectifier	29D	4R	—	—	Rectifier
1A3	Diode	5C	5AP	1.4	0.15	Rectifier
1A4P	Remote-Cutoff Pentode	24B	4M	2.0F	0.06	Class A Amplifier
1A5GT	Power Pentode	13D	6X	1.4F	0.05	Class A Amplifier
1A6	Pentagrid Converter	24B	6L	2.0F	0.06	Converter
1A7GT	Pentagrid Converter	14A	7Z	1.4F	0.05	Converter
1AC5	Power Converter	29A	8CP	1.25F	0.04	Class A Amplifier
1AD5	Sharp-Cutoff Pentode	29A	8CP	1.25F	0.04	Class A Amplifier
1AX2	Half-Wave Rectifier	7A	9Y	1.4F	0.65	Pulsed Rectifier in TV Receivers
1B3GT	Half-Wave Rectifier	14E	3C	1.25F	0.2	Pulsed Rectifier in TV Receivers
1B4P	Sharp-Cutoff Pentode	24B	4M	2.0F	0.06	Class A Amplifier
1B5/25S	Twin Diode—Medium-Mu Triode	22 or 13H	6M	2.0F	0.06	Triode Unit as Class A Amplifier
1B7GT	Pentagrid Converter	14A	7Z	1.4F	0.10	Converter
1C5GT	Power Pentode	13D	6X	1.4F	0.10	Class A Amplifier
1C6	Pentagrid Converter	24B	6L	2.0F	0.12	Converter
1C7G	Pentagrid Converter	23	7Z	2.0F	0.12	Converter
1D5GP	Remote-Cutoff Pentode	23	5Y	2.0F	0.06	Class A Amplifier
1D5GT	Remote-Cutoff Tetrode	23	5R	2.0F	0.06	Class A Amplifier
1D7G	Pentagrid Converter	23	7Z	2.0F	0.06	Converter
1D8GT	Diode-Triode-Power Pentode	14A	8AJ	1.4F	0.10	Pentode Unit as Class A Amplifier
						Triode Unit as Class A Amplifier
1DN5	Diode—Semiremote-Cutoff Pentode	5C	6BW	1.4F	0.5	Pentode Unit as Class A Amplifier

Replacement Use

the Manual (see Table of Contents on inside front cover). Basing diagrams are included in numerical-alphabetical order at the end of the chart.

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power Load Ohms	Power Output Watts	RCA Type
Starting-Supply Voltage per Plate, 300 min. peak volts DC Output Current, 75 max., 30 min. mA						Peak Plate Current, 200 max. mA DC Output Voltage, 300 max. volts.				**0Z4**
Starting-Supply Voltage per Plate, 300 min. peak volts DC Output Current, 75 max., 30 min. mA						Peak Plate Current, 200 max. mA DC Output Voltage, 300 max. volts.				0Z4G
Max. Peak Plate Inverse Volts, 330 Max. Peak Plate mA, 5						Max. DC Output mA, 0.5 Max. Peak Heater-Cathode Volts, 140				**1A3**
For other characteristics, refer to Type 1D5GP										1A4P
85 90	— 4.5V — 4.5V	85 90	0.7 1.1	3.5 4.0	300000 300000	800 850	—	25000 25000	0.100 0.115	**1A5GT**
135 180	— 3V — 3V	67.5 67.5	2.5 2.4	1.2 1.3	400000 500000	Anode-Grid (2): 180 max. volts 2.3 mA Oscillator-Grid (1) Resistor.				1A6
90	0V	45	0.7	0.6	600000	Anode-Grid (2): 90 volts, 1.2 mA Oscillator-Grid (1) Resistor, 0.2 MΩ Conversion Transcond., 250 micromhos				**1A7GT**
45 67.5	— 3V — 4.5V	45 67.5	0.2 0.4	1.0 2.0	170000 150000	600 750	— —	40000 25000	0.015 0.050	1AC5
30 67.5	0V 0V	30 67.5	0.16 0.75	0.45 1.85	700000 700000	430 735	— —	— —	— —	1AD5
Max. Peak Inverse Plate Volts, 25000 Max. Peak Plate mA, 45						Max. Average Plate mA, 0.5				1AX2
Max. Peak Inverse Plate Volts, 26000 Max. Peak Plate mA, 50						Max. Average Plate mA, 0.5				1B3GT
For other characteristics, refer to Type 1E5GP										1B4P
For other characteristics, refer to Type 1H6G										1B5/25S
For other characteristics, refer to Type 1A7GT										1B7GT
90	— 7.5V	90	3.5	7.8	115000	1550	—	8000	0.24	1C5GT
For other characteristics, refer to Type 1C7G										1C6
135 180	— 3V — 3V	67.5 67.5	2.5 2.0	1.3 1.5	600000 700000	Anode-Grid (2): 180 max. volts, 4.0 mA Oscillator-Grid (1) Resistor. Conversion Transcond., 325 micromhos.				1C7G
90 180	{ — 3V min. }	67.5 67.5	0.9 0.8	2.2 2.3	600000 1 M	720 750	—	—	—	1D5GP
For other characteristics, refer to Type 1D5GP										1D5GT
For other characteristics, refer to Type 1A6										1D7G
90	— 9V	90	1.0	5.0	—	925	—	12000	0.200	1D8GT
90	0V	—	—	1.1	43500	575	25	—	—	1D8GT
67.5	0V	67.5	0.55	2.1	600000	630	—	—	—	**1DN5**

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Amperes	Use — Values to right give operating conditions and characteristics for indicated typical use
1E5GP	Sharp-Cutoff Pentode	23	5Y	2.0F	0.06	Class A Amplifier
1E7GT	Twin Power Pentode	13D	8C	2.0F	0.24	Class A Amplifier
1E8	Pentagrid Converter	29A	8CN	1.25F	0.04	Converter
1F4	Power Pentode	26	5K	2.0F	0.12	Class A Amplifier
1F5G	Power Amplifier Pentode	25	6X	2.0F	0.12	Class A Amplifier
1F6	Twin Diode—Sharp-Cutoff Pentode	23	6W	2.0F	0.06	Pentode Unit as Class A Amplifier
1F7G	Twin Diode—Sharp-Cutoff Pentode	23	7AF	2.0F	0.06	Pentode Unit as Class A Amplifier
1G4GT	Medium-Mu Triode	13D	5S	1.4F	0.05	Class A Amplifier
1G5G	Power Pentode	25	6X	2.0F	0.12	Class A Amplifier
1G6GT	High-Mu Twin Power Triode	13D	7AB	1.4F	0.10	Class B Amplifier
1H4G	Medium-Mu Triode	22	5S	2.0F	0.06	Class A Amplifier Class B Amplifier
1H5GT	Diode—High-Mu Triode	14A	5Z	1.4F	0.05	Triode Unit as Class A Amplifier
1H6G	Twin Diode—Medium-Mu Triode	22	7AA	2.0F	0.06	Triode Unit as Class A Amplifier
1J3	Half-Wave Rectifier	14E	3C	1.25F	0.2	Pulsed Rectifier in TV Receivers
1J5G	Power Pentode	25	6X	2.0F	0.12	Class A Amplifier
1J6G 1J6GT	Twin-Triode Amplifiers	22 13F	7AB	2.0F	0.24	Class B Amplifier
1K3	Half-Wave Rectifier	14B	3C	1.25F	0.2	Pulsed Rectifier in TV Receivers
1L6	Pentagrid Converter	5C	7DC	1.4F	0.05	Converter
1LA4	Power Pentode	12B	5AD	1.4F	0.05	Amplifier
1LA6	Pentagrid Converter	12B	7AK	1.4F	0.05	Converter
1LB4	Power Pentode	12B	5AD	1.4F	0.05	Class A Amplifier
1LC5	Sharp-Cutoff Pentode	12B	7AO	1.4F	0.05	Class A Amplifier
1LC6	Pentagrid Converter	12B	7AK	1.4F	0.05	Converter
1LD5	Diode—Sharp-Cutoff Pentode	12B	6AX	1.4F	0.05	Pentode Unit as Class A Amplifier
1LE3	Medium-Mu Triode	12B	4AA	1.4F	0.05	Class A Amplifier
1LG5	Remote-Cutoff Pentode	12B	7AO	1.4F	0.05	Class A Amplifier
1LH4	Diode—High-Mu Triode	12B	5AG	1.4F	0.05	Triode Unit as Class A Amplifier
1LN5	Sharp-Cutoff Pentode	12B	7AO	1.4F	0.05	Class A Amplifier
1N2A	Half-Wave Rectifier	19A	3C	1.25F	0.2	Pulsed Rectifier in TV Receivers
1N5GT	Sharp-Cutoff Pentode	14A	5Y	1.4F	0.05	Class A Amplifier
1N6G	Diode—Power Pentode	29A	7AM	1.4F	0.05	Pentode Unit as Class A Amplifier
1P5GT	Remote-Cutoff Pentode	14A	5Y	1.4F	0.05	Class A Amplifier
1Q5GT	Beam Power Tube	13D	6AF	1.4F	0.1	Class A Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
90	— 3V	67.5	0.7	1.6	1 M	600	—	—	—	1E5GP
180	— 3V	67.5	0.6	1.7	1.5 M	650				
135	— 7.5V	135	3.5	10.5	—	—	—	24000	0.575	1E7GT
45	0V	45	1.1	0.6	400000	Oscillator Grid (1) Resistor, 0.1 MΩ				1E8
67.5	0V	67.5	1.5	1.0	400000	Conversion Transcond., 150 micromhos				
			For other characteristics, refer to Type 1F5G							1F4
90	— 3V	90	1.1	4.0	240000	1400	—	20000	0.11	1F5G
135	— 4.5V	135	2.4	8.0	—	—	—	—	0.31	
			For other characteristics, refer to Type 1F7G							1F6
180	— 1.5V	67.5	0.7	2.2	—	—	—	—	—	1F7G
90	— 6V	—	—	2.3	10700	825	8.8	—	—	1G4GT
90	— 6V	90	2.5	8.5	133000	1500	—	8500	0.25	1G5G
135	—13.5V	135	2.5	9.7	160000	1550		9000	0.55	
90	0V	—	11	—	—	—	—	12000	0.350	1G6GT
180	—13.5V	—	—	3.1	10300	900	9.3	—	—	1H4G
157.5	—15V	—	—	1.0□	—	—	—	8000	2.1†	
90	0V	—	—	0.15	240000	275	65	—	—	1H5GT
135	— 3V	—	—	0.8	35000	575	20	—	—	1H6G
	Max. Peak Inverse Plate Volts, 26000 (Abs.) Max. Peak Plate mA, 50					Max. Average Plate mA, 0.5				1J3
135	—16.5V	135	2.0	7.0	105000	950	—	13500	0.45	1J5G
135	0V	—	—	Power Output is for one tube at				10000	2.1	1J6G
135	— 3V			stated plate-to-plate load				10000	1.9	1J6GT
	Max. Peak Inverse Plate Volts, 26000 (Abs.) Max. Peak Plate mA, 50					Max. Average Plate mA, 0.5				**1K3**
90	0V	45	0.6	0.5	650000	Anode-Grid (2): 90 max. volts, 1.2 mA Oscillator Grid (1) Resistor, 0.2 MΩ Conversion Transcond, 300 micromhos				**1L6**
			For other characteristics, refer to Type 1A5GT							1LA4
90*	0V	65	0.6	0.55	750000	Total Cathode mA, 4 Conversion Transcond. (for grid-No. 4 bias of —3 volts), 10 micromhos				1LA6
			For other characteristics, refer to Pentode Unit of Type 1D8GT							1LB4
45	0V	45	0.35	1.10	700000	750	—	—	—	1LC5
90	0V	45	0.30	1.15	1 M	775	—	—	—	
45	0V	35	0.75	0.70	300000	Anode-Grid (2): 50 max. volts, 1.4 mA Oscillator-Grid (1) Resistor, 0.2 MΩ Conversion Transcond., 275 micromhos				1LC6
90	0V	35	0.70	0.75	650000					
90	0V	45	0.1	0.6	750000	575	—	—	—	1LD5
90	0V	—	—	4.5	11200	1300	14.5	—	—	1LE3
90	— 3V	—	—	1.4	19000	760	14.5	—	—	
90	0V	45	0.4	1.7	1 M	800	—	—	—	1LG5
90	— 1.5V	90	0.9	3.7	500000	1150	—	—	—	
			For other characteristics, refer to Type 1H5GT							1LH4
90	0V	90	0.35	1.6	1.1 M	800	—	—	—	1LN5
	Max. Peak Inverse Plate Volts (Total DC and Peak), 28000 Max. Peak Plate mA, 50					Max. Average Plate mA, 0.5				1N2A
0V	90	0.3	1.2	1.5 M	750	—	—	—	90	**1N5GT**
90	— 4.5V	90	0.6	3.1	300000	800	—	25000	0.1	1N6G
90	0V	90	0.7	2.3	800000	750	—	—	—	1P5GT
110	— 6.6V	110	1.4	10	100000	2200	—	8000	0.4	1Q5GT

† For two tubes at stated plate-to-plate load. **□ For two tubes.**

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Heater or Filament (F) Amperes	Use — Values to right give operating conditions and characteristics for indicated typical use
1R5	Pentagrid Converter	5C	7AT	1.4F	0.05	Converter
1S4	Power Pentode	5C	7AV	1.4F	0.1	Class A Amplifier
1S5	Diode—Sharp-Cutoff Pentode	5C	6AU	1.4F	0.05	Pentode Unit as AF Amplifier
1T4	Remote-Cutoff Pentode	5C	6AR	1.4F	0.05	Class A Amplifier
1T5GT	Beam Power Tube	13D	6X	1.4F	0.05	Class A Amplifier
1T6	Diode—Sharp-Cutoff Pentode	29A	8DA	1.25F	0.04	Pentode Unit as Class A Amplifier
1U4	Sharp-Cutoff Pentode	5C	6AR	1.4F	0.05	Class A Amplifier
1U5	Diode—Sharp-Cutoff Pentode	5C	6BW	1.4F	0.05	Pentode Unit as Class A Amplifier
1V	Half-Wave Rectifier	22 or 13H	4G	6.3	0.3	With Capacitive-Input Filter
1X2A	Half-Wave Rectifier	7A	9Y	1.25F	0.2	Pulsed Rectifier in TV Receivers
2A3	Power Triode	27B	4D	2.5F	2.5	Class A Amplifier Push-Pull Class AB_1 Amplifier
2A5	Power Pentode	28	6B	2.5	1.75	Amplifier
2A6	Twin Diode—High-Mu Triode	24B	6G	2.5	0.8	Triode Unit as Amplifier
2A7	Pentagrid Converter	24B	7C	2.5	0.8	Converter
2AF4A 2AF4B	Medium-Mu Triode	5B	7DK	2.35	0.6	Class A Amplifier
2B7	Twin Diode—Remote-Cutoff Pentode	24B	7D	2.5	0.8	Pentode Unit as Amplifier
2BN4	Medium-Mu Triode	5C	7EG	2.3	0.6	Class A Amplifier
2DZ4	Medium-Mu Triode	5B	7DK	2.35	0.6	Class A Amplifier
2E5	Electron-Ray Tube	22 or 13H	6R	2.5	0.8	Visual Indicator
2EN5	Twin Diode	5C	7FL	2.1	0.45	Horizontal Phase Detector
3A2	Half-Wave Rectifier	7A	9DT	3.15	0.22	Pulsed Rectifier in TV Receivers
3A3 3A3/3B2	Half-Wave Rectifier	14E	8EZ	3.15	0.22	Pulsed Rectifier in TV Receivers
3A8GT	Diode-Triode—Pentode	29G	8AS	1.4F 2.8F	0.1 0.05	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
3AF4A	Medium-Mu Triode	5B	7DK	3.15	0.45	Class A Amplifier
3AW3	Half-Wave Rectifier	14B	8EZ	3.15	0.22	Pulsed Rectifier in TV Receivers
3B2	Half-Wave Rectifier	21C	8GH	3.15	0.22	Pulsed Rectifier in TV Service
3BA6	Remote-Cutoff Pentode	5C	7BK	3.15	0.6	Class A Amplifier
3BC5	Sharp-Cutoff Pentode	5C	7BD	3.15	0.6	Class A Amplifier
3BE6	Pentagrid Converter	5C	7CH	3.15	0.6	Converter
3BN4	Medium-Mu Triode	5C	7EG	3.0	0.45	Class A Amplifier
3BU8	Sharp-Cutoff Twin Pentode	6E	9FG	3.15	0.6	Class A Amplifier (With both sections operating)
3CE5	Sharp-Cutoff Pentode	5C	7BD	3.15	0.6	Class A Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power Load Ohms	Power Output Watts	RCA Type
45 90	0V 0V	45 67.5	2.1 3.5	0.7 1.5	400000 500000	Conversion Transcond., 210 μmhos Conversion Transcond., 280 μmhos				1R5
45 90	— 4.5V — 7V	45 67.5	0.8 1.4	3.8 7.4	100000 100000	1250 1575	—	8000 8000	0.065 0.27	1S4
Plate Supply, 90 V applied through 1 MΩ resistor. Screen Supply, 90 V applied through 3.1 MΩ resistor. Grid Bias, 0 volts. Grid Resistor, 10 megohms. Voltage Gain, 66 approx.										1S5
45 90	0V 0V	45 67.5	0.7 1.4	1.7 3.5	350000 500000	700 900	—	—	—	1T4
90	— 6V	90	0.8	6.5	250000	1150	—	14000	0.17	1T5GT
45 67.5	0V 0V	45 67.5	0.21 0.4	0.75 1.6	500000 400000	475 600	— —	— —	— —	1T6
90	0V	90	0.50	1.1	1 M	900	—	—	—	1U4
67.5	0V	67.5	0.4	1.6	600000	625	—	—	—	1U5
Max. AC Plate Volts (RMS), 325 Max. DC Output mA, 45				Min. Total Effective Plate-Supply Impedance: Up to 117 volts, 0 ohms; at 150 volts, 30 ohms; at 325 volts, 75 ohms						1V
Max. Peak Inverse Plate Volts, 20000 Max. Peak Plate mA, 45						Max. Average Plate mA, 0.5				1X2A
250	—45V	—	—	60.0	800	5250	4.2	2500	3.5	2A3
300 300	780Ω□ —62V	—	—	80.0□ 80.0□	—	—	—	5000 3000	10.0† 15.0†	
For other characteristics, refer to Type 6F6G										2A5
For other characteristics, refer to Type 6SQ7										2A6
For other characteristics, refer to Type 6A8										2A7
80	150Ω	—	—	17.5	2100	6500	13.5	—	—	2AF4A 2AF4B
For other characteristics, refer to Type 6B8G										2B7
150	220Ω	—	—	9	6300	6800	43	—	—	2BN4
80	—	—	—	15	2000	6700	14	—	—	2DZ4
For other characteristics, refer to Type 6E5										2E5
{Max. Peak Heater-Cathode Volts, ±200 {DC Volts Not to Exceed +100						Max. DC Plate mA,5				2EN5
Max. Peak Inverse Plate Volts, 18000 Max. Peak Plate mA, 80						Max. Average Plate mA, 1.5				3A2
Max. Peak Inverse Plate Volts, 30000 Max. Peak Plate mA, 88						Max. Average Plate mA, 1.7				3A3 3A3/3B2
90	0V	—	—	0.2	200000	325	65	—	—	3A8GT
90	0V	90	0.5	1.5	800000	750	—	—	—	
For other characteristics, refer to Type 2AF4B										3AF4A
For other characteristics, refer to Type 3A3/3B2										3AW3
Max. Peak Plate mA, 80 Max. Total DC & Peak Inverse Plate Volts, 35000 (Abs.)						Max. DC Inverse Plate Volts, 25000 Max. Average Plate mA, 1.1				3B2
100 250	68Ω 68Ω	100 100	4.4 4.2	10.8 11	250000 1 M	4300 4400	— —	— —	— —	3BA6
100 250	180Ω	100 150	1.4 2.1	4.7 7.5	600000 800000	4900 5700	—	—	—	3BC5
250	Self-Excited	100	6.8	2.9	1 M	Conversion Transcond., 475 μmhos Grid-No. 1 Resistor, 20000 ohms				3BE6
For other characteristics, refer to Type 6BN4										3BN4
100 100	— —	67.5 67.5	6.5 3.3	— 2.2	— —	— —	— —	— —	— —	3BU8
For other characteristics, refer to Type 6CE5										3CE5

† For two tubes at stated plate-to-plate load. □ For two tubes.

RCA Type	Name	Out-line	Basing Dia-gram	Heater or Filament (F) Volts	Amperes	Use Values to right give operating conditions and characteristics for indicated typical use
3CF6	Sharp-Cutoff Pentode	5C	7CM	3.15	0.6	Class A Amplifier
3DT6	Sharp-Cutoff Pentode	5C	7EN	3.15	0.6	Class A Amplifier
3DZ4	Medium-Mu Triode	5B	7DK	3.2	0.45	Class A Amplifier
3EA5	Sharp-Cutoff Tetrode	5C	7EW	2.9	0.45	Class A Amplifier
3EH7	Semiremote-Cutoff Pentode	6C	9AQ	3.4	0.6	Class A Amplifier
3EJ7	Sharp-Cutoff Pentode	6C	9AQ	3.4	0.6	Class A Amplifier
3GS8 3GS8/ 3BU8	Sharp-Cutoff Twin Pentode	6E	9LW	3.15	0.6	Class A Amplifier (With both sections operating)
3HA5	High-Mu Triode	5A	7GM	2.7	0.45	Class A Amplifier
3HS8	Sharp-Cutoff Twin Pentode	6E	9FG	3.15	0.6	Class A Amplifier (With both sections operating)
3JC6	Sharp-Cutoff Pentode	6B	9PM	3.5	0.6	Class A Amplifier
3LF4	Beam Power Tube	12B	6BA	1.4F 2.8F	0.1 0.05	Class A Amplifier
3Q4	Power Pentode	5C	7BA	1.4F 2.8F	0.1 0.05	Class A Amplifier
3Q5GT	Beam Power Tube	13D	7AP	1.4F 2.8F	0.1 0.05	Class A Amplifier
3S4	Power Pentode	5C	7BA	1.4F 2.8F	0.1 0.05	Class A Amplifier
3V4	Power Pentode	5C	6BX	1.4F 2.8F	0.1 0.05	Class A Amplifier
4BC5	Sharp-Cutoff Pentode	5C	7BD	4.2	0.45	Class A Amplifier
4BL8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9DC	4.6	0.6	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
4BU8	Sharp-Cutoff Twin Pentode	6E	9FG	4.2	0.45	Class A Amplifier (With both sections operating)
4CY5	Sharp-Cutoff Tetrode	5C	7EW	4.5	0.3	Class A Amplifier
4DT6	Sharp-Cutoff Pentode	5C	7EN	4.2	0.45	Class A Amplifier
4ES8	Variable-Mu Twin-Triode	6B	9AJ	4	0.6	Each Unit as Class A Amplifier Cascode-Type Amplifier
4GS8	Sharp-Cutoff Pentode	6E	9LW	4.2	0.45	Class A Amplifier
4GS8/ 4BU8	Sharp-Cutoff Twin Pentode	6E	9LW	4.2	0.45	Class A Amplifier (With both sections operating)
4GX7	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9QA	4.2	0.6	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
4GZ5	Power Pentode	5C	7CV	4	0.6	Class A Amplifier
4HA5	High-Mu Triode	5A	7GM	3.9	0.3	Class A Amplifier
5AS4	Full-Wave Rectifier	27A	5T	5.0F	3.0	With Capacitive-Input Filter
5AU4	Full-Wave Rectifier	19G	5T	5.0F	3.75	With Capacitive-Input Filter With Inductive-Input Filter
5AW4	Full-Wave Rectifier	19H	5T	5.0F	3.7	Rectifier
5AZ4	Full-Wave Rectifier	12C	5T	5.0F	2.0	

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
For other characteristics, refer to Type 6CF6										3CF6
150	56Ω	100	2.1	1.1	150000	515	—	—	—	3DT6
For other characteristics, refer to Type 2DZ4										3DZ4
250	—1V	140	0.95	10	150000	8000	—	—	—	**3EA5**
For other characteristics, refer to Type 6EH7										3EH7
190	— 2.35V	190	4.1	10	350000	15000	—	—	—	3EJ7
200	— 2.5V	200	4.1	10	350000	15000	—	—	—	
For other characteristics, refer to Type 4GS8/4BU8										3GS8 3GS8/ 3BU8
135	87Ω	—	10	19	1000	20000	80	—	—	3HA5
			—	11.5	5600	14500	72			
100	—	67.5	7	—	—	—	—	—	—	3HS8
100	—	67.5	4.4	2	—	—	—	—	—	
125	56Ω	125	3.2	13	180000	15000	—	—	—	3JC6
125	56Ω	125	3.4	14	180000	16000	—	—	—	
For other characteristics, refer to Type 3Q5GT										3LF4
For other characteristics, refer to Type 3V4										**3Q4**
110	— 6.6V	110	1.4	10.0	100000	2200	—	8000	0.40	**3Q5GT**
110	— 6.6V	110	1.1	8.5	110000	2000		8000	0.33	
90	— 7V	67.5	1.4	7.4	100000	1575	—	8000	0.27	**3S4**
90	— 7V	67.5	1.1	6.1	100000	1425	—	8000	0.235	
90	— 4.5V	90	2.1	9.5	100000	2150	—	10000	0.27	**3V4**
90	— 4.5V	90	1.7	7.7	120000	2000	—	10000	0.24	
250	180Ω	150	2.1	7.5	800000	5700	—	—	—	4BC5
For other characteristics, refer to Type 6BL8										4BL8
For other characteristics, refer to Type 3BU8										4BU8
125	— 1V	80	1.5	10	100000	8000	—	—	—	4CY5
150	56Ω	100	2.1	1.1	150000	515	—	—	—	4DT6
For other characteristics, refer to Type 6ES8										**4ES8**
For other characteristics, refer to Type 4GS8/4BU8										4GS8
100	:	67.5	6.0	—	Grid-No. 3 volts, each section, —10					4GS8/ 4BU8
100	:	67.5	3.6	2.0	Grid–No. 3 volts, each section, 0					
: Grid current adjusted for 100 microamperes DC										
For other characteristics, refer to Type 5GX7										**4GX7**
For other characteristics, refer to Type 6GZ5										4GZ5
135	— 1V	—	—	11.5	—	14500	72	—	—	4HA5
135	0Ω	—	—	19	—	20000	80	—	—	
Max. AC Volts per Plate (RMS), 550; Max. Peak Inverse Volts, 1550; Max. DC Output mA, 300; Max. Peak Plate mA, 1000; Min. Total Effect. Supply Imped. per Plate, 97 ohms										5AS4
Max. DC Output mA, 325 for AC Volts per Plate, 400 and Total Effect. Supply Imped. per Plate, 50 ohms; Max. Peak Inverse Volts, 1400; Max. Peak Plate mA per Plate, 1075										**5AU4**
Max. DC Output mA, 325 for AC Volts per Plate, 500 and Input Choke 10 henries; Max. Peak Inverse Volts, 1400; Max Peak Plate mA per Plate, 1075										
Max. Peak Inverse Volts, 1550; Max. Peak Plate mA per Plate, 750										5AW4
For ratings and characteristics, refer to Type 5Y3GT										**5AZ4**

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Heater or Filament (F) Amperes	Use Values to right give operating conditions and characteristics for indicated typical use
5BC3	Full-Wave Rectifier	17C	9QJ	5F	3	With Capacitive-Input Filter With Inductive-Input Filter
5BE8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9EG	4.7	0.6	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
5BT8	Twin-Diode—Sharp-Cutoff Pentode	6B	9FE	4.7	0.6	Class A Amplifier
5CL8	Medium-Mu Triode—	6B	9FX	4.7	0.6	Triode Unit as Class A Amplifier
5CM8	High-Mu Triode—Sharp-Cutoff Pentode	6B	9FZ	6.3	0.45	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
5DH8	High-Mu Triode—Sharp-Cutoff Pentode	6B	9EG	5.2	0.6	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
5GX6	Sharp-Cutoff Pentode	5C	7EN	4.7	0.6	Class A Amplifier
5GX7	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9QA	5.6	0.45	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
5HG8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9MP	5.3	0.45	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
5T4	Full-Wave Rectifier	4	5T	5.0F	2.0	With Capacitive-Input Filter With Inductive-Input Filter
5U4G	Full-Wave Rectifier	27B	5T	5.0F	3.0	With Capacitive-Input Filter
5V3	Full-Wave Rectifier	19E	5T	5.0F	3.8	With Capacitive-Input Filter With Inductive Input Filter
5V4G	Full-Wave Rectifier	25	5L	5	2	With Capacitive-Input Filter With Inductive-Input Filter
5W4 5W4GT	Full-Wave Rectifier	2B 13E	5T 5T	5.0F	1.5	With Capacitive-Input Filter
5X4G	Full-Wave Rectifier	27B	5Q	5.0F	3.0	
5Y3G	Full-Wave Rectifier	25	5T	5.0F	2.0	With Capacitive-Input Filter
5Y4G 5Y4GA 5Y4GT	Full-Wave Rectifier	25 19E 13E	5Q 5Q 5Q	5.0F	2.0	
5Z3	Full-Wave Rectifier	27B	4C	5.0F	3.0	
5Z4	Full-Wave Rectifier	2B	5L	5.0	2.0	With Capacitive-Input Filter With Inductive-Input Filter
6A3	Power Triode	27B	4D	6.3F	1.0	Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
Max. AC Volts per Plate (RMS), 500 Max. Peak Inverse Volts, 1700 Max. DC Output mA, 150 Max. Peak Plate mA per Plate, 1000 Min. Total Effect. Supply Imped. per Plate, 21 ohms										5BC3
Max. AC Volts per Plate (RMS), 600 Max. Peak Inverse Volts, 1700 Max. DC Output mA, 150 Max. Peak Plate mA per Plate, 1000 Min. Value of Input Choke, 10 henries										5BC3
150	56Ω	—	—	18	5000	8500	40	—	—	5BE8
250	68Ω	110	3.5	10	400000	5200	—	—	—	5BE8
200	180Ω	150	2.8	9.5	300000	6200	—	—	—	5BT8
125	— 1V	—	—	14	5000	8000	40	—	—	5CL8
For other characteristics, refer to 6CM8										5CM8
250	390Ω	—	—	7.3	12000	4400	53	—	—	5DH8
125	56Ω	125	3.8	13.5	150000	8600	—	—	—	5DH8
150	180Ω	100	3	3.7	140000	3700 (Grid-No. 1 to Plate) 750 (Grid-No. 3 to Plate)	—	—	—	5GX6
100	—	—	—	12.5	—	8700	40	—	—	5GX7
125	— 1V	—	—	13	4700	8500	—	—	—	5GX7
120	—	90	2.8	8.5	—	13000	—	—	—	5GX7
125	— 1V	125	2.5	8	200000	11000	—	—	—	5GX7
For other characteristics, refer to Type 6HG8										5HG8
Max. AC Volts per Plate (RMS), 450 Max. Peak Inverse Volts, 1550 Max. DC Output mA, 225 Max. Peak Plate mA, 675 Min. Total Effect. Supply Imped. per Plate, 150 ohms										5T4
Max. AC Volts per Plate (RMS), 550 Max. Peak Inverse Volts, 1550 Max. DC Output mA, 225 Max. Peak Plate mH, 675 Min. Value of Input Choke, 10 henries										5T4
Max. AC Volts per Plate (RMS), 450 Max. Peak Inverse Volts, 1550 Max. DC Output mA, 225 Max. Peak Plate mA, 675 Min. Total Effect. Supply Imped. per Plate, 170 ohms										5U4G
Max. AC Volts per Plate (RMS), 425 Max. Peak Inverse Volts, 1400 Max. DC Output mA, 350 Max. Peak Plate mA per Plate, 1200 Min. Total Effect. Supply Imped. per Plate, 56 ohms										5V3
Max. AC Volts per Plate (RMS), 500 Max. Peak Inverse Volts, 1400 Max. DC Output mA, 350 Max. Peak Plate mA per Plate, 1200 Min. Value of Input Choke, 10 henries										5V3
Max. AC Volts per Plate (RMS), 375 Max. Peak Inverse Volts, 1400 Max. DC Output mA, 175 Max. Peak Plate mA per Plate, 525 Min. Total Effect. Supply Imped. per Plate, 100 ohms										5V4G
Max. AC Volts per Plate (RMS), 500 Max. Peak Inverse Volts, 1400 Max. DC Output mA, 175 Max. Peak Plate mA per Plate, 525 Min. Value of Input Choke, 4 henries										5V4G
Max. Peak Inverse Volts, 1400 Max. DC Output mA, 100 Max. Peak Plate mA,300										5W4 5W4GT
For other ratings, refer to Type 5U4G										5X4G
Max. AC Volts per Plate (RMS), 350 Max. Peak Inverse Volts, 1400 Max. DC Output mA, 125 Max. Peak Plate mA, 440 Min. Total Effect. Supply Imped. per Plate, 50 ohms										5Y3G
Max. Peak Plate mA, 375 (5Y4G) Max. Peak Plate mA, 400 (5Y4GA, 5Y4GT) For other ratings, refer to Type 5Y3G										5Y4G 5Y4GA 5Y4GT
For other ratings, refer to Type 5U4G										5Z3
Max. AC Volts per Plate (RMS), 350 Max. Peak Inverse Volts, 1400 Max. DC Output mA, 125 Max. Peak Plate mA, 375 Min. Total Effect Supply Imped. per Plate, 50 ohms										5Z4
Max. AC Volts per Plate (RMS), 500 Max. Peak Inverse Volts, 1400 Max. DC Output mA, 125 Max. Peak Plate mA, 375 Min. Value of Input Choke, 5 henries										5Z4
For other characteristics, refer to Type 6B4G										6A3

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Amperes	Use Values to right give operating conditions and characteristics for indicated typical use
6A6	High-Mu Twin Power Triode	28	7B	6.3	0.8	Amplifier
6A7 6A7S	Pentagrid Converter	24B 24B	7C	6.3	0.3	Converter
6A8 6A8G 6A8GT	Pentagrid Converter	3 23 14A	8A 8A 8A	6.3	0.3	Converter
6AB5/ 6N5	Electron-Ray Tube	22 or 13H	6R	6.3	0.15	Visual Indicator
6AB7	Sharp-Cutoff Pentode	2A	8N	6.3	0.45	Class A Amplifier
6AC5GT	*High-Mu Power Triode*	13D	6Q	6.3	0.4	Class B Amplifier *Dynamic-Coupled Amplifier With 76 Driver*
6AC7	Sharp-Cutoff Pentode	2A	8N	6.3	0.45	Class A Amplifier
6AD6G	*Electron-Ray Tube*	29E	7AG	6.3	0.15	*Visual Indicator*
6AD7G	Low-Mu Triode—Power Pentode	25	8AY	6.3	0.85	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6AE5GT	Low-Mu Triode	13D	8Q	6.3	0.3	Class A Amplifier
6AE6G	Twin-Plate Control Tube	22	7AH	6.3	0.15	Remote Cutoff Triode *Sharp-Cutoff Triode*
6AE7GT	Twin-Input Triode	13D	7AX	6.3	0.5	Class A Amplifier
6AH4GT	Low-Mu Triode	13D	8EL	6.3	0.75	Vertical Deflection Amplifier
6AH6	Sharp-Cutoff Pentode	5C	7BK	6.3	0.45	Class A Amplifier
6AL3	Half-Wave Rectifier	7D	9CB	6.3	1.55	Television Damper Service
6AL7GT	Electron-Ray Tube	13C	8CH	6.3	0.15	Visual Indicator
6AM4	High-Mu Triode	6A	9BX	6.3	0.225	Class A Amplifier
6AM8	Diode—Sharp-Cutoff Pentode	6B	9CY	6.3 6.3	0.45 0.45	Diode Unit Pentode Unit as Class A Amplifier
6AN8	*Medium-Mu Triode—Sharp-Cutoff Pentode*	6B	9DA	6.3 6.3	0.45 0.45	Triode Unit as Class A Amplifier Pentrode Unit as Class A Amplifier
6AQ5	Beam Power Tube	5D	7BZ	6.3 6.3	0.45 0.45	Single Tube Class A Amplifier *Push-Pull Class A_1 Amplifier*
6AQ6	Twin-Diode—High-Mu Triode	5C	7BT	6.3	0.15	Triode Unit as Class A Amplifier
6AQ7GT	Twin-Diode—High-Mu Triode	13D	8CK	6.3	0.3	Triode Unit as Class A Amplifier
6AQ8	High-Mu Twin Triode	6B	9AJ	6.3	0.435	Each Unit as Class A Amplifier
6AR5	Power Pentode	5D	6CC	6.3	0.4	Class A Amplifier
6AS11	*Dual Triode—Sharp-Cutoff Pentode*	8B	12DP	6.3	1.05	Dual Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6AT8	Medium-Mu Triode	6B	9DW	6.3	0.45	Triode Unit as Class A Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power Load Ohms	Power Output Watts	RCA Type
For other characteristics, refer to Type 6N7GT										6A6
For other characteristics, refer to Type 6A8										**6A7** 6A7S
250	— 3V	100	2.7	3.5	360000	Anode-Grid (2): 250 max. V, 4.0 mA Oscillator-Grid (1) Res. Conversion Transcond., 550 μmhos				**6A8** 6A8G 6A8GT
Plate & Target Supply = 135 volts. Triode Plate Resistor = 0.25 MΩ Target Current = 2.0 mA Grid Bias, — 10.0 volts; Shadow Angle, 0°. Bias, 0 volts; Angle, 90°; Plate Current, 0.5 mA.										6AB5/ 6N5
Plate & Target Supply = 135 volts. Triode Plate Resistor = 1.0 MΩ Target Current = 1.9 mA Grid Bias, — 15.5 volts; Shadow Angle, 0°. Bias, 0 volts; Angle, 90°; Plate Current, 0.13 mA										6AB5/ 6N5
300	— 3V	200	3.2	12.5	700000	5000	—	—	—	6AB7
250	0V	—	—	5.0□	—	—	—	10000	8.0†	**6AC5GT**
250	Bias for both 6AC5GT and 76 is developed in coupling circuit *Average Plate Current of Driver = 5.5 milliamperes* Average Plate Current of 6AC5GT = 32 milliamperes							*7000*	3.7	**6AC5GT**
300	160Ω	150	2.5	10.0	1 M	9000	—	—	—	**6AC7**
Target Voltage, 150 volts. Control-Electrode Voltage, —50 volts; Shadow Angle, 135°; Target *Current, 1.2 mA Control-Electrode Voltage, 75 volts; Angle, 0°; Target Current, 3 mA*										6AD6G
250	—25V	—	—	3.7	19000	325	6	—	—	6AD7G
250	—16.5V	250	6.5	34.0	80000	2500	—	7000	3.2	6AD7G
95	*—15V*	—	—	*7.0*	*3500*	*1200*	4.2	—	—	6AE5GT
250	— 1.5V	—	—	6.5	25000	1000	25	—	—	6AE6G
250	—35V	—	—	0.01	—	—	—	—	—	6AE6G
250	— 1.5V	—	—	4.5	35000	950	33	—	—	6AE6G
250	— 9.5V	—	—	0.01	—	—	—	—	—	6AE6G
250	—13.5V	—	—	10.0	4650	3000	14	—	—	6AE7GT
Max. DC Plate Volts, 500 Max. DC Cathode mA, 60						Max. Peak Positive-Pulse Plate Volts, 2000 Max. Plate Dissipation, 7.5 watts				**6AH4GT**
300	160Ω	150	2.5	10.0	500000	9000	—	—	—	**6AH6**
Max. Peak Inverse Plate Volts, 7500 (Abs.) Max. Peak Plate mA, 550 Max. DC Plate mA, 220							Max. Plate Dissipation, 5 watts Max. Peak Heater-Cathode Volts, 6600			6AL3
Target Voltage, 315 volts Grid Voltage = 0 volts Cathode Bias Res., 3300 ohms approx.					*Grid Voltage for Pattern Cutoff, —7 volts approx.* Deflecting-Electrodes—No. 1, No. 2 and No. 3 Voltage, 0					**6AL7GT**
200	100Ω	—	—	10	8700	9800	85	—	—	**6AM4**
		Max. DC Plate mA, 5			*Max. Peak Heater-Cathode Volts, ±200*					6AM8
125	56Ω	125	3.2	12.5	—	7800	—	—	—	6AM8
150	— 3V	—	—	15	4500	4700	31	—	—	6AN8
125	56Ω	125	3.8	12	170000	7800	—	—	—	6AN8
180	— 8.5V	180	3.0	29.0	50000	3700	—	5500	2.0	6AQ5
250	—12.5V	250	4.5	45.0	50000	4100	—	5000	4.5	6AQ5
250	—15V	250	5.0□	70.0□	60000	—	—	10000	10.0†	6AQ5
100	— 1V	—	—	0.8	61000	1150	70	—	—	**6AQ6**
250	— 3V	—	—	1.0	58000	1200	70	—	—	**6AQ6**
250	— 2V	—	—	2.3	44000	1600	70	—	—	**6AQ7GT**
250	— 2.3V	—	—	10	—	5900	57	—	—	6AQ8
250	—18V	250	5.5	32.0	90000	2300	—	7600	3.4	**6AR5**
200	220Ω	—	—	9.2	4400	4400	41	—	—	**6AS11**
200	— 2V	—	—	7	12400	5500	68	—	—	**6AS11**
200	125	125	5.2	24	70000	10500	—	—	—	**6AS11**
125	— 1V	—	—	12	6000	6500	40	—	—	6AT8

† For two tubes at stated plate-to-plate load. □ For two tubes.

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Amperes	Use Values to right give operating conditions and characteristics for indicated typical use
6AU4GT	Half-Wave Rectifier	13G	4CG	6.3	1.8	Television Damper Service
6AU6	Sharp-Cutoff Pentode	5C	7BK	6.3 6.3	0.3 0.3	Class A Amplifier
6AU7	*Medium-Mu Twin Triode*	6B	9A	3.15 6.3	0.6 0.3	Each Unit as Class A Amplifier
6AU8	Medium-Mu Triode—Sharp-Cutoff Pentode	6E	9DX	6.3	0.6	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6AV5GT	Beam Power Tube	13D	6CK	6.3	1.2	Horizontal Deflection Amplifier
6AW8	High-Mu Triode—Sharp-Cutoff Pentode	6E	9DX	6.3	0.6	Triode Unit as Class A Amplifier Pentode Unit as Class A *Amplifier*
6AX4GT	Half-Wave Rectifier	13D	4CG	6.3	1.2	Television Damper Service
6AX8	Medium-Mu Triode—Semiremote Cutoff Pentode	6B	9AE	6.3	0.45	*Triode Unit as Class A Amplifier* Pentode Unit as Class A Amplifier
6AY3	Half-Wave Rectifier	11D	9HP	6.3	1.2	Television Damper Service
6AY11	Twin Diode—High-Mu Twin Triode	8A	12DA	6.3	0.69	Each Triode Unit as Class A Amplifier
6B4G	Power-Triode	27B	5S	6.3F	1.0	Class A Amplifier
6B5	Direct-Coupled Power Triode	26	6AS	6.3	0.8	Class A Amplifier
6B6G	Twin-Diode—High-Mu Triode	23	7V	6.3	0.3	Triode Unit as Amplifier
6B7 6B7S	Twin-Diode—Remote-Cutoff Pentode	24B 24B	*7D*	6.3	*0.3*	*Pentode Unit as Amplifier*
6B8	Twin-Diode—Semiremote-Cutoff Pentode	3	8E	6.3	0.3	Pentode Unit as Amplifier
6B8G	Twin Diode—Semiremote-Cutoff Pentode	*23*	*8E*	6.3	*0.3*	Pentode Unit as Class A Amplifier
6BD4	Sharp-Cutoff Beam Triode	21C	8FU	6.3	0.6	Voltage-Control
6BD4A	Sharp-Cutoff Beam Triode	21C	8FU	6.3	0.6	Voltage-Control
6BD6	Remote-Cutoff Pentode	5C	7BK	6.3	0.3	Class A Amplifier
6BF5	Beam Power Tube	5D	7BZ	6.3	1.2	Class A Amplifier
6BF6	*Twin-Diode—Medium-Mu Triode*	5C	7BT	*6.3*	*0.3*	*Triode Unit as Class A Amplifier*
6BG6G **6BG6GA**	Beam Power Tube	28B 21B	5BT 5BT	6.3	0.9	Horizontal Deflection Amplifier
6BH3	Half-Wave Rectifier	11D	9HP	6.3	1.6	Television Damper Service
6BK4 6BK4A	Beam Triode	21B	8GC	6.3	0.2	Voltage-Control
6BK5	Beam Power Tube	6E	9BQ	6.3	1.2	Class A Amplifier
6BK7A	Medium-Mu Twin Triode	6B	9AJ	6.3 6.3	0.45 0.45	Each Unit as Class A Amplifier
6BL4	Half-Wave Rectifier	13F	8GB	6.3	3.0	Television Damper Service

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
Max. Peak Inverse Plate Volts, 4500 (Absolute) Max. Peak Plate mA, 1050						Max. Average Plate mA, 175 Max. Plate Dissipation 6.0 watts				6AU4GT
100 250	150Ω 68Ω	100 150	2.1 4.3	5.0 10.6	500000 1 M	3900 5200	— —	— —	— —	6AU6
100 250	0V — 8.5V	— —	— —	11.8 10.5	6250 7700	3500 2200	19.5 17	— —	— —	6AU7
150	150Ω	—	—	9	8200	4900	40	—	—	6AU8
200	82Ω	125	3.4	15	150000	7000	—	—	—	6AU8
Max. DC Plate Volts, 550 Max. DC Cathode mA, 110					Max. Peak Positive-Pulse Plate Volts, 5500 (Abs.) Max. Plate Dissipation, 11 watts					6AV5GT
200	— 2V	—	—	4	—	4000	70	—	—	6AW8
150	150Ω	150	3.5	13	200000	9500	—	—	—	6AW8
6AW8A Features a plate current characteristic with a controlled knee										6AW8
Max. Peak Inverse Plate Volts, 4400 Max. Peak Plate mA, 750 Max. DC Plate mA, 125					Max. Peak Heater-Cathode Volts: {—4400** +300 **DC component must not exceed 900 volts					6AX4GT
150	560Ω	—	—	18	5000	8500	40	—	—	6AX8
250	120Ω	110	3.5	10	400000	4800	—	—	—	6AX8
Max. Peak Inverse Plate Volts, 5000 Max. Peak Plate mA, 1100 Max. DC Plate mA 175					Max. Plate Dissipation, 6.5 watts Max. Peak Heater-Cathode Volts: {— 5000 + 300					6AY3
250	— 2V	—	—	1.2	52700	1900	100	—	—	6AY11
250	—45V	—	—	60	800	5250	4.2	2500	3.5	6B4G
For other characteristics, refer to Type 6N6G										6B5
For other characteristics, refer to Type 6SQ7										6B6G
Input Triode: Plate Volts, 300 max; Grid Volts, 0; Plate mA, 8; AF Signal Volts (Peak), 21 Output Triode: Plate Volts, 300 max.; Plate mA, 45; Plate Res., 24000 ohms; Load Resistance, 7000 ohms; Power Output, 4 watts										6B7 6B7S
For other characteristics, refer to Type 12C8										6B8
250	— 3V	125	2.3	9	600000	1125	—	—	—	6B8G
Max. DC Plate Volts, 20000 Max. Unregulated DC Supply Volts, 40000						Max. DC Plate mA, 1.5 Max. Plate Dissipation, 20.0 watts				6BD4
Max. DC Plate Volts, 27000 Max. Unregulated DC Supply Volts, 55000						Max. DC Plate mA, 1.5 Max. Plate Dissipation, 25.0 watts				6BD4A
250	— 3V	100	3.0	9.0	800000	2000	—	—	—	6BD6
110	— 7.5V	110	4.0	36.0	12000	7500	—	2500	1.9	6BF5
250	— 9V	—	—	9.5	8500	1900	16	Power Output, 300 milliwatts		6BF6
Max. DC Plate Volts, 700 Max. DC Cathode mA, 110				Max. Peak Positive-Pulse Plate Volts, 6600 (Abs.) Max. Plate Dissipation, 20 watts						6BG6G 6BG6GA
Max. Peak Inverse Plate Volts, 5500 Max. Peak Plate mA, 1100 Max. DC Plate mA, 180					Max. Plate Dissipation, 6.5 watts Max. Peak Heater-Cathode Volts: {— 5500 + 300					6BH3
Max. DC Plate Volts, 27000 Max. Unregulated DC Supply Volts, 60000 Max. Plate Dissipation, 30 Watts (6BK4A)						Max. DC Plate mA, 1.6 Max. Plate Dissipation, 25 Watts (6BK4)				6BK4 6BK4A
250	— 5V	250	3.5	35	100000	8500	—	6500	3.5	6BK5
150	56Ω	—	—	18	4600	9300	43	Grid-No. 1 Volts for Cutoff, —11		6BK7A
Max. Peak Inverse Plate Volts, 4500 (Abs.) Max. Peak Plate mA, 1200 Max. DC Plate mA, 200					Max. Peak Heater-Cathode Volts: {—4500* (Abs.) +300 *DC component not to exceed —900 volts					6BL4

RCA Type	Name	Out-line	Basing Dia-gram	Heater or Filament (F) Volts	Amperes	Use Values to right give operating conditions and characteristics for indicated typical use
6BL7GT	Medium-Mu Twin Triode	13D	8BD	6.3	1.5	Vertical Deflection Amplifier
6BL8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9DC	6.3	0.45	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6BN4	Medium-Mu Triode	5C	7EG	6.3	0.2	Class A Amplifier
6BQ6GT	Beam Power Tube	14D	6AM	6.3	1.2	Horizontal Deflection Amplifier
6BQ7	Medium-Mu Twin Triode	6B	9AJ	6.3	0.4	Each Unit as Class A Amplifier
6BR8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9FA	6.3 6.3	0.45 0.45	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6BS3	Half-Wave Rectifier	11D	9HP	6.3	1.2	Television Damper Service
6BV8	Twin Diode—Medium-Mu Triode	6B	9FJ	6.3	0.6	Triode Unit as Class A Amplifier
6BW4	Full-Wave Rectifier	6E	9DJ	6.3	0.9	With Capacitive Input Filter With Inductive Input Filter
6BX7GT	Medium-Mu Twin Triode	13D	8BD	6.3	1.5	Vertical Deflection Oscillator Vertical Deflection Amplifier
6BY5GA	*Full-Wave Rectifier*	*18B*	*6CN*	*6.3*	*1.6*	*Television Damper Service*
6BZ8	Medium-Mu Twin Triode	6B	9AJ	6.3	0.4	Each Unit as Class A Amplifier
6C5 6C5GT	*Medium-Mu Triode*	2A 14A	6Q 6Q	6.3	0.3	Class A Amplifier
6C6	Sharp-Cutoff Pentode	24A	6F	6.3	0.3	Amplifier Detector
6C7	Twin-Diode—Medium-Mu Triode	24B	7G	6.3	0.3	Triode Unit as Class A Amplifier
6C8G	Medium-Mu Twin-Triode	23	8G	6.3	0.3	Each Unit as Class A Amplifier
6CA7	Power Pentode		8ET	6.3	1.5	Class A Amplifier Push-Pull Class AB, Amplifier
6CB5	*Beam Power Tube*	*28A*	*8GD*	6.3	2.5	*Horizontal Deflection Amplifier*
6CB6	Sharp-Cutoff Pentode	5C	7CM	6.3	0.3	Class A Amplifier
6CD6G	Beam Power Tube	28B	5BT	6.3	2.5	Horizontal Deflection Amplifier
6CE5	Sharp-Cutoff Pentode	5C	7BD	6.3	0.3	Class A Amplifier
6CF6	Sharp-Cutoff Pentode	5C	7CM	6.3	0.3	Class A Amplifier
6CG8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9GF	6.3 6.3	0.45 0.45	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6CH8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9FT	6.3	0.45	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6CK4	Low-Mu Triode	13F	8JB	6.3	1.25	Vertical Deflection Amplifier
6CL8	Medium-Mu Triode—Sharp-Cutoff Tetrode	6B	9FX	6.3	0.45	Triode Unit as Class A Amplifier Tetrode Unit as Class A Amplifier
6CM8	High-Mu Triode—Sharp-Cutoff Pentode	6B	9FZ	6.3	0.45	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
Max. DC Plate Volts, 500 Max. DC Cathode mA. (Each Unit), 60					Max. Peak Positive-Pulse Plate Volts, 2000 (Abs.) Max. Plate Dissipation (Each Unit), 10 watts					6BL7GT
100	— 2V	—	—	14	—	5000	20	—	—	6BL8
170	— 2V	170	2.8	10	400000	6200	—	—	—	
150	220Ω	—	—	9	6300	6800	43	—	—	6BN4
Max. DC Plate Volts, 550 Max. DC Cathode mA, 110					Max. Peak Positive-Pulse Plate Volts, 5500 (Abs.) Max. Plate Dissipation, 11 watts					6BQ6GT
150	220Ω	—	—	9.0	5800	6000	35	Grid-No. 1 Volts for Cutoff, —10		6BQ7
125	— 1V	—	—	13.5	7500	—	40	—	—	6BR8
125	— 1V	110	3.5	9.5	200000	5000	—	—	—	
Max. Peak Inverse Plate Volts, 5000 Max. Peak Plate mA, 1100 Max. DC Plate mA, 200					Max. Plate Dissipation, 6 watts Max. Peak Heater-Cathode Volts: { — 5000, + 300					6BS3
200	330Ω	—	—	11	5900	5600	33	—	—	**6BV8**
Max. AC Volts per Plate (RMS), 325 Max. Peak Inverse Volts, 1275 Total Effect. Supply Imped. per Plate, 82 ohms					Max. DC Output mA, 62.5 Max. Peak Plate mA, per Plate, 350					**6BW4**
Max. AC Volts per Plate (RMS), 450 Max. Peak Inverse Volts, 1275 Min. Value of Input Choke, 10 henries					Max. DC Output mA, 62.5 Max. Peak Plate mA per Plate, 350					
Max. DC Plate Volts, 500 Max. Plate Dissipation: 10 watts either plate; 12 watts both plates					Max. DC Cathode mA, 180					**6BX7GT**
Max. DC Plate Volts, 500 Max. DC Cath. mA, 180					Max. Peak Positive-Pulse Plate Volts, 2000 (Abs.) Max. Plate Dissipation: 10 watts either plate; 12 watts both plates					
Max. Peak Inverse Plate Volts, 3000 (Abs.) Max. Peak Plate mA, 525 Max. DC Plate mA, 175					Max. Peak Heater-Cathode Volts: { —450, +100					**6BY5GA**
125	100Ω	—	—	10	5600	8000	45	—	—	**6BZ8**
250	— 8V	—	—	8.0	10000	2000	20	—	—	**6C5** 6C5GT
For other characteristics, refer to Type 6J7										**6C6**
250	— 9V	—	—	4.5	16000	1250	20	—	—	6C7
250	— 4.5V	—	—	3.2	22500	1600	36	—	—	6C8G
265	—13.5V	250	15	100	15000	11000	—	2000	11	6CA7
450	232Ω	450	20	120	—	—	—	6500	40	
Max. DC Plate Volts, 700 Max. DC Cathode mA, 200					Max. Peak Positive-Pulse Plate Volts, 6800 (Abs.) Max. Plate Dissipation, 23 Watts					6CB5
125	56Ω	125	3.7	13	280000	8000	—	—	—	**6CB6**
Max. DC Plate Volts, 700 Max. DC Cathode mA, 200					Max. Peak Positive-Pulse Plate Volts, 7000 Max. Plate Dissipation, 20 watts					6CD6G
125	— 1V	125	2.3	11	300000	7600	—	—	—	6CE5
125	56Ω	125	3.7	12.5	300000	7800	—	—	—	6CF6
100	— 1V	—	—	12	6000	6500	40	—	—	6CG8
250	— 1V	125	2.2	9	300000	5500	—	—	—	
200	— 6V	—	—	13	5750	3300	19	—	—	**6CH8**
200	180Ω	150	2.8	9.5	300000	6200	—	—	—	
Max. DC Plate Volts, 550 Max. Peak Cathode mA, 350					Max. Peak Positive-Pulse Plate Volts, 2000 (Abs.) Max. Plate Dissipation, 12 watts					**6CK4**
125	— 1V	—	—	14	5000	8000	40	—	—	6CL8
125	— 1V	125	4	12	120000	6000	—	—	—	
250	— 2V	—	—	1.8	50000	2000	100	—	—	**6CM8**
250	180Ω	150	2.8	9.5	600000	6200	—	—	—	

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Amperes	Use: Values to right give operating conditions and characteristics for indicated typical use
6CW5	Power Pentode	6G	9CV	6.3	0.76	Vertical-Deflection Amplifier
6D6	Remote-Cutoff Pentode	24A	6F	6.3	0.3	Amplifier Mixer
6D7	Sharp-Cutoff Pentode	24A	7H	6.3	0.3	Amplifier Detector
6D8G	Pentagrid Converter	23	8A	6.3	0.15	Converter
6DC8	Twin Diode—Remote-Cutoff Pentode	6E	9HE	6.3	0.3	Class A Amplifier
6DL5	Power Pentode	5E	7DQ	6.3	0.2	Class A Amplifier
6DM4	Half-Wave Rectifier	13G	4CG	6.3	1.2	Damper Service
6DN6	Beam Power Tube	21B	5BT	6.3	2.5	Horizontal Deflection Amplifier
6DQ4	Half-Wave Rectifier	13F	4CG	6.3	1.2	Damper Service
6DQ6A 6DQ6B	Beam Power Tube	20	6AM	6.3	1.2	Horizontal Deflection Amplifier
6DT6	Sharp-Cutoff Pentode	5C	7EN	6.3	0.3	Class A Amplifier
6DW4	Half-Wave Rectifier	11D	9HP	6.3	1.2	Television Damper Service
6DW5	Beam Power Tube	6G	9CK	6.3	1.2	Vertical Deflection Amplifier
6DX8	High-Mu Triode—Sharp-Cutoff Pentode	6E	9HX	6.3	0.72	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6DZ7	Twin Power Pentode	19B	8JP	6.3	1.52	Class A Amplifier Both Units as Push-Pull Class AB_1 Amplifier
6E6	Twin Power Amplifier	26	7B	6.3	0.6	Push-Pull Class A Amplifier
6E7	Remote-Cutoff Pentode	24A	7H	6.3	0.3	Amplifier
6EA4	High-Mu Triode	10D	12FA	6.3	0.2	Shunt Voltage Regulator
6EA5	Sharp-Cutoff Tetrode	5C	7EW	6.3	0.2	Class A Amplifier
6EH7	Semiremote-Cutoff Pentode	6C	9AQ	6.3	0.3	Class A Amplifier
6EH8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9JG	6.3	0.45	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6EJ7	Sharp-Cutoff Pentode	6C	9AQ	6.3	0.3	Class A Amplifier
6ES8	Variable-Mu Twin Triode	6B	9AJ	6.3	0.365	Each Unit as Class A Amplifier Cascode-Type Amplifier
6EV7	High-Mu Twin Triode	6E	9LP	6.3	0.6	Relay Control
6EX6	Beam Power Tube	21B	5BT	6.3	2.25	Horizontal Deflection Amplifier
6EY6	Beam Power Tube	13F	7AC	6.3	0.68	Vertical Deflection Amplifier
6EZ5	Beam Power Tube	13F	7AC	6.3	0.8	Vertical Deflection Amplifier
6F5 6F5GT	High-Mu Triode	3 14A	5M 5M	6.3	0.3	Class A Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
Max. DC Plate Volts, 275; Max. DC Cathode mA, 110; Max. Peak Positive-Pulse Plate Volts, 2200; Max. Plate Dissipation, 12 watts										6CW5
For other characteristics, refer to Type 6U7G										6D6
For other characteristics, refer to Type 6J7										6D7
250	— 3V	100	2.7	3.5	360000	Anode-Grid (2): 250 max. volts, 4 mA Oscillator-Grid (1) Resistor. Conversion Transcond., 550 micromhos.				6D8G
250	— 2V	100	2.7	9	1 M	3800	—	—	—	6DC8
200	230Ω	200	4.2	23	—	—	—	8000	2.3	6DL5
250	320Ω	250	4.5	24	—	—	—	10000	3	6DL5
Max. Peak Inverse Plate Volts, 5000; Max. Peak Plate mA, 1100; Max. DC Plate mA, 175; Max. Peak Heater—Cathode Volts, —5000 (DC Component Not to Exceed 900 Volts); Max. Peak Heater—Cathode Volts, +300 (DC Component Not to Exceed 100 Volts)										6DM4
Max. DC Plate Volts, 700; Max. DC Cathode mA, 200; Max. Peak Positive-Pulse Plate Volts, 6600 (Abs.); Max. Plate Dissipation, 15 watts										6DN6
Max. Peak Inverse Volts, 5500; Max. Peak Plate mA, 1000; Max. DC Plate mA, 175; Max. Plate Dissipation, 6 watts										6DQ4
Max. DC Plate Volts, 770; Max. DC Cathode mA, 155 (6DQ6A); Max. DC Cathode mA, 175 (6DQ6B); Max. Peak Positive-Pulse Plate Volts, 6000 (Abs.); Max. Plate Dissipation, 18 watts										6DQ6A 6DQ6B
150	560Ω	100	2.1	1.1	150000	515	—	—	—	6DT6
Max. Peak Inverse Plate Volts, 5000; Max. Peak Plate mA, 1300; Max. DC Plate mA, 250; Max. Plate Dissipation, 8.5; Max. Peak Heater-Cathode Volts: — 5000, + 300										6DW4
Max. DC Plate Volts, 330; Max. DC Cathode mA, 65; Max. Peak Positive-Pulse Plate Volts, 2200; Max. Plate Dissipation, 11 watts										6DW5
200	— 1.7V	—	—	3	—	4000	65	—	—	6DX8
170	— 2.1V	170	3	18	100000	11000	—	—	—	6DX8
200	— 2.9V	200	3	18	130000	10400	—	—	—	6DX8
220	— 3.4V	220	3	18	150000	10000	—	—	—	6DX8
250	— 7.3V	250	5.5	48	38000	11300	—	—	—	6DZ7
400	—11V	250	13	100	—	—	—	9000	18	6DZ7
300	120Ω	250	15	80	—	—	—	9000	12	6DZ7
250	—27.5V	—	—	—	—	—	—	14000	1.60†	6E6
For other characteristics, refer to Type 6U7G										6E7
Max. DC Plate Volts, 27000; Max. Unregulated DC Plate Supply Volts, 60000; Max. Plate Dissipation, 30 watts; Max. DC Plate mA, 1.6										6EA4
250	— 1V	140	0.95	10	150000	8000	—	—	—	6EA5
200	— 2V	90	4.5	12	500000	12500	—	—	—	6EH7
125	— 1V	—	—	13.5	—	7500	40	—	—	6EH8
125	— 1V	125	4	12	170000	6000	—	—	—	6EH8
200	— 2.5V	200	4.1	10	350000	15000	—	—	—	6EJ7
90	— 1.2V	—	—	15	2500	12500	—	—	—	6ES8
180	—	—	—	15	—	12500	—	—	—	6ES8
250	0V	—	—	18.5	Grid Volts for Plate μA 100 = —9			2500-ohm relay		6EV7
150	0V	—	—	10.0	Grid Volts for Plate μA 100 = —5					6EV7
175	—30V	175	3.3	67	8500	7700	—	—	—	6EX6
250	—17.5V	250	3	44	60000	4400	—	—	—	6EY6
250	—20V	250	3.5	43	50000	4100	—	—	—	6EZ5
100	— 1V	—	—	0.4	85000	1150	100	—	—	6F5 6F5GT
250	— 2V	—	—	0.9	66000	1500	100	—	—	6F5 6F5GT

† For two tubes at stated plate-to-plate load.

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Amperes	Use: Values to right give operating conditions and characteristics for indicated typical use
6F6G **6F6GT**	Power Pentode	25 13F	7S 7S	6.3	0.7	Pentode Class A Amplifier Triode□ Class A Amplifier Pentode Push-Pull Class A Amplifier
6F7	Low-Mu Triode—Remote-Cutoff Pentode	24B	7E	6.3	0.3	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6F8G	Medium-Mu Twin Triode	23	8G	6.3	0.6	Each Unit as Class A Amplifier
6FE5	Beam Power Tube	13G	8KB	6.3	1.2	Class A Amplifier
6FG6	Refer to type EM84/6FG6					
6FQ7	Medium-Mu Twin Triode	6E	9LP	6.3	0.6	Each Unit as Class A Amplifier
6FV8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9FA	6.3	0.45	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6FW8	Medium-Mu Twin Triode	6B	9AJ	6.3	0.4	Each Unit as Class A Amplifier
6G6G	Power Pentode	22	7S	6.3	0.15	Pentode Class A Amplifier
6G11	Beam Power Tube—Sharp-Cutoff Pentode	8B	12BU	6.3	1.2	Beam Power Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6GB5	Beam Power Tube	10E	9NH	6.3	1.38	Horizontal Deflection Amplifier
6GF7	Dual Triode	11A	9QD	6.3	0.985	Vertical Deflection Oscillator Vertical Deflection Amplifier
6GH8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9AE	6.3	0.45	Triode Unit as Horiz. Defl. Osc. Pentode Unit as Horiz. Defl. Osc.
6GJ5	Novar Beam Power Tube	18A	9QK	6.3	1.2	Horizontal Deflection Amplifier
6GJ7	Medium-Mu Triode—Sharp-Cutoff Pentode	6J	9QA	6.3	0.41	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6GJ8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9AE	6.3	0.6	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6GT5	Beam Power Tube	17B	9NZ	6.3	1.2	Horizontal Deflection Amplifier
6GV8	High-Mu Triode—Power Pentode	6G	9LY	6.3	0.9	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6GW6	Beam Power Tube	20	6AM	6.3	1.2	Horizontal Deflection Amplifier
6GZ5	Power Pentode	5C	7CV	6.3	0.38	Class A Amplifier
6H6 6H6GT	Twin Diode	29B 13D	7Q 7Q	6.3	0.3	Voltage Doubler Half-Wave Rectifier
6HG8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9MP	6.3	0.34	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6J5 **6J5GT**	Medium-Mu Triode	2A 13D	6Q 6Q	6.3	0.3	Class A Amplifier
6J6	Medium-Mu Twin Triode	5C	7BF	6.3 6.3	0.45 0.45	Each Unit as Class A Amplifier Push-Pull Class C Amplifier

□ For two tubes.

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power Load Ohms	Power Output Watts	RCA Type
250	—16.5V	250	6.5	34.0	80000	2500	—	7000	3.2	6F6G **6F6GT**
285	—20V	285	7.0	38.0	78000	2550		7000	4.8	
250	—20V	—	—	31.0	2600	2600	6.8	4000	0.85	
315	—24V	285	12.0□	62.0□	—	—	—	10000	11.0†	
100	— 3V	—	—	3.5	16000	500	8	—	—	6F7
250	— 3V	100	1.5	6.5	850000	1100	—	—	—	
For other characteristics, refer to Type 6J5										**6F8G**
145	—16V	145	18	100	8000	9500	—	1000	5.6	**6FE5**
										6FG6
250	— 8V	—	—	9	7700	2600	20	—	—	6FQ7
125	— 1V	—	—	14	5000	8000	40	—	—	6FV8
125	— 1V	125	4	12	200000	6500	—	—	—	
100	1.2V	—	—	15	2500	13000	33	—	—	**6FW8**
180	— 9V	180	2.5	15.0	175000	2300	—	10000	1.1	6G6G
120	— 8V	110	4	49	10000	7500	—	2500	2.3	**6G11**
150	150Ω	150	3.5	15	20000	9500	—	—	—	
Max. DC Plate Volts, 275 Max. DC Cathode mA, 275					Max. Peak Positive-Pulse Plate Volts, 7700 Max. Plate Dissipation, 17 watts					6GB5
Max. DC Plate Volts, 330 Max. DC Cathode mA, 22					Max. Plate Dissipation, 1.5 watts					6GF7
Max. DC Plate Volts, 330 Max. DC Cathode mA, 50					Max. Peak Positive-Pulse, Plate Volts, 1500 (Abs.) Max. Plate Dissipation, 11 watts					
Max. DC Plate Volts, 330						Max. Plate Dissipation, 2.5 watts				6GH8
Max. DC Plate Volts, 350 Max. Peak Neg.-Pulse Grid Volts, 175				Max. Peak Cathode mA, 300 Max. DC Cathode mA, 20			Max. Plate Dissipation, 2.5 watts			
250	—22.5V	150	2.1	70	15000	7100	—	—	—	6GJ5
100	— 3V	—	—	15	—	9000	20	—	—	6GJ7
170	— 1.2V	120	3	10	350000	11000	Ampl. Factor, 55 (Grid No. 2 to Grid No. 1)			
125	— 1V	—	—	13.5	5000	8500	40	—	—	6GJ8
125	— 1V	125	4.5	12	150000	7500	—	—	—	
Max. DC Plate Volts, 770 Max. DC Cathode mA, 175 Max. Plate Dissipation, 17.5' watts					Max. Peak Neg.-Pulse Grid-No. 1 Volts, — 330 Max. Grid-No. 2 Volts, 220 Max. Peak Positive-Pulse Plate Volts, 6500					6GT5
100	— 0.8V	—	—	5	7600	6500	50	—	—	6GV8
170	—15	170	2.7	41	25000	7500	—	—	—	
250	—22.5V	150	2.1	70	15000	7100	—	—	—	6GW6
250	270Ω	250	2.7	16	—	—	—	15000	1.8	6GZ5
250	270Ω (bypassed)	250	2.7	16	150000	8400	—	15000	1.1	
Max. AC Supply Volts per Plate (RMS), 117 Min. Total Effect. Plate-Supply Imped. per Plate: half-wave, 30 ohms; full wave, 15 ohms							Max. DC Output mA, 8. min.			**6H6**
Max. AC Plate Volts (RMS), 150 Max. DC Output mA, 8 per Plate					Min. Total Effective Plate-Supply Impedance: up to 117 volts, 15 ohms; at 150 volts, 40 ohms					6H6GT
100	— 3V	—	—	14	—	5500	17	—	—	6HG8
170	— 1.2V	150	3.3	10	350000	12000	—	—	—	
90	0V	—	—	10	6700	3000	20	—	—	**6J5**
250	— 8V	—	—	9	7700	2600	20	—	—	**6J5GT**
100	50Ω (For both units)			8.5	7100	5300	38	—	—	6J6
150	—10V	—	—	30	Grid Current, 16 mA Driving Power, 0.35 watt			—	3.5	

† For two tubes at stated plate-to-plate load.

□ For two tubes.

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Amperes	Use: Values to right give operating conditions and characteristics for indicated typical use
6J7 6J7G 6J7GT	**Sharp-Cutoff Pentode**	3 23 14A	7R 7R 7R	6.3	0.3	Pentode Class A RF Amplifier
6J8G	**Triode-Heptode Converter**	23	8H	6.3	0.3	Triode Unit as Oscillator Heptode Unit as Mixer
6JB6	**Beam Power Tube**	18A	9QL	6.3	1.2	Horizontal Deflection Amplifier
6JE6	**Beam Power Tube**	18B	9QL	6.3	2.5	Horizontal Deflection Amplifier
6JG6	**Beam Power Tube**	17B	9QU	6.3	1.6	Horizontal Deflection Amplifier
6JM6	**Beam Power Tube**	16A	12FJ	6.3	1.2	Horizontal Deflection Amplifier
6JS6	**Beam Power Tube**	16B	12FY	6.3	2.25	Horizontal Deflection Amplifier
6JT6	**Beam Power Tube**	17C	9QU	6.3	1.2	Horizontal Deflection Amplifier
6JU8	**Quadruple Diode**	6E	9PQ	6.3	0.6	Phase Detector
6K5GT	**High-Mu Triode**	14A	5U	6.3	0.3	Class A Amplifier
6K7 6K7G 6K7GT	**Remote-Cutoff Pentode**	3 23 14A	7R 7R 7R	6.3	0.3	*Class A Amplifier*
6K8 6K8G 6K8GT	**Triode-Hexode Converter**	3 23 —	8K 8K 8K	6.3	0.3	Triode Unit as Oscillator *Hexode Unit as Mixer*
6K11	**Twin High-Mu Triode—Medium-Mu Triode**	8A	12BY	6.3	0.6	Twin Unit as Class A Amplifier Class A Amplifier
6KL8	**Diode—Sharp-Cutoff Pentode**	6E	9LQ	6.3	0.3	Pentode Unit as Class A Amplifier
6KY8	**High-Mu Triode Beam Power Tube**	11C	9QT	6.3	1.1	Triode Unit as Oscillator Beam Power Unit as Amplifier
6L5G	**Medium-Mu Triode**	22	6Q	6.3	0.15	Class A Amplifier
6L6G 6L6GB	**Beam Power Tube**	27B 19D	7AC 7AC	6.3	0.9	Single-Tube Class A Amplifier Push-Pull Class A Amplifier Push-Pull Class AB_1 Amplifier
6L7 6L7G	**Pentagrid Mixer**□	3 23	7T 7T	6.3	0.3	Mixer Service
6N6G	**Direct-Coupled Power Triode**	25	7AU	6.3	0.8	Class A Amplifier
6N7 **6N7GT**	**Medium-Mu Twin Power Triode**	2B 13D	8B 8B	6.3	0.8	Class A Amplier (as Driver) Class B Amplifier
6P5GT	**Medium-Mu Triode**	13D	6Q	6.3	0.3	Amplifier Detector
6P7G	**Low-Mu Triode—Remote-Cutoff Pentode**	23	7U	6.3	0.3	Amplifier and Converter
6Q7 6Q7G 6Q7GT	**Twin Diode High-Mu Triode**	3 23 14A	7V 7V 7V	6.3	0.3	Triode Unit as Class A Amplifier

□ For two tubes.

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
100	— 3V	100	0.5	2.0	1 M	1185	—	—	—	**6J7** 6J7G 6J7GT
250	— 3V	100	0.5	2.0	1 M	1225	—	—	—	
100	Triode-Grid Resistor, 50000 ohms			4	—	—	—	—	—	6J8G
250				5	—	—	—	—	—	
250	— 3V	100	2.8	1.4	1.5 M	Conversion Transcond., 290 micromhos				
Max. DC Plate Volts, 770; Max. Peak Cathode mA, 550; Max. Plate Dissipation, 17.5 watts					Max. Peak Neg.-Pulse Grid-No. 1 Volts, — 330; Max. Grid-No. 2 Volts, 220; Max. Peak Positive-Pulse Plate Volts, 6500					6JB6
For other characteristics, refer to Type 6JE6A										6JE6
For other characteristics, refer to Type 6JG6A										6JG6
For other ratings, refer to Type 6JB6										6JM6
175	—25V	125	4.5	125	5600	11300	Ampl. Factor (Grid No.2 to Plate) 3			6JS6
For other ratings, refer to Type 6JB6										6JT6
Max. Peak Inverse Plate Volts, 300; Max. Peak Plate mA, 54					Max. DC Output mA, 9; Max. Peak Heater-Cathode Volts, ± 300					6JU8
250	— 3V	—	—	1.1	50000	1400	70	—	—	6K5GT
250	— 3V	125	2.6	10.5	600000	1650	—	—	—	**6K7** 6K7G 6K7GT
100	Grid Res., 50000 ohms			3.8	Triode-Grid & Hexode-Grid Current, 0.15 mA					**6K8** 6K8G 6K8GT
100	— 3V	100	6.2	2.3	400000	Conversion Transcond., 325 micromhos				
250	— 3V	100	6.0	2.5	600000	Conversion Transcond., 350 micromhos				
250	— 2V	—	—	1.2	62500	1600	100	—	—	6K11
250	— 8.5V	—	—	10.5	7700	2200	17	—	—	
100	0	100	2.2	5.5	555000	4300	Grid-No. 1 Volts for plate current of 10 μA, 4.2			**6KL8**
Max. DC Plate Volts, 330; Max. DC Cathode mA, 22					Max. Plate Dissipation, 1.5 watts					6KY8
Max. DC Plate Volts, 300; Max. DC Cathode mA, 60					Max. Peak Positive-Pulse Plate Volts, 2200 (Abs.); Max. Plate Dissipation, 12 watts					
250	— 9V	—	—	8.0	9000	1900	17	—	—	6L5G
250	—14V	250	5.0	72.0	—	—	—	2500	6.5	6L6G 6L6GB
250	168Ω	250	5.4	75.0	—	—	—	2500	6.5	
270	—17.5V	270	11.0□	134.0□	—	—	—	5000	17.5†	
270	124Ω□	270	11.0□	134.0□	—	—	—	5000	18.5†	
360	—22.5V	270	5.0□	88.0□	—	—	—	6600	26.5†	
360	248Ω□	270	5.0□	88.0□	—	—	—	9000	24.5†	
250	— 6V	150	9.2	2.3	Oscillator-Grid (No. 3) Bias, —15 volts; Grid-No. 3 Peak Swing, 16 volts minimum; Conversion Transcond., 350 micromhos					**6L7** 6L7G
Output Triode: Plate Volts, 300; Plate mA, 45; Load, 7000 ohms; Triode: Plate Volts, 300; Grid Volts, 0; Input Plate mA, 8									4.0	6N6G
250	— 5V	—	—	6.0	11300	3100	35	20000 or more	exceeds 0.4	**6N7** **6N7GT**
300	— 6V	—	—	7.0	11000	3200	35			
300	0V	Power Output for 1 tube at stated plate-to-plate load						8000	10.0	
250	—13.5	—	—	5.0	9500	—	13.8	—	—	6P5GT
For other characteristics, refer to Type 6F7										6P7G
100	— 1V	—	—	0.8	58000	1200	70	—	—	**6Q7** 6Q7G 6Q7GT
250	— 3V	—	—	1.1	58000	1200	70	—	—	

† For two tubes at stated plate-to-plate load.

□ For two tubes.

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Heater or Filament (F) Amperes	Use Values to right give operating conditions and characteristics for indicated typical use
6Q11	Twin High-Mu Triode—Medium-Mu Triode	8A	12BY	6.3	0.6	Twin Unit as Class A Amplifier Class A Amplifier
6R7 6R7G 6R7GT	Twin Diode—Medium-Mu Triode	3 23 14A	7V 7V 7V	6.3	0.3	Triode Unit as Class A Amplifier
6S4	Medium-Mu Triode	8E	9AC	6.3 6.3	0.6 0.6	Vertical Deflection Amplifier
6S7 6S7G	Remote-Cutoff Pentode	3 23	7R 7R	6.3	0.15	Class A Amplifier
6S8GT	Triple Diode—High-Mu Triode	14C	8CB	6.3	0.3	Triode Unit as Class A Amplifier
6SA7 6SA7GT	Pentagrid Converter	2A 13D	8R 8AD	6.3	0.3	Converter
6SB7Y	Pentagrid Converter	2A	8R	6.3	0.3	Mixer
6SC7	High-Mu Twin Triode	2A	8S	6.3	0.3	Each Unit as Amplifier
6SF5 6SF5GT	High-Mu Triode	2A 13D	6AB 6AB	6.3	0.3	Class A Amplifier
6SF7	Diode—Remote-Cutoff Pentode	2A	7AZ	6.3	0.3	Pentode Unit as Class A Amplifier
6SG7	Semiremote-Cutoff Pentode	2A	8BK	6.3	0.3	Class A Amplifier
6SH7	Sharp-Cutoff Pentode	2A	8BK	6.3	0.3	Class A Amplifier
6SJ7 6SJ7GT	Sharp-Cutoff Pentode	2A 13D	8N 8N	6.3	0.3	Class A Amplifier
6SK7 6SK7GT	Remote-Cutoff Pentode	2A 13D	8N 8N	6.3	0.3	Class A Amplifier
6SN7GT 6SN7 GTA	Medium-Mu Twin Triode	13D 13D	8BD	6.3 6.3 6.3	0.6 0.6 0.6	Each Unit as Class A Amplifier Each Unit as Vertical Amplifier
6SQ7 6SQ7GT	Twin-Diode—High-Mu Triode	2A 13D	8Q 8Q	6.3	0.3	Triode Unit as Class A Amplifier
6SR7	Twin Diode—Medium-Mu Triode	2A	8Q	6.3	0.3	Triode Unit as Class A Amplifier
6SS7	Remote-Cutoff Pentode	2A	8N	6.3	0.15	Class A Amplifier
6ST7	Twin Diode—Medium-Mu Triode	2A	8Q	6.3	0.15	Triode Unit as Amplifier
6SZ7	Twin Diode—High-Mu Triode	2A	8Q	6.3	0.15	Triode Unit as Class A Amplifier
6T4	Medium-Mu Triode	5D	7DK	6.3	0.225	Oscillator in UHF TV Receivers Class A Amplifier
6T7G	Twin Diode—High-Mu Triode	22	7V	6.3	0.15	Triode Unit as Class A Amplifier
6T8	Triple Diode—High-Mu Triode	6B	9E	6.3 6.3	0.45 0.45	Triode Unit as Class A Amplifier
6U5	Electron-Ray Tube	13H	6R	6.3	0.3	Visual Indicator
6U7G	Remote-Cutoff Pentode	28J	7R	6.3	0.3	Class A Amplifier
6U8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9AE	6.3	0.45	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
6V6GT	Beam Power Tube	13D	7AC	6.3	0.45	Single-Tube Class A Amplifier Push-Pull Class AB_1 Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
250	— 2V	—	—	1.2	62500	1600	100	—	—	6Q11
150	0V	—	—	22	7000	2500	18	—	—	
250	— 9V	—	—	9.5	8500	1900	16	—	—	6R7 6R7G 6R7GT
Max. DC Plate Volts, 550 Max. DC Cathode mA, 30					Max. Peak Positive-Pulse Plate Volts, 2200 Max. Plate Dissipation, 8.5 watts					6S4
250	— 3V	100	2.0	8.5	1 M	1750	—	—	—	6S7 6S7G
250	— 2V	—	—	0.9	91000	1100	100	—	—	6S8GT
250	Self-Excited	100	8.5	3.5	1.0	Grid-No. 1 Resistor, 20000 ohms. Conversion Transcond., 450 micromhos				6SA7 6SA7GT
100	— 1V	100	10.2	3.6	500000	Grid-No. 1 Resistor, 20000 ohms. Conversion Transcond., 950 micromhos				6SB7Y
250	— 2V	—	—	2.0	53000	1325	70	—	—	6SC7
250	— 2V	—	—	0.9	66000	1500	100	—	—	6SF5 6SF5GT
100	— 1V	100	3.4	12.0	200000	1975	—	—	—	6SF7
250	— 1V	100	3.3	12.4	700000	2050				
100	— 1V	100	3.2	8.2	250000	4100	—	—	—	6SG7
250	— 2.5V	150	3.4	9.2	1 M	4000	—	—	—	
100	— 1V	100	2.1	5.3	350000	4000	—	—	—	6SH7
250	— 1V	150	4.1	10.8	900000	4900				
100	— 3V	100	0.9	2.9	700000	1575	—	—	—	6SJ7 6SJ7GT
250	— 3V	100	0.8	3.0	1 M	1650				
100	— 1V	100	4.0	13.0	120000	2350	—	—	—	6SK7 6SK7GT
250	— 3V	100	2.6	9.2	800000	2000				
100	0V	—	—	10.0	6700	3000	20	—	—	6SN7GT 6SN7 GTA
250	— 8V	—	—	9.0	7700	2600	20	—	—	
Max. DC Plate Volts, 450 Max. Peak Cathode mA, 70				Max. Plate Dissipation: 5 watts either plate; 7.5 watts both plates Max. Peak Positive Pulse Plate Volts, 1500						
100	— 1V	—	—	0.5	110000	925	100	—	—	6SQ7 6SQ7GT
250	— 2V			1.1	85000	1175	100	—	—	
250	— 9V	—	—	9.5	8500	1900	16	—	—	6SR7
250	— 3V	100	2.0	9.0	1 M	1850	—	—	—	6SS7
For other characteristics, refer to Type 6SR7										6ST7
100	— 1V	—	—	0.8	54000	1300	70	—	—	6SZ7
250	— 3V	—	—	1.0	53000	1200	70	—	—	
Max. DC Plate Volts, 200 Max. DC Cathode mA, 30						Max. Grid mA, 8 Max. Plate Dissipation, 3.5 watts				6T4
80	150Ω	—	—	18	—	7000	13	—	—	
250	— 3V	—	—	1.2	62000	1050	65	—	—	6T7G
300	4580Ω	—	—	Grid Resistor, 0.5 MΩ			Gain per stage, 40			
100	— 1V	—	—	0.8	54000	1300	70	—	—	6T8
250	— 3V	—	—	1.0	58000	1200	70	—	—	
Plate & Target Supply, 250 volts. Triode Plate Resistor, 1.0 MΩ Target Current, 4.0 mA Grid Bias, —22 volts; Shadow Angle, 0°. Bias, 0 volts; Angle, 90°; Plate Current, 0.24 mA										6U5
250	— 3V	100	2.0	8.2	800000	1600	—	—	—	6U7G
125	— 1V	—	—	13.5	—	7500	40	—	—	6U8
125	— 1V	110	3.5	9.5	200000	5000	—	—	—	
250	—12.5V	250	4.5	45.0	50000	4100	—	5000	4.5	6V6GT
315	—13V	225	2.2	34.0	80000	3750	—	8500	5.5	
250	—15V	250	5.0□	70.0□	—	—	—	10000	10.0†	
285	—19V	285	4.0□	70.0□	—	—	—	8000	14.0†	

† For two tubes at stated plate-to-plate load. □ For two tubes.

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Amperes	Use Values to right give operating conditions and characteristics for indicated typical use
6V7G	Twin Diode—Low-Mu Triode	23	7V	6.3	0.3	Triode Unit as Amplifier
6W7G	Sharp-Cutoff Pentode	23	7R	6.3	0.15	Class A Amplifier
6X5	Full-Wave Rectifier	2B	6S	6.3	0.6	With Capacitive-Input Filter
						With Inductive-Input Filter
6X8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9AK	6.3	0.45	Triode Unit as Class A Amplifier
						Pentode Unit as Class A Amplifier
6Y5	Full-Wave Rectifier	22 or 13H	6J	6.3	0.8	With Capacitive-Input Filter
6Y7G	High-Mu Twin Power Triode	22	8B	6.3	0.6	Class B Amplifier
6Z4	Refer to type 84/6Z4					
6Z5	Full-Wave Rectifier	22	6K	12.6 6.3	0.8 0.4	With Capacitive-Input Filter
6Z7G	High-Mu Twin Power Triode	22	8B	6.3	0.3	Class B Amplifier
6ZY5G	Full-Wave Rectifier	22	6S	6.3	0.3	With Capacitive-Input Filter
7A4	Medium-Mu Triode	12B	5AC	6.3	0.3	Amplifier
7A5	Beam Power Tube	12C	6AA	6.3	0.75	Class A Amplifier
7A6	Twin Diode	12B	7AJ	6.3	0.15	Detector Rectifier
7A7	Remote-Cutoff Pentode	12B	8V	6.3	0.3	Class A Amplifier
7A8	Octode Converter	12B	8U	6.3	0.15	Converter
7AD7	Power Pentode	12C	8V	6.3	0.6	Class A Amplifier
7AF7	Medium-Mu Twin Triode	12B	8AC	6.3	0.3	Each Unit as Class A Amplifier
7AG7	Sharp-Cutoff Pentode	12B	8V	6.3	0.15	Class A Amplifier
7AH7	Sharp-Cutoff Pentode	12B	8V	6.3	0.15	Class A Amplifier
7B4	High-Mu Triode	12B	5AC	6.3	0.3	Amplifier
7B5	Power Pentode	12C	6AE	6.3	0.4	Class A Amplifier
7B6	Twin Diode—High-Mu Triode	12B	8W	6.3	0.3	Triode Unit as Amplifier
7B7	Remote-Cutoff Pentode	12B	8V	6.3	0.15	Class A Amplifier
7B8	Pentagrid Converter	12B	8X	6.3	0.3	Converter
7C5	Beam Power Tube	12C	6AA	6.3	0.45	Class A Amplifier
7C6	Twin Diode—High-Mu Triode	12B	8W	6.3	0.15	Triode Unit as Class A Amplifier
7C7	Sharp-Cutoff Pentode	12B	8V	6.3	0.15	Class A Amplifier
7E6	Twin Diode—Medium-Mu Triode	12B	8W	6.3	0.3	Triode Unit as Amplifier
7E7	Twin Diode—Remote-Cutoff Pentode	12B	8AE	6.3	0.3	Pentode Unit as Class A Amplifier
7EY6	Beam Power Tube	13F	7AC	7.2	0.6	Vertical Deflection Amplifier
7F7	High-Mu Twin Triode	12B	8AC	6.3	0.3	Each Unit as Amplifier
7F8	Medium-Mu Twin Triode	12A	8BW	6.3	0.3	Each Unit as Class A Amplifier
7G7	Sharp-Cutoff Pentode	12B	8V	6.3	0.45	Class A Amplifier
7H7	Semiremote-Cutoff Pentode	12B	8V	6.3	0.3	Class A Amplifier
7HG8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9MP	7.2	0.3	Triode Unit as Class A Amplifier
						Pentode Unit as Class A Amplifier
7J7	Triode-Heptode Converter	12B	8BL	6.3	0.3	Triode Unit as Oscillator
						Heptode Unit as Mixer

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power Load Ohms	Power Output Watts	RCA Type
For other characteristics, refer to Type 85										6V7G
250	— 3V	100	0.5	2.0	1.5 M	1225	—	—	—	6W7G
Max. AC Volts per Plate (RMS), 325 Max. Peak Inverse Volts, 1250				Max. DC Output mA, 70 Max. Peak Plate mA, 245			Min. Total Effect. Supply Imped. per Plate, 525 ohms			6X5
Max. AC Volts per Plate (RMS), 400 Max. Peak Inverse Volts, 1250				Max. DC Output mA, 70 Max. Peak Plate mA, 245			Min. Value of Input Choke, 10 henries			6X5
125	— 1V	—	—	12	6000	6500	40	—	—	**6X8**
125	— 1V	125	2.2	9	300000	5500	—	—	—	**6X8**
Max. AC Volts per Plate (RMS), 350 Max. DC Output mA, 50										**6Y5**
For other characteristics, refer to Type 79										6Y7G
										6Z4
Max. AC Volts per Plate (RMS), 230 Max. DC Output mA, 60										6Z5
180	0V	Power Output is for one tube at stated plate-to-plate load						12000	4.2	6Z7G
Max. Peak Inverse Volts, 1250				Max. DC Output mA, 40 Max. Peak Plate mA, 120			Min. Total Effect. Supply Imped. per Plate, 225 ohms			6ZY5G
For other characteristics, refer to Type 6J5										**7A4**
110 125	— 7.5V — 9V	110 125	3.0 3.3	40.0 44.0	16000 17000	5800 6000	—	2500 2700	1.5 2.2	7A5
Max. AC Voltage per Plate, 150 Volts, RMS						Max. DC Output Current per plate, 8 mA				7A6
For other characteristics, refer to Type 6SK7										**7A7**
250	— 3V	100	3.2	3.0	700000	Anode-Grid (2): 250 max. volts, 4.2 mA Oscillator-Grid No. 1 Resistor. Conversion Transcond., 550 micromhos				7A8
300	68Ω	150	7.0	28.0	300000	9500	—	—	—	7AD7
250	—10V	—	—	9.0	7600	2100	16	—	—	**7AF7**
250	250Ω	250	2.0	6.0	1 M	4200	—	—	—	7AG7
250	250Ω	250	1.9	6.8	1 M	3300	—	—	—	7AH7
For other characteristics, refer to Type 6SF5										7B4
For other characteristics, refer to Type 6K6GT										7B5
For other characteristics, refer to Type 6SQ7										7B6
250	— 3V	100	1.7	8.5	—	750000	1750	—	—	7B7
For other characteristics, refer to Type 6A8										**7B8**
For other characteristics, refer to Type 6V6										**7C5**
250	— 1V	—	—	1.3	100000	1000	100	—	—	7C6
250	— 3V	100	0.5	2.0	2 M	1300	—	—	—	**7C7**
For other characteristics, refer to Type 6BF6										7E6
250	330Ω	100	1.6	7.5	700000	1300	—	—	—	7E7
For other characteristics, refer to Type 6EY6										**7EY6**
For other characteristics, refer to Type 6SL7GT										**7F7**
250	500Ω	—	—	6.0	—	3300	48	—	—	**7F8**
250	— 2V	100	2.0	6.0	800000	4500	—	—	—	7G7
100 250	— 1.5V 180Ω	100 150	2.6 3.2	7.5 10.0	350000 800000	4000 4000	— —	— —	— —	**7H7**
For other characteristics, refer to Type 6HG8										7HG8
250	Triode-Grid Resistor, 50000 ohms			5.0	Triode-Grid & Heptode-Grid Current, 0.4 mA					**7J7**
250	— 3V	100	2.8	1.4	1.5 M	Conversion Transcond., 290 μmhos				**7J7**

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Heater or Filament (F) Amperes	Use — Values to right give operating conditions and characteristics for indicated typical use
7K7	Twin Diode—High-Mu Triode	12B	8BF	6.3	0.3	Triode Unit as Class A Amplifier
7L7	Sharp-Cutoff Pentode	12B	8V	6.3	0.3	Class A Amplifier
7N7	Medium-Mu Twin-Triode	12C	8AC	6.3	0.6	Each Unit as Class A Amplifier
7Q7	Pentagrid Converter	12B	8AL	6.3	0.3	Converter
7R7	Twin Diode—Remote-Cutoff Pentode	12B	8AE	6.3	0.3	*Pentode Unit as Class A Amplifier*
7S7	Triode-Heptode Converter	12B	8BL	6.3	0.3	Triode Unit as Oscillator
						Heptode Unit as Mixer
7V7	Sharp-Cutoff Pentode	12B	8V	6.3	0.45	Class A Amplifier
7W7	Sharp-Cutoff Pentode	12B	8BJ	6.3	0.45	Class A Amplifier
7X7	Twin Diode—High-Mu Triode	12C	8BZ	6.3	0.3	Triode Unit as Class A Amplifier
7Y4	Full-Wave Rectifier	12B	5AB	6.3	0.5	With Capacitive-Input Filter
7Z4	Full-Wave Rectifier	12C	5AB	6.3	0.9	With Capacitive-Input Filter
8FQ7	Medium-Mu Twin Triode	6E	9LP	8.4	0.45	*Vertical and Horizontal Deflection Oscillators*
8GJ7	Medium-Mu Triode—Sharp-Cutoff Pentode	8J	9QA	8	0.3	Triode Unit as Class A Amplifier
						Pentode Unit as Class A Amplifier
9A8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9DC	9	0.3	Triode Unit as Class A Amplifier
						Pentode Unit as Class A Amplifier
9BR7	*Twin Diode—High-Mu Triode*	6B	9CF	4.7 9.4	0.6 0.3	*Triode Unit as Class A Amplifier*
9CL8	Medium-Mu Triode—Sharp-Cutoff Tetrode	6B	9FX	9.5	0.3	Triode Unit as Class A Amplifier
						Tetrode Unit as Class A Amplifier
9GV8	High-Mu Triode—Power Pentode	6G	9LY	9.5	0.6	Triode Unit as Class A Amplifier
						Pentode Unit as Class A Amplifier
10	Power Triode	27B	4D	7.5F	1.25	Class A Amplifier
10C8	High-Mu Triode—Sharp-Cutoff Pentode	6B	9DA	10.5	0.3	Triode Unit as Class A Amplifier
						Pentode Unit as Class A Amplifier
10CW5	Power Pentode	6G	9CV	10.6	0.45	Vertical Deflection Amplifier
10DX8	High-Mu Triode—Sharp-Cutoff Pentode	6E	9HX	10.2	0.45	Triode Unit as Class A Amplifier
						Pentode Unit as Class A Amplifier
10GF7	Dual Triode	11A	9QD	9.7	0.6	Vertical Deflection Amplifier
						Vertical Deflection Oscillator
11	Detector Amplifier	4F	4F	1.1F	0.25	Class A Amplifier
11Y9	Dual Pentode	6L	10L	11	0.45	Unit No. 1 as Class A Amplifier
						Unit No. 2 as Class A Amplifier
12A5	Power Pentode	22 or 13H	7F	6.3 12.6	0.6 0.3	Class A Amplifier
12A7	Rectifier—Power Pentode	24B	7K	12.6	0.3	Pentode Unit as Class A Amplifier
						Half-Wave Rectifier
12A8GT	Pentagrid Converter	14A	8A	12.6	0.15	Converter
12AC6	Remote-Cutoff Pentode	5C	7BK	10.0 to 15.9	0.15 approx. at 12.6 V	Class A Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
250	— 2V	—	—	2.3	44000	1600	70	—	—	7K7
100	— 1V	100	2.4	5.5	100000	3000	—	—	—	7L7
250	— 1.5V	*100*	*1.5*	4.5	*1 M*	*3100*	—	—	—	7L7
For other characteristics, refer to Type 6SN7GT										**7N7**
250	— 2V	100	8.5	3.5	1 M	Grid No. 1 Resistor, 20000 ohms Conversion Transcond., 450 μmhos				7Q7
250	— 1V	100	2.1	5.7	1 M	3200	—	—	—	7R7
100	Triode-Grid Resistor, 50000 ohms			3.0	—	—	—	—	—	7S7
250				5.0	—	—	—	—	—	7S7
250	— 2V	*100*	*3.0*	*1.8*	*1.25 M*	*Conversion Transcond., 525 μmhos*				7S7
300	160Ω	150	3.9	10.0	300000	5800	—	—	—	**7V7**
For other characteristics, refer to Type 7V7										**7W7**
250	— 1V	—	—	*1.9*	67000	1500	100	—	—	**7X7**
Max. Peak Inverse Volts, 1250				Max. DC Output mA,70			Max. Peak Plate mA, 180			7Y4
Max. Peak Inverse Volts, 1250				Max. DC Output mA, 100 Max. Peak Plate mA, 300			Min. Total Effec. Supply Imped. per Plate, 75 ohms			7Z4
For other characteristics, refer to Type 6FQ7										8FQ7
For other characteristics, refer to Type 6GJ7										8GJ7
100	— 2V	—	—	14	—	5000	20	—	—	9A8
170	— 2V	170	2.8	10	400000	6200	Ampl. Factor. (Grid No. 2 to Grid No. 1), 47			9A8
250	200Ω	—	—	10	*10900*	*4000*	60	—	—	**9BR7**
125	56Ω	—	—	15	5000	8000	40	—	—	9CL8
125	— 1V	125	4	12	100000	5800	—	—	—	9CL8
For other characteristics, refer to Type 6GV8										9GV8
425	—40V	—	—	18.0	5000	1600	8.0	10200	1.6	10
250	390Ω	—	—	7.3	12000	4400	53	—	—	**10C8**
135	100Ω	135	3.2	11.5	190000	8000	—	—	—	**10C8**
For other ratings, refer to Type 6CW5										10CW5
For other characteristics, refer to Type 6DX8										10DX8
For other ratings, refer to Type 6GF7										10GF7
135	—10.5V	—	—	3	15500	440	—	—	—	11
170	— 2.6	170	6.5	30	—	21000	Ampl. Factor (Grid-No. 1 to Grid-No. 2), 38			11Y9
150	— 2.3	150	3	10	—	8500	Ampl. Factor (Grid-No. 1 to Grid No. 2), 35			11Y9
180	—25V	180	8.0	45.0	35000	2400	—	3300	3.4	12A5
135	—13.5V	135	2.5	9.0	100000	975	—	13500	0.55	12A7
	Maximum AC Plate Voltage........125 Volts, RMS Maximum DC Output Current........30 Milliamperes									12A7
For other characteristics, refer to Type 6A8GT										12A8GT
12.6	—	12.6	.2	.55	500000	730	{Grid-No. 1 Supply Volts, 0 {Grid-No. 1 Res., 2.2 megohms}			**12AC6**

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Amperes	Use Values to right give operating conditions and characteristics for indicated typical use
12AD6	Pentagrid Converter	5C	7CH	10.0 to 15.9	0.15 approx. at 12.6 V	Converter
12AE6	Twin Diode—Medium-Mu Triode	5C	7BT	10.0 to 15.9	0.15 approx. at 12.6 V	Triode Unit as Class A Amplifier
12AE6A	Twin Diode—Medium-Mu Triode	5C	7BT	10.0 to 15.9	0.15 approx. at 12.6 V	Triode Unit as Class A Amplifier
12AE7	Dual Triode	6B	9A	10.0 to 15.9	0.45 approx. at 12.6 V	Unit No. 1 as Class A Amplifier Unit No. 2 as Class A Amplifier
12AF6	Remote-Cutoff Pentode	5C	7BK	10.0 to 15.9	0.15 approx. at 12.6 V	Class A Amplifier
12AH7GT	Medium-Mu Twin Triode	13C	8BE	12.6	0.15	Each Unit as Class A Amplifier
12AJ6	Twin Diode—Medium-Mu Triode	5C	7BT	10.0 to 15.9	0.15 approx. at 12.6 V	Triode Unit as Class A Amplifier
12AL8	Medium-Mu Triode—Power Tetrode	6E	9GS	10.0 to 15.9	0.55 approx. at 12.6 V	Triode Unit as Class A Amplifier Tetrode Unit as Class A Amplifier
12AU7	Medium-Mu Twin Triode	6B	9A	6.3 12.6	0.3 0.15	Each Unit as Class A Amplifier
12AV7	Medium-Mu Twin Triode	6B	9A	6.3 12.6	0.45 0.225	Each Unit as Class A Amplifier
12AX4-GT 12AX4-GTA	Half-Wave Rectifier	13D 13D	4CG	12.6 12.6	0.6 0.6	Television Damper Service
12AX7	High-Mu Twin-Triode	6B	9A	6.3 12.6	0.3 0.15	Each Unit as Class A Amplifier
12AY3	Half-Wave Rectifier	11D	9HP	12.6	0.6	Television Damper Service
12AZ7	High-Mu Twin-Triode	6B	9A	6.3 12.6	0.45 0.225	Each Unit as Class A Amplifier
12B8GT	High-Mu Triode—Remote-Cutoff Pentode	—	8T	12.6	0.3	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
12BA7	Pentagrid Converter	6E	8CT	12.6	0.15	Converter
12BD6	Remote-Cutoff Pentode	5C	7BK	12.6	0.15	Class A Amplifier
12BF6	Twin Diode—Medium-Mu Triode	5C	7BT	12.6	0.15	Triode Unit as Class A Amplifier
12BH7	Medium-Mu Twin Triode	6E	9A	6.3 12.6	0.6 0.3	Vertical Deflection Amplifier
12BK5	Beam Power Tube	6E	9BQ	12.6	0.6	Class A Amplifier
12BL6	Remote-Cutoff Pentode	5C	7BK	10.0 to 15.9	0.15 approx. at 12.6V	Class A Amplifier
12BR7	Twin Diode—High-Mu Triode	6B	9CF	6.3 12.6	0.45 0.225	Triode Unit as Class A Amplifier
12BS3	Half-Wave Rectifier	11D	9HP	12.6	0.6	Television Damper Service
12BV7	Sharp-Cutoff Pentode	6E	9BF	6.3 12.6	0.6 0.3	Class A Amplifier
12BW4	Full-Wave Rectifier	6E	9DJ	6.3	0.9	With Capactive Input Filter With Inductive Input Filter
12BY7	Sharp-Cutoff Pentode	6E	9BF	6.3 12.6	0.6 0.3	Class A Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
12.6	Self-excited	12.6	1.5	0.45	1 M	Grid-No. 1 Resistor, 33000 ohms; Conversion Transcond., 260 micromhos				**12AD6**
12.6	0V	—	—	0.75	15000	1000	15	—	—	12AE6
12.6	*0V*	—	—	*1*	*13000*	*1300*	*16.7*	—	—	**12AE6A**
12.6	Grid Res. 1.5 megohms			1.9	3150	4000	13.0	—	—	12AE7
12.6	Grid Res. 1 megohm			7.5	985	6500	6.4	—	—	12AE7
12.6	—	12.6	0.45	1.1	350000	1500	{Grid-No. 1 Supply Volts, 0; Grid-No. 1 Res., 2.2 megohms}			**12AF6**
180	— 6.5V	—	—	7.6	8400	1900	16	—	—	12AH7 GT
12.6	{Grid-No. 1 Supply Volts, 0; Grid-No. 1 Res., 2.2 megohms}			0.75	45000	1200	55	—	—	12AJ6
12.6	— 0.9V (across 2.2 megohm res.)			.5	13000	1000	13	—	—	**12AL8**
Grid-No. 2 (Control Grid) Volts, —.5 (across 2.2 megohm res.); Grid-No. 1 (Space-Charge Grid) Volts, 12.6; Transcond. (Grid-No. 2 to Plate), 15000 μmhos					Ampl. Factor (Grid-No. 2 to Plate) 7.2; Grid-No. 1 mA, 75 Plate mA, 40; Plate Resistance, 480 ohms					**12AL8**
100	0V	—	—	11.8	6250	3100	19.5	—	—	12AU7
250	— 8.5V	—	—	10.5	7700	2200	17	—	—	12AU7
150	56Ω	—	—	18	48000	8500	41	Cutoff Volts, —12		**12AV7**
Max. Peak Inverse Plate Volts, 4400; Max. Peak Plate mA, 750; Max. DC Plate mA, 125					Max. Peak Heater-Cathode Volts: {—4400, +300}; DC component must not exceed 900 volts					12AX4-GT 12AX4-GTA
100	— 1V	—	—	0.5	80000	1250	100	—	—	12AX7
250	— 2V	—	—	1.2	62500	1600	100	—	—	12AX7
For other ratings, refer to Type 6AY3										12AY3
100	270Ω	—	—	3.7	15000	4000	60	—	—	12AZ7
250	200Ω	—	—	10.0	10900	5500	60	—	—	12AZ7
90	0V	—	—	2.8	37000	2400	90	—	—	12B8GT
90	— 3V	90	2	7	200000	1800	—	—	—	12B8GT
For other characteristics, refer to Type 6BA7										**12BA7**
For other characteristics, refer to Type 6BD6										**12BD6**
250	— 9V	—	16	*1900*	*9.5*	*8500*	—	Power Output, 300 milliwatts		**12BF6**
Max. DC Plate Volts, 450; Max. DC Plate mA, 20					Absolute Max. Peak Positive-Pulse Plate Volts, 1500; Max. Plate Dissipation (Each Unit), 3.5 watts					12BH7
250	— 5V	250	3.5	35	100000	8500	—	6500	3.5	**12BK5**
12.6	Grid-No. 1 Supply Volts, 0	12.6	0.5	1.35	500000	1350	Grid-No. 1 and Grid-No. 3 Volts for transcond. of 10 micromhos, —5			**12BL6**
100	270Ω	—	—	3.7	15000	4000	60	—	—	**12BR7**
250	*200Ω*	—	—	*10*	*10900*	5500	*60*	—	—	**12BR7**
For other ratings, refer to Type 6BS3										12BS3
250	68Ω	150	6	27	85000	13000	—	—	—	**12BV7**
250	— 8V	180	—	0.5	—	—	—	—	—	**12BV7**
For other characteristics, refer to 6BW4										**12BW4**
250	100Ω	180	5.75	26	93000	11000	—	—	—	12BY7

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Heater or Filament (F) Amperes	Use — Values to right give operating conditions and characteristics for indicated typical use
12C8	Twin Diode—Semiremote-Cutoff Pentode	3	8E	12.6	0.15	Pentode Unit as RF Amplifier
12CN5	Remote-Cutoff Pentode	5D	7CV	10.0 to 15.9	0.45 approx. at 12.6V	Class A Amplifier
12CT8	Medium-Mu Triode—Sharp-Cutoff Pentode	6E	9DA	12.6	0.3	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
12CX6	Remote-Cutoff Pentode	5C	7BK	10.0 to 15.9	0.15 approx. at 12.6V	Class A Amplifier
12DE8	Diode—Remote-Cutoff Pentode	6B	9HG	10.0 to 15.9	0.2 approx. at 12.6V	Pentode Unit as Class A Amplifier
12DK7	Twin Diode—Power Tetrode	6E	9HZ	10.0 to 15.9	0.5 approx. at 12.6V	Tetrode Unit as Class A Amplifier
12DL8	Twin Diode—Power Tetrode	6E	9HR	10.0 to 15.9	0.55 approx. at 12.6V	Tetrode Unit as Class A Amplifier
12DM4 12DM4A	Half-Wave Rectifier	13F 13G	4CG	12.6	0.6	Television Damper Service
12DQ6A	Beam Power Tube	20	6AM	12.6	0.6	Horizontal Deflection Amplifier
12DQ6B	Beam Power Tube	20	6AM	12.6	0.6	Horizontal Deflection Amplifier
12DQ7	Power Pentode	6E	9BF	6.3 12.5	0.6 0.3	Class A Amplifier
12DS7 12DS7A	Twin Diode—Power Tetrode	6E 6E	9JU	10.0 to 15.9	0.4 approx. at 12.6V	Tetrode Unit as Class A Amplifier Didoe Units
12DU7	Twin Diode—Power Tetrode	6B	9JX	10.0 to 15.9	0.25 approx. at 12.6V	Tetrode Unit as Class A Amplifier
12DV8	Twin Diode—Power Tetrode	6E	9HR	10.0 to 15.9	0.375 approx. at 12.6V	Class A Amplifier
12DW7	Dual Triode	6B	9A	12.6 6.3	0.15 0.3	Unit No. 1 as Class A Amplifier Unit No. 2 as Class A Amplifier
12DY8	Medium-Mu Triode—Remote-Cutoff Tetrode	6B	9JD	10.0 to 15.9	0.35 approx. at 12.6V	Triode Unit as Class A Amplifier Tetrode Unit as Signal Seeker Relay
12DZ6	Remote-Cutoff Pentode	5C	7BK	10.0 to 15.9	0.19 approx. at 12.6V	Class A Amplifier
12EA6	Remote-Cutoff Pentode	5C	7BK	10.0 to 15.9	0.19 approx. at 12.6V	Class A Amplifier
12EC8	Medium-Mu Triode—Semiremote-Cutoff Pentode	6B	9FA	10.0 to 15.9	0.225 approx. at 12.6V	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
12ED5	Beam Power Tube	5D	7CV	12.6	0.45	Class A Amplifier
12EG6	Pentagrid Amplifier	5C	7CH	10.0 to 15.9	0.15 approx. at 12.6V	Class A Amplifier
12EH5	Power Pentode □	5D	7CV	12.6	0.6	Push-Pull Class AB, Amplifier

□ For two tubes.

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
250	— 3V	125	2.3	10	600000	1325	—	—	—	12C8
12.6	—	12.6	3.5	4.5	40000	3800	Grid-No. 1 Supply Volts, 0; Grid-No. 1 Res., 2.2 megohms			**12CN5**
150	150Ω	—	—	9	8200	4900	40	—	—	12CT8
200	82Ω	125	3.4	15	150000	7000	—	—	—	12CT8
12.6	Grid-No. 1 Supply Volts, 0	12.6	1.4	3	40000	3100	Grid-No. 1 Volts for Plate Current of 10 μA, —4.5			**12CX6**
12.6	—	12.6	0.5	1.3	300000	1500	Grid No. 1 Supply Volts, 0; Grid-No. 1 Res., 2.2 megohms			12DE8
12.6	—	12.6	1	6	4000	5000	—	3500	0.010	12DK7
12.6	Grid-No. 2 (Control Grid) Volts, —0.5 (across 2.2 megohm resistor); *Grid-No. 1* (Space-Charge Grid) Volts, 12.6; Transcond. (Grid-No. 2 to Plate), 15000 μmhos						Ampl. Factor (Grid-No. 2 to Plate) 7.2; *Grid-No. 1 mA*, 75; Plate mA, 40; Plate Resistance, 480 ohms			12DL8
For other characteristics, refer to Type 6DM4										12DM4 12DM4A
Max. DC Plate Volts, 700; Max. DC Cathode mA, 140							*Max. Peak Positive-Pulse Plate Volts, 6000 (Abs.)*; Max. Plate Dissipation, 15 watts			12DQ6A
For other ratings, refer to Type 6DQ6B										12DQ6B
200	68Ω	125	5.6	26	53000	10500	—	—	—	**12DQ7**
12.6	12.6V	—0.5 (across 2.2 megohm resistor)	75 (Grid-No. 1)	35	500	19000 (Grid-No. 2 to Plate)	9.1 (Grid-No. 2 to Plate)	—	—	**12DS7** 12DS7A
Diode Plate mA, with 10 Volts Applied, 3 mA										**12DS7** 12DS7A
12.6	—	12.6	1.5	12	6000	6200	—	2700	0.025	**12DU7**
Grid-No. 2 (Control Grid) Resistor, 4.7 megohms; Grid-No. 1 (Space-Charge Grid) Volts, 12.6; Transcond. (Grid-No. 2 to Plate), 8500 μmhos							Ampl. Factor (Grid-No. 2 to Plate) 7.6; Grid-No. 1 mA, 53; Plate mA, 9; Plate Resistance, 900 ohms			12DV8
250	— 2V	—	—	1.2	62500	—	100	—	—	**12DW7**
250	— 8.5V	—	—	10.5	7700	2200	17	—	—	**12DW7**
12.6	—	—	—	1.2	10000	2000	20	—	—	**12DW7**
10	—	10	—	5 min.	Grid No. 1 resistor 10 megohms. Plate Load 700 ohms					**12DY8**
15	— 6V	15	—	3 max.	—	—	—	Plate Load 700 ohms		**12DY8**
12.6	Grid-No.1 Supply Volts, 0	12.6	2.2	4.5	25000	3800	—	—	—	**12DZ6**
12.6	—	12.6	1.4	3.2	32000	3800	Grid-No. 1 Supply Volts, 0; Grid-No. 1 Res., 10 megohms			12EA6
12.6	4700Ω (Grid Res.)	—	—	2.4	6000	4700	25	—	—	12EC8
12.6	—	12.6	0.28	0.66	750000	2000	Grid No. 1 Res., 33000 ohms			12EC8
1.25	— 4.5V	125	7	37	14000	8500	—	4500	1.5	**12ED5**
12.6	— 0.6V†	12.6	2.8	.55	150000	800‡	‡Between Grid No. 3 & Plate; †Bias voltage across res. 2.2 megohms			12EG6
140	68Ω	120	11	47	—	—	—	6000	3.8	12EH5

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Amperes	Use — Values to right give operating conditions and characteristics for indicated typical use
12EK6	Remote-Cutoff Pentode	5C	7BK	10.0 to 15.9	0.19 approx. at 12.6V	Class A Amplifier
12EL6	Twin Diode—High-Mu Triode	5C	7FB	10.0 to 15.9	0.15 approx. at 12.6V	Class A Amplifier
12EM6	Diode—Power Tetrode	6E	9HV	10.0 to 15.9	0.5 approx. at 12.6V	Class A Amplifier
12EN6	Beam Power Tube	13D	7AC	12.6	0.6	Vertical Deflection Amplifier
12F5GT	High-Mu Triode	14A	5M	12.6	0.15	Amplifier
12F8	Twin Diode—Remote-Cutoff Pentode	6B	9FH	10.0 to 15.9	0.15 approx. at 12.6V	Pentode Unit as Class A Amplifier
12FK6	Twin Diode—Low-Mu Triode	5C	7BT	10.0 to 15.9	0.15 approx. at 12.6 V	Triode Unit as Class A Amplifier
12FM6	Twin Diode—Medium-Mu Triode	5C	7BT	10.0 to 15.9	0.15 approx. at 12.6V	Triode Unit as Class A Amplifier
12FR8	Diode—Medium-Mu Triode Remote-Cutoff Pentode	6K	9KU	12.6	0.32	Triode Unit as Class A Amplifier
12FV7	Medium-Mu Twin Triode	6E	9A	6.3 12.6	0.9 0.45	Each Unit as Class A Amplifier
12FX8	Medium-Mu Triode—Pentagrid Converter	6D	9KV	10.0 to 15.9	0.3 approx. at 12.6V	Triode Unit as Class A Amplifier Pentagrid Unit as Converter
12FX8A	Medium-Mu Triode-Pentagrid Converter	6D	9KV	10.0 to 15.9	0.27 approx. at 12.6V	Triode Unit as Class A Amplifier Pentagrid Unit as Converter
12GA6	Pentagrid Converter	5C	7CH	10.0 to 15.9	0.15 approx. at 12.6V	Converter
12GC6	Beam Power Tube	20	8JX	12.6	0.6	Horizontal Deflection Amplifier
12GJ5	Beam Power Tube	18A	9QK	12.6	0.6	Horizontal Deflection Amplifier
12GN7	Sharp-Cutoff Pentode	6E	9BF	6.3 12.6	0.6 0.3	Class A Amplifier
12GT5 12GT5A	Beam Power Tube	17B	9NZ	12.6	0.6	Horizontal Deflection Amplifier
12H6	Twin Diode	29B	7Q	12.6	0.15	Voltage Doubler Half-Wave Rectifier
12J5GT	Medium-Mu Triode	13D	6Q	12.6	0.15	Amplifier
12J7GT	Sharp-Cutoff Pentode	14A	7R	12.6	0.15	Amplifier
12J8	Twin Diode—Power Tetrode	6B	9GC	10.0 to 15.9	0.325 approx. at 12.6V	Tetrode Unit as Class A Amplifier
12JB6	Beam Power Tube	18A	9QL	12.6	0.6	Horizontal-Deflection Amplifier
12JN8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9FA	12.6	0.225	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
12JT6	Beam Power Tube	17C	9QU	12.6	0.6	Horizontal Deflection Amplifier
12K5	Power Tetrode	5D	7EK	10.0 to 15.9	0.4 approx. at 12.6V	Class A Amplifier
12K7GT	Remote-Cutoff Pentode	14A	7R	12.6	0.15	Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
12.6	—	12.6	1.7	4	50000	4200	Grid-No. 1 Supply Volts, 0 Grid-No. 1 Res. (Bypassed), 2.2 megohms			**12EK6**
12.6	0V	—	—	0.75	45000	1200	55	—	—	12EL6
12.6	—	12.6	1	6	4000	5000	Grid-No. 1 Res., 2.2 megohms			12EM6
Max. Peak Pos.-Pulse Volts, 1200 Max. Peak Neg.-Pulse Grid Volts, 250 Max. Peak Cathode mA, 175							Max. Plate Dissipation, 7 watts Max. DC Plate Volts, 300			12EN6
For other characteristics, refer to Type 6F5GT										12F5GT
12.6	0V	12.6	0.38	1	330000	1000	Grid-No. 1 Volts for transcond. of 10 micromhos, —5			**12F8**
12.6	Grid Supply Volts, 0 Grid Res. (Bypassed), 2.2 megohms			1.3	6200	1200	7.4	—	—	**12FK6**
12.6	0V	—	—	1	7700	1300	10	—	—	**12FM6**
12.6	—0.8V	12.6	0.7	1.9	400000	2700	—	—	—	12FR8
100	— 2V	—	—	16	2250	9600	21.5	—	—	**12FV7**
12.6	—	—	—	1.3	7150	1400	10	Grid Res., 2.2 megohms		12FX8
12.6	—	12.6	1.25	0.29	500000	Grid No. 3 Res., 2.2 megohms Conversion Transcond., 300 μmhos				12FX8
12.6	— 0.8	—	—	1.3	7150	1400	10	—	—	**12FX8A** •
12.6	— 0.5	12.6	1.25	0.29	500000	Grid No. 3 Res., 2.2 megohms Conversion Transcond., 300 μmhos				**12FX8A** •
12.6	1.6V	12.6	0.8	0.3	1 M	Grid No. 1 Res., 33000 ohms Conversion. Transcond., 140 μmhos				12GA6
Max. DC Plate Volts, 770 Max. DC Cathode mA, 175						Max. Peak Positive-Pulse Plate Volts, 6500 Max. Plate Dissipation 17.5 watts				**12GC6**
For other characteristics, refer to Type 6GJ5										12GJ5
50	0V	125	24	70	—	—	—	—	—	12GN7
250	0V	150	6.5	28	50000	36000	—	—	—	12GN7
Max. DC Plate Volts, 770 Max. DC Cathode mA, 175						Max. Peak Positive-Pulse Plate Volts, 6500 Max. Plate Dissipation, 17.5 watts				**12GT5** 12GT5A
Max. AC Supply Volts per Plate (RMS), 117 Min. Total Effect. Plate-Supply Imped. per Plate: half-wave, 30 ohms; full wave, 15 ohms Max. AC Plate Volts (RMS), 150 Max. DC Output mA, 8 per Plate						Max. DC Output mA, 8. min. Min. Total Effective Plate-Supply Impedance: up to 117 volts, 15 ohms; at 150 volts, 40 ohms				**12H6**
For other characteristics, refer to Type 6J5GT										**12J5GT**
For other characteristics, refer to Type 6J7GT										12J7GT
12.6	— 0V	12.6	1.5	12	6000	5500	—	2700	0.02	**12J8**
For other ratings, refer to Type 6JB6										12JB6
125	— 1V	—	—	13.5	5400	8500	46	—	—	**12JN8**
125	— 1V	125	4	12	200000	7500	—	—	—	**12JN8**
Max. DC Plate Supply Volts, 770 Max. DC Cathode Current mA, 175						Max. Peak Positive-Pulse Plate Volts, 6500 Max. Plate Dissipation, 17.5				12JT6
DC Plate Volts, 12.6 Grid-No. 1 (Space- Charge Grid) Volts, 12.6 DC Plate mA, 40	Grid-No. 2 (Control Grid) Volts, —.5 Grid-No. 1 mA, 75					Plate Resistance, 480 ohms Amplification Factor, Grid-No. 2 to Plate, 7.2 Transcond., Grid-No. 2 to Plate, 15000 μmhos				**12K5**
For other characteristics, refer to Type 6K7GT										**12K7GT**

RCA Type	Name	Outline	Basing Diagram	Filament (F) Heater or Volts	Amperes	Use — Values to right give operating conditions and characteristics for indicated typical use
12K8	Triode-Hexode Converter	3	8K	12.6	0.15	Oscillator Mixer
12KL8	Diode—Sharp-Cutoff Pentode	6E	9LQ	12.6	0.15	Pentode Unit as Class A Amplifier
12L6GT	Beam Power Tube	13D	7AC	12.6	0.6	Class A Amplifier
12Q7GT	Twin Diode—High-Mu Triode	14A	7V	12.6	0.15	Triode Unit as Amplifier
12R5	Beam Power Tube	5D	7CV	12.6	0.6	Vertical Deflection Amplifier
12S8GT	Triple Diode—High-Mu Triode	14B	8CB	12.6	0.15	Triode Unit as Class A Amplifier
12SA7 12SA7 GT	Pentagrid Converter	2A 13D	8R 8AD	12.6	0.15	Converter
12SC7	High-Mu Twin Triode	2A	8S	12.6	0.15	Each Unit as Class A Amplifier
12SF5 12SF5 GT	High-Mu Triode	2A 13D	6AB 6AB	12.6	0.15	Class A Amplifier
12SF7	Diode—Remote-Cutoff Pentode	2A	7AZ	12.6	0.15	Pentode Unit as Amplifier
12SG7	Semiremote-Cutoff Pentode	2A	8BK	12.6	0.15	Class A Amplifier
12SH7	Remote-Cutoff Pentode	3	8BK	12.6	0.15	Class A Amplifier
12SJ7 12SJ7 GT	Sharp-Cutoff Pentode	2A 13D	8N 8N	12.6	0.15	Class A Amplifier
12SK7 12SK7 GT	Remote-Cutoff Pentode	2A 13D	8N 8N	12.6	0.15	Class A Amplifier
12SN7 GT	Medium-Mu Twin Triode	13D	8BD	12.6	0.3	*Each Unit as Class A Amplifier*
12SQ7 12SQ7 GT	Twin Diode—High-Mu Triode	2A 13D	8Q 8Q	12.6	0.15	Triode Unit as Class A Amplifier
12SR7 12SR7 GT	Twin Diode—Medium-Mu Triode	2A 13D	8Q 8Q	12.6	0.15	Triode Unit as Class A Amplifier
12U7	Medium-Mu Twin Triode	6B	7CK	10.0 to 15.9	0.15 approx. at 12.6V	Each Unit as Class A Amplifier
12Z3	Half-Wave Rectifier	22	4G	12.6	0.3	With Capacitive-Input Filter
13EM7	Dual Triode	13A	8BD	13	0.45	Unit No. 1 as Vertical Deflection Amplifier Unit No. 2 as Vertical Deflection Amplifier
13GB5	Beam Power Tube	10E	9NH	13.3	0.6	Horizontal Deflection Amplifier
13GF7	Dual Triode	11A	9QD	13	0.45	Vertical Deflection Amplifier Vertical Deflection Oscillator
13J10	Pentode—Beam Power Tube	8B	12BT	13.2	0.45	Pentode Unit as Class A Amplifier Beam Power Unit as Gated-Beam Discriminator
14A4	Medium-Mu Triode	12B	5AC	12.6	0.15	Class A Amplifier
14A5	Beam Power Tube	12B	6AA	12.6	0.15	Class A Amplifier
14A7	Remote-Cutoff Pentode	12B	8V	12.6	0.15	Class A Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
For other characteristics, refer to Type 6K8										12K8
For other characteristics, see Type 6KL8										**12KL8**
110	— 7.5V	110	4.0	49	13000	8000	—	2000	2.1	**12L6GT**
200	180Ω	125	2.2	46	28000	8000	—	4000	3.8	**12L6GT**
For other characteristics, refer to Type 6Q7GT										12Q7GT
Max. DC Plate Volts, 150 Max. Peak Cathode mA, 155 Max. Plate Dissipation, 4.5 watts					Max. Peak Neg.-Pulse Grid-No. 1 Volts, 150 Max. Grid-No. 2 Volts, 150 Max. Peak Positive-Pulse Plate Volts, 1500 (Abs.)					**12R5**
250	— 2V	—	—	0.9	91000	1100	100	—	—	12S8GT
For other characteristics, refer to Type 6SA7										**12SA7** **12SA7 GT**
For other characteristics, refer to Type 6SC7										**12SC7**
For other characteristics, refer to Type 6SF5										**12SF5** 12SF5 GT
For other characteristics, refer to Type 6SF7										**12SF7**
For other characteristics, refer to Type 6SG7										**12SG7**
For other characteristics, refer to Type 6SH7										**12SH7**
For other characteristics, refer to Type 6SJ7										**12SJ7** 12SJ7 GT
For other characteristics, refer to Type 6SK7										**12SK7** **12SK7 GT**
For other characteristics, refer to Type 6J5										12SN7 GT
For other characteristics, refer to Type 6SQ7										**12SQ7** **12SQ7 GT**
For other characteristics, refer to Type 6SR7										12SR7 12SR7 GT
12.6	0V	—	—	1	12500	1600	20	—	—	12U7
Max. DC Output mA,55										12Z3
Max. DC Plate Volts, 330 Max. DC Cathode mA, 22					Max. Plate Dissipation, 1.5 watts					13EM7
Max. DC Plate Volts, 330 Max. DC Cathode mA, 50					Max. Peak Positive-Pulse Plate Volts, 1500 Max. Plate Dissipation, 10 watts					13EM7
For other ratings, refer to Type 6GB5										13GB5
For other ratings, refer to Type 6GF7										13GF7
250	— 8V	250	2.5	35	100000	6500	—	5000	4.2	13J10
Max. Supply Volts, 330 Max. Grid-No. 2 Volts, 110					Max. Peak Positive Grid-No. 1 Volts, 60 Max. DC Cathode mA, 13					13J10
For other characteristics, refer to Type 6J5										14A4
250	—12.5V	250	5.5	32	70000	3000	—	7500	2.8	14A5
100	— 1V	100	4.0	13.0	120000	2350	—	—	—	14A7
250	— 3V	100	2.6	9.2	800000	2000	—	—	—	14A7

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Heater or Filament (F) Amperes	Use — Values to right give operating conditions and characteristics for indicated typical use
14AF7	Medium-Mu Twin-Triode	12B	8AC	12.6	0.15	Each Unit as Class A Amplifier
14B6	Twin Diode—High-Mu Triode	12B	8W	12.6	0.15	Triode Unit as Class A Amplifier
14B8	Pentagrid Converter	12B	8X	12.6	0.15	Converter
14C5	Beam Power Tube	12C	6AA	12.6	0.225	Class A Amplifier
14C7	Sharp-Cutoff Pentode	12B	8V	12.6	0.15	Class A Amplifier
14E6	Twin Diode—Medium-Mu Triode	12B	8W	12.6	0.15	Triode Unit as Class A Amplifier
14E7	Twin Diode—Remote-Cutoff Pentode	12B	8AE	12.6	0.15	Pentode Unit as Class A Amplifier
14F7	High-Mu Twin Triode	12B	8AC	12.6	0.15	Each Unit as Class A Amplifier
14F8	Medium-Mu Twin Triode	12A	8BW	12.6	0.15	Each Unit as Class A Amplifier
14H7	Semiremote-Cutoff Pentode	12B	8V	12.6	0.15	Class A Amplifier
14J7	Triode-Heptode Converter	12B	8BL	12.6	0.15	Converter
14JG8	Twin Diode—High-Mu Triode	6B	9KR	14	0.15	Triode Unit as Class A Amplifier
14N7	Medium-Mu Twin Triode	12C	8AC	12.6	0.3	Each Unit as Class A Amplifier
14Q7	Pentagrid Converter	12B	8AL	12.6	0.15	Converter
14R7	Twin Diode—Remote-Cutoff Pentode	12B	8AE	12.6	0.15	Pentode Unit as Class A Amplifier
15	Sharp-Cutoff Pentode	24B	5F	2.0	0.22	Class A Amplifier
15CW5	Power Pentode	6G	9CV	15	0.3	Vertical Deflection Amplifier
15HB6	Power Pentode	6G	9NW	14.7	0.3	Vertical Deflection Amplifier
15KY8	High-Mu Triode—Beam Power Tube	11C	9QT	15	0.45	Triode Unit as Oscillator; Beam Power Unit as Amplifier
16AQ3	Diode	7D	9CB	16.4	0.6	
17AX4GT	Half-Wave Rectifier	13D	4CG	16.8	0.45	Television Damper Service
17AY3	Half-Wave Rectifier	11D	9HP	16.8	0.45	Television Damper Service
17BH3	Half-Wave Rectifier	11D	9HP	17	0.6	Television Damper Service
17BS3	Half-Wave Rectifier	11D	9HP	16.8	0.45	Television Damper Service
17BZ3	Half-Wave Rectifier	8D	12FX	16.8	0.45	Television Damper Service
17DM4	Half-Wave Rectifier	13G	4CG	16.8	0.45	Television Damper Service
17DQ6A	Beam Power Tube	20	6AM	16.8	0.45	Horizontal Deflection Amplifier
17EW8	High-Mu Twin Triode	6B	9AJ	17.5	0.15	Each Unit as Class A Amplifier
17GJ5	Novar-Beam Power Tube	18A	9QK	16.8	0.45	Horizontal Deflection Amplifier
17GT5	Beam Power Tube	17B	9NZ	16.8	0.45	Horizontal Deflection Amplifier
17H3	Half-Wave Rectifier	6E	9FK	17.5	0.3	Television Damper Service
17JB6	Beam Power Tube	18A	9QL	16.8	0.45	Horizontal Deflection Amplifier
17JG6	Beam Power Tube	17B	9QU	16.8	0.6	Horizontal Deflection Amplifier
17JT6	Beam Power Tube	17C	9QU	16.8	0.45	Horizontal Deflection Amplifier
17LD8	Medium-Mu Triode—Sharp-Cutoff Pentode	10F	9QT	16.8	0.45	Triode Unit as Class A Amplifier; Pentode Unit as Class A Amplifier
17X10	Pentode—Beam Power Tube	8C	12BT	16.8	0.45	Beam Power Unit as Class A Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
For other characteristics, refer to Type 7AF7										14AF7
For other characteristics, refer to Type 6SQ7										14B6
For other characteristics, refer to Type 6A8										14B8
315	—13V	225	2.2	34.0	80000	3750	—	8500	5.5	14C5
For other characteristics, refer to Type 6SJ7										14C7
For other characteristics, refer to Type 6BF6										14E6
250	330Ω	100	1.6	7.5	700000	1300	—	—	—	14E7
For other characteristics, refer to Type 6SL7GT										14F7
250	500Ω	—	—	6.0	—	3300	48	—	—	14F8
For other characteristics, refer to Type 7H7										14H7
For other characteristics, refer to Type 7J7										14J7
250	— 2V	—	—	2	41000	2200	90	—	—	**14JG8**
For other characteristics, refer to Type 6SN7GT										14N7
For other characteristics, refer to Type 6SA7										**14Q7**
For other characteristics, refer to Type 7R7										14R7
135	— 1.5V	67.5	0.3	1.85	800000	750	—	—	—	15
For other ratings, refer to Type 6CW5										15CW5
Max. DC Plate Volts, 350 Max. Peak Positive-Pulse Plate Volts, 2500					Max. Plate Dissipation, 10 watts					**15HB6**
For other ratings, refer to Type 6KY8										15KY8
Max. Supply Volts, 250 Max. DC Plate mA, 220					Max. Peak Negative-Pulse Plate Volts, — 6000 Max. Plate Dissipation, 5 watts					16AQ3
Max. Peak Inverse Plate Volts, 4400 Max. Peak Plate mA, 750 Max. DC Plate mA, 125					Max. Peak Heater-Cathode Volts: {—4000, +300} DC component must not exceed 900 volts					17AX4 GT
For other ratings, refer to Type 6AY3										17AY3
For other ratings, refer to Type 6BH3										**17BH3**
For other ratings, refer to Type 6BS3										17BS3
Max. Peak Inverse Plate Volts, 4500 Max. Peak Plate mA, 1200					Max. Plate Dissipation, 6.5 watts Max. Peak Heater Cathode Volts: {— 4500, + 900}					17BZ3
For other ratings, refer to Type 6DM4										17DM4
Max. DC Plate Volts, 700 Max. DC Cathode mA, 140					Max. Peak Positive-Pulse Plate Volts, 6000 (Abs.) Max. Plate Dissipation, 15 watts					17DQ6A
100 200	— 1.1V — 2.1V	— —	— —	10 4.5	— —	4600 5800	50 48	— —	— —	17EW8
For other ratings, refer to Type 6GJ5										17GJ5
For other ratings, refer to Type 6GT5										**17GT5**
Max. Peak Inverse Plate Volts, 2000 Max. Peak Plate mA, 450					Max. Average Plate mA, 75 Max. Plate Dissipation, 3 watts					17H3
For other ratings, refer to Type 6JB6										17JB6
For other characteristics, refer to Type 17JG6A										17JG6
For other ratings, refer to Type 6JT6										17JT6
150	— 5V	—	—	3.3	11300	1900	21.5	—	—	17LD8
120	*— 8V*	*110*	*4*	*46*	*11700*	*7100*	—	—	—	17LD8
145	— 6V	110	3	36	30000	8600	—	3000	—	**17X10**

RCA Type	Name	Out-line	Basing Dia-gram	Heater or Filament (F) Volts	Heater or Filament (F) Amperes	Use Values to right give operating conditions and characteristics for indicated typical use.
18A5	Beam Power Tube	13F	6CK	18.5	0.3	*Horizontal Deflection Amplifier*
18FW6	*Remote-Cutoff Pentode*	5C	7CC 7CC	18.0 18.0	0.1 0.1	Class A Amplifier
18FX6	*Pentagrid Converter*	5C	7CH 7CH	18.0 18.0	0.1 0.1	Converter
18FY6	Twin Diode—High-Mu Triode	5C	7BT 7BT	18.0 18.0	0.1 0.1	Triode Unit as Class A Amplifier
19	High-Mu Twin Power Triode	22 or 13H	6C	2.0F	0.26	Amplifier
19AU4 **19AU4 GTA**	Half-Wave Rectifier	13G	4CG	18.9	0.6	*Television Damper Service*
19BG6G **19BG6 GA**	Beam Power Tube	27B —	5BT	18.9	0.3	Horizontal Deflection Amplifier
19HS6	*Sharp-Cutoff Pentode*	5C	7BK	*18.4*	**0.15**	*Class A Amplifier*
19J6	*Medium-Mu Twin Triode*	5C	7BF	*18.9*	*0.15*	*Each Unit as Class A Amplifier*
19T8	Triple Diode—High-Mu Triode	—	9E	18.9	0.15	Triode Unit as Class A Amplifier
19X8	Medium-Mu Triode—Sharp-Cutoff Pentode	6B	9AK	18.4	0.15	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
20	*Power Triode*		4D	3.3F	0.132	Class A Amplifier
20EQ7	Diode—Remote-Cutoff Pentode	6E	9LQ	20.0	0.1	Pentode Unit as Class A Amplifier
21EX6	Beam Power Tube	21B	5BT	21.5	0.6	Horizontal Deflection Amplifier
21HB5	Beam Power Tube	15B	12BJ	21	0.45	Horizontal Deflection Amplifier
21HJ5	Beam Power Tube	15C	12FL	21.5	0.6	Horizontal Deflection Amplifier
22	Sharp-Cutoff Tetrode	29K	4K	3.3F	0.132	Screen-Grid RF Amplifier
22BH3	*Half-Wave Rectifier*	*11D*	*9HP*	*22.4*	*0.45*	*Television Damper Service*
22JG6	Beam Power Tube	17B	9QU	22	0.45	Horizontal Deflection Amplifier
24A	Sharp-Cutoff Tetrode	29K	5E	2.5	1.75	Screen-Grid RF Amplifier
25A6 25A6GT	Power Pentode	2B 13D	7S 7S	25.0	0.3	Class A Amplifier
25A7GT	Rectifier—Power Pentode	13D	8F	25.0	0.3	Pentode Unit as Class A Amplifier Half-Wave Rectifier
25AC5 GT	High-Mu Power Triode	13D	6Q	25.0	0.3	Amplifier
25B5	*Direct-Coupled Power Amplifier*		*6D*	*25.0*	*0.3*	*Amplifier*
25B6G	Power Pentode	25	7S	25.0	0.3	Class A Amplifier
25B8GT	High-Mu Triode—Remote-Cutoff Pentode	13D	8T	25.0	0.15	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
25BK5	Beam Power Tube	6E	9BQ	25	0.3	Class A Amplifier
25BQ6 GT	Beam Power Tube	14D	6AM	25.0	0.3	Horizontal Deflection Amplifier
25C6G	Beam Power Tube	25	7AC	25.0	0.3	Class A Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power: Load Ohms	Power: Output Watts	RCA Type
Max. DC Plate Volts, 350 Max. DC Cathode mA, 90					Max. Peak Pos.-Pulse Plate Volts, 3000 Max. Plate Dissipation, 9 watts					**18A5**
100	68Ω	100	4.4	11	250000	4400	—	—	—	18FW6
100	— 1.5V	100	6.2	2.3	400000	Grid No. 1 Resistor, 20000 ohms Conversion Transcond., 480 μmhos				18FX6
100	— 1V	—	—	0.6	77000	1300	100	—	—	18FY6
For other characteristics, refer to Type 1J6GT										19
For other ratings, refer to Type 6AU4GTA										19AU4 **19AU4 GTA**
Max. DC Plate Volts, 700 Max. DC Plate Current, 110 mA.					Max. Peak Positive-Pulse Plate Volts, 6600 (Abs.) Max. Plate Dissipation, 20 watts					19BG6G **19BG6 GA**
75	0V	75	—	—	—	—	50	—	—	**19HS6**
150	0V	75	2.8	8.8	500000	9500	—	—	—	
100	50Ω (For both units at the specified conditions)			8.5	7100	5300	38	—	—	**19J6**
For other characteristics, refer to Type 6T8A										**19T8**
For other characteristics, refer to Type 6X8										19X8
135	—22.5V	—	—	6.5	6300	525	3.3	6500	0.110	20
For other characteristics, refer to Type 6EQ7										20EQ7
For other ratings, refer to Type 6EX6										21EX6
Max. DC Plate Supply Volts, 770 Max. Peak Positive-Pulse Plate Volts, 6000						Max. DC Cathode mA, 230 Max. Plate Dissipation, 18 watts				21HB5
Max. DC Plate Supply Volts, 770 Max. Peak Positive-Pulse Plate Volts, 7000						Max. DC Cathode mA, 280 Max. Plate Dissipation, 24 watts				21HJ5
135	— 1.5V	67.5	1.3 (Max.)	3.7	325000	500	—	—	—	22
For other ratings, refer to Type 6BH3										22BH3
For other characteristics, refer to Type 22JG6A										22JG6
250	— 3V	90	1.7 (Max.)	4.0	600000	1050	—	—	—	24A
95	—15V	95	4	20	45000	2000	—	4500	0.9	25A6 25A6GT
100	—15V	100	4.0	20.5	50000	1800	—	4500	0.77	25A7 GT
Max. AC Plate Volts (RMS), 117				Max. DC Output mA, 75			Max. Peak Plate mA, 450			
110	+15V (Grid mA, 7)			15	15200	3800	58	—	—	25AC5 GT
For other characteristics, refer to Type 25N6G										25B5
200	—23V	135	1.8	62.0	18000	5000	—	2500	7.1	25B6G
100	— 1V	—	—	0.6	75000	1500	112	—	—	25B8GT
100	— 3V	100	2.0	7.6	185000	2000	—	—	—	
For other characteristics, refer to Type 6BK5										**25BK5**
Max. DC Plate Volts, 600 Max. DC Cathode mA, 112.5					Absolute Max. Peak Positive-Pulse Plate Volts, 6000 (Abs.) Max. Plate Dissipation, 11 Watts					25BQ6 GT
For other characteristics, refer to Type 6Y6G										25C6G

RCA Type	Name	Outline	Basing Diagram	Heater or Filament (F) Volts	Heater or Filament (F) Amperes	Use Values to right give operating conditions and characteristics for indicated typical use
25CD6 GA	Beam Power Tube	21B	5BT 5BT	25 25	0.6	Horizontal Deflection Amplifier
25EC6	Beam Power Tube	21A	5BT	25.0	0.6	Horizontal Deflection Amplifier
25L6	Beam Power Tube	2B	7AC	25.0	0.3	Amplifier
25L6GT	Beam Power Tube	13D	7AC	25.0	0.3	Amplifier
25N6G	Direct-Coupled Power Amplifier	—	7W	25.0	0.3	Class A Amplifier
25W4GT	Half-Wave Rectifier	13D	4CG	25.0	0.3	Television Damper Service
25Y5	Rectifier-Doubler	22 or 13H	6E	25.0	0.3	Half-Wave Rectifier
25Z5	Rectifier-Doubler	22 or 13H	6E	25.0	0.3	Rectifier-Doubler
25Z6 **25Z6GT**	Rectifier-Doubler	2B 13D	7Q 7Q	25.0 25.0	0.3 0.3	Voltage Doubler Half-Wave Rectifier
26	Medium-Mu Triode	26	4D	1.5F	1.05	Class A Amplifier
27	Low-Mu Triode	22 or 13H	5A	2.5	1.75	Class A Amplifier
30	Medium-Mu Triode	22 or 13H	4D	2.0F	0.06	Amplifier
31	Power Triode	22 or 13H	4D	2.0F	0.13	Class A Amplifier
32	Sharp-Cutoff Tetrode	29K	4K	2.0F	0.06	Class A Amplifier
32ET5	Power Pentode	5D	7CV	32.0	0.1	Class A Amplifier
32L7GT	Rectifier—Beam Power Tube	14A	8Z	32.5	0.3	Class A Amplifier Half-Wave Rectifier
33	Power Pentode	25	5K	2.5F	0.26	Class A Amplifier
33GY7	Diode—Beam Power Tube	15A	12FN	33.6	0.45	Diode Unit as Television Damper Service Beam Power Unit as Horizontal Deflection Amplifier
34	Remote-Cutoff Pentode	29K	4M	2.0F	0.06	Screen-Grid RF Amplifier
34GD5	Beam Power Tube	5D	7CV 7CV	34.0 34.0	0.1 0.1	Class A Amplifier
35	Remote-Cutoff Tetrode	29K	5E	2.5	1.75	Screen-Grid RF Amplifier
35A5	Beam Power Tube	12C	6AA	35.0	0.15	Single-Tube Class A Amplifier
35B5	Beam Power Tube	5D	7BZ	35.0	0.15	Class A Amplifier
35DZ8	High-Mu Triode—Power Pentode	6H	9JE	35.0	0.15	Triode Unit as Class A Amplifier Pentode Unit as Class A Amplifier
35GL6	Beam Power Tube	5D	7FZ	35.0	0.15	Class A Amplifier
35Y4	Half-Wave Rectifier Heater Tap for Pilot	12C	5AL Pilot Between Pins 1 and 4	35.0	0.15	With Capacitive-Input Filter
35Z3	Half-Wave Rectifier	12C	4Z	35.0	0.15	With Capacitive-Input Filter
35Z4GT	Half-Wave Rectifier	13D	5AA	35.0	0.15	With Capacitive-Input Filter
36	Sharp-Cutoff Tetrode	24B	5E	6.3	0.3	Screen-Grid RF Amplifier

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power Output Ohms	Power Output Watts	RCA Type
Max. DC Plate Volts, 700 Max. DC Plate mA, 200				Max. Peak Positive-Plus Plate Vols, 7000 Max. Plate Dissipation, 20 Watts						25CD6 GA
Max. DC Plate Volts, 700 Max. DC Cathode mA, 200					Max. Peak Positive-Pulse Plate Volts, 700 (Abs.) Max. Plate Dissipation 10 watts.					**25EC6**
110 200	— 7.5V — 8V	110 110	4 2	49 50	13000 30000	9000 9500	— —	2000 3000	2.1 4.3	25L6
For other characteristics, refer to Type 50L6GT										**25L6GT**
Output Triode: Plate Volts, 180; Plate mA, 46; Load, 4000 ohms Triode: Plate Volts, 100; Grid Volts, 0; A-F Signal Volts (Peak), 29.7; Plate mA, 5.8									3.8	25N6G
Max. Peak Inverse Plate Volts, 3850 (Abs.) Max. Peak Plate mA, 750 Max. DC Plate mA, 125					Max. Peak Heater-Cathode Volts: { —500 (Abs.) +200 DC Component must not exceed 100 volts					**25W4GT**
Max. DC Output mA per Plate, 75										25Y5
For other ratings, refer to Type 25Z6										**25Z5**
Max. AC Volts per Plate (RMS), 117 Max. DC Output mA, 75				Min. Total Effective Plate-Supply Impedance: Half-Wave, 30 ohms; Full-Wave, 15 ohms						25Z6
Max. AC Volts per Plate (RMS), 235 Max. DC Output mA per Plate, 75				Min. Total Effect. Supply Imped. per Plate: at 117 volts 15 ohms; at 150 volts, 40 ohms; at 235 volts, 100 ohms						**25Z6GT**
180	—14.5V	—	—	6.2	7300	1150	8.3	—	—	26
250	—21V	—	—	5.2	9250	975	9.0	—	—	27
For other characteristics, refer to Type 1H4G										30
180	—30V	—	—	12.3	3600	1050	3.8	5700	0.375	31
180 (Max.)	— 3V	67.5	0.4	1.7	1 M	650	—	—	—	32
110	— 7.5V	110	2.8	30	21500	5500	—	2800	1.2	32ET5
90	— 7V	90	2.0	27.0	17000	4800	—	2600	1.0	32L7GT
Maximum AC Plate Voltage........125 Volts, RMS Maximum DC Output Current........60 Milliamperes										32L7GT
180	—18V	180	5.0	22.0	55000	1750	—	6000	1.4	33
Max. Peak Inverse Plate Volts, 4200 Max. Peak Plate mA, 810						Max. Plate Dissipation, 3.8 watts Max. Peak Heater-Cathode Volts: { — 4200 + 200				33GY7
Max. DC Plate Supply Volts, 400 Max. Peak Positive-Pulse Plate Volts, 5000						Max. DC Cathode mA, 155 Max. Plate Dissipation, 9 watts				33GY7
180	— 3V min.	67.5	1.0	2.8	1 M	620	—	—	—	34
110	— 7.5V	110	3	35	13000	5700	—	2500	1.4	34GD5
250	— 3V min.	90	2.5	6.5	—	1050	—	—	—	35
For other characteristics, refer to Type 35L6GT										35A5
For other characteristics, refer to Type 35C5										**35B5**
120	1500Ω	—	—	0.8	—	1400	100	—	—	35DZ8
140	180Ω	120	6	45	—	7500	—	2500	2.0	35DZ8
110	— 7.5V	110	3	45	12000	7500	—	2500	1.8	**35GL6**
For other characteristics, refer to Type 35W4										**35Y4**
For other ratings, refer to Type 35Z5GT										**35Z3**
Max. DC Output mA, 100				Min. Total Effective Plate-Supply Impedance: Up to 117 volts, 15 ohms; at 235 volts, 100 ohms						**35Z4GT**
100 250	— 1.5V — 3V	55 90	— 1.7	1.8 3.2	550000 550000	850 1080	—	—	—	36

RCA Type	Name	Out-	Basing Diagram	Heater or Filament (F) Volts	Amperes	Use — Values to right give operating conditions and characteristics for indicated typical use
36AM3	Half-Wave Rectifier	5D	5BQ	36.0	0.1	With Capacitive-Input Filter
36AM3A	Half-Wave Rectifier	5D	5BQ 5BQ	36.0 36.0	0.1 0.1	With Capacitive-Input Filter
37	Medium-Mu Triode	22 or 13H	5A	6.3	0.3	Class A Amplifier
38	Power Pentode	24B	5F	6.3	0.3	Class A Amplifier
39/44	Remote-Cutoff Pentode	24B	5F	6.3	0.3	Class A Amplifier
40	Medium-Mu Triode	26	4D	5.0F	0.25	Class A Amplifier
41	Power Pentode	22 or 13H	6B	6.3	0.4	Amplifier
42	Power Pentode	28	6B	6.3	0.7	Amplifier
43	Power Pentode	28	6B	25.0	0.3	Amplifier
45	Power Triode	26	4D	2.5F	1.5	Class A Amplifier
45Z3	Half-Wave Rectifier	5C	5AM	45.0	0.075	Half-Wave Rectifier
45Z5GT	Half-Wave Rectifier Heater Tap for Pilot	13D	6AD Pilot Between Pins 2 and 3	45.0	0.15	With Capacitive-Input Filter
46	Dual-Grid Power Amplifier	27B	5C	2.5F	1.75	Class A Amplifier
47	Power Pentode	27B	5B	2.5F	1.75	Class A Amplifier
48	Power Tetrode	27B	6A	30.0	0.4	Class A Amplifier
49	Dual-Grid Power Amplifier	26	5C	2.0F	0.12	Class A Amplifier
50	Power Triode	29L	4D	7.5F	1.25	Class A Amplifier
50A5	Beam Power Tube	12C	6AA	50.0	0.15	Class A Amplifier
50C6G	Beam Power Tube	25	7AC	50.0	0.15	Single-Tube Class A Amplifier
50FE5	Beam Power Tube	13G	8KB	50.0	0.15	Class A Amplifier
50FK5	Power Pentode	5D	7CV	50.0	0.1	Class A Amplifier
50X6	Rectifier-Doubler	12C	7DX	50.0	0.15	Rectifier-Doubler
50Y6GT	Rectifier-Doubler	13D	7Q	50.0	0.15	Rectifier-Doubler
50Y7GT	Rectifier-Doubler Heater Tap for Pilot	13D	8AN Pilot Between Pins 6 and 7	50.0	0.15	Voltage Doubler Half-Wave Rectifier
50Z7G	Rectifier-Doubler Heater Tap for Pilot	22	8AN Pilot Between Pins 6 and 7	50.0	0.15	Voltage Doubler Half-Wave Rectifier
53	High-Mu Twin Power Triode	26	7B	2.5	2.0	Amplifier
70L7GT	Rectifier-Beam Power Tube	13F	8AA	70.0	0.15	Amplifier Unit as Class A Amplifier Half-Wave Rectifier
75	Twin Diode—High-Mu Triode	24B	6G	6.3	0.3	Amplifier
78	Remote-Cutoff Pentode	24B	6F	6.3	0.3	Amplifier Mixer
80	Full-Wave Rectifier	26	4C	5.0F	2.0	With Capacitive-Input Filter With Inductive-Input Filter
84/6Z4	Full-Wave Rectifier	22 or 13H	5D	6.3	0.5	With Capacitive-Input Filter With Inductive-Input Filter

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Current mA	Plate Current mA	AC Plate Resistance Ohms	Transconductance Micromhos	Amplification Factor	Power Ohms	Power Output Watts	RCA Type
AC Plate Volts (RMS), 117; Max. DC Output mA, 82					Max. Peak Inverse Volts, 365; Tube Voltage Drop for Plate mA, 150, 20 volts					36AM3
Max. AC Plate Volts (RMS), 120; Max. DC Output mA, 82					Max. Peak Inverse Volts, 365; Tube Voltage Drop for Plate mA, 150, 16 volts					36AM3A
250	—18V	—	—	7.5	8400	1100	9.2	—	—	37
250	—25V	250	3.8	22.0	100000	1200	—	10000	2.50	38
250	— 3V min.	90	1.4	5.8	1.0	1050	—	—	—	39/44
180	— 3V	—	—	0.2	150000	200	30	—	—	40
For other characteristics, refer to Type 6K6GT										41
For other characteristics, refer to Type 6F6G										42
For other characteristics, refer to Type 25A6										43
275	—56V	—	—	36.0	1700	2050	3.5	4600	2.00	45
Max. Peak Inverse Volts, 350				Max. DC Output mA, 65			Max. Peak Plate mA, 390			45Z3
For other ratings, refer to Type 35Z5GT										45Z5GT
250	—33V	—	—	22	2380	2350	5.6	6400	1.25	46
250	450Ω	250	6.0	31	60000	2500	—	7000	2.7	47
125	—20V	100	9.5	56	—	3900	—	1500	2.5	48
135	—20V	—	—	6.0	4175	1125	4.7	11000	0.17	49
450	—84V	—	—	55	1800	2100	3.8	4350	4.6	50
For other characteristics, refer to Type 50L6GT										50A5
135	—13.5V	135	3.5	58	9300	7000	—	2000	3.6	50C6G
200	—14V	135	2.2	61	18300	7100	—	2600	6	50C6G
For other characteristics, refer to Type 6FE5										50FE5
110	62Ω	115	8.5	32	14000	12800	—	3000	1.2	50FK5
For other ratings, refer to Type 25Z6GT										50X6
For other ratings, refer to Type 25Z6GT										50Y6GT
Max. AC Volts per Plate (RMS), 117; Max. DC Output mA, 65				Min. Total Effective Plate-Supply Impedance per Plate, 15 ohms						50Y7GT
Max. AC Volts per Plate (RMS), 235; Max. DC Output mA per Plate, 65				Min. Total Effec. Plate-Supply Imped. per Plate: At 117 volts, 15 ohms; at 150 volts, 40 ohms; at 235 volts, 100 ohms						50Y7GT
Max. DC Output mA, 65										50Z7G
Max. DC Output mA per Plate, 65										50Z7G
For other characteristics, refer to Type 6N7										53
110	— 7.5V	110	3.0	40.0	15000	7500	—	2000	1.8	70L7GT
Max. Peak Inverse Volts, 350				Max. DC Output mA, 70; Min. Total Effect. Plate-Supply Imped., 15 ohms			Max. Peak Plate mA, 420			70L7GT
For other characteristics, refer to Type 6SQ7										75
For other characteristics, refer to Type 6K7										78
AC Volts per Plate (RMS), 350; Max. Peak Inverse Volts, 1400				DC Output mA, 125; Max. Peak Plate mA, 440			Min. Total Effect. Supply Imped. per Plate, 50 ohms			80
AC Volts per Plate (RMS), 500; Max. Peak Inverse Volts, 1400				Max. DC Output mA, 125; Max. Peak Plate mA, 440			Min. Value of Input Choke, 10 henries			80
AC Volts per Plate (RMS), 325; Max. Peak Inverse Volts, 1250				DC Output mA, 60; Max. Peak Plate mA, 180			Total Effect. Supply Imped. per Plate, 150 ohms			84/6Z4
AC Volts per Plate (RMS), 450; Max. Peak Inverse Volts, 1250				Max. DC Output mA, 60; Max. Peak Plate mA, 180			Value of Input Choke, 10 henries			84/6Z4

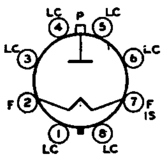
RCA Type	Name	Out-line	Basing Diagram	Heater or Filament (F) Volts	Amperes	Use Values to right give operating conditions and characteristics for indicated typical use
117L7 GT/ M7GT	Rectifier-Beam Power Tube	13F	8AO	117	0.09	Amplifier Unit as Class A Amplifier
						Half-Wave Rectifier
117N7 GT	Rectifier-Beam Power Tube	13F	8AV	117	0.09	Amplifier Unit as Class A Amplifier
						Half-Wave Rectifier
117P7 GT	Rectifier-Beam Power Tube	13F	8AV	117	0.09	
117Z3	Half-Wave Rectifier	5D	4CB	117	0.04	With Capacitive-Input Filter
117Z4 GT	Half-Wave Rectifier	29F	5AA	117	0.04	With Capacitive-Input Filter
117Z6 GT	Rectifier-Doubler	13D	7Q	117	0.075	Voltage Doubler
						Half-Wave Rectifier
7027	Beam Power Tube	19F	8HY	6.3	0.9	Push-Pull Class AB_1 Amplifier
						Push-Pull Class AB_1 Amplifier
7247	Dual Triode	6B	9A	12.6	0.15	Unit No. 1 as Class A Amplifier
				6.3	0.3	Unit No. 2 as Class A Amplifier
7591	Beam Power Tube	13D	8KQ	6.3	0.8	Class A Amplifier
						Push-Pull Class AB_1 Amplifier
7695	Beam Power Tube	13D	9PX	50	0.15	Class A Amplifier
						Push-Pull Class AB_1 Amplifier
EM84/ 6FG6	Electron—Ray Tube	6F	9GA	6.3	0.27	Visual Indicator

Plate Volts	Grid Bias or Cathode Resistor	Screen Grid Volts	Screen Grid Cur-rent mA	Plate Cur-rent mA	AC Plate Resist-ance Ohms	conduct-conduct-ance micromhos	Amplifi-cation Factor	Power Ohms	Out-put Watts	RCA Type
105	— 5.2V	105	4	43	17000	5300	——	4000	0.85	117L7 GT/ M7GT
Max. AC Plate Volts (RMS), 117 Maffl. Peak Inverse Volts, 350				Max. DC Output mA, 75 Max. Peak Plate mA, 450			Min. Total Effect. Plate-Supply Imped., 15 ohms			
100	— 6V	100	5	51	16000	7000	——	3000	1.2	117N7 GT
Max. AC Plate Volts (RMS), 117 Max. Peak Inverse Volts, 350				Max. DC Output mA, 75 Max. Peak Plate mA, 450			Min. Total Effect. Plate-Supply Impedance, 15 ohms			
For other characteristics, refer to Type 117L7/M7GT										117P7 GT
Max. Peak Inverse Volts, 330				Max. DC Output mA, 90 Max. Peak Plate mA, 540			Min. Total Effect. Plate-Supply Imped., 20 ohms			117Z3
Max. Peak Inverse Volts, 350				Max. DC Output mA, 90 Max. Peak Plate mA, 540			Min. Total Effect. Plate-Supply Imped., 30 ohms			117Z4 GT
AC Volts per Plate (RMS), 117 DC Output mA, 60				Min. Total Effective Plate-Supply Impedance per Plate: Half-Wave, 30 ohms; Full-Wave, 15 ohms						117Z6 GT
AC Volts per Plate (RMS), 235 DC Output mA per Plate, 60				Min. Total Effect. Supply Imped. per Plate: At 117 volts, 15 ohms; at 150 volts, 40 ohms; at 235 volts, 100 ohms						
450	—30V	350	3.4□	95□	——	——	——	6000	50	7027
400	200Ω	300	7□	112□	——	——	——	6600	32	
380	180Ω	380	5.6□	138□	——	——	——	4500	36	
410	220Ω	——	Cath. mA, 134		——	——	——	8000	24	
250	— 2V	——	——	1.2	62500	1600	100	——	——	7247
250	— 8.5V	——	——	10.5	7700	2200	17	——	——	
300	—10V	300	8	60	29000	10200	——	3000	11	7591
450	200Ω	400	11.5	82	——	——	——	9000	28†	
130	—11V	130	5	100	7000	11000	——	1100	4.5	7695
140	50Ω	140	9	210	——	——	——	1500	10†	
Triode Plate Supply Volts, 250 Triode-Plate Resistance, 1 MΩ Triode Grid-Supply Volts, —22				Triode Plate mA, 0.06			Fluorescent-Target Volts, 250 Triode-Grid Resistance, 0.47 MΩ Fluorescent Target mA, 1.6			EM84/ 6FG6
Max. Length of Dark Part of Target, when triode grid resistor = 0, 1.14 inch										

† For two tubes at stated plate-to-plate load. □ For two tubes.

Basing Diagrams for RCA Replacement and Discontinued Types

For Key: Basing Diagrams, see inside back cover.



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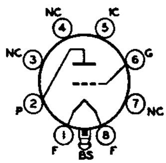

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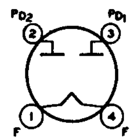

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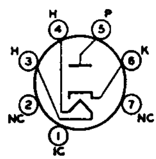

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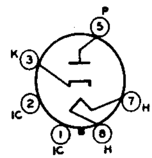

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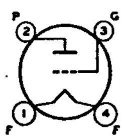

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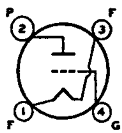

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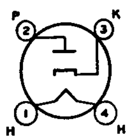

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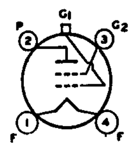

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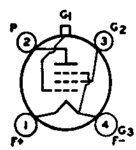

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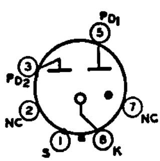

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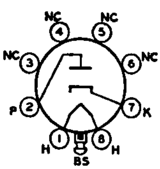

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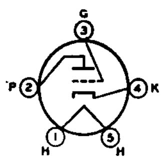

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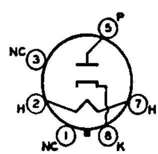

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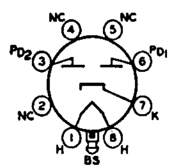

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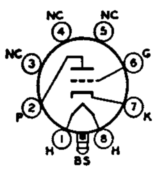

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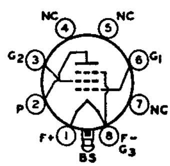

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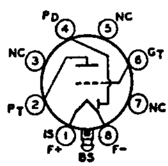

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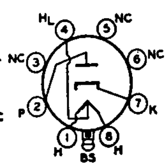

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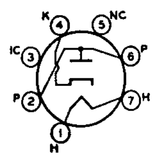

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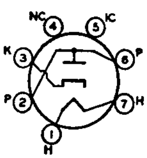

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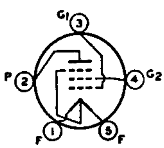

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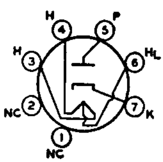

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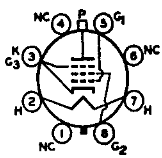

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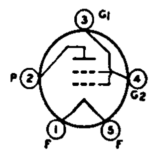

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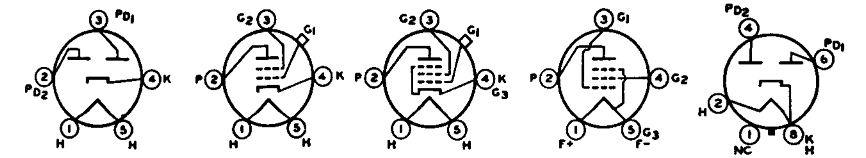

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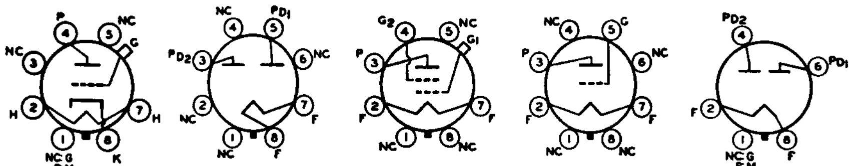

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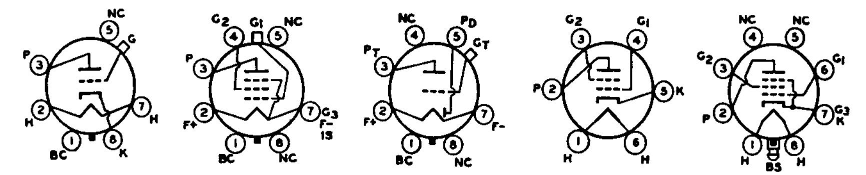

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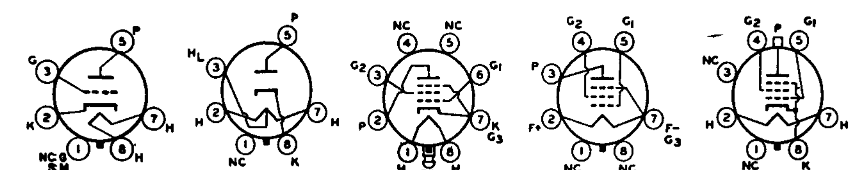

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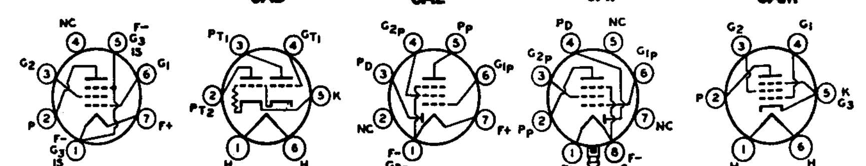

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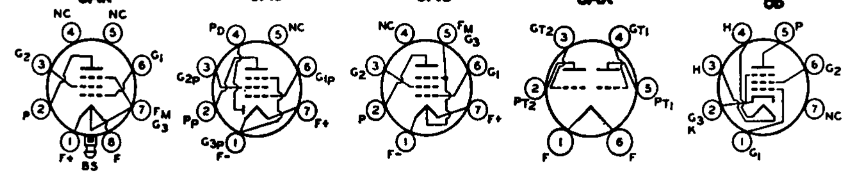

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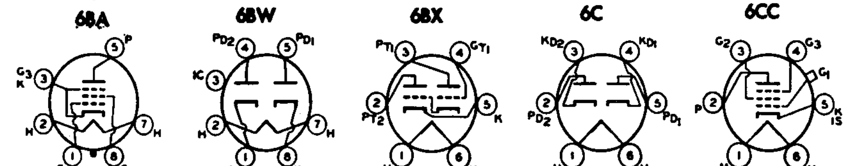

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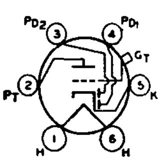
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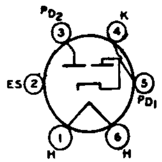
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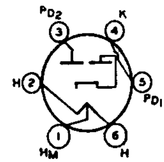
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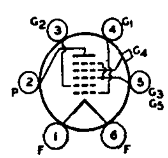
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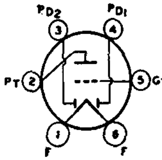
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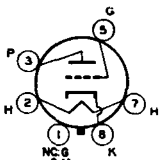
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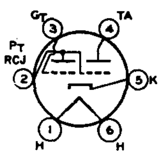
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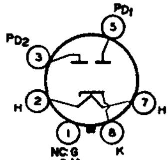
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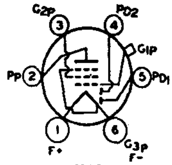
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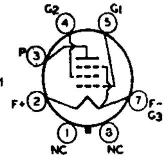
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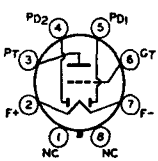
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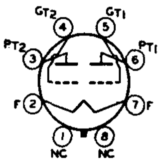
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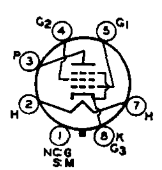
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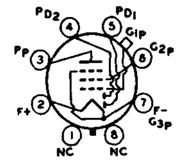
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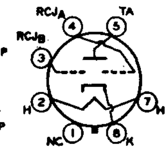
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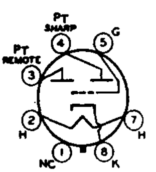
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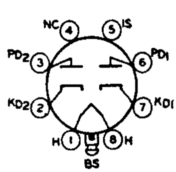
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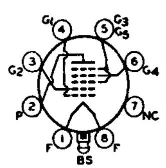
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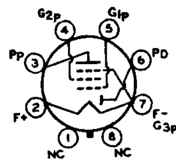
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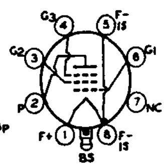
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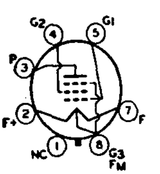
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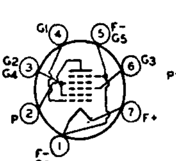
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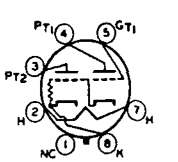
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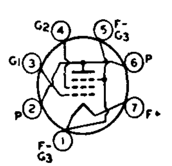
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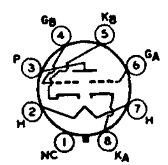
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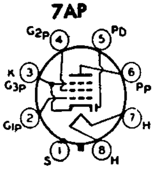
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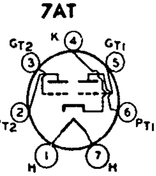
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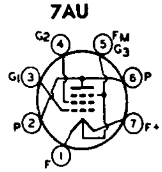
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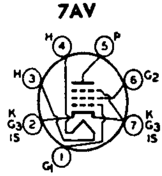
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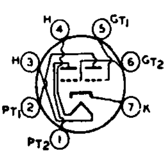
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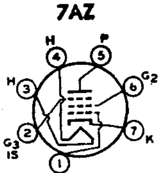
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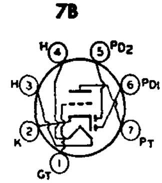
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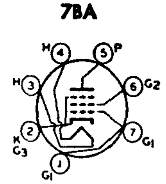
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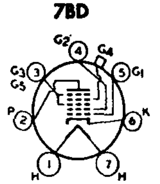
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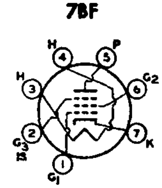
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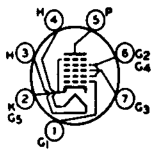
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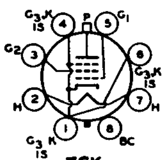
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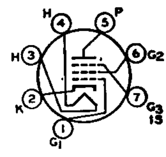
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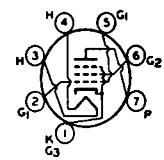
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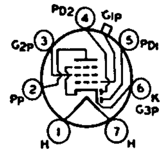
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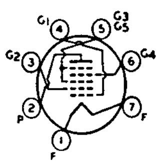
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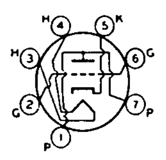
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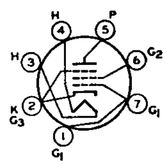
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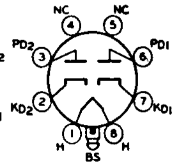
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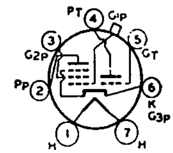
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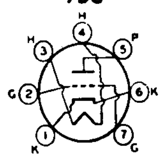
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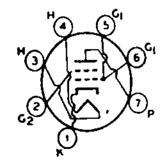
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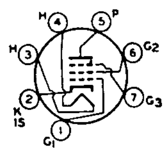
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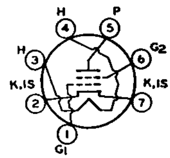
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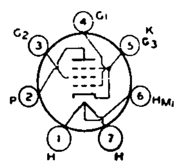
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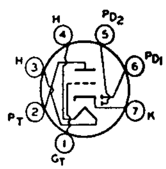
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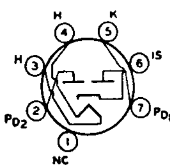
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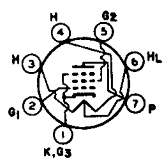
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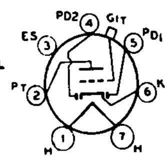
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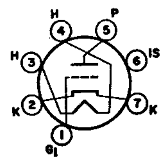
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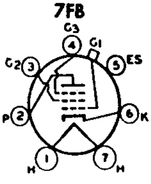
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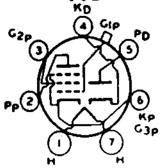
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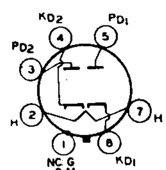
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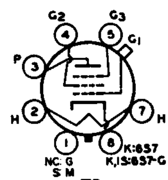
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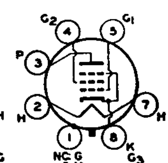
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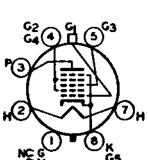
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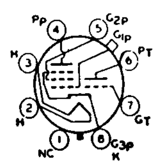
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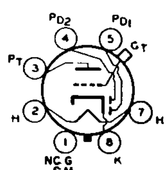
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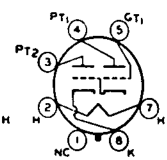
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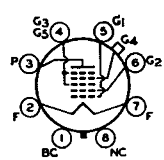
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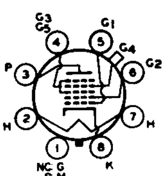
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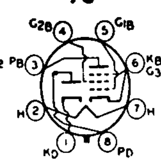
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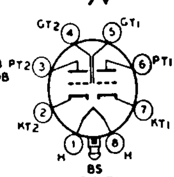
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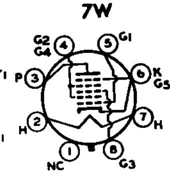
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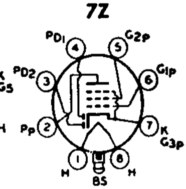
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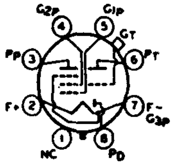
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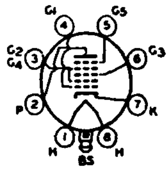
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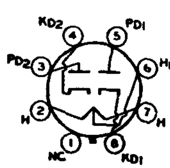
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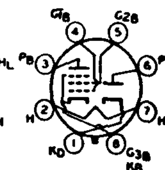
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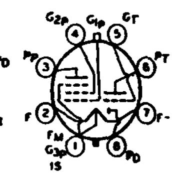
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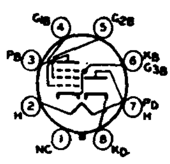
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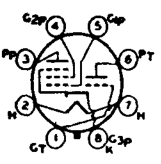
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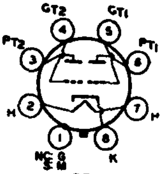
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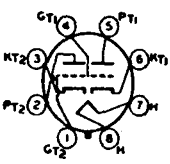
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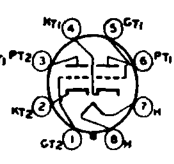
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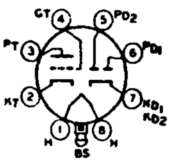
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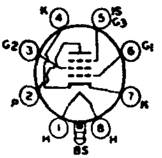
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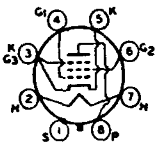
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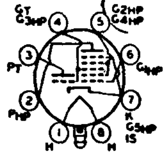
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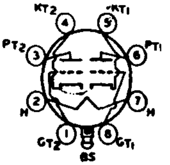
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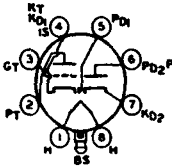
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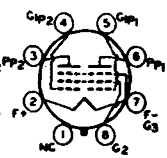
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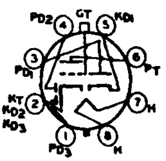
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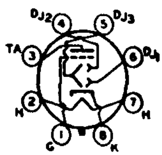
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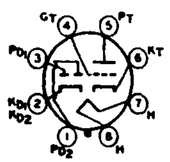
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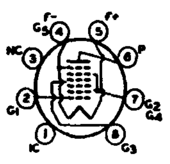
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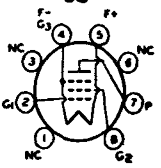
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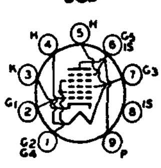
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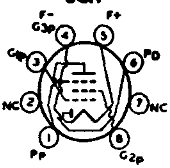
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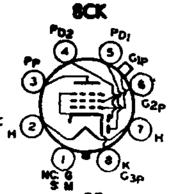
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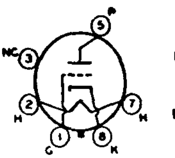
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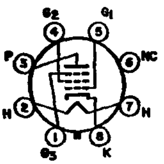
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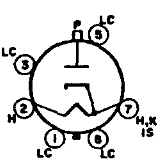
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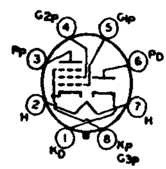
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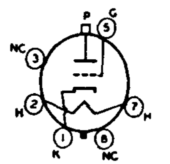
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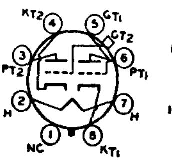
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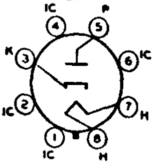
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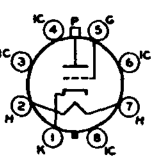
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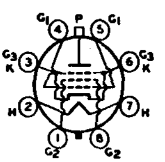
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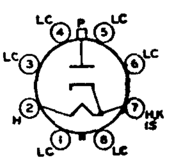
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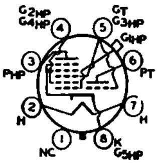
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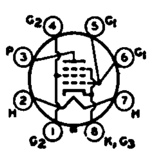
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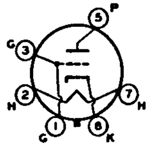
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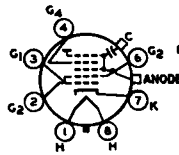
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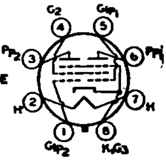
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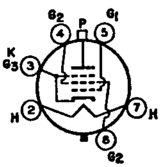
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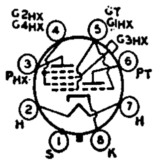
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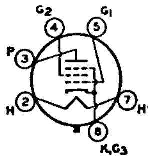
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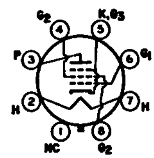
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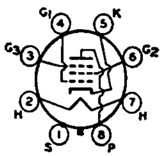
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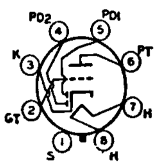
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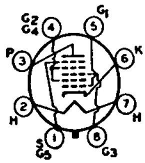
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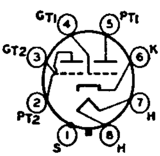
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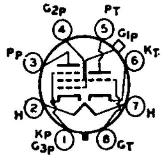
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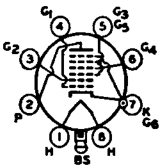
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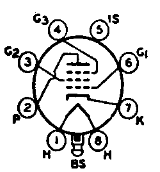
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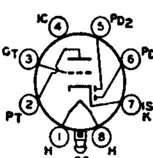
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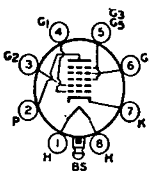
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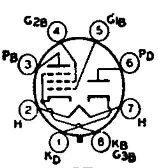
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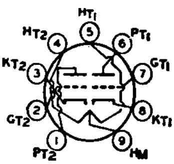
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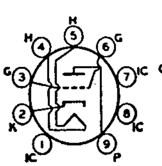
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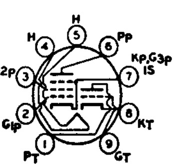
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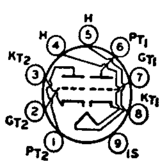
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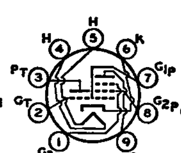
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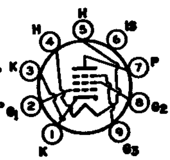
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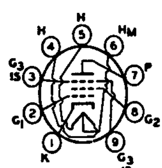
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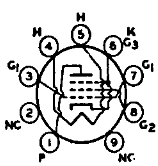
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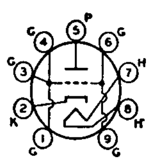
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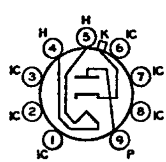
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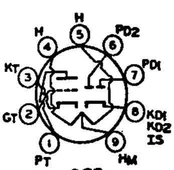
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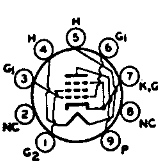
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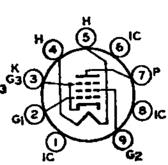
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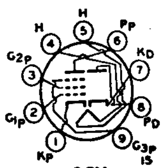
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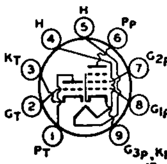
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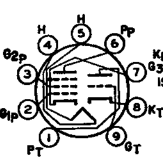
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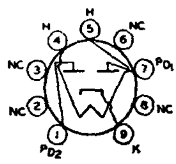

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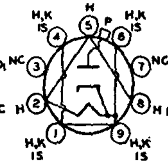

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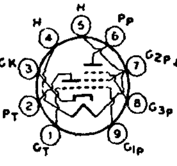

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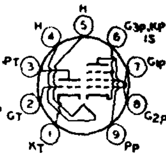

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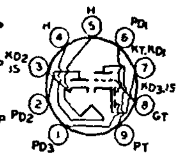

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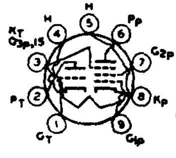

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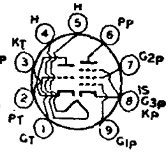

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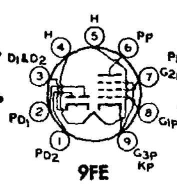

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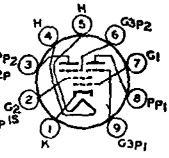

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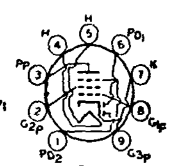

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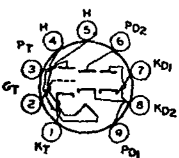

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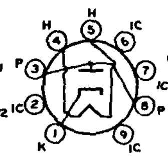

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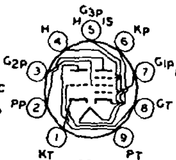

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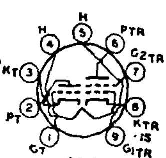

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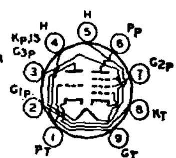

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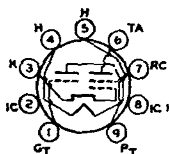

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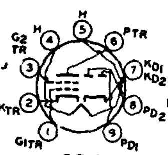

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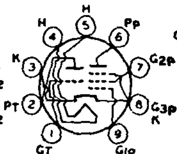

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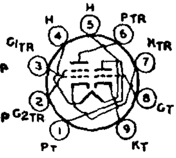

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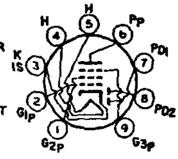

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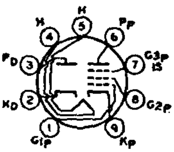

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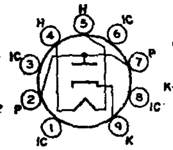

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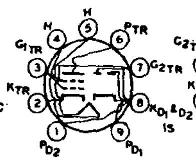

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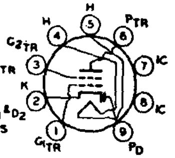

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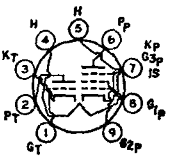

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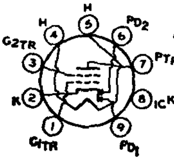

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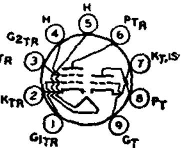

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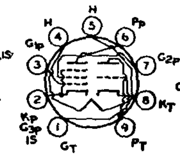

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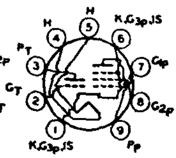
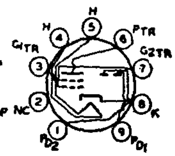
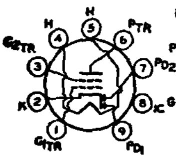


9JG



9JU



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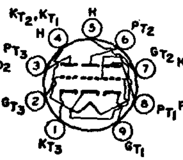
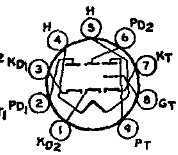


9KA



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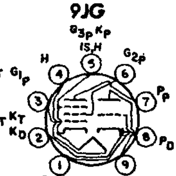
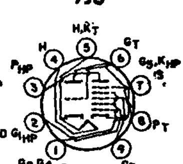


9KU



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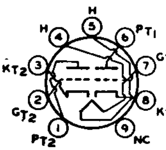

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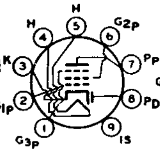

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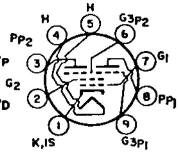

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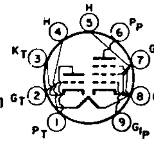

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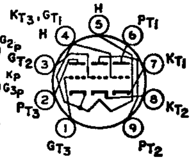
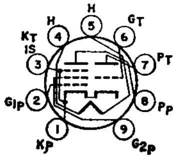


9MB



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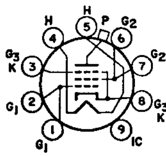

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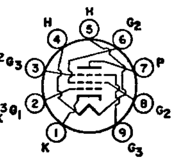

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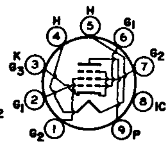

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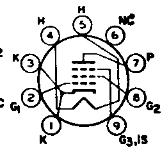

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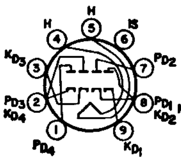

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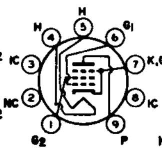

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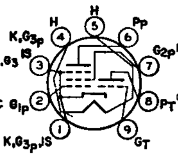

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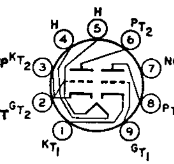

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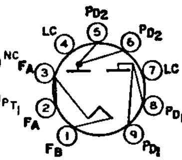

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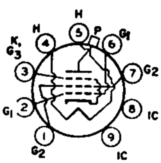

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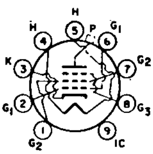

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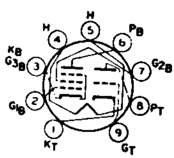
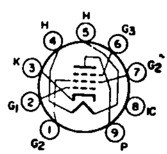
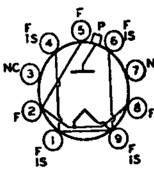


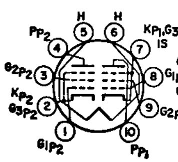
9QT



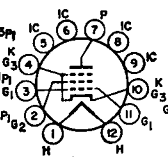
9QU



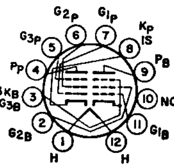
9Y



10L



12BJ



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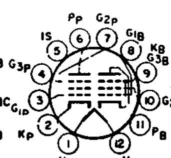

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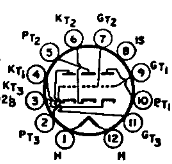
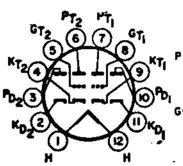
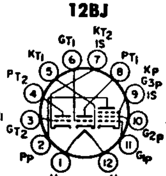


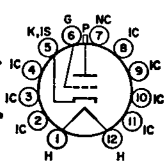
12BY



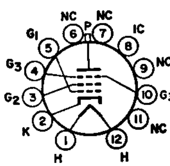
12DA



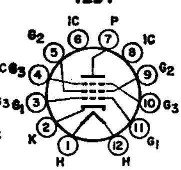
12DP



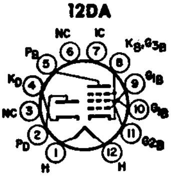
12FA



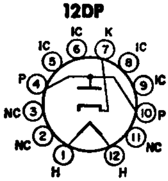
12FJ



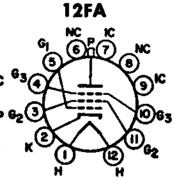
12FL



12FN.

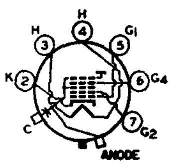


12FX

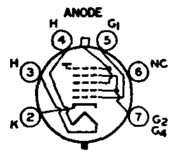


12FY

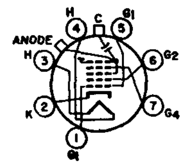
Basing Diagrams for RCA Picture Tubes



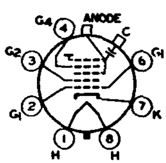
7FA
ANODE = $G_3 + G_5 + CL$
FOCUSING ELECTRODE = G_4



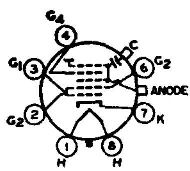
7FG
ANODE = $G_3 + G_5 + CL$
AUTOMATIC FOCUSING



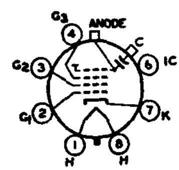
7GR
ANODE = $G_3 + G_5 + CL$
FOCUSING ELECTRODE = G_4



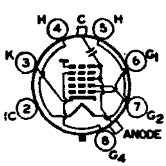
8HR
ANODE = $G_3 + G_5 + CL$
FOCUSING ELECTRODE = G_4



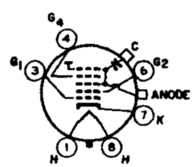
8JK
ANODE = $G_3 + G_5 + CL$
FOCUSING ELECTRODE = G_4



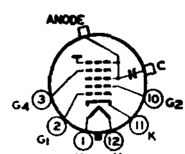
8JR
ANODE = $G_4 + CL$
FOCUSING ELECTRODE = G_3



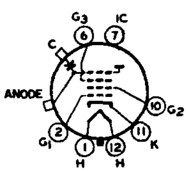
8KP
ANODE = $G_3 + G_5 + CL$
FOCUSING ELECTRODE = G_4



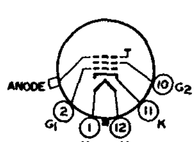
8KW
ANODE = $G_3 + G_5 + CL$
FOCUSING ELECTRODE = G_4



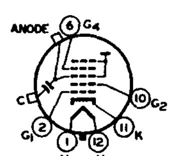
12AB
ANODE = $G_3 + G_5 + CL$
FOCUSING ELECTRODE = G_4



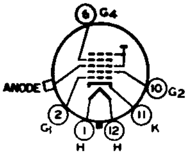
12C
ANODE = $G_4 + CL$
FOCUSING ELECTRODE = G_3



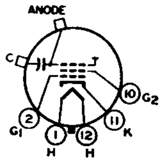
12D
ANODE = $G_3 + CL$



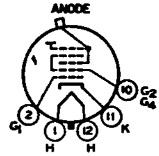
12L
ANODE = $G_3 + G_5 + CL$
FOCUSING ELECTRODE = G_4



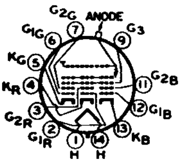
12M
ANODE = $G_3 + G_5$ + CL
FOCUSING ELECTRODE = G_4



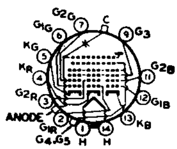
12N
ANODE = G_3 + CL



12S
ANODE = $G_3 + G_5$ + CL
AUTOMATIC FOCUSING

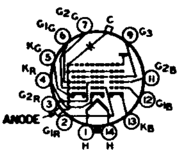


14AH
ANODE = $G_4 + G_5$ + CL + R
FOCUSING ELECTRODE = G_3

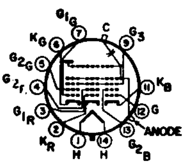


14AL

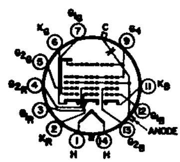
(14AL)
CAP OVER PIN No. 1
= $G_4 + G_5$
CAP OVER PIN No. 2 = ANODE
= G_6 + CL & HIGH-VOLTAGE TERMINAL. Connect High-Voltage Supply to this Cap and also connect 50,000-ohm resistor between this Cap and the Cap over Pin No. 1.
FOCUSING ELECTRODE = G_3



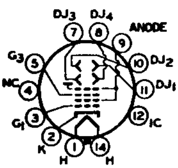
14AU
ULTOR = $G_4 + G_5$ + CL
FOCUSING ELECTRODE = G_3



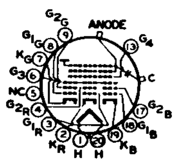
14AE
ANODE = $G_2 + G_4$ + CL
FOCUSING ELECTRODE = G_3



14BH
ANODE = $G_3 + G_5$ + CL
FOCUSING ELECTRODE = G_4



14R
ANODE = $G_2 + G_4$ + CL
FOCUSING ELECTRODE = G_3



20A
ANODE = $G_5 + G_6$ + CL
FOCUSING ELECTRODE = G_3

Notes for RCA Picture Tube Characteristics Chart

● *G* Glass round.
● *M* Metal round.
■ *G* Glass rectangular.
■ *M* Metal rectangular.
E Electrostatic.
M Magnetic.
a Faceplate is spherical, unless otherwise specified.
b All types utilize magnetic deflection except for type 7JP4 which employs electrostatic deflection.
c The anode is defined as the electrode, or the electrode in combination with one or more additional electrodes connected within the tube to it, to which is applied the highest dc voltage for accelerating the electrons in the beam.
d Projection type.
e Typical deflection factors (volts dc/in.) for anode voltage of 6000 volts:
DJ1 & DJ2 (nearer screen) 186 to 246
DJ3 & DJ4 (nearer base) 150 to 204
f Has low grid-No.2 voltage rating: for Cathode-Drive Service.
g This type has an internal magnetic shield.
h Cylindrical faceplate.
j Bipanel type.
k Treated to reduce specular reflection.
m Integral implosion protection.
n This type has a flat, aluminized, filterglass phosphor-dot screen plate.
p Three heaters paralleled internally.
q This type has an integral protective window.
r Three heaters series connected internally.
s Automatic focus.
t Hi-Lite screen, rare-earth phospor.
u Filled-rim-type safety feature.
v 21-inch round color picture tube similar to 21FBP22.

RCA PICTURE TUBE CHARACTERISTICS CHART

RCA Type	Aluminized Screen	Heater Volts/mA	Envelope[a]	Greatest Deflection Angle[b] (Approx.) Degrees	Focusing Method	Approx. Tube Weight Pounds	Maximum Over-all Length Inches	Basing*	Design Maximum Anode[c] Volts	PM Ion-Trap Magnet Required
Silverama Types for Black-and-White TV										
5TP4[d]	Yes	6.3/600	● G	50	E	1.2	12.125	12C	29500	No
7JP4	No	6.3/600	● G	(e)	E	3	14.875	14R	6500	No
8DP4	No	6.3/600	■ G	90	E	3	10.750	12AB	9000	Yes
9WP4	Yes	12/75	■ G[m]	90	E	3.1	8.28	7GR	12000	No
10BP4A	No	6.3/600	● G	55	M	10	18.000	12N	13000	Yes
10FP4A	Yes	6.3/600	● G	55	M	10	18.000	12N	13000	No
11CP4	Yes	6.3/450	■ G	110	E	4	9.188	8HR	15000	No
11GP4	Yes	6.3/450	■ G[u]	110	E		9.035	8HR	15000	No
11HP4A	Yes	6.3/450	■ G[m]	110	E	4	9.188	8HR	15000	No
12BNP4A	Yes	6.3/450	■ G[m]	110	E	5	9.598	8HR	16000	No
12CNP4	Yes	4.2/450	■ G[m]	110	E	5.5	9.54	7GR	14000	No
12KP4A	Yes	6.3/600	● G	55	M	12	18.000	12N	13000	No
14ATP4	Yes	8.4/450	■ G	90	E	8.5	13.500	12L	15500	No
14WP4	Yes	6.3/600	■ G	90	E	8.5	13.500	12L	15500	No
16AYP4	Yes	6.3/450	■ G	114	E	8.5	10.561	8HR	20000	No
16BGP4	Yes	6.3/450	■ G[m]	114	E	9.5	10.811	8HR	20000	No
16CHP4A[f]	Yes	6.3/450	■ G[m]	114	E	9.5	10.811	8HR	20000	No
16DP4A	No	6.3/600	● G	60	M	15	21.000	12D	16500	Yes
16CMP4A	Yes	6.3/450	■ G[m]	114	E	9.5	10.811	8HR	18000	No
16LP4A	No	6.3/600	● G	52	M	14.5	22.625	12N	15500	Yes
16RP4B	Yes	6.3/600	■ G	70	M	16	19.125	12N	17500	No
16TP4	No	6.3/600	■ G	70	M	16	18.500	12N	15500	Yes
16WP4A	No	6.3/600	● G	70	M	16.5	18.125	12N	17500	Yes
17BJP4	Yes	6.3/600	■ G	90	E	15	15.000	12L	17500	No
17BP4D	Yes	6.3/600	■ G	70	M	18	19.562	12N	17500	No
17CDP4	Yes	8.4/450	■ G	110	E	10	12.812	8HR	17500	No
17CFP4	Yes	6.3/600	■ G	90	E	10	15.375	12L	17500	No
17CP4	No	6.3/600	■ M[k]	70	M	10	19.000	12D	17500	Yes
17CSP4	Yes	6.3/600	■ G	110	E	10	12.625	7FA	17500	No
17CYP4	Yes	6.3/600	■ G	90	E	10	14.375	12L	17500	No
17DAP4	Yes	2.68/450	■ G	110	E	10	10.875	8JK	17500	No
17DKP4	Yes	6.3/600	■ G	110	E	10	10.938	8JR	23000	No
17DQP4[f]	Yes	6.3/450	■ G	110	E	10	12.375	7FA	17500	No
17DRP4[g]	Yes	2.68/450	■ G	110	E	10	11.000	8JK	17500	No
17DSP4	Yes	6.3/600	■ G	110	E	10	11.438	8HR	20000	No
17DXP4	Yes	6.30/450	■ G	110	E	10	10.938	8JR	17500	No
17EFP4	Yes	6.30/450	■ G	110	E	10	11.438	8HR	20000	No
17EMP4	Yes	6.3/450	■ G[m]	114	E		11.562	8HR	22000	No
17HP4C	Yes	6.3/600	■ G	70	E	18	19.562	12L	17500	No
17LP4B	Yes	6.3/600	■ G[h]	70	E	19	19.562	12L	17500	No
17QP4B	Yes	6.3/600	■ G[h]	70	M	19	19.562	12N	20000	No
17TP4	No	6.3/600	■ M[k]	70	E	10	19.312	12M	17500	Yes
19ABP4	Yes	2.68/450	■ G	114	E	14	11.125	8JK	20000	No
19AHP4	Yes	6.3/450	■ G	114	E	13.5	11.625	8HR	17500	No
19AJP4[f]	Yes	6.3/450	■ G	114	E	14	11.625	7FA	20000	No

*** Basing diagrams for RCA picture tubes are shown on pages 534 and 535.**

RCA PICTURE TUBE CHARACTERISTICS CHART (Cont'd)

RCA Type	Aluminized Screen	Heater Volts/mA	Envelope[a]	Greatest Deflection Angle[b] (Approx.) Degrees	Focusing Method	Approx. Tube Weight Pounds	Maximum Over-all Length Inches	Basing*	Design Maximum Anode[c] Volts	PM Ion-Trap Magnet Required
Silverama Types for Black-and-White TV										
19AUP4	Yes	6.3/600	■ G[j,k]	114	E	18.5	11.938	8HR	20000	No
19AVP4	Yes	6.3/600	■ G	114	E	14	11.625	8HR	23000	No
19AYP4	Yes	6.3/450	■ G	114	E	14	11.625	8HR	23000	No
19BDP4[f]	Yes	6.3/600	■ G	92	E	15	15.625	12L	20000	No
19BTP4	Yes	6.3/600	■ G	114	E	14	11.062	8JR	23000	No
19CHP4[f]	Yes	6.3/600	■ G	114	E	14	11.875	8HR	20000	No
19CMP4[f]	Yes	6.3/450	■ G	114	E	14	11.875	8HR	20000	No
19CXP4[f]	Yes	6.3/600	■ G	114	E	14	11.875	7FA	20000	No
19DQP4	Yes	6.3/450	■ G[m]	114	E	15	11.875	8HR	23000	No
19DRP4	Yes	6.3/600	■ G[m]	114	E	15	11.875	8HR	23000	No
19DSP4[f]	Yes	6.3/600	■ G[m]	114	E	15	11.875	8HR	20000	No
19EBP4	Yes	6.3/600	■ G[u]	114	E		11.875	8HR	23000	No
19EGP4[f]	Yes	6.3/450	■ G[u]	114	E		11.875	8HR	21000	No
19ENP4A	Yes	6.3/450	■ G[m]	114	E	15	11.875	8HR	21000	No
19FEP4B[f]	Yes	6.3/450	■ G[m]	114	E	15	11.875	8HR	23500	No
19FLP4	Yes	6.3/450	■ G[m]	114	E	15	11.625	8HR	23000	No
20DP4D	Yes	6.3/600	■ G	70	M	30	22.125	12N	20000	No
20HP4E	Yes	6.3/600	■ G	70	E	30	22.125	12L	17500	No
20SP4	Yes	6.3/450	■ G[m]	114	E	16.5	12.519	8HR	23000	No
20TP4	Yes	6.3/450	■ G[m]	114	E	16.5	12.519	8HR	23000	No
21AMP4B	Yes	6.3/600	■ G	90	M	24	20.375	12N	20000	No
21AVP4C	Yes	6.3/600	■ G	72	E	24	23.406	12L	22000	No
21AWP4A	Yes	6.3/600	■ G	72	M	24	23.406	12N	20000	No
21CBP4A	Yes	6.3/600	■ G	90	E	24	18.375	12L	22000	No
21CQP4	Yes	6.3/600	■ G	110	E	20	14.812	7FA	20000	No
21DEP4A	Yes	6.3/600	■ G	110	E	20	15.000	8HR	22000	No
21DFP4	Yes	6.3/600	■ G	110	E	23	14.750	8HR	20000	No
21DHP4	Yes	6.3/450	■ G	110	E	20	15.000	8HR	20000	No
21DLP4	Yes	6.3/600	■ G	90	E	24	17.375	12L	22000	No
21DSP4[f]	Yes	6.3/600	■ G	90	E	24	18.375	12L	22000	No
21EP4C	Yes	6.3/600	■ G[h]	70	M	29	23.406	12N	20000	No
21EQP4	Yes	6.3/600	■ G	110	E	23	12.875	8JR	20000	No
21FAP4	Yes	6.3/600	■ G	110	E	20	13.125	8JR	22000	No
21FDP4	Yes	6.3/600	■ G	110	E	20	13.375	8KW	20000	No
21FP4D	Yes	6.3/600	■ G[h]	70	E	29	23.406	12L	20000	No
21FVP4	Yes	6.3/450	■ G[m]	114	E	19	12.937	8HR	23000	No
21GAP4A	Yes	6.3/450	■ G	114	E	19	12.937	8HR	23500	No
21WP4B	Yes	6.3/600	■ G	70	M	24	22.812	12N	20000	No
21XP4B	Yes	6.3/600	■ G	70	E	24	22.812	12L	20000	No
21YP4B	Yes	6.3/600	■ G	70	E	24	23.406	12L	20000	No
21ZP4C	Yes	6.3/600	■ G	70	M	24	23.406	12N	20000	No
23AHP4	Yes	6.3/600	■ G	92	E	25	18.375	12L	22000	No
23ARP4	Yes	6.3/600	■ G	110	E	25	15.156	8HR	22000	No
23ASP4	Yes	6.3/600	■ G	92	E	25	17.375	12L	22000	No
23BGP4[f]	Yes	6.3/600	■ G[j]	110	E	33	15.562	8HR	22000	No

* Basing diagrams for RCA picture tubes are shown on pages 534 and 535.

RCA PICTURE TUBE CHARACTERISTICS CHART (Cont'd)

RCA Type	Aluminized Screen	Heater Volts/mA	Envelope[a]	Greatest Deflection Angle[b] (Approx.) Degrees	Focusing Method	Approx. Tube Weight Pounds	Maximum Over-all Length Inches	Basing*	Design Maximum Anode[c] Volts	PM Ion-Trap Magnet Required
Silverama Types for Black-and-White TV										
23BJP4[f]	Yes	6.3/600	■ G	92	E	27	18.500	12L	25000	No
23BLP4[f]	Yes	6.3/600	■ G[j,k]	92	E	35	18.875	12L	25000	No
23BQP4	Yes	6.3/450	■ G[j]	110	E	33	15.562	8HR	23000	No
23CBP4	Yes	6.3/450	■ G[j,k]	110	E	33	15.562	8HR	23000	No
23CGP4	Yes	6.3/450	■ G	92	E	27	18.375	12L	22000	No
23CP4	Yes	6.3/600	■ G[j]	110	E	33	15.562	8HR	22000	No
23CQP4	Yes	6.3/450	■ G	114	E	24	14.062	8HR	23500	No
23DAP4[f]	Yes	6.3/600	■ G	94	E	27	17.391	8HR	23000	No
23DBP4[f]	Yes	6.3/600	■ G	110	E	25	15.156	8HR	22000	No
23EKP4	Yes	6.3/450	■ G[m]	92	E	29	18.375	12L	25000	No
23ENP4[f]	Yes	6.3/600	■ G[m]	92	E	29	18.500	12L	25000	No
23EP4[f]	Yes	6.3/600	■ G[j]	110	E	33	15.562	8KP	22000	No
23EQP4	Yes	6.3/450	■ G[m]	114	E	28	14.812	8HR	23000	No
23ETP4	Yes	6.3/600	■ G[m]	110	E	28	15.156	8HR	23000	No
23EYP4	Yes	6.3/600	■ G[m]	92	E		18.500	12L	25000	No
23EZP4	Yes	6.3/450	■ G[m]	94	E	30	17.390	8HR	23500	No
23FBP4[f]	Yes	6.3/600	■ G[k,m]	92	E	29	18.500	12L	25000	No
23FP4A	Yes	6.3/600	■ G	114	E	24	14.062	8HR	23500	No
23FRP4[f]	Yes	6.3/450	■ G[u]	110	E	29	14.531	8HR	23000	No
23FSP4	Yes	6.3/600	■ G[u]	110	E	29	15.156	8HR	23000	No
23GJP4A	Yes	6.3/450	■ G[m]	110	E	28	14.531	8HR	23000	No
23GSP4	Yes	6.3/600	■ G[m]	110	E	28	15.156	8HR	23000	No
23HFP4A	Yes	6.3/450	■ G[m]	110	E	28	15.156	8HR	23000	No
23HGP4	Yes	6.3/450	■ G[m]	110	E	28.5	15.156	8HR	23000	No
23HUP4A[f]	Yes	6.3/450	■ G[m]	110	E	28	14.656	8HR	23500	No
23HWP4A	Yes	6.3/450	■ G[m]	110	E	28	15.156	8HR	22000	No
23JP4[f]	Yes	6.3/450	■ G[j]	110	E	33	15.875	7FA	22000	No
23NP4[f]	Yes	6.3/600	■ G	114	E	24	14.812	8HR	22000	No
23YP4	Yes	6.3/600	■ G[j]	92	E	35	18.750	12L	22000	No
24AEP4	Yes	6.3/600	■ G	90	E	32.5	19.500	12L	22000	No
24AHP4	Yes	6.3/600	■ G	110	E	26.5	16.188	8HR	22000	No
24ATP4[f]	Yes	6.3/600	■ G	90	E	32.5	19.500	12L	22000	No
24AUP4	Yes	6.3/600	■ G	90	E	32.5	18.500	12L	22000	No
24BAP4[f]	Yes	6.3/600	■ G	110	E	26.5	16.188	8HR	22000	No
24BEP4	Yes	6.3/600	■ G	110	E	26.5	15.125	8KW	20000	No
24CP4B	Yes	6.3/600	■ G	90	M	32.5	21.500	12N	22000	No
27RP4A	Yes	6.3/600	■ G	90	M	44	23.438	12N	22000	Yes
Color Picture Tubes										
15GP22[n]	Yes	6.3/1800[p]	● G	45	E	25	26.125	20A	22000	No
15LP22	Yes	6.3/900	■ G[k,q]	90	E	12.5	15.566	14BH	22500	No
15NP22	Yes	6.3/900	■ G[m]	90		11.3	15.375	14BH	22500	No
19EYP22[t]	Yes	6.3/800[r]	■ G[q,k]	90	E	24	18.423	14BE	27500	No
19GVP22	Yes	6.3/900	■ G	90	E	21	18.231	14BE	27500	No
19GWP22	Yes	6.3/900	■ G[k,q]	90	E	24	18.423	14BE	27500	No

* Basing diagrams for RCA picture tubes are shown on pages 534 and 535.

RCA PICTURE TUBE CHARACTERISTICS CHART (Cont'd)

RCA Type	Aluminized Screen	Heater Volts/mA	Envelope[a]	Greatest Deflection Angle[b] (Approx.) Degrees	Focusing Method	Approx. Tube Weight Pounds	Maximum Over-all Length Inches	Basing*	Design Maximum Anode[c] Volts	PM Ion-Trap Magnet Required
Color Picture Tubes										
21AXP22A	Yes	6.3/1800[p]	● M	70	E	28	25.312	14AH	27500	No
21CYP22A	Yes	6.3/1800[p]	● G	70	E	36.5	25.406	14AL	27500	No
21FBP22	Yes	6.3/1800[p]	● G	70	E	36.5	25.406	14AU	27500	No
21FJP22	Yes	6.3/1800[p]	● G[kq]	70	E	41	25.594	14AU	27500	No
21GUP22	Yes	6.3/1900	● G	70	E	36.5	25.406	14AU	27500	No
21GVP22	Yes	6.3/1900	● G[k,q]	70	E	41	25.594	14BE	27500	No
22JP22	Yes	6.3/900	■ G[k,q]	90	E	32.5	19.579	14BE	27500	No
23EGP22A	Yes	6.3/1350	■ G[k,q]	92	E		20.094	14BE	27500	No
25AP22A[t]	Yes	6.3/800[r]	■ G[kq]	90	E	42	21.299	14BE	27500	No
25BP22A[t]	Yes	6.3/800[r]	■ G	90	E	37	21.107	14BE	27500	No
25XP22	Yes	6.3/900	■ G[k,q]	90	E	42	21.299	14BE	27500	No
25YP22	Yes	6.3/900	■ G	90	E	37	21.107	14BE	27500	No
Test Picture Tubes										
5AXP4	No	6.3/600	● G	53	E[e]	1.5	11.000	12S	20000	No
8XP4	Yes	6.3/600	■ G	90	E[e]	3	11.750	12S	22000	No
8YP4	Yes	6.3/600	■ G	110	E[e]	2	9.000	7FG	22000	No
1828P22[v]	Yes	6.3/1800[p]	● G	70	E	36.5	25.406	14AU	27500	No
1830P22	Yes	6.3/900	■ G[q]	90	E	24	18.423	14BE	27500	No

* Basing diagrams for RCA picture tubes are shown on pages 534 and 535.

RCA VOLTAGE-REGULATOR AND VOLTAGE-REFERENCE TUBES

These tubes are designed for voltage-regulation requiring a relatively constant dc output voltage across a load independent of load and line-voltage variations.

VOLTAGE-REGULATOR TUBES †

RCA Type	DC Operating Volts	DC Operating Current Range mA	Anode Starting Volts	Anode Starting (mA)	Regulation Volts	Ambient Operating Temperature Range (°C)	Max Length (in)	Max Diameter (in)	Diagram Terminal
0A2	150	5 to 30	185	75	6	−55 to +90	2-5/8	3/4	5BO
0A3	75	5 to 40	105	100	6.5	−55 to +90	4-1/8	1-9/16	4AJ
0A3A	75	5 to 40	105	100	6.5	−55 to +90	3-1/16	1-9/32	4AJ
0B2	105	5 to 30	133	75	4	−55 to +90	2-5/8	3/4	5BO
0C2	75	5 to 30	115	75	4.5	−55 to +90	2-5/8	3/4	5BO
0C3	105	5 to 40	133	100	4	−55 to +90	4-1/8	1-9/16	4AJ
0C3A	105	5 to 40	127	100	4	−55 to +90	3-1/16	1-9/32	4AJ
0D3	150	5 to 40	185	100	5.5	−55 to +90	4-1/8	1-9/16	4AJ
0D3A	150	5 to 40	180	100	5.5	−55 to +90	3-1/16	1-9/32	4AJ
991	59	0.4 to 2	87	—	8	—	1-9/16	5/8	*
6073	150	5 to 30	185	75	6	−55 to +90	2-5/8	3/4	5BO
6073/0A2	150	5 to 30	185	75	6	−55 to +90	2-5/8	3/4	5BO
6074	105	5 to 30	133	75	4	−55 to +90	2-5/8	3/4	5BO
6074/0B2	105	5 to 30	133	75	4	−55 to +90	2-5/8	3/4	5BO
6626/0A2WA	150	5 to 30	165	75	5	−55 to +90	2-5/8	3/4	5BO

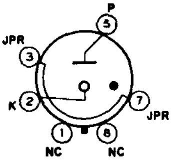
VOLTAGE-REFERENCE TUBES †
(for exceptional voltage stability)

RCA Type	DC Operating Volts	DC Operating Current Range mA	Anode Starting Volts	Anode Starting (mA)	Regulation Volts	Ambient Operating Temperature Range (°C)	Max Length (in)	Max Diameter (in)	Diagram Terminal
5651	87	1.5 to 3.5	115	—	3	−55 to +90	2-1/8	3/4	5BO
5651A	85.5	1.5 to 3.5	115	—	3	−55 to +90	2-1/8	3/4	5BO
5783	86	1.5 to 3.5	115	—	3	−55 to +90	1-5/8	0.4	5783

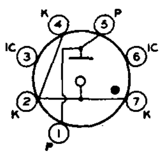
SERIES-VOLTAGE-REGULATOR TUBES **
(for high-current applications)

RCA Type	Heater Volts	Heater Amperes	DC Plate Volts	DC• Plate Amperes	Plate dissipation (watts)	Amplification• Factor	Plate Resistance• (ohms)	Max Length (in)	Max Diameter (in)	Terminal Diagram
6AS7G	6.3	2.5	250	0.125	13	2	280	5-5/16	2-1/16	8BD
6080	6.3	2.5	250	0.125	13	2	280	4-1/16	1-23/32	8BD
6082	26.5	0.6	250	0.125	13	2	280	4-1/16	1-23/32	8BD
6336A	6.3	5	400	0.4	30	2.7	200	4-3/4	2.07	8BD

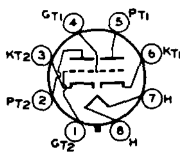
** Indirectly-heated-cathode, vacuum, low-mu twin triodes.
† Cold-cathode, glow-discharge types.
• Each triode unit.
* Bayonet candelabra double contact base.



4AJ



5BO



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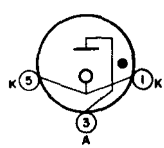

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Electron Tube Testing

THE electron-tube user—service man, experimenter, or nontechnical radio listener—is interested in knowing the condition of his tubes, since they govern the performance of the device in which they are used. In order to determine the condition of a tube, some method of test is necessary. Because the operating capabilities and design features of a tube are indicated and described by its electrical characteristics, a tube is tested by measuring its characteristics and comparing them with values established as standard for that type. Tubes which read abnormally high with respect to the standard for the type are subject to criticism just the same as tubes which are too low.

Certain practical limitations are placed on the accuracy with which a tube test can be correlated with actual tube performance. These limitations make it impractical for the service man and dealer to employ complex and costly testing equipment having laboratory accuracy. Because the accuracy of the tube-testing device need be no greater than the accuracy of the correlation between test results and receiver performance, and since certain fundamental characteristics are virtually fixed by the manufacturing technique of leading tube manufacturers, it is possible to employ a relatively simple test in order to determine the serviceability of a tube.

In view of these factors, dealers and service men will find it economically expedient to obtain adequate accuracy and simplicity of operation by employing a device which indicates the status of a single characteristic. Whether the tube is satisfactory or unsatisfactory is judged from the test result of this single characteristic. Consequently, it is very desirable that the characteristic selected for the test be one which is truly representative of the tube's over-all condition.

The following information and circuits are given to describe and illustrate general theoretical and practical tube-tester considerations and not to provide information on the construction of a home-made tube tester. In addition to the problem of determining what tube characteristic is most representative of performance capabilities in all types of receivers, the designer of a home-made tester faces the difficult problem of determining satisfactory limits for his particular tester. Getting information of this nature, if it is to be accurate and useful, is a big job. It requires the testing of many tubes of each type, testing of many types, and correlation of the data with performance in many kinds of equipment.

Short-Circuit Test

The fundamental circuit of a short-circuit tester is shown in Fig. 129. Although this circuit is suitable for tet-

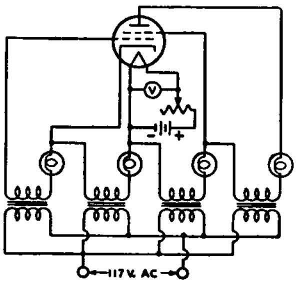


Fig. 129—Fundamental circuit of a short-circuit tester.

rodes and types having less than four electrodes, tubes of more electrodes may be tested by adding more indicator lamps to the circuit. Voltages are applied between the various electrodes with lamps in series with the electrode leads. The value of the voltages applied will depend on the type of tube being tested and its maximum ratings. Any two shorted electrodes complete a circuit and light one or more lamps. Since two electrodes may be just touching to give a high-resistance short, it is desirable that the indicating lamps operate on very low current. It is also desirable to maintain the filament or heater of the tube at its operating temperature during the short-circuit test, because short-circuits in a tube may sometimes occur only when the electrodes are heated. However, a short-circuit tester having too high a sensitivity may indicate very-high-resistance shorts that do not adversely affect tube operation.

Selection of a Suitable Characteristic for Test

Some characteristics of a tube are far more important in determining its operating worth than are others. The cost of building a device to measure any one of the more important characteristics may be considerably higher than that of a device which measures a less representative characteristic. Consequently, three methods of test will be discussed, ranging from relatively simple and inexpensive equipment to more elaborate, more accurate, and more costly devices.

An **emission test** is perhaps the simplest method of indicating a tube's condition. (Refer to *Diodes*, in **Electrons, Electrodes, and Electron Tubes** section, for a discussion of electron emission.) Since emission falls off as the tube wears out, low emission is indicative of the end of tube serviceability. However, the emission test is subject to limitations because it tests the tube under static conditions and does not take into account the actual operation of the tube. On the one hand, coated filaments, or cathodes, often develop active spots from which the emission is so great that the relatively small grid area adjacent to these spots cannot control the electron stream. Under these conditions, the total emission may indicate the tube to be normal although the tube is unsatisfactory. On the other hand, coated types of filaments are capable of such large emission that the tube will often operate satisfactorily after the emission has fallen far below the original value.

Fig. 130 shows the fundamental circuit diagram for an emission test. All of the electrodes of the tube, except the cathode, are connected to the plate. The filament, or heater, is operated at rated voltage; after the tube has reached con-

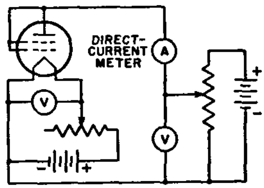


Fig. 130—Fundamental circuit of an emission tester

stant temperature, a low positive voltage is applied to the plate and the electron emission is read on the meter. Readings which are well below the average for a particular tube type indicate that the total number of available electrons has been so reduced that the tube is no longer able to function properly.

A **transconductance test** takes into account a fundamental operating principle of the tube. (This fact will be seen from the definition of transconductance in the Section on **Electron Tube Characteristics.**) It follows that transconductance tests, when properly made, permit better correlation between test results and actual performance than does a straight emission test.

There are two forms of transconductance test which can be utilized in a tube tester. In the first form (illustrated by Fig. 131 giving a fundamental circuit with a tetrode under test), appropriate operating voltages are applied to the electrodes of the tube. A plate current depending upon the electrode voltages will then be indicated by the meter. If the bias on the grid is then shifted by the application of a different grid voltage, a new plate-current reading is obtained. The difference between the two plate-current readings is indicative of the transconductance of the tube. This

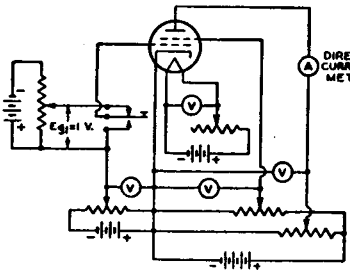


Fig. 131—Fundamental circuit of a transconductance tester using the "grid-shift" method.

method of transconductance testing is commonly called the "grid-shift" method, and depends on readings under static conditions. The fact that this form of test is made under static conditions imposes limitations not encountered in the second form of test made under dynamic conditions.

The dynamic transconductance test illustrated in Fig. 132 gives a fundamental circuit with a tetrode under test. This method is superior to the static transconductance test in that ac voltage

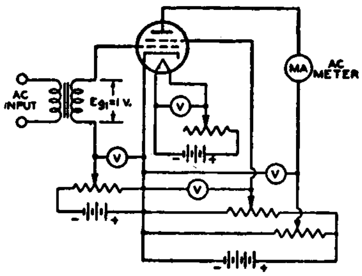


Fig. 132—Fundamental circuit of a dynamic transconductance tester.

is applied to the grid. Thus, the tube is tested under conditions which approximate actual operating conditions. The alternating component of the plate current is read by means of an ac ammeter of the dynamometer type. The transconductance of the tube is equal to the ac plate current divided by the input-signal voltage. If a one-volt rms signal is applied to the grid, the plate-current-meter reading in milliamperes multiplied by one thousand is the value of transconductance in micromhos.

The **power-output test** probably gives the best correlation between test results and actual operating performance of a tube. In the case of voltage amplifiers, the power output is indicative of the amplification and output voltages obtainable from the tube. In the case of power-output tubes, the performance of the tube is closely checked. Consequently, although more complicated to set up, the power-output test will give closer correlation with actual performance than any other single test.

Fig. 133 shows the fundamental circuit of a power-output test for class A operation of tubes. The diagram illustrates the method for a pentode. The ac output voltage developed across the

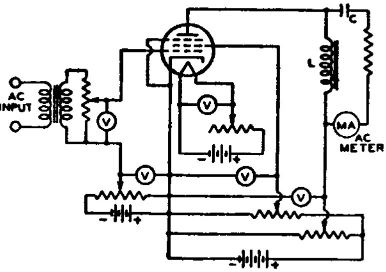


Fig. 133—Fundamental circuit of a power-output tester for class A operation of tubes.

plate-load impedance (L) is indicated by the current meter. The current meter is isolated as far as the dc plate current is concerned by the capacitor (C). The power output can be calculated from the current reading and known load resistance. In this way, it is possible to determine the operating condition of the tube quite accurately.

Fig. 134 shows the fundamental circuit of a power-output test for class B operation of tubes. With ac voltage applied to the grid of the tube, the current in the plate circuit is read on a dc milliammeter. The power output of the tube is approximately equal to:

$$(I_b^2 \times R_L)/0.405,$$

where P_o is the power output in watts, I_b is the dc current in amperes, and R_L is the load resistance in ohms.

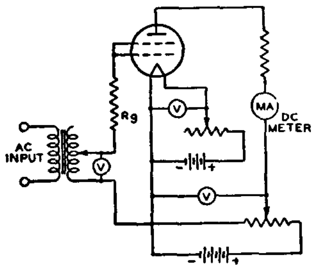


Fig. 134—Fundamental circuit of a power-output tester for class B operation of tubes.

Essential Tube-Tester Requirements

1. The tester should provide for making a short-circuit test before measurement of the tube's characteristics.

2. It is important that some means of controlling the voltages applied to the electrodes of the tube be provided. If the tester is ac operated, a line-voltage control permits the supply of proper electrode voltages.

3. It is essential that the rated voltage applied to the filament or heater be maintained accurately.

4. It is suggested that the characteristics test follow one of the methods described. The method selected and the quality of the parts used in the test will depend upon the user's requirements.

Tube-Tester Limitations

A tube-testing device can only indicate the difference between a given tube's characteristics and those which are standard for that particular type. Since the operating conditions imposed upon a tube of a given type may vary within wide limits, it is impossible for a tube-testing device to evaluate tubes in terms of performance capabilities for all applications. The tube tester, therefore, cannot be looked upon as a final authority in determining whether or not a tube is always satisfactory. Actual operating test in the equipment in which the tube is to be used will give the best possible indication of a tube's worth.

Resistance-Coupled Amplifiers

RESISTANCE-COUPLED, audio-frequency voltage amplifiers utilize simple components and are capable of providing essentially uniform amplification over a relatively wide frequency range.

Suitable Tubes

In this section, data are given for over 45 types of tubes suitable for use in resistance-coupled circuits. These types include low- and high-mu triodes, twin triodes, triode-connected pentodes, and pentodes. The accompanying key to tube types will assist in locating the appropriate data chart.

Circuit Advantages

For most of the types shown, the data pertain to operation with cathode bias; for all of the pentodes, the data pertain to operation with series screen-grid resistor. The use of a cathode-bias resistor where feasible and a series screen-grid resistor where applicable offers several advantages over fixed-voltage operation.

The advantages are: (1) effects of possible tube differences are minimized; (2) operation over a wide range of plate-supply voltages without appreciable change in gain is feasible; (3) the low frequency at which the amplifier cuts off is easily changed; and (4) tendency toward motorboating is minimized.

Number of Stages

These advantages can be enhanced by the addition of suitable decoupling filters in the plate supply of each stage of a multi-stage amplifier. With proper filters, three or more amplifier stages can be operated from a single power-supply unit of conventional design with-

Type	Chart No.	Type	Chart No.
3AU6	2	6EU7	9
3AV6	9	6FQ7/	
3BC5/		6CG7	8
3CE5	11	6SL7GT	5
3CB6	11	6SN7GTB	8
4AU6	2	6T8A	5
4AV6	9	7AU7	3
4BQ7A	10	8CN7	5
4BZ7	10	8FQ7/	
4CB6	11	8CG7	8
5BK7A	10	9AU7	3
5BQ7A	10	12AT6	5
5T8	5	12AT7	4
6AB4	4	12AU6	2
6AG5	11	12AU7A	3
6AT6	5	12AV6	9
6AU6A	2	12AX7A	9
6AV6	9	12AY7	1
6BC5	11	12SL7GT	5
6BK7B	10	12SN7GTA	8
6BQ7A	10	19T8	5
6BZ7	10	20EZ7	9
6C4	3	5879P	6
6CB6	11	5879T	7
6CB6A	11	7025	9
6CN7	5	7199P	12
		7199T	13

T = Triode Unit or Triode Connection
P = Pentode Unit or Pentode Connection

KEY TO CHARTS

out encountering any difficulties due to coupling through the power unit. When decoupling filters are not used, not more than two stages should be operated from a single power-supply unit.

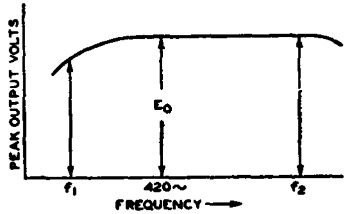
Symbols Used in Resistance-Coupled Amplifier Charts

C = Blocking Capacitor (μf).
C_k = Cathode Bypass Capacitor (μf).
C_{g2} = Screen-Grid Bypass Capacitor (μf).
E_{bb} = Plate-Supply Voltage (volts). Voltage at plate equals plate-supply voltage minus drop in R_p and R_k.
R_k = Cathode Resistor (ohms).
R_{g2} = Screen-Grid Resistor (megohms).
R_g = Grid Resistor (megohms) for following stage.
R_p = Plate Resistor (megohms).
V.G. = Voltage Gain.
E_o = Output Voltage (peak volts). This voltage is obtained across R_g (for following stage) at any frequency within the flat region of the output vs. frequency curve, and is for the condition where the signal level is adequate to swing the grid of the resistance-coupled amplifier tube to the point where its grid starts to draw current.

Note: The listed values for E_o are the peak output voltages available when the grid is driven from a low-impedance source. The listed values for the cathode resistors are optimum for any signal source. With a high-impedance source, protection against severe distortion and loss of gain due to input loading may be obtained by the use of a coupling capacitor connected directly to the input grid and a high-value resistor connected between the grid and ground.

General Circuit Considerations

In the discussions which follow, the frequency (f_2) is that value at which the high-frequency response begins to fall off. The frequency (f_1) is that value at which the low-frequency response drops below a satisfactory value, as discussed below. A variation of 10 per cent in values of resistors and capacitors has only slight effect on performance. One-half-watt resistors are usually suitable for R_{g2}, R_g, R_p, and R_k resistors. Capacitors C and C_{g2} should have a working voltage equal to or greater than E_{bb}. Capacitor C_k may have a low working voltage in the order of 10 to 25 volts.



Triode Amplifier
Heater-Cathode Type

Capacitors C and C_k have been chosen to give an output voltage equal to 0.8 E_o for a frequency (f_1) of 100 Hz. For any other value of f_1, multiply values of C and C_k by $100/f_1$. In

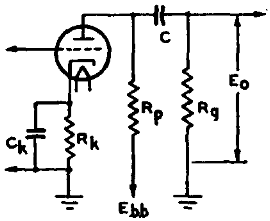


Diagram No. 1

the case of capacitor C_k, the values shown in the charts are for an amplifier with dc heater excitation; when ac is used, depending on the character of the associated circuit, the gain, and the value of f_1, it may be necessary to increase the value of C_k to minimize hum disturbances. It may be desirable to operate the heater at a positive voltage of from 15 to 40 volts with respect to the cathode. The voltage output at f_1 of "n" like stages equals $(0.8)^n \times E_o$, where E_o is the peak output voltage of final stage. For an amplifier of typical construction, the value of f_2 is well above the audio-frequency range for any value of R_p.

Pentode Amplifier

Heater-Cathode Type

Capacitors C, C_k, and C_{g2} have been chosen to give an output voltage equal to $0.7 \times E_o$ for a frequency (f_1) of 100 cycles. For any other value of f_1, multiply values of C, C_k, and C_{g2} by $100/f_1$. In the case of capacitor C_k, the values shown in the charts are for an amplifier with dc heater excitation; when ac is used, depending on the character of the associated circuits, the voltage gain, and the value of f_1, it may be necessary to increase the value of C_k to minimize hum disturbances. It may be desirable to operate the heater at a positive voltage of from 15 to 40 volts with respect to the cathode. The voltage output at f_1 for "n" like stages equals $(0.7)^n \times E_o$ where E_o is peak output voltage of final stage. For an amplifier of typical construction, and for R_p values of 0.1, 0.25, and 0.5 megohm, approximate values of f_2 are 20000, 10000, and 5000 Hz, respectively.

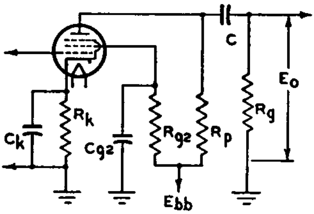


Diagram No. 2

1

12AY7•

See Circuit Diagram 1

E_{bb}	R_p	R_g	R_{g2}	R_k	C_{g2}	C_k	C	E_o*	V.G.
	0.1	0.24	—	1800	—	—	—	13	24
90	0.24	0.51	—	3700	—	—	—	14	26
	0.51	1.0	—	7800	—	—	—	16	27
	0.1	0.24	—	1300	—	—	—	31	27
180	0.24	0.51	—	2800	—	—	—	33	29
	0.51	1.0	—	5700	—	—	—	33	30
	0.1	0.24	—	1200	—	—	—	58	28
300	0.24	0.51	—	2300	—	—	—	30	30
	0.51	1.0	—	4800	—	—	—	56	31

• One triode unit. * Peak volts.

▲ Coupling capacitors should be selected to give desired frequency response. Cathode resistors should be adequately bypassed.

E_{bb}	R_p	R_g	R_{g2}	R_k	C_{g2}	C_k	C	E_o*	V.G.
	0.22	0.22	0.340	2700	0.057	5.8	0.0081	16	79
	0.22	0.47	0.370	2900	0.050	5.4	0.0055	22	104
	0.22	1.0	0.380	3100	0.050	5.3	0.0034	25	125
	0.47	0.47	1.00	6000	0.027	2.8	0.0042	13	105
90	0.47	1.0	1.00	6200	0.023	2.7	0.0027	17	137
	0.47	2.2	1.00	6300	0.027	2.8	0.0019	25	161
	1.0	1.0	1.90	10800	0.017	1.7	0.0025	10	139
	1.0	2.2	2.40	13100	0.017	1.7	0.0017	19	184
	0.22	0.22	0.520	1340	0.059	8.8	0.0081	31	143
	0.22	0.47	0.520	1390	0.059	8.7	0.0053	43	192
	0.22	1.0	0.520	1420	0.059	8.6	0.0032	48	223
	0.47	0.47	1.05	2700	0.039	5.5	0.0041	34	189
180	0.47	1.0	1.15	2880	0.037	5.4	0.0027	43	249
	0.47	2.2	1.20	2960	0.036	5.4	0.0019	50	294
	1.0	1.0	2.40	5500	0.028	3.2	0.0023	33	230
	1.0	2.2	2.70	6000	0.022	2.8	0.0015	40	323
	0.22	0.22	0.530	780	0.077	13.2	0.0082	53	200
	0.22	0.47	0.540	783	0.077	13.2	0.0053	65	270
	0.22	1.0	0.540	800	0.077	13.1	0.0033	74	316
	0.47	0.47	1.15	1590	0.057	8.4	0.0045	56	275
300	0.47	1.0	1.22	1650	0.049	7.4	0.0027	72	357
	0.47	2.2	1.31	1720	0.045	7.2	0.0017	82	418
	1.0	1.0	2.50	3300	0.036	5.3	0.0022	57	352
	1.0	2.2	2.80	3500	0.031	4.2	0.0015	72	466

2

3AU6
4AU6
6AU6A
12AU6

See Circuit Diagram 2

	0.047	0.047	—	1600	—	3.2	0.061	9	10
	0.047	0.1	—	1800	—	2.5	0.033	11	11
	0.047	0.22	—	2000	—	2.0	0.015	14	11
	0.1	0.1	—	3000	—	1.6	0.032	10	11
90	0.1	0.22	—	3800	—	1.1	0.015	15	11
	0.1	0.47	—	4500	—	1.0	0.007	18	11
	0.22	0.22	—	6800	—	0.7	0.015	14	11
	0.22	0.47	—	9500	—	0.5	0.0065	20	11
	0.22	1.0	—	11500	—	0.43	0.0035	24	11
	0.047	0.047	—	920	—	3.9	0.062	20	11
	0.047	0.1	—	1200	—	2.9	0.037	26	12
	0.047	0.22	—	1400	—	2.5	0.016	29	12
	0.1	0.1	—	2000	—	1.9	0.032	24	12
180	0.1	0.22	—	2800	—	1.4	0.016	33	12
	0.1	0.47	—	3600	—	1.1	0.007	40	12
	0.22	0.22	—	5300	—	0.8	0.015	31	12
	0.22	0.47	—	8300	—	0.56	0.007	44	12
	0.22	1.0	—	10000	—	0.48	0.0035	54	12
	0.047	0.047	—	870	—	4.1	0.065	38	12
	0.047	0.1	—	1200	—	3.0	0.034	52	12
	0.047	0.22	—	1500	—	2.4	0.016	68	12
	0.1	0.1	—	1900	—	1.9	0.032	44	12
300	0.1	0.22	—	3000	—	1.3	0.016	68	12
	0.1	0.47	—	4000	—	1.1	0.007	80	12
	0.22	0.22	—	5300	—	0.9	0.015	57	12
	0.22	0.47	—	8800	—	0.52	0.007	82	12
	0.22	1.0	—	11000	—	0.46	0.0035	92	12

3

6C4
7AU7•
9AU7•
12AU7A•

See Circuit Diagram 1

• One triode unit. * Peak volts.

(4)

6AB4
12AT7•

See Circuit Diagram 1

E_{bb}	R_p	R_g	R_{g2}	R_k	C_{g2}	C_k	C	E_o*	V.G.
	0.1	0.1	—	2680	—	2.4	0.026	8	24
	0.1	0.22	—	3060	—	2.00	0.014	11	25
	0.1	0.47	—	3390	—	1.84	0.0074	13	28
	0.22	0.22	—	5500	—	1.33	0.0136	10	25
90	0.22	0.47	—	6300	—	1.01	0.0067	14	28
	0.22	1.0	—	6930	—	0.92	0.0038	15	28
	0.47	0.47	—	10900	—	0.63	0.007	13	26
	0.47	1.0	—	12500	—	0.52	0.0043	14	28
	0.47	2.2	—	13500	—	0.47	0.0031	18	28
	0.1	0.1	—	1407	—	3.6	0.029	20	31
	0.1	0.22	—	1674	—	3.0	0.016	28	33
	0.1	0.47	—	1786	—	2.6	0.0083	31	34
	0.22	0.22	—	2890	—	1.75	0.0140	24	33
180	0.22	0.47	—	3860	—	1.34	0.0077	35	33
	0.22	1.0	—	4660	—	1.14	0.0047	42	33
	0.47	0.47	—	6960	—	0.83	0.0075	31	31
	0.47	1.0	—	8450	—	0.67	0.0046	39	32
	0.47	2.2	—	9600	—	0.55	0.0032	45	32
	0.1	0.1	—	974	—	4.0	0.028	37	34
	0.1	0.22	—	1404	—	3.1	0.015	57	34
	0.1	0.47	—	2169	—	2.5	0.0083	78	33
	0.22	0.22	—	2510	—	1.9	0.015	50	33
300	0.22	0.47	—	4200	—	1.3	0.0074	78	33
	0.22	1.0	—	4950	—	1.1	0.0046	85	32
	0.47	0.47	—	5700	—	0.90	0.0076	57	33
	0.47	1.0	—	8720	—	0.62	0.0041	81	32
	0.47	2.2	—	9700	—	0.57	0.0030	88	32

(5)

5T8
6AT6
6CN7
6SL7GT•
6T8A
8CN7
12AT6
12SL7GT•
19T8

See Circuit Diagram 1

	0.1	0.1	—	4200	—	2.5	0.025	5.4	22
	0.1	0.22	—	4600	—	2.2	0.014	7.5	27
	0.1	0.47	—	4800	—	2.0	0.0065	9.1	30
	0.22	0.22	—	7000	—	1.5	0.013	7.3	30
90	0.22	0.47	—	7800	—	1.3	0.007	10	34
	0.22	1.0	—	8100	—	1.1	0.0035	12	37
	0.47	0.47	—	12000	—	0.83	0.006	10	36
	0.47	1.0	—	14000	—	0.7	0.0035	14	39
	0.47	2.2	—	15000	—	0.6	0.002	16	41
	0.1	0.1	—	1900	—	3.6	0.027	19	30
	0.1	0.22	—	2200	—	3.1	0.014	25	35
	0.1	0.47	—	2500	—	2.8	0.0065	32	37
	0.22	0.22	—	3400	—	2.2	0.014	24	38
180	0.22	0.47	—	4100	—	1.7	0,0065	34	42
	0.22	1.0	—	4600	—	1.5	0.0035	38	44
	0.47	0.47	—	6600	—	1.1	0.0065	29	44
	0.47	1.0	—	8100	—	0.9	0.0035	38	46
	0.47	2.2	—	9100	—	0.8	0.002	43	47
	0.1	0.1	—	1500	—	4.4	0.027	40	34
	0.1	0.22	—	1800	—	3.6	0.014	54	38
	0.1	0.47	—	2100	—	3.0	0.0065	63	41
	0.22	0.22	—	2600	—	2.5	0.013	51	42
300	0.22	0.47	—	3200	—	1.9	0.0065	65	46
	0.22	0.1	—	3700	—	1.6	0.0035	77	48
	0.47	0.47	—	5200	—	1.2	0.006	61	48
	0.47	1.0	—	6300	—	1.0	0.0035	74	50
	0.47	2.2	—	7200	—	0.9	0.002	85	51

• One triode unit. * Peak volts.

⑥ As Pentode: **5879** See Circuit Diagram 2

E_{bb}	R_p	R_g	R_{g2}	R_k	C_{g2}	C_k	C	E_o*	V.G.
90	0.1	0.1	0.35	1700	0.044	4.6	0.020	13	29
	0.1	0.22	0.35	1700	0.046	4.5	0.012	17	39
	0.1	0.47	0.35	1700	0.047	4.4	0.006	20	47
	0.22	0.22	0.80	3000	0.034	3.2	0.010	15	43
	0.22	0.47	0.80	3000	0.035	3.1	0.005	21	59
	0.22	1.0	0.80	3000	0.036	3.0	0.003	24	67
	0.47	0.47	1.9	7000	0.021	1.8	0.005	21	59
	0.47	1.0	1.9	7000	0.022	1.7	0.003	25	75
	0.47	2.2	1.9	7000	0.023	1.7	0.002	28	87
180	0.1	0.1	0.35	700	0.060	7.4	0.020	24	39
	0.1	0.22	0.35	700	0.062	7.3	0.012	28	56
	0.1	0.47	0.35	700	0.064	7.2	0.006	33	65
	0.22	0.22	0.80	1200	0.045	5.5	0.010	24	65
	0.22	0.47	0.80	1200	0.046	5.3	0.005	31	87
	0.22	1.0	0.80	1200	0.048	5.2	0.003	34	101
	0.47	0.47	1.9	2500	0.033	3.5	0.005	27	98
	0.47	1.0	1.9	2500	0.034	3.4	0.003	32	122
	0.47	2.2	1.9	2500	0.035	3.3	0.002	37	140
300	0.1	0.1	0.35	300	0.075	10.8	0.020	25	51
	0.1	0.22	0.35	300	0.077	10.6	0.012	32	68
	0.1	0.47	0.35	300	0.080	10.5	0.006	35	83
	0.22	0.22	0.80	600	0.056	7.9	0.010	28	81
	0.22	0.47	0.80	600	0.057	7.5	0.005	37	109
	0.22	1.0	0.80	600	0.058	7.4	0.003	41	123
	0.47	0.47	1.3	1200	0.044	5.3	0.005	34	125
	0.47	1.0	1.3	1200	0.046	5.2	0.003	42	152
	0.47	2.2	1.3	1200	0.047	5.1	0.002	48	174

⑦ As Triode: **5879** See Circuit Diagram 1

E_{bb}	R_p	R_g	R_{g2}	R_k	C_{g2}	C_k	C	E_o*	V.G.
90	0.047	0.047	—	1800	—	2.9	0.060	9	10
	0.047	0.1	—	2100	—	2.4	0.033	12	11
	0.047	0.22	—	2200	—	2.3	0.016	14	21
	0.1	0.1	—	3200	—	1.8	0.027	10	12
	0.1	0.22	—	3900	—	1.3	0.015	13	13
	0.1	0.47	—	4300	—	1.0	0.007	16	13
	0.22	0.22	—	6200	—	0.87	0.015	12	13
	0.22	0.47	—	8100	—	0.53	0.006	16	13
	0 22	1.00	—	9000	—	0.49	0.003	19	14
180	0.047	0.047	—	1200	—	3.5	0.063	21	12
	0.047	0.1	—	1600	—	2.6	0.033	29	13
	0.047	0.22	—	1800	—	2.4	0.016	35	13
	0.1	0.1	—	2200	—	1.9	0.031	26	13
	0.1	0.22	—	2900	—	1.35	0.015	33	14
	0.1	0.47	—	3400	—	1.1	0.007	40	14
	0.22	0.22	—	4500	—	0.92	0.015	28	14
	0.22	0.47	—	6400	—	0.61	0.006	39	14
	0.22	1.00	—	8200	—	0.52	0.003	47	14
300	0.047	0.047	—	1100	—	3.9	0.063	42	13
	0.047	0.1	—	1500	—	2.8	0.033	65	13
	0.047	0.22	—	1700	—	2.5	0.016	71	14
	0.1	0.1	—	2000	—	2.1	0.032	45	15
	0.1	0.22	—	3400	—	1.4	0.015	74	15
	0.1	0.47	—	3700	—	1.1	0.007	83	15
	0.1	0.22	—	4300	—	0.97	0.015	50	15
	0.22	0.47	—	7200	—	0.63	0.007	88	15
	0.22	1.00	—	7400	—	0.63	0.003	94	15

* Peak volts

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6FQ7/6CG7
6SN7GTB•
8FQ7/8CG7
12SN7GTA•

See Circuit Diagram 1

E_{bb}	R_p	R_g	R_{g2}	R_k	C_{g2}	C_k	C	E_o*	V.G.
90	0.047	0.047	—	1870	—	3.1	0.063	14	13
	0.047	0.1	—	2230	—	2.5	0.031	18	14
	0.047	0.22	—	2500	—	2.1	0.016	20	14
	0.1	0.1	—	3370	—	1.8	0.034	15	14
	0.1	0.22	—	4100	—	1.3	0.015	20	14
	0.1	0.47	—	4800	—	1.1	0.006	23	15
	0.22	0.22	—	7000	—	0.80	0.013	16	14
	0.22	0.47	—	9100	—	0.65	0.007	22	14
	0.22	1.00	—	10500	—	0.60	0.004	25	15
180	0.047	0.047	—	1500	—	3.6	0.066	33	14
	0.047	0.1	—	1860	—	2.9	0.055	41	14
	0.047	0.22	—	2160	—	2.2	0.015	47	15
	0.1	0.1	—	2750	—	1.8	0.028	35	15
	0.1	0.22	—	3550	—	1.4	0.015	45	15
	0.1	0.47	—	4140	—	1.3	0.007	51	16
	0.22	0.22	—	5150	—	1.0	0.016	36	16
	0.22	0.47	—	7000	—	0.71	0.007	45	16
	0.22	1.00	—	7800	—	0.61	0.004	51	16
300	0.047	0.047	—	1300	—	3.6	0.061	59	14
	0.047	0.1	—	1580	—	3.0	0.032	73	15
	0.047	0.22	—	1800	—	2.5	0.015	83	16
	0.1	0.1	—	2500	—	1.9	0.031	68	16
	0.1	0.22	—	3130	—	1.4	0.014	82	16
	0.1	0.47	—	3900	—	1.2	0.0065	96	16
	0.22	0.22	—	4800	—	0.95	0.015	68	16
	0.22	0.47	—	6500	—	0.69	0.0065	85	16
	0.22	1.00	—	7800	—	0.58	0.0035	96	16

9

3AV6
4AV6
6AV6
6EU7•
12AV6
12AX7A•
20EZ7•
7025•

See Circuit Diagram 1

E_{bb}	R_p	R_g	R_{g2}	R_k	C_{g2}	C_k	C	E_o*	V.G.
90	0.1	0.1	—	4400	—	2.7	0.023	5	29
	0.1	0.22	—	4700	—	2.4	0.013	6	35
	0.1	0.47	—	4800	—	2.3	0.007	8	41
	0.22	0.22	—	7000	—	1.6	0.012	6	39
	0.22	0.47	—	7400	—	1.4	0.006	9	45
	0.22	1.0	—	7600	—	1.3	0.003	11	48
	0.47	0.47	—	12000	—	0.9	0.006	9	48
	0.47	1.0	—	13000	—	0.8	0.003	11	52
	0.47	2.2	—	14000	—	0.7	0.002	13	55
180	0.1	0.1	—	1800	—	4.0	0.025	18	40
	0.1	0.22	—	2000	—	3.5	0.013	25	47
	0.1	0.47	—	2200	—	3.1	0.006	32	52
	0.22	0.22	—	3000	—	2.4	0.012	24	53
	0.22	0.47	—	3500	—	2.1	0.006	34	59
	0.22	1.0	—	3900	—	1.8	0.003	39	63
	0.47	0.47	—	5800	—	1.3	0.006	30	62
	0.47	1.0	—	6700	—	1.1	0.003	39	66
	0.47	2.2	—	7400	—	1.0	0.002	45	68
300	0.1	0.1	—	1300	—	4.6	0.027	43	45
	0.1	0.22	—	1500	—	4.0	0.013	57	52
	0.1	0.47	—	1700	—	3.6	0.006	66	57
	0.22	0.22	—	2200	—	3.0	0.013	54	59
	0.22	0.47	—	2800	—	2.3	0.006	69	65
	0.22	1.0	—	3100	—	2.1	0.003	79	68
	0.47	0.47	—	4300	—	1.6	0.006	62	69
	0.47	1.0	—	5200	—	1.3	0.003	77	73
	0.47	2.2	—	5900	—	1.1	0.002	92	75

• One triode unit. * Peak volts.

E_{bb}	R_p	R_g	R_{g2}	R_k	C_{g2}	C_k	C	E_o*	V.G.
90	0.047	0.047	—	1580	—	4.0	0.058	9	18
	0.047	0.10	—	1760	—	3.5	0.032	13	19
	0.047	0.22	—	1820	—	3.0	0.015	16	20
	0.1	0.1	—	2920	—	2.1	0.029	12	19
	0.1	0.22	—	3570	—	1.7	0.015	17	20
	0.1	0.47	—	4020	—	1.4	0.0075	20	20
	0.22	0.22	—	6040	—	0.98	0.0135	16	19
	0.22	0.47	—	7500	—	0.78	0.0075	21	20
	0.22	1.0	—	8800	—	0.63	0.0036	25	20
180	0.047	0.047	—	694	—	6.0	0.062	25	23
	0.047	0.1	—	817	—	4.4	0.032	32	24
	0.047	0.22	—	905	—	4.0	0.0155	35	25
	0.10	0.1	—	1596	—	2.80	0.030	30	23
	0.10	0.22	—	1630	—	2.30	0.0152	32	24
	0.10	0.47	—	1860	—	2.00	0.0073	38	24
	0.22	0.22	—	3950	—	1.24	0.0150	35	22
	0.22	0.47	—	4500	—	0.96	0.0072	41	23
	0.22	1.0	—	5530	—	0.79	0.0038	49	23
300	0.047	0.047	—	438	—	6.70	0.062	38	26
	0.047	0.1	—	542	—	5.50	0.032	48	27
	0.047	0.22	—	644	—	4.30	0.016	57	27
	0.10	0.10	—	1009	—	3.5	0.031	42	25
	0.10	0.22	—	1332	—	2.5	0.015	56	26
	0.10	0.47	—	1609	—	2.1	0.0074	64	25
	0.22	0.22	—	2623	—	1.5	0.015	50	24
	0.22	0.47	—	3900	—	1.1	0.0073	70	24
	0.22	1.0	—	4920	—	0.88	0.0039	84	24

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4BQ7A• 4BZ7• 5BK7A• 5BQ7A• 6BK7B• 6BQ7A• 6BZ7•

See Circuit Diagram 1

E_{bb}	R_p	R_g	R_{g2}	R_k	C_{g2}	C_k	C	E_o*	V.G.
90	0.22	0.22	0.480	3800	0.046	5.5	0.0084	10	89
	0.22	0.47	0.480	3800	0.049	5.5	0.0054	16	114
	0.22	1.0	0.500	4400	0.045	5.3	0.0034	23	128
	0.47	0.47	1.04	7200	0.033	2.9	0.0044	10	111
	0.47	1.0	1.04	7700	0.033	2.8	0.0029	15	133
	0.47	2.2	1.10	8400	0.031	2.6	0.0020	18	152
	1.0	1.0	2.50	16000	0.018	1.4	0.0023	10	118
	1.0	2.2	2.50	18600	0.016	1.2	0.0017	11	139
180	0.22	0.22	0.550	1600	0.072	9.5	0.0090	30	161
	0.22	0.47	0.620	1800	0.062	8.5	0.0053	36	208
	0.22	1.0	0.650	1900	0.062	8.5	0.0034	43	239
	0.47	0.47	1.00	3400	0.059	6.0	0.0048	34	183
	0.47	1.0	1.00	3500	0.059	6.0	0.0031	41	229
	0.47	2.2	1.00	3800	0.059	5.8	0.0020	46	262
	1.0	1.0	2.60	7300	0.029	2.7	0.0022	33	227
	1.0	2.2	2.60	7400	0.029	2.7	0.0016	38	281
300	0.22	0.22	0.600	980	0.085	13.0	0.0085	51	223
	0.22	0.47	0.680	1090	0.084	12.0	0.0055	64	288
	0.22	1.0	0.700	1150	0.081	11.0	0.0033	74	334
	0.47	0.47	1.25	2000	0.064	7.9	0.0045	52	285
	0.47	1.0	1.34	2150	0.061	7.6	0.0029	67	363
	0.47	2.2	1.53	2350	0.057	7.1	0.0019	79	416
	1.0	1.0	2.60	4000	0.044	5.2	0.0023	51	334
	1.0	2.2	3.00	4700	0.038	4.3	0.0015	69	427

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3BC5/3CE5 3CB6 4CB6 6AG5 6BC5 6CB6 6CB6A

See Circuit Diagram 2

• One triode unit. * Peak volts.

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7199

Pentode Unit

See Circuit Diagram 2

E_{bb}	R_p	R_g	R_{g2}	R_k	C_{g2}	C_k	C	E_o*	V.G.
90	0.22	0.22	0.560	3700	0.046	4.50	0.0090	12	73
	0.22	0.47	0.600	3900	0.043	4.30	0.0055	17	95
	0.22	1.0	0.640	4200	0.039	4.00	0.0033	19	109
	0.47	0.47	0.870	6000	0.036	2.70	0.0046	16	95
	0.47	1.0	0.980	6700	0.044	3.00	0.0030	22	113
	0.47	2.2	1.00	6700	0.043	2.80	0.0020	25	131
	1.0	1.0	2.00	12200	0.021	1.44	0.0028	15	119
	1.0	2.2	2.20	12800	0.024	1.74	0.0016	21	167
180	0.22	0.22	0.530	1570	0.069	7.50	0.0088	32	82
	0.22	0.47	0.600	1730	0.064	7.40	0.0064	38	164
	0.22	1.0	0.650	1820	0.061	7.30	0.0034	45	190
	0.47	0.47	1.12	3200	0.053	5.30	0.0046	35	147
	0.47	1.0	1.40	3500	0.042	5.10	0.0028	40	209
	0.47	2.2	1.57	3740	0.040	5.40	0.0019	45	250
	1.0	1.0	2.50	6500	0.039	2.80	0.0024	34	179
	1.0	2.2	3.40	7500	0.026	2.30	0.0015	39	277
300	0.22	0.22	0.600	9200	0.086	11.2	0.0085	52	182
	0.22	0.47	0.670	1010	0.076	10.5	0.0052	66	236
	0.22	1.0	0.720	1100	0.076	10.0	0.0033	77	257
	0.47	0.47	1.25	1950	0.060	7.0	0.0044	41	221
	0.47	1.0	1.43	3210	0.053	6.4	0.0027	72	296
	0.47	2.2	1.45	2200	0.055	6.3	0.0019	82	345
	1.0	1.0	3.00	4100	0.040	4.2	0.0022	57	295
	1.0	2.2	3.30	4340	0.037	3.6	0.0016	74	378

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7199

Triode Unit

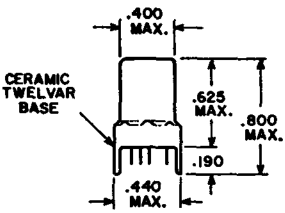
See Circuit Diagram 1

E_{bb}	R_p	R_g	R_{g2}	R_k	C_{g2}	C_k	C	E_o*	V.G.
90	0.047	0.047	—	1292	—	3.3	0.060	8	12
	0.047	0.1	—	1401	—	2.3	0.032	10	13
	0.047	0.22	—	1470	—	2.4	0.016	11	13
	0.10	0.1	—	2630	—	1.60	0.029	9	13
	0.10	0.22	—	3090	—	1.24	0.015	12	13
	0.10	0.47	—	3440	—	1.10	0.008	14	14
	0.22	0.22	—	6550	—	0.70	0.015	12	12
	0.22	0.47	—	8270	—	0.51	0.0077	16	12
	0.22	1.0	—	9130	—	0.44	0.0045	18	12
180	0.047	0.047	—	723	—	4.0	0.061	16	14
	0.047	0.1	—	836	—	3.5	0.032	20	14
	0.047	0.22	—	948	—	2.9	0.016	24	15
	0.10	0.1	—	1543	—	2.0	0.031	17	14
	0.10	0.22	—	2002	—	1.6	0.016	24	14
	0.10	0.47	—	2522	—	1.2	0.0082	30	13
	0.22	0.22	—	4390	—	0.79	0.015	24	13
	0.22	0.47	—	6122	—	0.57	0.0078	33	12
	0.22	1.0	—	8060	—	0.47	0.0046	41	12
300	0.047	0.047	—	534	—	4.0	0.061	27	15
	0.047	0.1	—	726	—	3.6	0.031	38	15
	0.047	0.22	—	840	—	3.0	0.015	44	15
	0.10	0.1	—	1117	—	2.3	0.031	26	15
	0.10	0.22	—	1613	—	1.7	0.0155	41	14
	0.10	0.47	—	2043	—	1.31	0.0078	51	14
	0.22	0.22	—	3133	—	0.93	0.015	36	13
	0.22	0.47	—	4480	—	0.69	0,0079	51	13
	0.22	1.0	—	4930	—	0.56	0.0045	55	13

* Peak volts

Outlines

METAL TYPES



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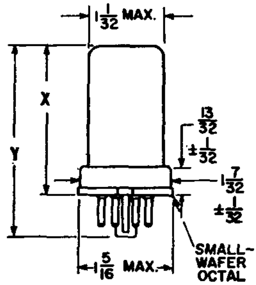
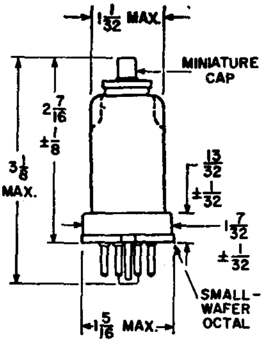


—2—

	Y	X
2A	2-5/8	2-1/16
2B	3-1/4	2-11/16



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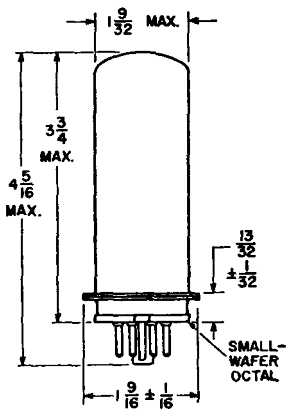
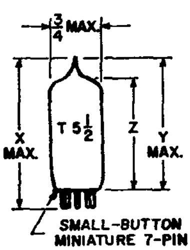


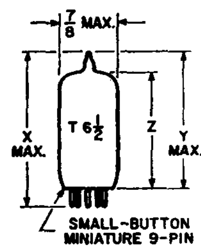
—4—

GLASS TYPES



—5—

	X	Y	Z
5A	1-5/8	1-3/8	1 ± 3/32
5B	1-3/4	1-1/2	1-1/8 ± 3/32
5C	2-1/8	1-7/8	1-1/2 ± 3/32
5D	2-5/8	2-3/8	2 ± 3/32
5E	2-3/8	2-1/8	1-3/4 ± 3/32

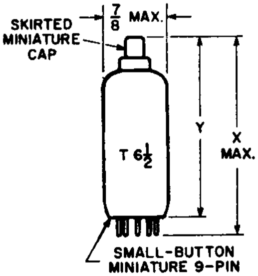


—6—

	X	Y
6A	1-3/4	1-1/2
6B	2-3/16	1-15/16
6C	2-13/32	2-5/32
6D	2-7/16	2-13/16
6E	2-5/8	2-3/8
6F	2-3/4	2-1/2
6G	3-1/16	2-13/16
6H	3-1/8	2-7/8
6J	2	1-3/4
6K	2-7/16	2-3/16
6L	2-7/8	2-5/8

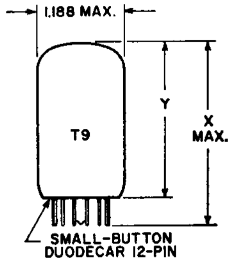
	Z
6A	1-1/8 ± 3/32
6B	1-9/16 ± 3/32
6C	1-25/32 ± 3/32
6D	1-13/16 ± 3/32
6E	2 ± 3/32
6F	2-1/8 ± 3/32
6G	2-7/16 ± 3/32
6H	2-1/2 ± 3/32
6J	—
6K	1-29/32
6L	—

All measurements in inches.



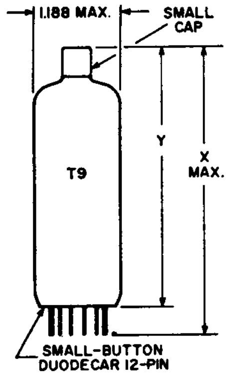
-7-

	X	Y
7A	2-27/32	2-7/16 ± 1/8
7B	3-1/16	2-15/32 MAX.
7C	3-9/32	2-7/8 ± 1/8
7D	3-1/2	3-1/4 MAX.
7E	2-17/32	2-1/8 ± 1/8
7F	2-29/32	2-5/8 MAX.
7G	2-23/32	2-1/8 ± 1/8



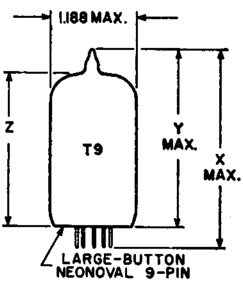
-8-

	X	Y
8A	1.875	1.250-1.500
8B	2.375	1.750-2.000
8C	2.625	2.000-2.250
8D	2.875	2.250-2.500
8E	3.050	2.770 MAX.
8F	3.125	2.500-2.750
8G	3.375	2.750-3.000



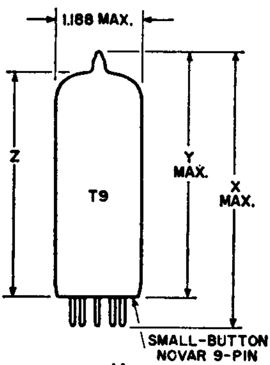
-9-

	X	Y
9A	3.375	2.750-3.000
9B	3.625	3.000-3.250
9C	4.110	3.766 MAX.



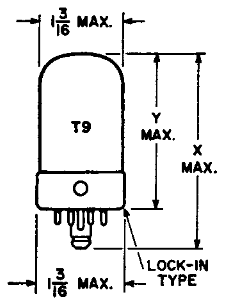
-10-

	X	Y	Z
10A	2.630	2.320	1.770-2.010
10B	2.900	2.620	2.070-2.310
10C	2.930	2.620	2.070-2.310
10D	3.230	2.920	2.370-2.610
10E	4.125	3.750	
10F	3.110	2.730	
10G	3.080	2.770	



-11-

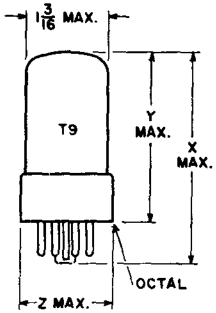
	X	Y	Z
11A	3.000	2.620	2.100-2.280
11B	3.080	2.700	2.050-2.230
11C	3.110	2.730	2.210-2.390
11D	3.410	3.010	2.510-2.690
11E	2.960	2.580	2.060-2.240



-12-

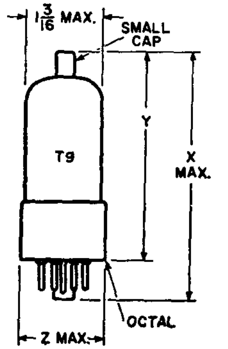
	X	Y
12A	2-9/32	1-3/4
12B	2-25/32	2-1/4
12C	3-5/32	2-5/8

All measurements in inches.



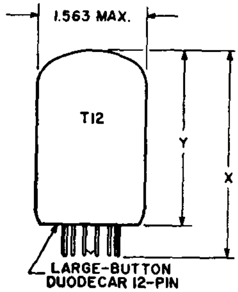
—13—

	X	Y	Z
13A	2-7/8	2-5/16	1-9/32
13B	3	2-7/16	1-9/32
13C	3-1/16	2-1/2	1-9/32
13D	3-5/16	2-3/4	1-5/16
13E	3-3/8	2-13/16	1-9/32
13F	3-7/16	2-7/8	1-9/32
13G	3-13/16	3-1/4	1-9/32
13H	4-3/16	3-9/16	1-3/16



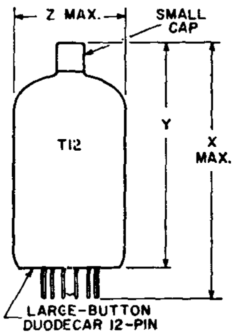
—14—

	X	Y	Z
14A	3-5/16	2-3/4	1-5/16
14B	3-9/16	3	1-9/32
14C	3-5/8	3-1/16	1-9/32
14D	3-7/8	3-5/16	1-9/32
14E	4-1/16	3-1/2	1-9/32
14F	3-13/16	3-1/4	1-9/32



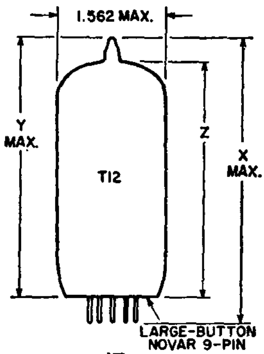
—15—

	X	Y
15A	2.875	2.250-2.500
15B	3.375	3.000 MAX.
15C	3.625	3.000-3.250
15D	3.125	2.750
15E	3.875	3.250-3.500



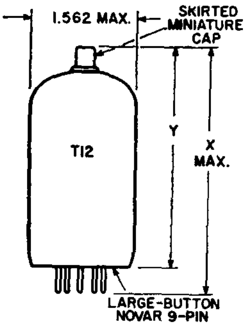
—16—

	X	Y	Z
16A	3.625	3.000-3.250	1.563
16B	4.125	3.500-3.750	1.563
16C	4.875	4.250-4.500	1.563
16D	4.375	4.000 MIN.	1.563
16E	4.375	3.750-4.000	1.563



—17—

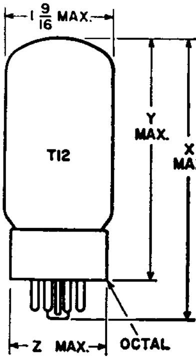
	X	Y	Z
17A	3.180	2.800	2.280-2.460
17B	3.410	3.030	2.510-2.690
17C	4.160	3.780	3.260-3.440
17D	3.550	3.170	
17E	3.710	3.330	2.810-2.900



—18—

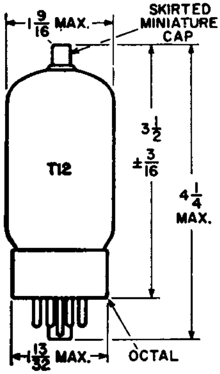
	X	Y
18A	3.55	3.04 ± 0.13
18B	4.60	4.09 ± 0.13

All measurements in inches.

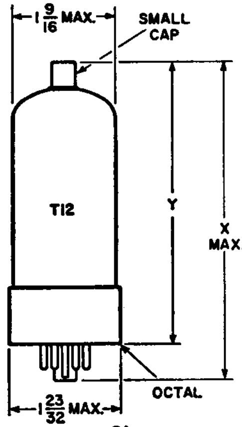


-19-

	X	Y	Z
19A	3-9/16	3	1-9/32
19B	3-7/8	3.5/16	1-13/32
19C	4	3-7/16	1-13/32
19D	4-1/4	3-11/16	1-3/8
19E	4-5/8	4-1/16	1-3/8
19F	4-5/8	4-1/16	1-5/8
19G	4-3/4	4-3/16	1-11/16
19H	5-3/16	4-5/8	1-3/8
19J	4-1/2	3-7/8	—

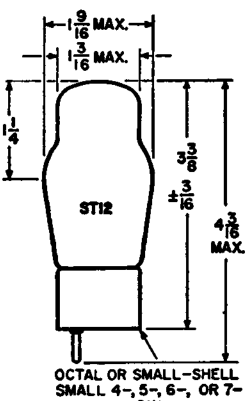


-20-

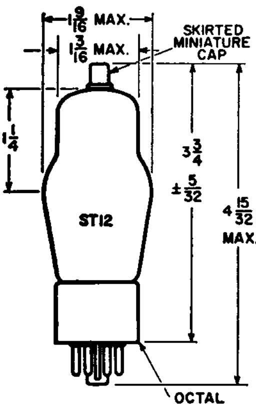


-21-

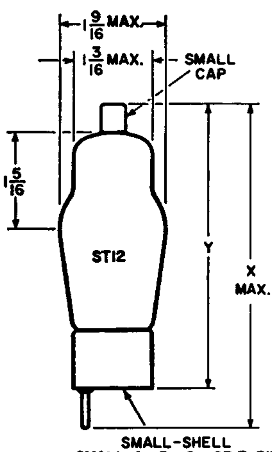
	X	Y
21A	4-3/4	4 ± 3/16
21B	5	4-7/16
21C	5-7/32	4-1/4
21D	5	4-1/4



-22-



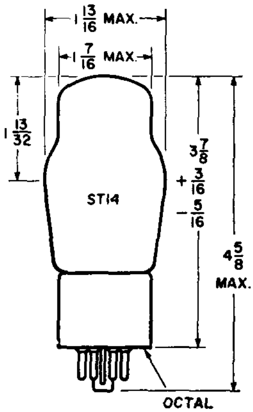
-23-



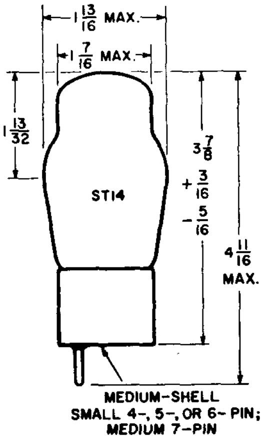
-24-

	X	Y
24A	4-15/16	4-3/16 ± 1/8
24B	4-17/32	3.25/32 ± 1/8

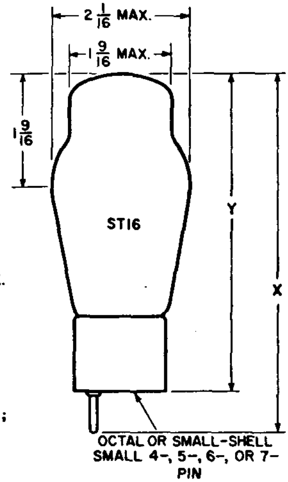
All measurements in inches.



-25-

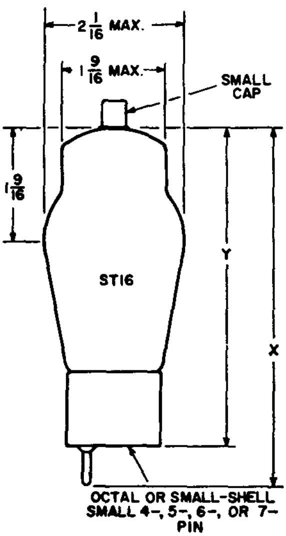


-26-



-27-

	X	Y
27A	5-1/8	4-3/8 ± 3/16
27B	5-3/8	4-9/16 ± 3/16

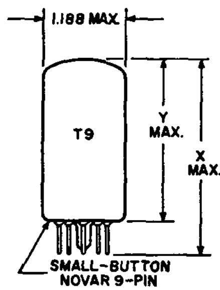


-28-

	X	Y
28A	5-1/8	4-7/16 ± 5/32
28B	5-11/16	4-31/32 ± 5/32

	MAX. LENGTH	MAX. DIAMETER
29A	1-3/4	0.4
29B	1-3/4	1-5/16
29C	2-5/16	1-5/16
29D	2-5/8	1-1/16
29E	2-7/8	1-5/16
29F	3	1-5/16
29G	3-7/16	1-15/16
29H	4	1-3/16
29J	4-7/8	1-9/16
29K	5-1/32	1-13/16
29L	6-1/4	2-7/16
29M	3-15/32	1-7/16

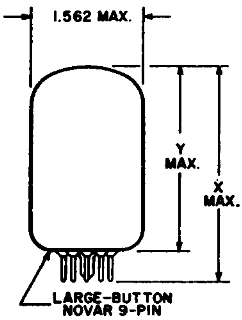
-29-



-30-

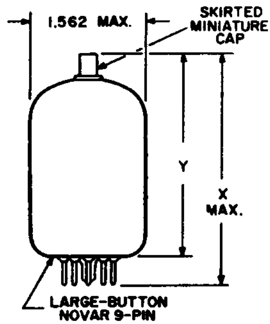
	X	Y
30A	2.380	2.000
30B	3.005	2.625
30C	3.080	2.700
30D	3.110	2.730
30E	2.125	1.750
30F	3.380	3.000

All measurements in inches.



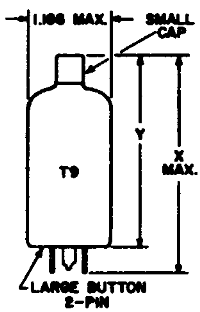
—31—

	X	Y
31A	2.880	2.500
31B	3.130	2.750
31C	3.880	3.500
31D	3.380	3.000



—32—

	X	Y
32A	3.505	2.875-3.125
32B	4.130	3.500-3.750
32C	4.380	3.750-4.000



—33—

	X	Y
33A	3.06 Max	2.52-2.68

All measurements in inches.

Circuits

THE circuits included in this Manual illustrate some of the more important applications of RCA receiving tubes; they are not necessarily examples of commercial practice. These circuits have been conservatively designed and are capable of excellent performance. The brief description provided with each circuit explains the functional relationships of the various stages and points out intended applications, major performance characteristics, and significant design features of the over-all circuit. Detailed descriptive information on individual circuit stages (for example, amplifiers, detectors, or oscillators) is given in the section on **Electron-Tube Applications** earlier in this Manual, as well as in many textbooks on electron-tube circuits.

Electrical specifications are given for circuit components to assist those interested in home construction. Layouts and mechanical details are omitted because they vary widely with the requirements of individual set builders and with the sizes and shapes of the components employed.

Circuits designed for operation from both ac and dc voltage supplies should be installed in non-metallic cabinets or properly insulated from metallic cabinets. Potentiometer shafts and switches should make use of insulated (plastic) knobs. In practical use, no metallic part of an "ac/dc" chassis should be exposed to touch, accidental or otherwise. When such circuits are tested outside of their cabinets, a line isolation transformer such as the RCA WP-25A Isotap should be used.

Performance of these circuits depends as much on the quality of the components selected and the care employed in layout and construction as on the circuits themselves. Good signal reproduction from receivers and amplifiers requires the use of good-quality speakers, transformers, chokes, and input sources (microphones, phonograph pickups, etc.).

Coils for the receiver circuits may be purchased at local parts dealers by specifying the characteristics required: for rf coils, the circuit position (antenna or interstage), tuning range desired, and tuning capacitances employed; for if coils or transformers, the intermediate frequency, circuit position (1st if, 2nd if, etc.), and, in some cases, the associated tube types; for oscillator coils, the receiver tuning range, the intermediate frequency, the type of converter tube, and the type of winding used (tapped or transformer-coupled).

The voltage ratings specified for capacitors are the minimum dc working voltages required. Paper, mica, or ceramic capacitors having higher voltage ratings than those specified may be used except insofar as the physical sizes of such capacitors may affect equipment layout. However, if electrolytic capacitors having substantially higher voltage ratings than those specified are used, they may not "form" completely at the operating voltage, with the result that the effective capacitances of such units may be below their rated value. The wattage ratings specified for resistors assume methods of construction that provide adequate ventilation; compact installations having poor ventilation may require resistors of higher wattage ratings.

Circuits which work at very high frequencies or which are required to handle very wide bandwidths demand more than ordinary skill and experience in construction. Placement of component parts is quite critical and may require considerable experimentation. All rf leads to components including bypass capacitors must be kept short and must be properly dressed to minimize undesirable coupling and capacitance effects. Correct circuit alignment and oscillator tracking may require the use of a cathode-ray oscilloscope, a high-impedance vacuum-tube voltmeter, and a signal generator capable of supplying a properly modulated signal at the appropriate frequencies. Unless the builder has had considerable experience with broad-band, high-frequency circuits, he should not undertake the construction of such circuits.

LIST OF CIRCUITS

Page

MANUFACTURERS OF SPECIAL COMPONENTS AND MATERIALS
REFERRED TO IN PARTS LIST

Allen-Bradley Co.
1201 S. 2nd Street
Milwaukee, Wis.

Alpha Wire Corp.
711 Lidgerwood Avenue
Elizabeth, N. J.

Arco Electronics, Inc.
Community Drive
Great Neck, N. Y.

Knight Products
Allied Radio Corp.
100 N. Western Avenue
Chicago, Ill.

J. W. Miller Co.
5917 S. Main Street
Los Angeles, Calif.

Moldite Electronics Corp.
250 South Street
Newark, N. J.

Ohmite Manufacturing Co.
3635 W. Howard Street
Skokie, Ill.

Stancor Electronics, Inc.
3501 W. Addison Street
Chicago, Ill.

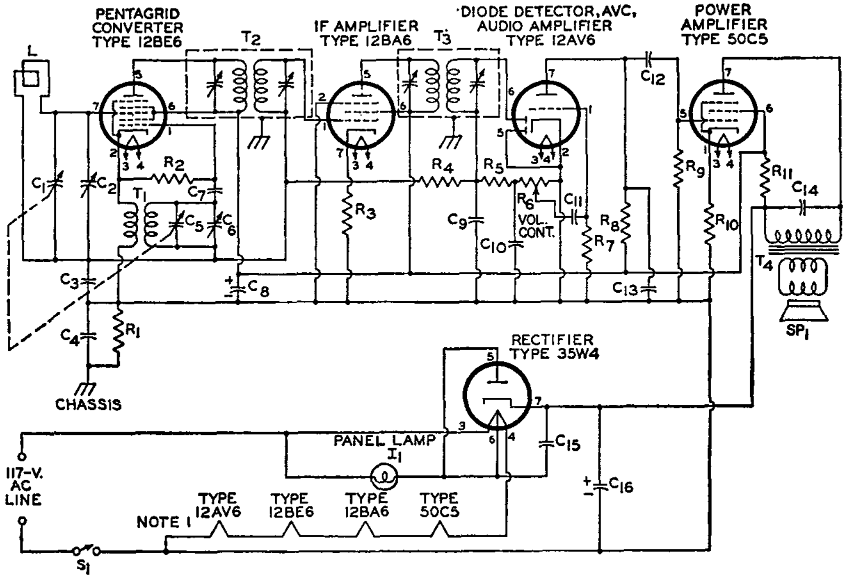
Thordarson-Meissner, Inc.
Electronic Center
7th and Bellmont
Mt. Carmel, Ill.

Triad Distributor Div.
Litton Industries
305 N. Briant Street
Huntington, Ind.

United Transformer Corp.
Div. Thompson-Ramo-Wooldridge,
150 Varick Street
New York, N. Y.

Note: Components and materials identified by RCA stock numbers may be obtained through authorized RCA distributors.

26-1 AC/DC SUPERHETERODYNE RADIO RECEIVER



Parts List

C_1, C_5=Ganged tuning capacitors; C_1, 10-365 pF, C_5, 7-115 pF
C_2=Trimmer capacitor, 4-30 pF
C_3=0.05 μF, paper, 50 V
C_4=0.1 μF, paper, 400 V
C_6=Trimmer capacitor, 2-17 pF
C_7=56 pF, ceramic
C_8=30 μF, electrolytic, 150 V
C_9, C_{10}=150 pF, ceramic
C_{11}, C_{14}=0.02 μF, paper, 400 V
C_{12}=0.002 μF, paper, 400 V
C_{13}=330 pF, mica
C_{15}=0.05 μF, paper, 400 V
C_{16}=50 μF, electrolytic, 150 V
I_1=Panel lamp, No. 40 or 47
L=Loop antenna or ferrite-rod antenna, 540-1600 kHz (with specified values of capacitance for C_1 and C_2)
R_1=0.22 megohm, 0.5 watt
R_2=33000 ohms, 0.5 watt
R_3=100 ohms, 0.5 watt
R_4=3.3 megohms, 0.5 watt
R_5=47000 ohms, 0.5 watt
R_6=Volume control, potentiometer, 0.5 megohm
R_7=4.7 megohms, 0.5 watt
R_8, R_9=0.47 megohm, 0.5 watt
R_{10}=150 ohms, 0.5 watt
R_{11}=1200 ohms, 1 watt
S_1=On-off switch; single-pole, single-throw
SP_1=Speaker
T_1=Oscillator coil for use with 7-115 pF tuning capacitor and 455-kHz intermediate-frequency transformer
T_2, T_3=Intermediate-frequency transformers, 455 kHz (permeability-tuned type may be used)
T_4=Output transformer for matching impedance of voice coil to 2500-ohm load

Note: The following tube types are recommended for a 100-mA-heater tube complement: 18FX6A converter, 18FW6A if amplifier, 18FY6A detector and audio amplifier, 34GD5A power amplifier, and 36AM3B rectifier.

Circuit Description

This basic five-tube superheterodyne radio receiver operates directly from an ac power line or a dc supply of 117 volts. AC power inputs are converted to dc power by the 35W4 half-wave rectifier circuit. The receiver uses a series heater arrangement. With ON-OFF switch S_1 closed, the heater string is connected directly across the 117-volt input terminals. A 6.3-volt panel lamp I_1 connected between heater pins 3 and 6 of the 35W4 rectifier tube lights to indicate that power is applied to the receiver.

A ferrite-rod or loop antenna L and tuning capacitor C_1 select amplitude-modulated rf signals from the desired broadcast-band (550 to 1600 kHz) radio station and couple these signals to grid No. 3 (pin 7) of the 12BE6 pentagrid converter. A local-oscillator signal, developed by the resonant circuit formed by oscillator coil T_1 and variable capacitors C_5 and

26-1 AC/DC SUPERHETERODYNE RADIO RECEIVER (Cont'd)

Circuit Description (Cont'd)

C_6, is also applied to the 12BE6 pentagrid converter, at grid No. 1 (pin 1). The modulated-rf and local-oscillator signals are mixed across the nonlinear impedance of the converter tube to produce the 455-kHz intermediate frequency used in the receiver. The antenna and oscillator tuning capacitors C_1 and C_5 are mechanically ganged so that the antenna and oscillator resonant circuits can be adjusted together to maintain the 455-kHz difference frequency for any dial setting in the broadcast-frequency band. Trimmer capacitors C_2 and C_6 are adjusted to assure that the desired tracking relationship is maintained across the band. Positive feedback to sustain oscillations is inductively coupled by T_1 from the cathode of the 12BE6 converter to the local-oscillator resonant circuit.

A single if stage, which uses a high-transconductance 12BA6 remote-cutoff pentode, provides the required amplification of the intermediate-frequency signals. This stage is made selective at 455 kHz by the double-tuned input and output transformers T_2 and T_3. Audio-signal components are extracted from the if signal by the second-detector circuit, which consists of the pin 6 diode section in the 12AV6 tube and associated components. (The pin 5 diode section of the 12AV6 is not used and is shorted to the tube cathode, pin 2.) The audio output from the detector is developed across the VOL. CONT. potentiometer R_6, which provides manual adjustment of the output sound level of the receiver. The detector also develops a negative dc voltage proportional to the rf input across a 150-picofarad capacitor C_9 for automatic volume control in the receiver. This avc voltage is used as bias for the converter and if amplifier and automatically controls the gain of these stages.

The audio-signal voltage at the wiper arm of the VOL. CONT. potentiometer is amplified by the triode (audio-voltage-amplifier) section of the 12AV6 and is then used to drive the 50C5 audio output stage. The output stage develops the audio power required to produce an audible output from the speaker. Audio output transformer T_4 matches the 2500-ohm plate-load impedance of the 50C5 to the speaker voice coil.

26-2 AM/FM SUPERHETERODYNE RADIO RECEIVER

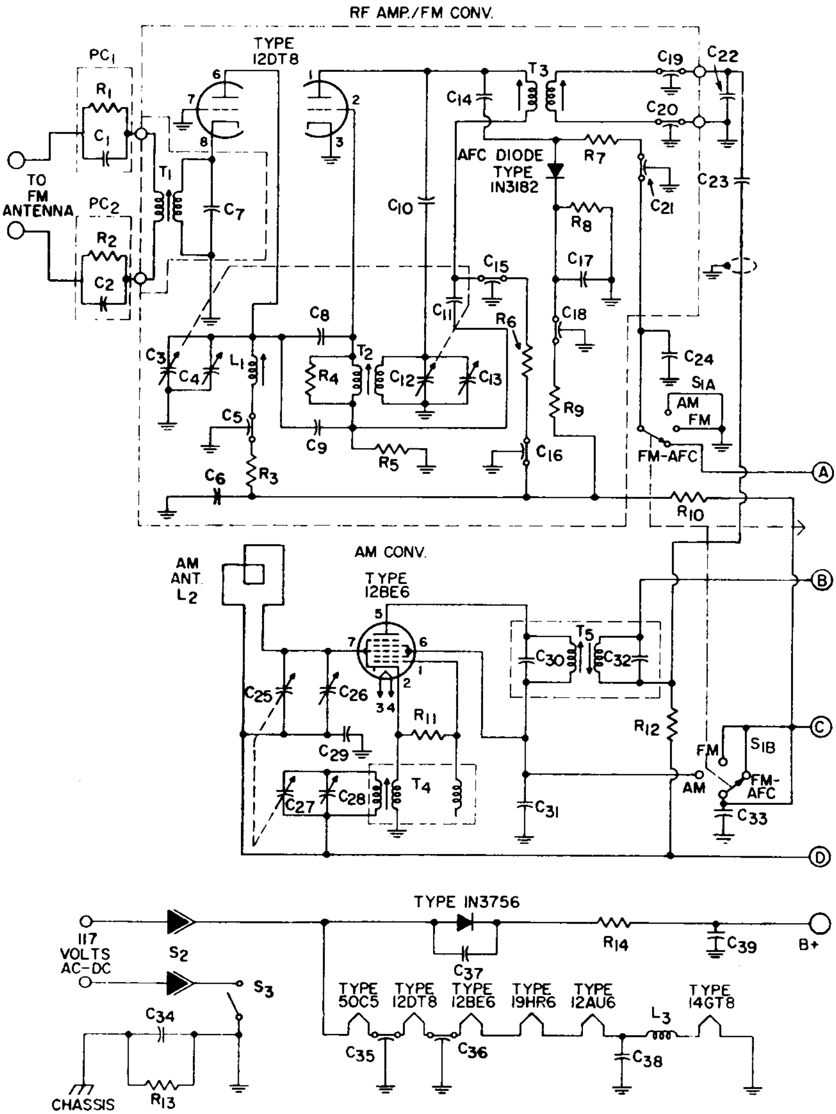
Circuit Description

This AM/FM radio receiver operates directly from either an ac power line or a dc supply of 117 volts. AC power inputs are converted to dc power by a 1N3756 silicon-rectifier half-wave power supply. The receiver uses a series heater string, which is connected across the 117-volt input when ON-OFF switch S_3 and interlock S_2 are closed. The interlock assures that power is automatically disconnected when the receiver is removed from the chassis.

AM or FM operation of the receiver is selected by means of switch S_1. For AM operation (S_1 set to AM position), amplitude-modulated rf signals in the AM broadcast band (550 to 1600 kHz) from the desired radio broadcast station are selected by antenna L_2 and tuning capacitor C_{25}. These signals are amplified and converted to the 455-kHz AM intermediate frequency by the 12BE6 pentagrid converter. Tuning capacitors C_{25} and C_{27} are mechanically ganged so that the antenna and local-oscillator sections of the converter can be tuned simultaneously to maintain the 455-kHz difference frequency for any station setting. Trimmer adjustments are provided by variable capacitors C_{26} and C_{28}.

26-2

AM/FM SUPERHETERODYNE RADIO RECEIVER (Cont'd)



Parts List

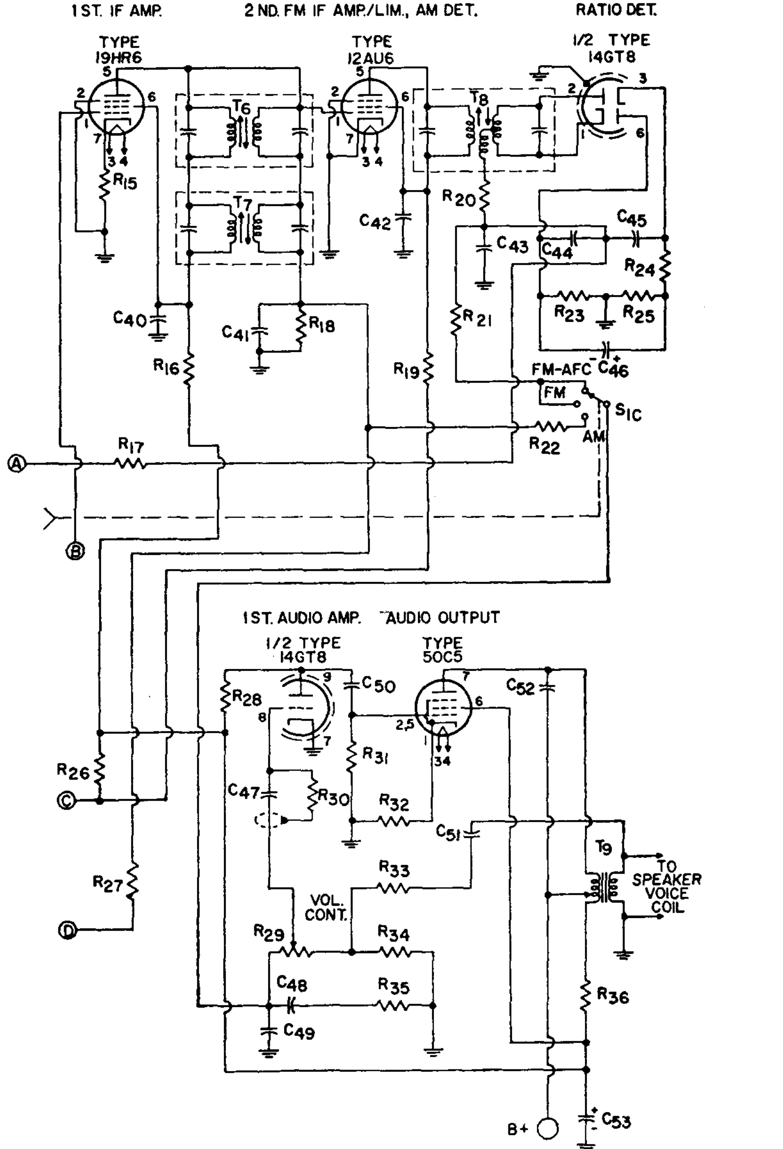
C_1 = Part of PC_1
C_2 = Part of PC_2
C_3, C_{12} = Ganged tuning capacitors; tune L_1 and T_2 to 88-108 MHz

C_4, C_{13} = Trimmer capacitors, 1-7 pF
C_5, C_{16}, C_{18} = 1000 pF, feed-through, 500 V
C_6 = 0.1 μF, ceramic, 500 V

C_7 = 36 pF, ceramic, 500 V
C_8, C_{14} = 6.8 pF, ceramic, 500 V
C_9 = 11 pF, ceramic, 500 V
C_{10} = 68 pF, ceramic, 500 V

26-2

AM/FM SUPERHETERODYNE RADIO RECEIVER (Cont'd)



Parts List (Cont'd)

C_{11}=21 pF, ceramic, 500 V
C_{15}=500 pF, feedthrough, 500 V
C_{17}=0.22 μF, ceramic disc, 500 V

C_{19}, C_{20}=2 pF, feedthrough, 500 V
C_{21}, C_{35}, C_{36}=2000 pF, feedthrough, 500 V

C_{22}=IF transformer tuning capacitor; value, with cable capacitance, tunes T_3 to 10.7 MHz

26-2 AM/FM SUPERHETERODYNE RADIO RECEIVER (Cont'd)

Parts List (Cont'd)

C_{23}=4700 pF, ceramic, 500 V
C_{24}=0.15 μF, paper, 200 V
C_{25}, C_{27}=Ganged tuning capacitors; tune T_4 to 540-1650 kHz
C_{26}, C_{28}=Trimmer capacitors, 12 pF
C_{29}, C_{33}, C_{38}, C_{47}=0.01 μF, ceramic, 500 V
C_{30}=Part of T_5
C_{31}, C_{49}=1000 pF, ceramic, 500 V
C_{32}=Part of T_5
C_{34}=0.1 μF, ceramic, 500 V
C_{37}=0.047 μF, paper, 400 V
C_{39}=80 μF, electrolytic, 150 V
C_{40}, C_{42}=2700 pF, ceramic, 500 V
C_{41}, C_{43}=100 pF, ceramic, 500 V, NPO
C_{44}, C_{45}=330 pF, mica, 500 V
C_{46}=2 μF, electrolytic, 50 V
C_{48}=0.01 μF, paper, 200 V
C_{50}=5600 pF, ceramic, 500 V
C_{51}=0.1 μF, paper, 200 V
C_{52}=0.022 μF, paper, 200 V
C_{53}=50 μF, electrolytic, 150 V
L_1, L_3=1 μH, rf coil
L_2=Antenna, air-loop type with back cover
PC_1, PC_2=Printed circuit; includes 0.5 megohm, 0.25-watt resister and 470-picofarad, 500-volt capacitor; RCA Stock No. 104328
R_1=Part of PC_1
R_2=Part of PC_2
R_3=2200 ohms, 0.5 watt
R_4=1200 ohms, 0.5 watt
R_5, R_{21}=33000 ohms, 0.5 watt
R_6, R_{11}=22000 ohms, 0.5 watt
R_7, R_{28}, R_{31}=0.47 megohm, 0.5 watt
R_8=3900 ohms, 0.5 watt
R_9, R_{22}=47000 ohms, 0.5 watt
R_{10}=220 ohms, 0.5 watt
R_{12}, R_{17}=1 megohm, 0.5 watt
R_{13}=0.22 megohm, 0.5 watt
R_{14}=100 ohms, wire-wound, 4 watts
R_{15}, R_{20}=68 ohms, 0.5 watt
R_{16}=4700 ohms, 0.5 watt
R_{18}=0.33 megohm, 0.5 watt
R_{19}, R_{24}=1000 ohms, 0.5 watt
R_{23}, R_{25}=6800 ohms, 0.5 watt
R_{26}=220 ohms, 0.5 watt
R_{27}=3.3 megohms, 0.5 watt
R_{29}=Volume control, potentiometer, 1 megohm, part of assembly with S_3
R_{30}=4.7 megohms, 0.5 watt
R_{32}=150 ohms, 0.5 watt
R_{33}=1500 ohms, 0.5 watt
R_{34}=820 ohms, 0.5 watt
R_{35}=3900 ohms, 0.5 watt
R_{36}=560 ohms, 0.5 watt
S_1=AM-FM-AFC selector; 3-section slide switch
S_2=Interlock
S_3=ON-OFF switch, part of assembly with R_{29}
T_1=FM antenna transformer
T_2=FM oscillator transformer
T_3, T_6=FM if transformer, 10.7 MHz
T_4=AM oscillator coil; with specified values of tuning and trimmer capacitance, tunes to 540 to 1600 kHz
T_5, T_7=AM if transformer, 455 kHz
T_8=Ratio-detector transformer, 10.7 MHz
T_9=Audio output transformer, matches impedance of speaker voice coil to 2500-ohm tube load

Circuit Description (Cont'd)

With switch S_1 in the FM or FM-AFC position, the FM tuner selects rf signals in the FM broadcast band (88 to 108 MHz) from the desired FM radio station, amplifies these signals, and converts them to the 10.7-MHz FM intermediate frequency. The rf-amplifier and converter stages of the tuner each use one section of a 12DT8 high-mu twin triode. Ganged tuning of the rf-amplifier and converter tuning capacitors, C_3 and C_{12}, assures that the converter local-oscillator frequency tracks the input tuning at 10.7 MHz above the center frequency of the FM channel selected. Trimmer adjustments are provided by variable capacitors C_4 and C_{13}.

The 19HR6 if amplifier is used in both FM and AM modes of operation. Depending upon the setting of selector switch S_1, this stage amplifies the frequency-modulated 10.7-MHz intermediate-frequency output from the FM converter or the amplitude-modulated 455-kHz intermediate-frequency signal from the AM converter. Additional amplification of FM if signals is provided by the 12AU6 pentode stage, which is used as a combination second FM if amplifier and noise limiter. A portion of the 12AU6 stage is also used as a second detector circuit to extract the audio-signal components from the 455-kHz AM if signals. For this demodulation function, the cathode and control grid of the 12AU6 are used as the detector diode. The 10.7-MHz FM if signals are demodulated and amplitude distortion is removed by a ratio dector that uses the diode sections of a 14GT8 twin diode—high-mu triode. Good selectivity in the if amplifier and detector at 10.7 MHz is provided by the double-tuned transformers T_3, T_6, and T_8, and at 455 kHz by the double-tuned transformers T_5 and T_7.

Depending upon the mode of operation, a section of S_1 selects the audio output from the AM detector or from the FM ratio detector. The selected audio output is amplified by

26-2 AM/FM SUPERHETERODYNE RADIO RECEIVER (Cont'd)

Circuit Description (Cont'd)

an audio voltage amplifier which uses the high-mu triode section of a 14GT8 and a 50C5 audio output stage. The output stage provides the power necessary to produce the required speaker output. Transformer T_9 matches the 2500-ohm plate impedance of the 50C5 to the speaker voice coil. Manual adjustment of the receiver output is provided by the VOL. CONT. potentiometer R_{29} in the control-grid circuit of the audio voltage amplifier.

A negative dc voltage proportional to the input signal level is developed across R_{18} and C_{41} during either AM or FM operation of the receiver. This voltage is applied as bias to the control grid (pin 1) of the 19HR6 if amplifier and the signal grid (pin 7) of the 12BE6 AM converter to provide automatic gain control of the receiver in each mode of operation. With S_1 in the FM-AFC position, the 1N3182 AFC diode rectifies the voltage across the tertiary winding of the ratio-detector transformer T_8. The resultant frequency-sensitive voltage, applied to the plate resonant circuits of the FM rf-amplifier and converter stages, provides automatic frequency control in the FM tuner.

Note: See general considerations for construction of high-frequency and broad-band circuits on page 561.

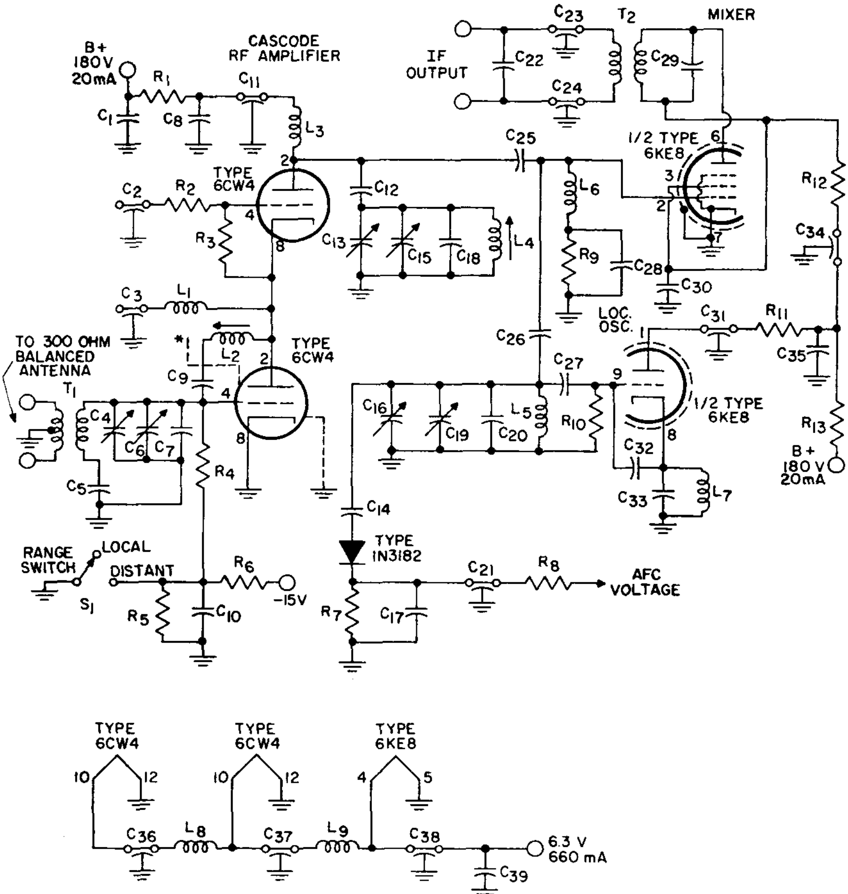
26-3 FM TUNER

Circuit Description

This three-stage FM tuner features a pair of 6CW4 nuvistor triodes operated in a low-noise, high-gain cascode rf-amplifier stage. The mixer and local-oscillator sections of the tuner use the pentode and triode sections, respectively, of a 6KE8 triode-pentode. The dc operating power for the tuner is obtained from a 180-volt, 20-milliampere supply. Power for the tube heaters is obtained from a 6.3-volt, 660-milliampere ac source.

The tuner uses a 300-ohm balanced antenna. Antenna transformer T_1 matches the 300-ohm antenna impedance to the input circuit of the cascode rf amplifier. Antenna tuning capacitor C_4 is adjusted to select the desired FM channel. The frequency-modulated rf signals are amplified by the cascode rf stage and coupled to the control grid of the mixer stage. The local oscillator generates a signal, at a frequency 10.7 MHz above the center frequency of the selected FM channel, which is also applied to the control grid of the mixer stage. The rf and local-oscillator signals are mixed to produce the desired 10.7-MHz FM intermediate frequency. Ganged tuning of the antenna, mixer, and local-oscillator tuning capacitors, C_4, C_{13}, and C_{16}, assures that the local-oscillator frequency tracks the input tuning at 10.7 MHz above the selected FM channel. Capacitors C_6, C_{15}, and C_{19} are trimmer adjustments for the tuner. The double-tuned transformer T_2 selects the 10.7-MHz FM if signals at the plate of the mixer stages and couples them to the if-amplifier/limiter section of the FM receiver.

26-3 FM TUNER (Cont'd)



* A metal shield should be provided between grid and plate terminals on the 6CW4 socket.

Parts List

C_1, C_8, C_{35}, C_{39} = 0.01 μF, ceramic disc, 400 V
C_2, C_{31} = 2000 pF, feed-through, 400 V
C_3, C_{11}, C_{21}, C_{34}, C_{36}, C_{37}, C_{38} = 1000 pF feedthrough, 400 V
C_4, C_{13}, C_{16} = Ganged tuning capacitor; 6.6-23 pF, 400 V; Miller No. 1461-BS or equiv.
C_5, C_9, C_{28} = 1000 pF, ceramic, 400 V
C_6, C_{15}, C_{19} = Trimmer capacitors, 1-7.5 pF, ceramic, 400 V
C_7, C_{18}, C_{33} = 10 pF, ceramic, 400 V
C_{10} = 2000 pF, ceramic disc, 400 V
C_{12}, C_{30} = 2000 pF, ceramic, 400 V
C_{14}, C_{32} = 6.8 pF, ceramic, 400 V
C_{17} = 0.22 μF, ceramic, 400 V
C_{20} = 18 pF, ceramic, 400 V
C_{22} = Capacitor inserted in place of tuning capacitor in secondary winding of T_2; value with cable capacitance tunes output circuit of tuner to 10.7 MHz
C_{23}, C_{24} = 2 pF feedthrough, 400 V
C_{25} = 22 pF, ceramic, 400 V
C_{26} = 2.2 pF, ceramic, 400 V
C_{27} = 47 pF, ceramic, 400 V
C_{29} = Part of T_2
L_1 = RF coil, 5 turns of No. 22 enamel wire close-wound on ¼-inch-diameter coil form
L_2 = RF coil, 12 turns of No. 22 enamel wire close-wound on ¼-inch-diameter slug-tuned coil form; tuning slug = ⅜-inch-long Moldite No. 5101 ferrite or equiv.
L_3 = RF choke, 4 μH, J. W. Miller No. 70F396A1 or equiv.
L_4 = RF coil, 3 turns of No. 16 enamel wire wound double-spaced on ¼-inch-diameter slug-tuned coil form; tuning slug = ⅜-inch-long Moldite No. 5101 ferrite or equiv.
L_5 = RF coil, 1-½ turns of No. 16 enamel wire close-

26-3 FM TUNER (Cont'd)

Parts List (Cont'd)

wound on 1/4-inch-diameter slug-tuned coil form: tuning slug=3/8-inch-long Moldite No. 5101 ferrite or equiv.
L_6=RF choke, 2μH, Ohmite No. Z144 or equiv.
L_7=RF coil; 0.4 μH; 20 turns of No. 26 enamel wire close-wound on a 0.47 megohm, 0.5-watt Allen-Bradley resistor or resistor of equivalent physical size
L_8, L_9=RF chokes; 1μH; 25 turns of No. 24 enamel wire close-wound on a 0.47-megohm, 1-watt Allen-Bradley resistor or resistor of equivalent physical size
R_1, R_{13}=220 ohms, 0.5 watt
R_2=5 ohms, 0.5 watt
R_3, R_9=0.47 megohm, 0.5 watt
R_4, R_6, R_8=47000 ohms, 0.5 watt
R_5=0.1 megohm, 0.5 watt
R_7=3900 ohms, 0.5 watt
R_{10}=22000 ohms, 0.5 watt
R_{11}=4700 ohms, 0.5 watt
R_{12}=15000 ohms, 0.5 watt
S_1=AM/FM range switch; open position is used for local stations, closed position for distant stations
T_1=Antenna transformer; primary: 2 turns of No. 32 wire with type B nylon insulation, Alpha No. 1860 or equivalent, center-tapped; secondary: 3 turns of No. 16 enamel wire; wound double-spaced on 1/4-inch-long coil form; tuning slug = 3/8-inch-long Moldite No. 5101 ferrite or equiv.
T_2=FM if transformer, 10.7 MHz; J. W. Miller 1451 or equiv.; capacitor in secondary should be replaced by C_{22}

Note: See general considerations for construction of high-frequency and broadband circuits on page 561.

26-4 THREE-STAGE IF AMPLIFIER/LIMITER AND DETECTOR

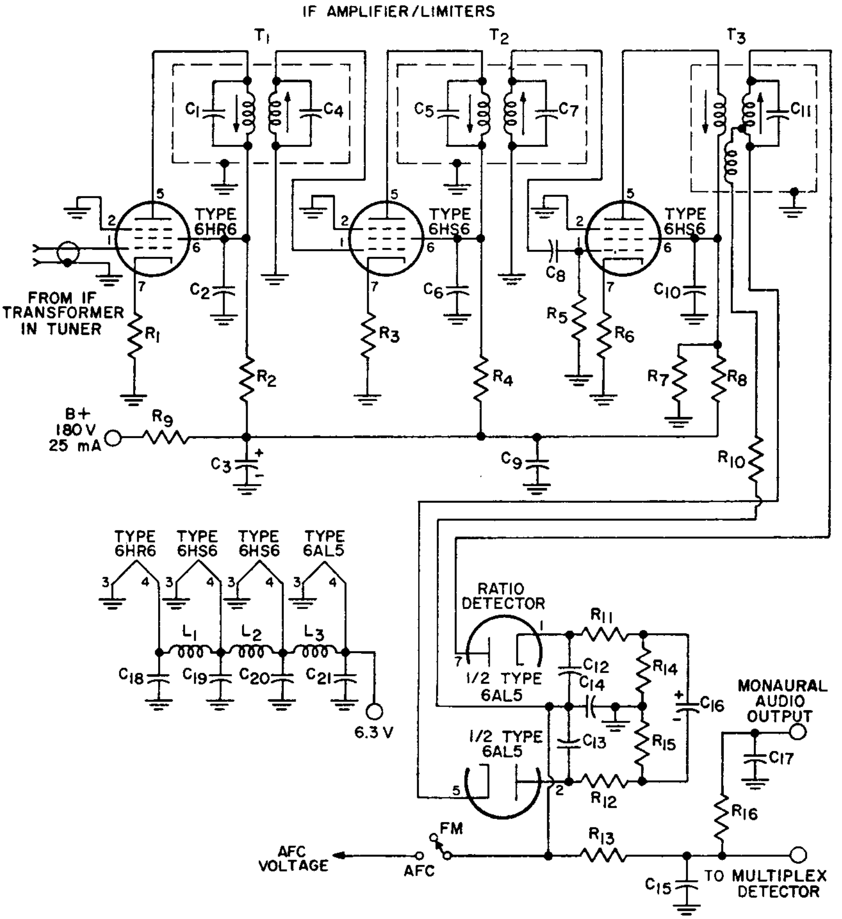
For Monaural or Stereo Tuner

Circuit Description

This three-stage if amplifier/limiter and detector circuit, when used with a front-end circuit such as that shown in circuit 25-3, makes possible an over-all tuner gain of 35 dB. The over-all bandwidth of the if-amplifier stages, between the 6-dB-down points, is 300 kHz, and the peak separation of the detector is 440 kHz. The circuit provides a signal-to-noise ratio of 20 dB for an input of 2.8 microvolts or 30 dB for an input of 4.1 microvolts. The 6HR6 and 6HS6 pentodes used in the if-amplifier stages have very high transconductance and a grid-No.1-to-plate capacitance substantially less than 0.01 picofarad and are, therefore, especially suited for use in FM if amplifiers and television sound if amplifiers. These pentodes operate from a 180-volt, 25-milliampere dc supply. Heater power for the pentodes and for the 6AL5 twin diode used in the ratio detector is obtained from a 6.3-volt ac source.

The frequency-modulated, 10.7-MHz intermediate-frequency signal from the mixer stage in the FM tuner is applied to the control grid of the first if-amplifier stage. This signal is amplified by the three transformer-coupled amplifier stages and applied by transformer T_3 to the ratio detector. The doubled-tuned coupling transformers T_1, T_2, and T_3 provide the selectivity at 10.7 MHz and the band-pass characteristics required for optimum transfer of the frequency-modulated signal. Circuit stability is improved by the use of unbypassed cathode resistors in each amplifier stage. The first two if stages are basically amplifiers, although they provide some saturation limiting of large-level signals. The 3300-ohm screen-grid dropping resistors (R_2 and R_4) reduce the screen-grid voltages in these stages to obtain the desired limiting characteristics. The 6HR6 pentode used in the first if amplifier is a remote-cutoff tube and, if desired, this stage may be operated with agc bias. The 6HS6 pentodes used in the second and third if stages are sharp-cutoff tubes. In addition, the screen-grid voltage divider network (R_7 and R_8) for the third stage substantially reduces the screen-grid voltage so that the stage will provide both cutoff and saturation limiting of large-level signals. The limiting in the if stages helps remove any amplitude modulation from the frequency-mod-

26-4 THREE-STAGE IF AMPLIFIER/LIMITER AND DETECTOR (Cont'd)



Parts List

C_1, C_4 = Part of T_1
C_2, C_6 = 2200 pF, ceramic disc, 400 V
C_3 = 50 μF, electrolytic, 450 V
C_5, C_7 = Part of T_2
C_8 = 47 pF, ceramic disc, 400 V
C_9, C_{18}, C_{19}, C_{20}, C_{21} = 0.01 μF, ceramic disc, 400 V
C_{10} = 1500 pF ceramic disc, 400 V
C_{11} = Part of T_3
C_{12}, C_{13}, C_{15} = 330 pF, ceramic disc, 400 V
C_{14} = 100 pF, ceramic disc, 400 V
C_{16} = 2 μF, electrolytic, 400 V
C_{17} = 1000 pF, ceramic disc, 400 V
L_1, L_2, L_3 = 1 μH
R_1, R_3 = 68 ohms, 0.5 watt
R_2, R_4, R_{13} = 3300 ohms, 0.5 watt
R_5 = 0.1 megohm, 0.5 watt
R_6, R_{10} = 100 ohms, 0.5 watt
R_7 = 15000 ohms, 0.5 watt
R_8 = 22000 ohms, 0.5 watt
R_9 = 2200 ohms, 3 watts
R_{11} = 1200 ohms, 0.5 watt
R_{12} = 390 ohms, 0.5 watt
R_{14}, R_{15} = 6800 ohms, 0.5 watt
R_{16} = 68000 ohms, 0.5 watt
T_1, T_2 = IF transformers, 10.7 MHz
T_3 = Ratio-detector transformer, 10.7 MHz

Note: Tube shields may be required if regeneration is encountered. See general considerations for construction of high-frequency and broad-band circuits on page 561.

26-4 THREE-STAGE IF AMPLIFIER/LIMITER AND DETECTOR (Cont'd)

Circuit Description (Cont'd)

ulated signals.

The 6AL5 ratio-detector circuit provides additional noise limiting of the FM signal and demodulates this signal to recover the audio information. The detector circuit provides the input to the audio amplifiers of a monaural receiver or to the multiplex detector in a stereo system. The RC network (R_{16} and C_{17}) in the monaural output lead provides the desired de-emphasis of high audio frequencies.

26-5 FM STEREO MULTIPLEX ADAPTER

Circuit Description

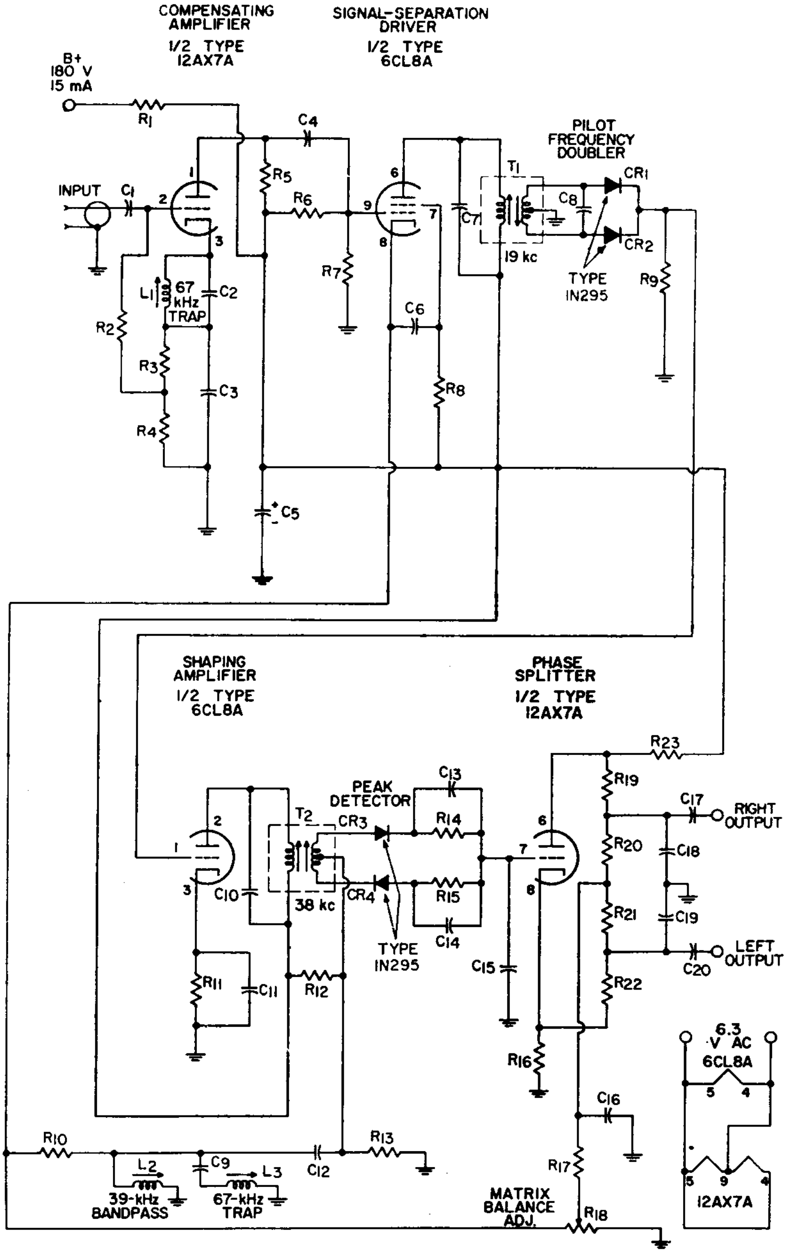
This FM stereo multiplex adapter demodulates composite multiplex signals from an FM tuner and separates these signals into left- and right-channel inputs for stereo audio-output stages. The dc operating power for the 12AX7A and 6CL8A twin triodes used in the adapter circuit is obtained from a 180-volt, 15-milliampere supply. Power for the dual heaters of the 12AX7A and the single heater of the 6CL8A is obtained from a 6.3-volt source.

The composite signal applied to the multiplex adapter from the ratio detector (or discriminator) in an FM receiver includes a 19-kHz pilot-frequency (multiplex-reference) component and sum (L + R) and difference (L − R) components of left- and right-channel audio signals. The L + R signal is the demodulated in-phase combination of the left- and right-channel audio information used to modulate the main carrier frequency of the receiver. The L − R signal is the out-of-phase combination of the left- and right-channel information and is used to amplitude-modulate a 38-kHz subcarrier. This subcarrier is suppressed in the FM tuner so that only the L − R sideband components of the amplitude-modulated signal remain.

The composite input signal is amplified by the 12AX7A triode section in the input stage of the adapter. The high input impedance of this stage prevents excessive loading of the ratio detector. The 67-kHz trap (L_1 and C_2) in the cathode circuit of this stage eliminates any SCA (storecast allocation) signal components that may be included in the composite signal. The composite signal is coupled from the plate of the input stage to the control grid of the 6CL8A triode section used in a signal-separation driver. This stage operates as a cathode follower for the L + R audio components and the L − R subcarrier sideband components. The L + R audio components are developed across the MATRIX BALANCE ADJ. potentiometer R_{18} and coupled from the wiper arm of this potentiometer to the output resistor matrix network R_{19} through R_{22}. A 3300-picofarad capacitor C_{16} in the coupling circuit filters out any 19-kHz pilot-frequency components or 38-kHz subcarrier sideband components that may be developed across potentiometer R_{18}. The L − R sideband components are coupled from the cathode of the signal-separation driver to the center tap of the secondary winding of the transformer T_2 in the peak detector. The 38-kHz band-pass coil L_2 and the 67-kHz series-resonant trap C_9 and L_3 assure maximum signal transfer of the L − R sideband components with minimum interference from storecast signals.

The 19-kHz double-tuned transformer T_1 in the plate circuit of the signal-separation driver presents a highly selective load to the 19-kHz pilot-frequency component included in the composite multiplex signal and couples this 19-kHz component to the pilot-frequency doubler. The doubler

26-5 FM STEREO MULTIPLEX ADAPTER (Cont'd)



26-5 FM STEREO MULTIPLEX ADAPTER (Cont'd)

Parts List

C_1, C_{11}, C_{12}, C_{13}, C_{14}, C_{17}, C_{20}=0.01 μF, ceramic, 500 V
C_2, C_9=2200 pF, film, 500 V, N150
C_3, C_{18}, C_{19}=270 pF, ceramic, 500 V, N750
C_4=0.047 μF, paper, 200 V
C_5=40 μF, electrolytic, 450 V
C_6=0.22 μF, paper, 400 V
C_7, C_8=1500 pF, film, 500 V, N150
C_{10}=1000 pF, film, 500 V, N150
C_{15}=470 pF, ceramic, 500 V
L_1, L_3=RF coil, 67-kHz trap, RCA stock No. 111047 or equiv.
L_2=RF coil, 38-kHz bandpass, RCA stock No. 111048 or equiv.
R_1=330 ohms, 1 watt
R_2=0.56 megohm, 0.5 watt
R_3=1500 ohms, 0.5 watt
R_4=15000 ohms, 0.5 watt
R_5=68000 ohms, 0.5 watt
R_6=3.9 megohms, 0.5 watt
R_7=1 megohm, 0.5 watt
R_8, R_{10}=10000 ohms, 0.5 watt
R_9, R_{14}, R_{15}=47000 ohms, 0.5 watt
R_{11}=4700 ohms, 0.5 watt
R_{12}=1.2 megohms, 0.5 watt
R_{13}=0.15 megohm, 0.5 watt
R_{16}, R_{17}, R_{23}=22000 ohms, 0.5 watt
R_{18}=Potentiometer, balance adjustment, 10000 ohms, RCA stock No. 111044 or equiv.
R_{19}, R_{20}, R_{21}, R_{22}=0.1 megohm, 0.5 watt
T_1=19-kHz transformer, RCA stock No. 111045 or equiv.
T_2=38-kHz transformer, RCA stock No. 111046 or equiv.

Note: See general considerations for construction of high-frequency and broadband circuits on page 561.

Circuit Description (Cont'd)

circuit, which consists of two 1N295 diodes (CR_1 and CR_2) in a full-wave rectifier configuration, doubles the pilot frequency to regenerate the 38-kHz subcarrier required for demodulation of the L − R sideband components.

The 38-kHz output of the doubler is amplified by the 6CL8A triode section used in the shaping amplifier and reshaped to a sine wave by the tuned primary of the peak detector transformer T_2. In the secondary of T_2, the 38-kHz subcarrier is recombined with the L − R sideband components from the cathode of the signal-separation driver. This combined signal is then demodulated by the 1N295 detector diodes CR_3 and CR_4 to obtain the L − R audio signal.

The L − R audio signal is applied to the control grid of the 6CL8A section used in a phase-splitter circuit. The cathode and plate outputs of the phase splitter are equal in amplitude and opposite in phase so that one output represents an L − R signal and the other output represents a − L + R signal. These signals are applied to the output-resistor matrix network where they are added to the L + R audio signal from the cathode circuit of the signal-separation driver. In the summation of the L + R and L − R audio signal, the R components are canceled, and the resultant obtained is the left-channel audio output. The summation of the L + R and − L + R signals results in cancellation of the L components so that only the right-channel audio output is obtained. These outputs are then applied to the stereo receiver left- and right-channel audio-output stages, respectively.

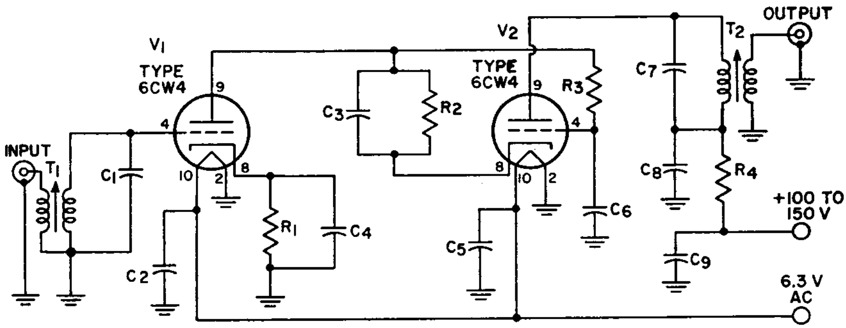
26-6 PREAMPLIFIER FOR AMATEUR RECEIVER

For 15-, 10-, and 6-Meter (21-, 30-, and 50-MHz)
Amateur Bands and 27-MHz Citizens Band

Circuit Description

In this preamplifier, two 6CW4 high-mu nuvistor triodes are used in a high-gain, low-noise cascode rf-amplifier stage that adds 25 to 35 dB of gain ahead of a receiver operated on the 6-, 10-, or 15-meter amateur band or on the 27-MHz citizens band. This added gain, together with the low noise figure (approximately 5 dB) of the preamplifier, substantially increases both the sensitivity and the signal-to-noise ratio of the receiver. The preamplifier operates from a dc plate supply of 150 volts at 5 milliamperes. The tube heaters require an ac power input of 6.3 volts at 0.26

26-6 PREAMPLIFIER FOR AMATEUR RECEIVER (Cont'd)



ALIGNMENT DATA		
Operating Frequency	**Tune T_1 to:**	**Tune T_2 to:**
21 MHz	21.25 MHz	21.22 MHz
27 MHz	30 MHz	27 MHz
30 MHz	32 MHz	29.5 MHz
50 MHz	51 MHz	50 MHz

Parts List

C_1, C_7 = See Note 1
C_2, C_3, C_4, C_5, C_6, C_8, C_9 = 0.001 μF, 500 V, ceramic
R_1, R_2 = 100 ohms, 0.5 watt
R_3 = 0.47 megohm, 0.5 watt
R_4 = 1000 ohms, 0.5 watt
T_1 = Input transformer (slug-tuned); matches preamplifier to 52-ohm input line (for 300-ohm input line, double number of turns in primary); wound from #32 copper enamel wire on slug-tuned form having ¼-inch outer diameter: primary, 1½ turns; secondary, 18 turns for operation at 21, 27, or 30 MHz or 10 turns for operation at 50 MHz
T_2 = Output transformer (slug-tuned); matches preamplifier to 72-ohm output lines (use of other than a 72-ohm line between preamplifier output and receiver input is not recommended); wound from #32 copper enamel wire on slug-tuned form having ¼-inch outer diameter; primary, 18 turns for operation at 21, 27, or 30 MHz or 10 turns for operation at 50 MHz, secondary, 1½ turns.

Notes: 1. For operation at 21 or 27 MHz, use 6.8-pF 500-volt capacitors for C_1 and C_7; for operation at 30 MHz, use 5-pF 500-volt capacitors for C_1 and C_7; for operation at 50 MHz, use 5-pF 500-volt capacitor for C_1 and 6.8-pF 500-volt capacitor for C_7.
2. See general considerations for construction of high-frequency and broadband circuits on page 561.

Circuit Description (Cont'd)

ampere. These small power requirements can usually be provided by the receiver.

Input transformer T_1 matches the high input impedance of the preamplifier to a 72-ohm or 300-ohm antenna. When a 72-ohm antenna is used, the primary of T_1 consists of a 1½-turn link wound about the hot end of the secondary coil. For a 300-ohm antenna, a 3-turn link is used. The secondary of T_1 is an 18-turn coil for operation at 10 or 15 meters or on the citizens band. At 6 meters, a 10-turn secondary coil is used. The unit is normally connected to the antenna cable by means of a coaxial connector. If a balanced antenna system is used, however, terminal strips for the twin leads may be used instead of the coaxial connector. In this latter case, the input link (primary of T_1) is not grounded.

Nuvistors V_1 and V_2 are operated in a stacked (cascode) arrangement in series with the B+ supply. The input is coupled by T_1 to the control grid of V_1, which is essentially a grounded-cathode amplifier. The output of V_1 is applied to the cathode of V_2, which is basically a grounded-grid amplifier. The inherent stability

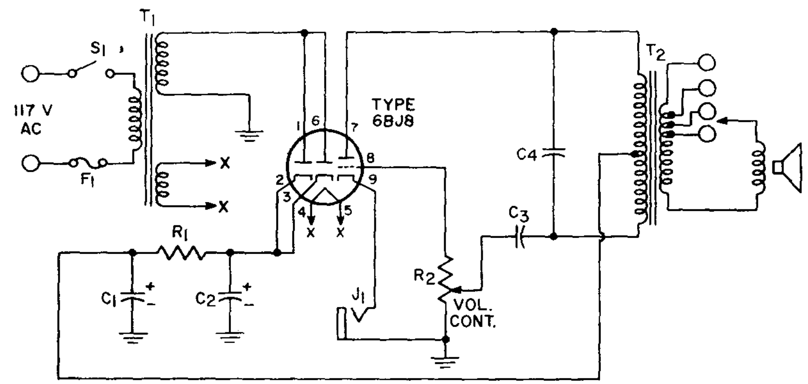
26-6 PREAMPLIFIER FOR AMATEUR RECEIVER (Cont'd)

Circuit Description (Cont'd)

of this type of arrangement, together with the ample decoupling and by-passing networks included in the circuit, provides assurance that the preamplifier will not break into oscillation.

The output of V_2 is developed across the primary coil of output transformer T_2. This coil is identical to the secondary coil of input transformer T_1. The secondary of T_2 consists of a 1½-turn link about the primary coil. This link matches the output of the preamplifier to a 75-ohm receiver input cable. (The maximum length of coaxial cable between receiver and preamplifier should not exceed 12 inches.)

26-7 CODE-PRACTICE OSCILLATOR



Note: Any two terminals of the secondary of T_2 that give the desired tone may be selected. Adjustment of volume control may cause a slight change in tone.

Parts List

C_1, C_2=20 µF, electrolytic, 150 V
C_3=0.001 µF, paper, 200 V
C_4=0.03 µF, paper, 200 V
F=⅛ ampere
J_1=Input jack for key
R_1=1500 ohms, 1 watt
R_2=Potentiometer, 0.1 megohm, 0.5 watt
T_1=Power transformer, 125 volts rms, 15 ma; 6.3 volts, 0.6 ampere
T_2=Output transformer, universal

Circuit Description

This code-practice oscillator operates from a 117-volt ac power line. When ON-OFF switch S_1 is closed, the 117-volt ac input power is stepped up to 125 volts across the upper secondary winding of power transformer T_1 and is stepped down to 6.3 volts across the lower secondary winding. The 6.3-volt winding provides the operating power for the heater of the 6BJ8 twin diode-tride used in the circuit. The diode sections of the 6BJ8 are connected to operate as a single diode in a half-wave rectifier circuit that converts the ac power across the 125-volt winding of T_1 to dc operating power for the 6BJ8 triode section. This triode section is used as the amplifier tube in a simple audio-oscillator stage.

Operation of the oscillator stage is controlled by a telegraph key, which is connected into the circuit by means of jack J_1. When the key is closed, the triode section of the 6BJ8 supplies energy to the oscillator resonant circuit formed by capacitor C_4 and the effective inductance of the primary of output transformer T_2. This circuit then resonates to pro-

26-7 CODE-PRACTICE OSCILLATOR (Cont'd)

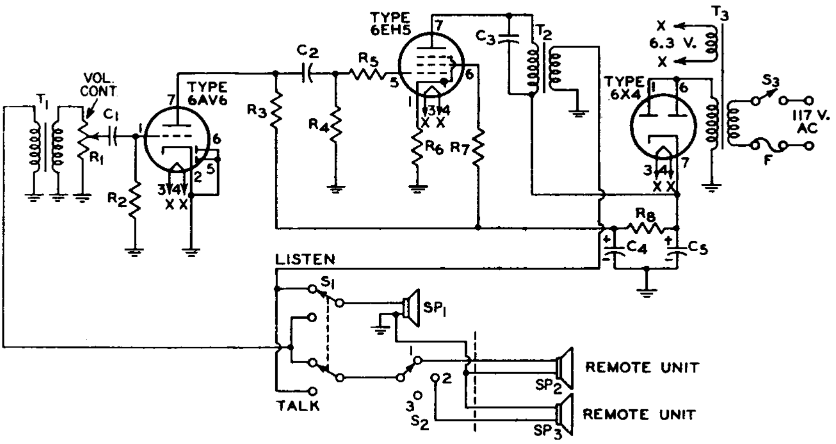
Circuit Description (Cont'd)

duce an audio signal that is coupled by transformer T_2 to the speaker to produce an audible indication of the keying. Positive feedback to sustain oscillation is developed by the auto-transformer action of the tapped primary of transformer T_2.

Output transformer T_2 is a universal type which contains multiple taps on the secondary winding. These taps enable the transformer to match the oscillator output impedance to different values of speaker voice-coil impedance. The speaker impedance and transformer terminals used, however, affect the effective inductance in the primary of T_1 and, thus, the tone of the audio output. Volume-control potentiometer R_2 adjusts the level of the audio output. Adjustment of potentiometer R_2 varies the loading on the oscillator resonant circuit and may also cause a slight change in the tone of the audio output.

26-8 INTERCOMMUNICATION SET

With Master Unit and Two or More Remote Units



Notes: 1. The leads from the LISTEN-TALK switch S_1 to T_1 and T_2 should be kept as far apart as possible to prevent undesirable regenerative effects.
2. Connections to the remote speaker units should be made with low-resistance wire, preferably with shielded "intercom" cable.

Parts List

C_1, C_2=0.0022 μF, paper, 200 V.
C_3=0.005 μF, paper, 200 V.
C_4, C_5=60 μF, electrolytic, 150 V.
F_1=Fuse, 1 ampere
R_1=Volume control, potentiometer, 0.5 megohm, audio taper, attached to switch S_3
R_2=6.8 megohms, 0.5 watt
R_3, R_4=0.47 megohm, 0.5 watt
R_5=10000 ohms, 0.5 watt
R_6, R_7=68 ohms, 0.5 watt
R_8=2200 ohms, 1 watt
S_1=Talk-listen switch, double-pole, double-throw
S_2=Station Selector, rotary switch
S_3=On-off switch, single-pole, single-throw; attached to volume-control potentiometer
SP_1, SP_2, SP_3=Speaker; permanent-magnet; voice-coil impedance, 3 to 4 ohms
T_1=Input transformer, 4-ohm primary, 25000-ohm secondary, Knight 54A1492 or equiv.
T_2=Output transformer, 3000-ohm primary, 4-ohm secondary, Knight 54A2371 or equiv.
T_3=Power transformer, 125 volts rms, 50 mA., 6.3 volts rms. 2 amperes, Knight 54A1411 or equiv.

26-8 INTERCOMMUNICATION SET (Cont'd)

Circuit Description

This simple "intercom" set can be used to achieve reliable voice communications, at normal speaking levels, between any two points in a normal-size house. The system consists of a master unit, centrally located at the hub of household activity, interconnected by low-loss cabling to remote units located at points (e.g., garage, attic, and cellar) beyond the range of normal voice levels. An audio amplifier, which includes a 6AV6 voltage-amplifier stage and a 6EH5 power-output stage, provides the amplification necessary to overcome the attenuation of voice levels by system cabling. A 6X4 half-wave rectifier circuit converts the 117-volt ac input power to the dc power required for operation of the amplifier stages. A 6.3-volt secondary winding on the power transformer (T_3) in the rectifier circuit provides heater power for the amplifier and rectifier tubes.

The speaker at each intercom station is used for both talk and listen functions. The talk-listen switch S_1 at the master location establishes the talk or listen mode for all stations. The voice communications are initiated from the master unit. Switch S_1 is depressed to the TALK position, and the initiator talks into the master-unit speaker. The audio (voice-signal) voltage that is then developed across the speaker voice coil is coupled by input transformer T_1 to the control grid of the 6AV6 audio amplifier. Selector switch S_2 connects the desired remote unit into the intercom system. With S_1 depressed to the TALK position, the remote unit speaker is automatically connected to the audio amplifier output for listen-mode operation. When S_1 is in the LISTEN position, the master-unit speaker is connected in the listen mode, and the remote-unit speaker is connected to the amplifier input. A reply from the remote unit is then coupled from the remote speaker by transformer T_1 to the control grid of the 6AV6 audio amplifier.

Transformer T_1 matches the voice-coil impedance of the 4-ohm permanent-magnet speaker (of either master or remote unit) to the 25000-ohm input impedance of the 6AV6 amplifier stage. This stage and the 6EH5 audio output stage amplify the audio (voice) signals received from one location (the master unit or one of the remote units) to develop the audio power required to produce an audible output from the speaker at another location. Output transformer T_2 matches the 3000-ohm plate-circuit impedance of the output stage to the 4-ohm voice-coil impedance of the speaker (master-unit or remote-unit) to which the communication is directed, as determined by the settings of switches S_1 and S_2. The VOL. CONT. potentiometer R_1 in the input circuit of the 6AV6 audio amplifier stage provides the volume-control adjustment for the system.

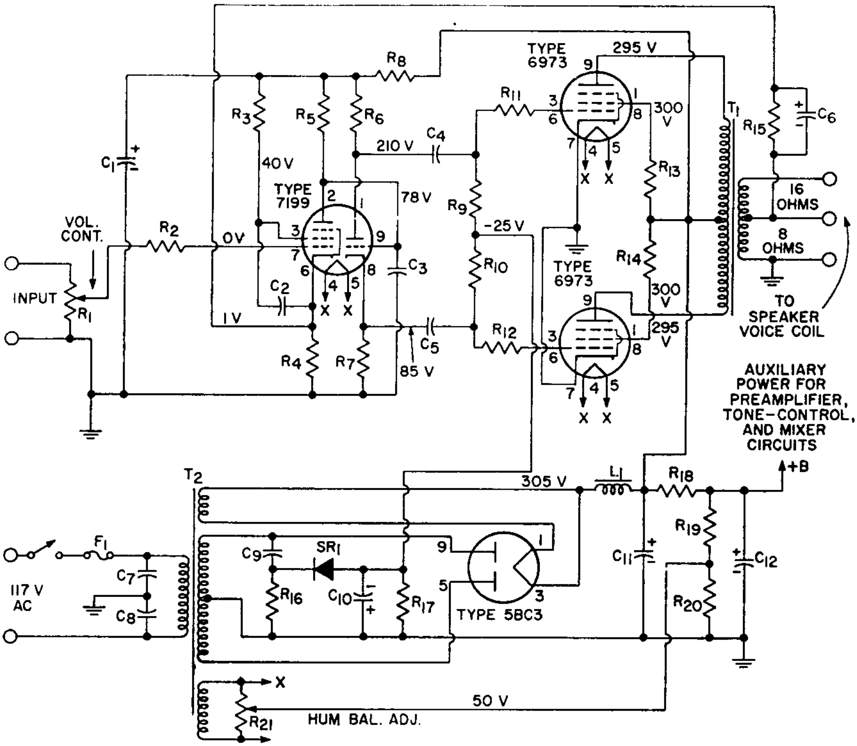
26-9 HIGH-FIDELITY AUDIO AMPLIFIER

Class AB_1; Power Output, 15 Watts

Circuit Description

This high-fidelity audio power amplifier can deliver 15 watts of rms output power with less than 0.4 per cent total harmonic distortion and less than 1.5 per cent intermodulation distortion. The frequency response of the amplifier is flat within ± 0.5 dB from 20 Hz to 60 kHz, and the sensitivity is such that the rated output of 15 watts is obtained for an input of 1.2 volts rms. The total hum and noise, with the input shorted, is 84 dB below 15 watts. The circuit operates from a 117-volt ac power line. The transformer-coupled ac input power is converted to dc operating

26-9 HIGH-FIDELITY AUDIO AMPLIFIER (Cont'd)



Parts List

- C_1=40 μF, electrolytic, 450 V.
- C_2, C_4, C_5=0.25 μF, paper, 400 V.
- C_3=3.3 pF, ceramic or mica, 600 V.
- C_6=150 pF, ceramic or mica, 400 V.
- C_7, C_8=0.05 μF, paper, 400 V.
- C_9=0.02 μF, paper 600 V.
- C_{10}=100 μF, electrolytic, 50 V.
- C_{11}=80 μF, electrolytic, 450 V.
- C_{12}=40 μF, electrolytic, 450 V.
- F_1=Fuse, 3 amperes
- L_1=Choke, 3 H, 160 mA, dc resistance 75 ohms or less, Triad C-13X or equiv.
- R_1=Volume control, potentiometer, 1 megohm
- R_2=10000 ohms, 0.5 watt
- R_3=0.82 megohm, 0.5 watt
- R_4=820 ohms, 0.5 watt
- R_5=0.22 megohm, 0.5 watt
- R_6, R_7=15000 ohm ±5 per cent, 2 watts
- R_8=3900 ohms, 2 watts
- R_9, R_{10}=0.1 megohm, 0.5 watt
- R_{11}, R_{12}=1000 ohms, 0.5 watt
- R_{13}, R_{14}=100 ohms, 0.5 watt
- R_{15}=8200 ohms, 0.5 watt
- R_{16}=15000 ohms, 1 watt
- R_{17}=68000 ohms, 0.5 watt
- R_{18}=4700 ohms, 2 watts
- R_{19}=0.27 megohm, 1 watt
- R_{20}=47000 ohms, 0.5 watt
- R_{21}=Hum balance adjustment, potentiometer, 100 ohms, 0.5 watt
- SR_1=Selenium rectifier, 20 mA, 135 volts rms
- T_1=Output transformer, (having 8-ohm tap for feedback connection) for matching impedance of voice coil to 6600-ohm plate-to-plate tube load; 50 watts; frequency response, 10 to 50000 Hz; Stancor A-8056 or equiv.
- T_2=Power transformer, 360-0-360 volts rms, 120 mA; 6.3 V., 3.5 A; 5 V., 3 A; Stancor 8410 or equiv. (see Note 1)

Notes: 1. For stereo operation from a single power supply, the power transformer T_2 must be replaced by one that has a higher current rating. A Stancor Type 6315 or equivalent (370-0-370 volt rms, 275 mA) is recommended.

2. If the amplifier oscillates or "motorboats," reverse ground and feedback connections in secondary of output transformer T_1.

26-9 HIGH-FIDELITY AUDIO AMPLIFIER (Cont'd)

Circuit Description (Cont'd)

power for the amplifier stages by the 5BC3 full-wave rectifier. Heater power for the amplifier tubes and the rectifier are obtained from the 6.3-volt and 5-volt secondary windings, respectively, on the rectifier power transformer (T_2).

A high-gain pentode voltage amplifier is used as the input stage for the audio power amplifier. The output of this stage is direct-coupled to the control grid of a triode split-load type of phase inverter. The use of direct coupling between these stages minimizes phase shift and, consequently, increases the amount of inverse feedback that may be used without danger of low-frequency instability. A low-noise 7199 tube, which contains a high-gain pentode section and a medium-mu triode section in one envelope, fulfills the active-component requirement for both the pentode input stage and the triode phase inverter. Potentiometer R_1 in the input circuit of the 7199 pentode section is the volume control for the amplifier.

The plate and cathode outputs of the phase inverter, which are equal in amplitude and opposite in phase, are used to drive a pair of pentode-connected 6973 beam-power tubes used in a class AB_1 push-pull output stage. The 6973 output tubes are biased for class AB_1 operation by the fixed negative voltage applied to the control-grid circuit from the rectifier circuit. Fixed bias is used because a class AB amplifier provides highest efficiency and least distortion for this bias method.

Transformer T_1 couples the audio-amplifier output to the speaker. The taps on the secondary of this transformer match the plate-to-plate impedance of the output stage to the voice-coil impedance of an 8- or 16-ohm speaker. Negative feedback of 19.5 dB is coupled from the secondary of the output transformer (speaker voice coil) to the cathode of the input stage to reduce distortion and to improve circuit stability.

Fixed-bias operation of the output stage requires that the power supply provide very good voltage regulation because the plate current of the 6973 tubes varies considerably with the signal level. The conventional choke-input type of power supply used provides the required regulation. The fixed bias for the output stage is obtained from one-half the high-voltage secondary winding of power transformer T_2 through a capacitance-resistance voltage divider and the 20-milliampere, 135-volt selenium rectifier. Potentiometer R_{21} connected across the 6.3-volt secondary winding of transformer T_2 provides a hum balance adjustment for the audio power amplifier. The wiper arm of this potentiometer is connected to the junction of a resistive voltage divider across the output of the power supply. The resulting positive bias voltage applied to the tube heaters minimizes heater-to-cathode leakage and substantially reduces hum.

26-10 HIGH-FIDELITY AUDIO AMPLIFIER

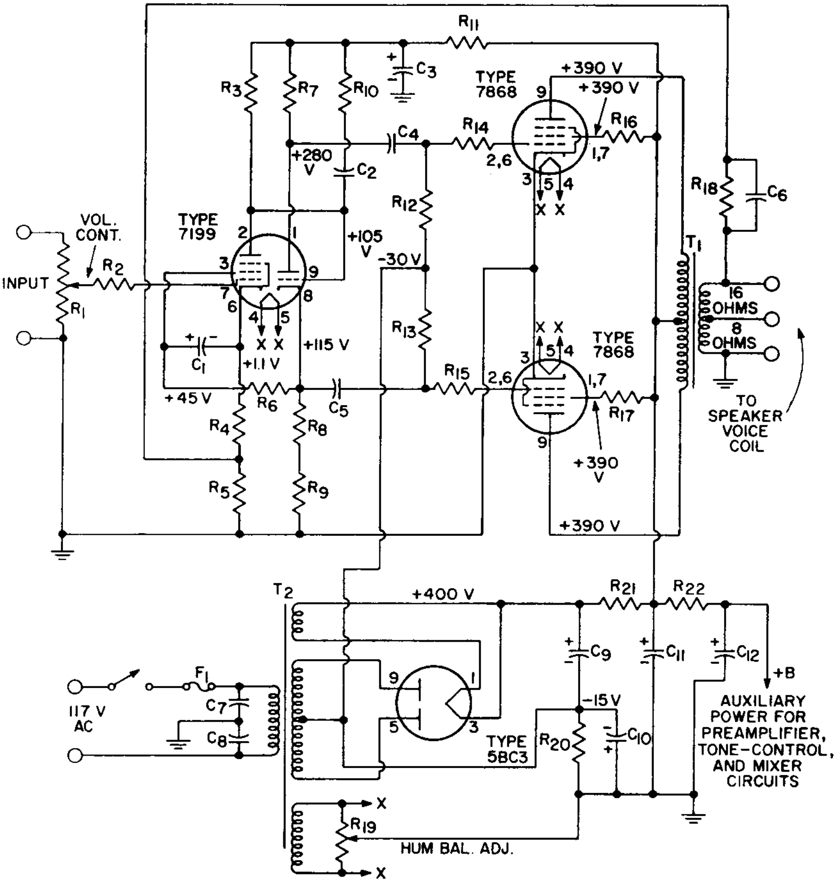
Class AB_1; Power Output, 30 Watts

Circuit Description

This audio power amplifier can deliver 30 watts of rms output power with less than 0.7 per cent total harmonic distortion and less than 1.5 per cent intermodulation distortion. The frequency response of the amplifier is flat within ±0.5 dB from 15 Hz to 40 kHz. The total hum and noise, with the input shorted, is 85 dB below 30 watts. The rated output of 30 watts is obtained for an input of 1 volt rms.

The 30-watt amplifier is essentially identical to the 15-watt ampli-

26-10 HIGH-FIDELITY AUDIO AMPLIFIER (Cont'd)



Parts List

C_1=25 μF, electrolytic, 50 V
C_2=22 pF, ceramic or mica, 600 V
C_3=80 μF, electrolytic, 450 V
C_4, C_5=0.25 μF, paper, 600 V
C_6=0.01 μF, paper, 600 V
C_7, C_8=0.05 μF, paper, 600 V
C_9, C_{11}=40 μF, electrolytic, 500 V
C_{10}=100 μF, electrolytic, 50 V
C_{12}=20 μF, electrolytic, 450 V
F_1=Fuse, 3 amperes, 150 V
R_1=Volume control, potentiometer, 1 megohm
R_2=10000 ohms, 0.5 watt
R_3=0.22 megohm, 0.5 watt
R_4=820 ohms, 0.5 watt
R_5=10 ohms, 0.5 watt
R_6=0.18 megohm, 0.5 watt
R_7, R_8=15000 ohms $\pm$5 per cent, 2 watts
R_9=1000 ohms, 0.5 watt
R_{10}=22000 ohms, 0.5 watt
R_{11}=2000 ohms, 2 watts
R_{12}, R_{13}=0.1 megohm, 0.5 watt
R_{14}, R_{15}=1000 ohms, 0.5 watt
R_{16}, R_{17}=56 ohms, 0.5 watt
R_{18}=270 ohms, 0.5 watt
R_{19}=Hum balance adjustment, potentiometer, 100 ohms, 0.5 watt
R_{20}=120 ohms, 100 watts
R_{21}=50 ohms, 10 watts
R_{22}=10000 ohms, 2 watts
T_1=Output transformer (having 16-ohm tap for feedback connection) for matching impedance of voice coil to 6600-ohm plate-to-plate tube load; 50 watts; frequency response, 10 to 50000 Hz; Stancor A-8056 or equiv.
T_2=Power transformer, 375-0-375 volts rms, 160 mA; 6.3 V., 5 A; 5 V., 3 A; Thordarson type T22R33 or equivalent (see Note 1).

Notes: 1. For stereo operation from a single power supply, the power transformer T_2 must be replaced by one that has a higher current rating. A Stancor Type 6315 or equivalent (370-0-370 volts rms, 275 mA) is recommended.
2. If amplifier oscillates or "motorboats," reverse ground and feedback connections in secondary of output transformer T_1.

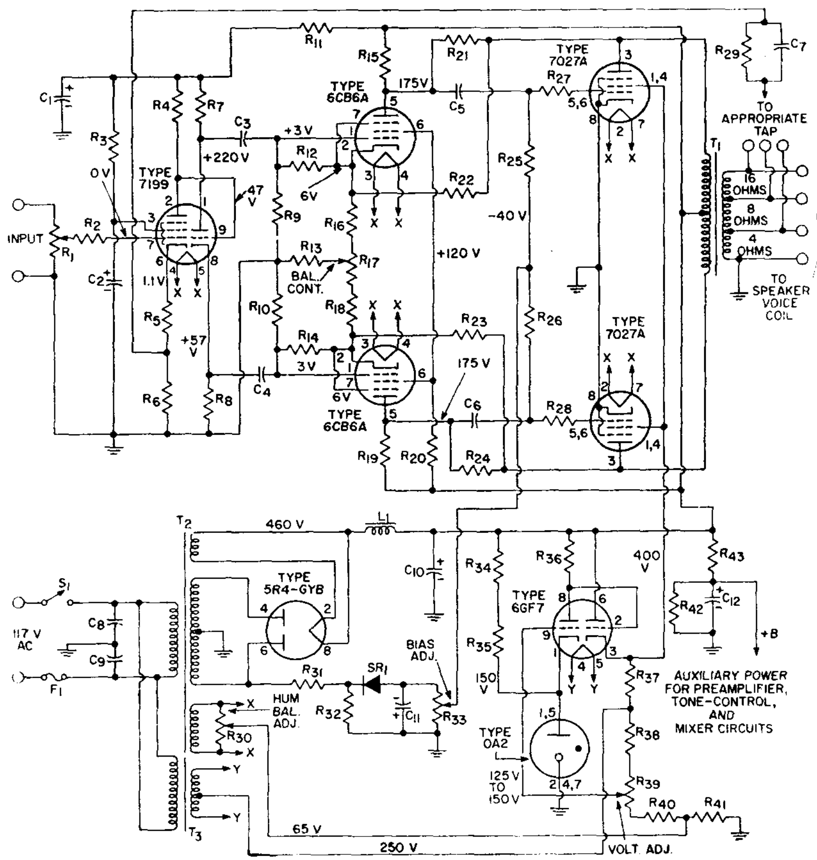
26-10 HIGH-FIDELITY AUDIO AMPLIFIER (Cont'd)

Circuit Description (Cont'd)

fier (circuit 26-9) except that it uses 7868 beam power tubes in the output stage to develop the higher audio power output and uses a resistive network in the negative leg of the power supply, rather than a separate rectifier, to supply the fixed-bias voltage for the output stage. A potentiometer (R_{19}) connected across the 6.3-volt heater winding also provides the hum balance adjustment for the 30-watt amplifier.

26-11 HIGH-FIDELITY AUDIO AMPLIFIER

Class AB_1; Power Output, 50 Watts



Preliminary Adjustments

The following adjustments should be made before operation:

(1) With rectifier out of socket, adjust Bias Adj. R_{33} for —40 volts between the wiper arm and ground bus.

(2) With speaker connected, adjust Screen-Grid Voltage Adj. R_{39} for 400 volts between pin 3 of 6GF7 and ground bus.

(3) With input shorted, adjust Hum Bal. Adj. R_{30} for minimum hum from speaker.

(4) With input open and Vol. Cont. set for maximum volume, adjust Bal. Cont. R_{17} for minimum hum from speaker.

26-11 HIGH-FIDELITY AUDIO AMPLIFIER (Cont'd)

Parts List

C_1, C_2=40 µF, electrolytic, 450 V
C_3, C_4=0.02 µF, paper, 400 V
C_5, C_6=1 µF, paper, 400 V
C_7=0.002 µF to 4-ohm tap; 0.0015 µF to 8-ohm tap; or, 0.001 µF to 16-ohm tap; paper, 400 V
C_8, C_9=0.05 µF, paper, 600 V
C_{10}=20 µF, electrolytic, 450 V
F_1=Fuse, 5 amperes
L_1=Choke, 8 H, 250 mA, dc resistance 60 ohms, or less
R_1=Volume control, potentiometer, 0.5 megohm
R_2=4700 ohms, 0.5 watt
R_3=0.82 megohm, 0.5 watt
R_4=0.22 megohm, 0.5 watt
R_5=820 ohms, 0.5 watt
R_6=10 ohms, 0.5 watt
R_7 R_8= 15000 ohms, 2 watts
R_9, R_{10}=1.5 megohms, 0.5 watt
R_{11}=33000 ohms, 2 watts
R_{12}, R_{14}=1.3 megohms, 0.5 watt
R_{13}=47 ohms, 0.5 watt
R_{15}, R_{19}=0.15 megohm, 0.5 watt
R_{16}, R_{18}=390 ohms, 0.5 watt
R_{17}=AC balance control, potentiometer, 500 ohms
R_{20}=0.15 megohm, 1 watt
R_{21}, R_{24}=0.33 megohm, 1 watt
R_{22}, R_{23}=0.12 megohm, 2 watts
R_{25}, R_{26}=0.1 megohm, 0.5 watt
R_{27}, R_{28}=47000 ohms, 0.5 watt
R_{29}=600 ohms to 4-ohm tap; 820 ohms to 8-ohm tap; or, 1200 ohms to 16-ohm tap; 0.5 watt
R_{30}=Hum balance adjustment, potentiometer, 100 ohms
R_{31}=0.12 megohm, 5 watts
R_{32}, R_{34}, R_{35}, R_{37}=33000 ohms, 2 watts
R_{33}=Bias adjustment, potentiometer 5000 ohms,
R_{38}=10000 ohms, 1 watt
R_{39}=Screen-grid voltage adjustment, potentiometer, 25000 ohms, 2 watts
R_{40}=15000 ohms, 2 watts
R_{41}=12000 ohms, 2 watts
R_{42}=0.22 megohm, 2 watts
R_{43}=22000 ohms, 2 watts
SR_1=Selenium rectifier, 20 mA, 135 volts rms
T_1=Output transformer for matching impedance of voice coil to 5000-ohm plate-to-plate tube load; 50 watts; frequency response, 10 to 50000 Hz; United Transformer Corp. LS55 or equiv.
T_2=Power transformer, 600-0-600 volts rms, 200 mA, 6.3 V., 5 A; 5 V., 3 A; Thordarson 22R36 or equiv. (see Note 1)
T_3=Filament transformer, 6.3 volts, center tapped, 1 ampere; Thordarson 21F08 or equiv.

Notes: 1. For stereo operation from a single power supply, the following changes are required: (a) The power transformer T_2 must be replaced by one that has a higher current rating; a United Transformer Corporation Type H-93 or equivalent (600-0-600 volts rms, 300 mA) is recommended. (b) The 5000-ohm Bias Adj. potentiometer R_{33} should be replaced by two 10000-ohm potentiometers (one for each channel) connected in parallel. (c) A second 5R4-GYB rectifier tube should be connected in parallel with the one used for monaural operation. (Connect the 5R4-GYB tubes so that the two sections of each tube are in parallel with the corresponding sections of the other tube; do not use separate tubes for each section of the rectifier circuit.)

2. If the amplifier oscillates or "motorboats," reverse ground and feedback connections in secondary of output transformer T_1.

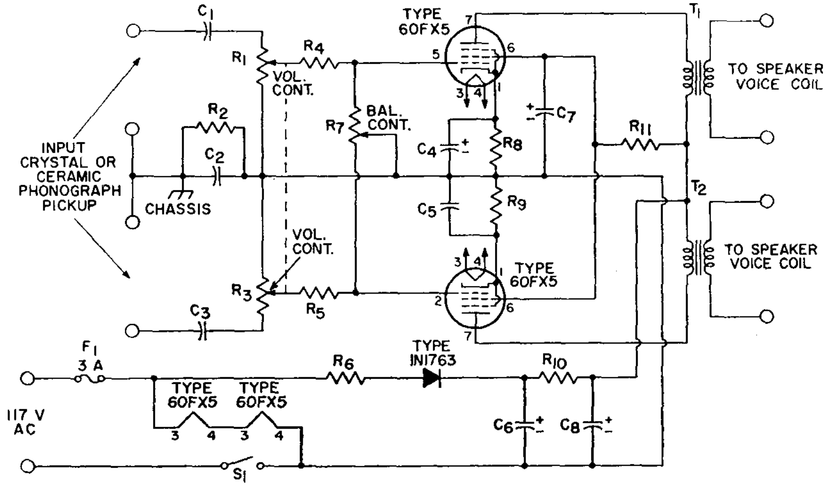
Circuit Description

This four-stage audio power amplifier can deliver 50 watts of rms power output with less than 0.1 per cent total harmonic distortion and less than 1 per cent intermodulation distortion. The frequency response of the amplifier is flat within ±0.5 dB from 10 Hz to 50 kHz. Sensitivity is 0.4 volt rms input for 50 watts output. The total hum and noise is 70 dB below 50 watts.

The 50-watt amplifier, like the 15-watt and 30-watt high-fidelity amplifiers (circuits 26-9 and 26-10), uses a 7199 low-noise triode-pentode as an input amplifier and phase-splitter, but has a push-pull driver stage, which uses 6CB6 sharp-cutoff pentodes. The superior performance of this amplifier can also be attributed, in part, to the use of a 450-volt plate supply and a 400-volt electronically regulated grid-No. 2 supply for the 7027A beam power tubes in the output stage and to the use of inverse-feedback loops from the plates to the grids of the output tubes, from the plates of the output tubes to the cathodes of the driver tubes, and from the voice-coil winding of the output transformer to the cathode of the input amplifier. Additional features are the operation of all heaters at a positive voltage with respect to ground and use of a balancing adjustment (R_{30}) in the heater-supply circuit to minimize hum, a grid-No. 2 voltage adjustment (R_{39}), a grid-No. 1 bias adjustment (R_{33}) for the 7027A output tubes, and an ac-balance adjustment (R_{13}) which may be used to balance the outputs of the push-pull stages. Operation of the 50-watt amplifier is essentially the same as that of the 15- and 30-watt amplifiers.

26-12 TWO-CHANNEL STEREOPHONIC AMPLIFIER

Power Output, 1 Watt Each Channel



Parts List

C_1, C_3=0.22 μF, paper, 400 V
C_2=0.1 μF, paper, 400 V
C_4, C_5=50 μF, electrolytic, 25 V
C_6=50 μF, electrolytic, 150 V
C_7, C_8=50 μF ,electrolytic, 150 V
F_1=Fuse, 3 amperes
R_1, R_2=Volume control, potentiometer, 1.5 megohms, ganged, audio taper
R_3, R_4=47000 osms, 0.5 watt
R_5=Balance control, potentiometer, 2 megohms audio taper
R_6, R_7=60 ohms, 1 watt
R_8=220 ohms, 2 watts
R_9=280 ohms, 2 watts
R_{10}=12 ohms, 1 watt
R_{11}=0.22 megohm, 0.5 watt
S_1=ON-OFF switch, single-pole, single-throw
T_1 T_2=Output transformer for matching impedance of voice coil to 3000-ohm tube load; Triad S-16X or equiv.

Circuit Description

This ac/dc two-channel (stereo) amplifier operates from either an ac power line or dc supply of 117 volts. AC power inputs are converted to dc power by the 1N1763 silicon-diode half-wave rectifier circuit. The heaters of the 60FX5 power pentodes (one for each channel) used in the amplifier are connected in series directly across the input power line.

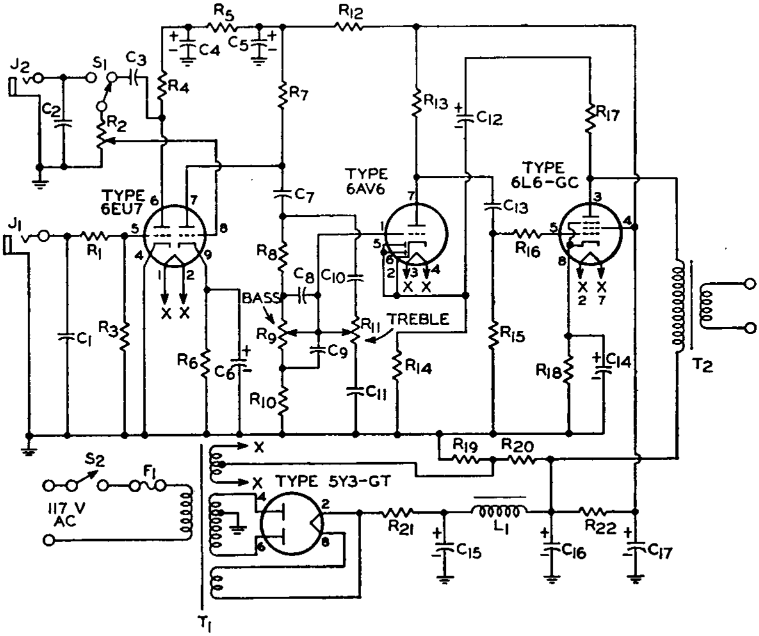
In stereo units that use high-output ceramic stereo cartridges, the high power sensitivity of the 60FX5 tubes at low supply voltage eliminates the need for preamplifier stages. The 60FX5 provides a power output of 1.3 watts to a 3000-ohm transformer primary with only 3 volts peak drive on grid No. 1. With a transformer having a good impedance match and 85-per-cent efficiency, each channel of the stereo amplifier supplies 1.1 watts of useful power output at the speaker.

No special mounting or layout precautions are necessary for this amplifier other than the value and placement of the isolating capacitor C_2 between B— and the chassis. This capacitor should be connected to the same point on the chassis at which the common cartridge lead is tied. A value of 0.1 microfarad for the isolating capacitor is suggested so that full output is obtained from the pickup.

As with all single-ended amplifier circuits, especially ac/dc units, adequate screen-grid bypassing is necessary to minimize hum. Screen-grid filtering is obtained through use of a 220-ohm dropping resistor R_9 and a 50-microfarad electrolytic capacitor C_6. Although, in the circuit shown, separate cathode-bias resistors are used for better dynamic balance, a single 30-ohm common cathode-bias resistor bypassed with a 50-microfarad electrolytic capacitor may also be used.

26-13 MICROPHONE AND PHONOGRAPH AMPLIFIER

Power Output, 8 Watts



Parts List

C_1, C_2=100 pF, disc-ceramic, 300 V
C_3=0.05 μF, paper, 200 V
C_4=8 μF, electrolytic, 450 V
C_5=16 μF, electrolytic, 450 V
C_6=25 μF, electrolytic, 450 V
C_7=0.1 μF, paper, 200 V
C_8=0.001 μF, disc-ceramic, 300 V
C_9=0.01 μF, disc-ceramic, 300 V
C_{10}=470 pF, disc-ceramic, 300 V
C_{11}=4700 pF, disc-ceramic, 300 V
C_{12}=4 μF, electrolytic, 450 V
C_{13}=0.05 μF, paper, 600 V
C_{14}=25 μF, electrolytic, 25 V
C_{15}, C_{16}, C_{17}=20 μF, electrolytic, 450 V
F_1=Fuse, 1 ampere
J_1=Jack for high-impedance crystal microphone input; max. input: 2 millivolts peak
J_2=Jack for crystal phonopickup input
L_1=Filter choke, 5 H, 200 mA, United Transformer Corp. R20 or equiv.
R_1, R_{16}=10000 ohms, 0.5 watt
R_2=Volume Control, potentiometer, 1 megohm
R_3=2.2 megohms, 0.5 watt
R_4, R_8, R_{20}=0.22 megohm, 0.5 watt
R_5=27000 ohms, 0.5 watt
R_6=1200 ohms, 0.5 watt
R_7, R_{13}=0.1 megohm, 0.5 watt
R_9, R_{11}=Tone control, potentiometer, 0.5 megohm
R_{10}=22000 ohms, 0.5 watt
R_{12}=12000 ohms, 0.5 watt
R_{14}=1800 ohms, 0.5 watt
R_{15}=0.47 megohm, 0.5 watt
R_{17}=0.15 megohm, 0.5 watt
R_{18}=180 ohms, 2 watts
R_{19}=47000 ohms, 1 watt
R_{21}=50 ohms, 10 watts
R_{22}=8200 ohms, 2 watts
S_1=Microphone-phonograph selector; wafer switch; single-pole, double-throw
S_2=ON-OFF switch, single-pole, single-throw
T_1=Power transformer, 300-0-300 V., 90 mA.; 6.3 V., 3.5 A., center tapped; 5 V., 2 A, Thordarson 22R04 or equiv.
T_2=Output transformer for matching impedance of voice coil to 4000-ohm tube load; 10 watts; United Transformer Corp. S14 or equiv.

Circuit Description

This microphone and phonograph amplifier can deliver up to 8 watts of audio output power for an input of 200 millivolts rms at J_2 (phonograph input) or an input of 6.8 millivolts rms at J_1 (microphone input). The amplifier uses a 6EU7 twin-triode input amplifier, a 6AV6 driver stage, and a 6L6GC single-ended output stage to increase the signal power from a high-impedance crystal microphone or crystal phonograph pickup to the desired level. The transformer-coupled ac input power is converted to dc operating power for these stages by a 5Y3GT full-wave recti-

26-13 MICROPHONE AND PHONOGRAPH AMPLIFIER (Cont'd)

Circuit Description (Cont'd)

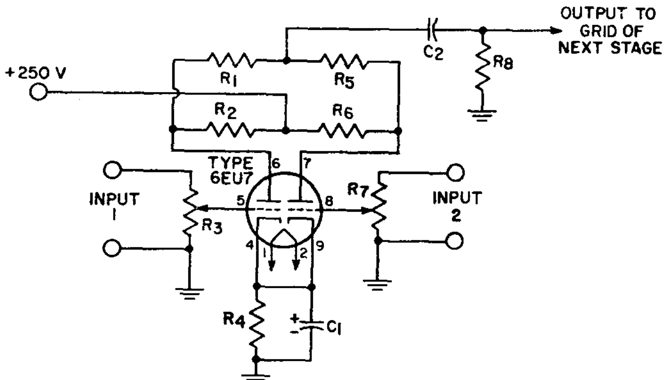
fier circuit. A 5-volt winding on power transformer T_1 provides the heater power for the rectifier tube, and a 6.3-volt winding provides heater power for the other tubes in the amplifier. The center tap on the 6.3-volt winding is connected to the junction of a resistive voltage divider (R_{19} and R_{20}) across the output of the power supply. The resulting positive bias applied to the tube heaters substantially reduces heater-to-cathode leakage and, consequently, minimizes hum.

The signals from a crystal microphone are usually much smaller than those from a crystal phonograph pickup. Microphone signals, therefore, are amplified by both sections of the 6EU7 twin-triode amplifier. The signals are coupled from J_1 to the pin 5 control grid of the 6EU7. The plate output from this triode section is then coupled through switch S_1 (microphone position) and volume-control potentiometer R_2 to the pin 8 control grid of the 6EU7. With selector switch S_1 in the phonograph position, phonograph inputs are coupled directly from J_2 across volume-control potentiometer R_2 to the pin 8 control grid, and the first section of the 6EU7 is bypassed.

The outputs from the pin 7 plate of the 6EU7 are coupled across the frequency-sensitive tone-control network to the control grid of the 6AV6 driver stage. The bass and treble controls R_9 and R_{11} are adjusted to assure optimum low- and high-frequency response characteristics for the amplifier. The two diode plate sections of the 6AV6 are shorted to the tube cathode and thereby are made inoperative. The output of the driver stage is applied to the 6L6GC output stage which develops the audio power required to drive a speaker. Transformer T_2 matches the 4000-ohm plate impedance of the output stage to the speaker voice-coil impedance.

26-14 TWO-CHANNEL AUDIO MIXER

Voltage Gain from Each Grid of 6EU7 to Output is Approximately 20



Parts List

C_1=10 μF, electrolytic, 25 V
C_2=0.05 μF, paper, 400 V
R_1, R_5, R_8=1 megohm, 0.5 watt
R_2 R_6=0.1 megohm, 0.5 watt
R_3, R_7=Potentiometers, 0.1 megohm, audio taper
R_4=1200 ohms, 0.5 watt

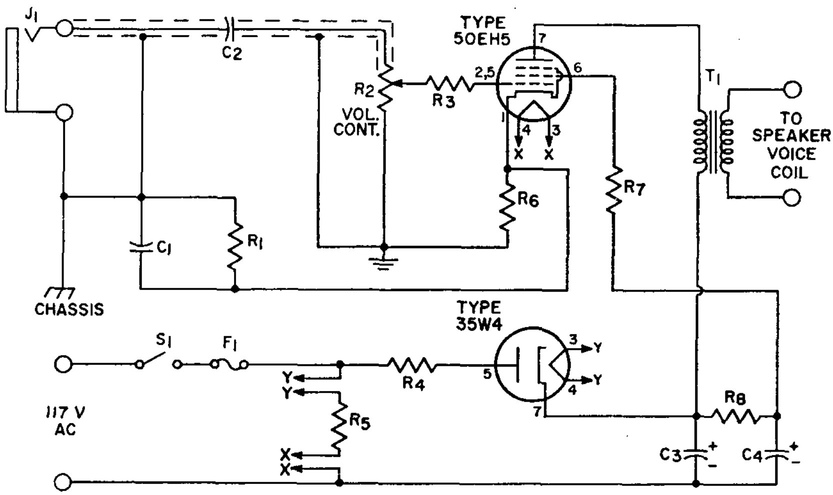
26-14 TWO-CHANNEL AUDIO MIXER (Cont'd)

Circuit Description

This high-fidelity mixer circuit can be used to combine audio-frequency program material from two sources. Each signal channel consists of a one-stage voltage amplifier using one section of a 6EU7 low-noise twin-triode. Each section of the mixer can provide a voltage gain of about 20, and can handle an input signal of about 0.2 volt rms without overloading. The dc plate supply of +250 volts (nominal value) for the mixer stages can usually be obtained from an auxiliary tap on the power supply for the audio power amplifiers.

26-15 PHONOGRAPH AMPLIFIER

Power Output, 1 Watt



Parts List

C_1=0.082 μF, paper, 400 V
C_2=0.02 μF, paper, 400 V
C_3, C_4=40 μF, electrolytic, 150 V
F_1=Fuse, 1 ampere
J_1=Input connector, shielded, for crystal phonograph pickup
R_1=0.22 megohm, 0.5 watt
R_2=Volume control, potentiometer, 0.5 megohm, audio taper
R_3=10000 ohms, 0.5 watt
R_4=22 ohms, 0.5 watt
R_5=210 ohms, 10 watts
R_6, R_7=56 ohms, 0.5 watt
T_1=Output transformer for matching impedance of voice coil to 3000-ohm tube load

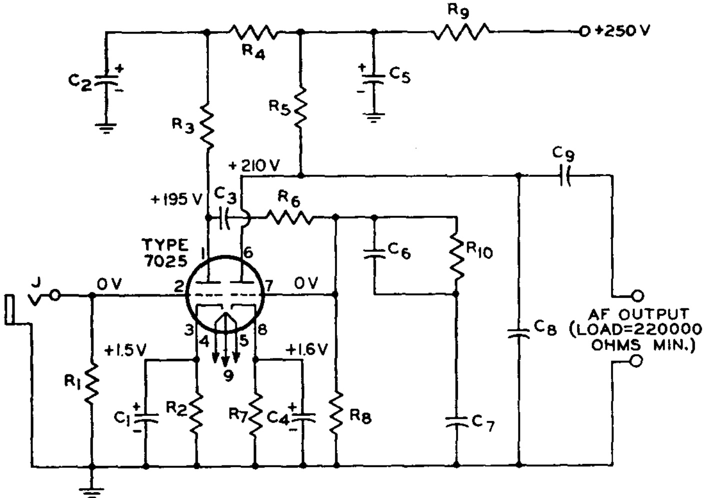
Circuit Description

This single-stage phonograph amplifier operates directly from either an ac power line or a dc supply of 117 volts. AC power inputs are converted to dc power by the 35W4 half-wave rectifier circuit. The heaters of the amplifier and rectifier tube are connected in series, together with a 210-ohm voltage-dropping resistor, R_5, directly across the input power line.

The amplifier uses a 50EH5 power pentode to develop up to 1 watt of audio output power from the input supplied from a crystal phonograph pickup. The input is applied at J_1 and coupled through a length of shielded cable to the input circuit of the pentode amplifier. Volume-control adjustment for the amplifier is provided by potentiometer R_2. The output coupling transformer T_1 matches the 3000-ohm plate load impedance of the 50EH5 to the voice-coil impedance of the speaker.

26-16 PREAMPLIFIER FOR MAGNETIC PHONOGRAPH PICKUP

With RIAA Equalization



Sensitivity=3 millivolts rms input for output of 0.55 volt at frequency of 1000 Hz.

Parts List

C_1, C_4=25 μF, electrolytic, 25 V
C_2, C_5=20 μF, electrolytic, 450 V
C_3=0.1 μF, paper, 600 V
C_6=0.0033 μF±5 per cent, paper, 600 V
C_7=0.01 μF±5 per cent, paper, 600 V
C_8=180 pF±5 per cent, ceramic or mica, 500 V (includes capacitance of output cable)
C_9=0.22 μF, ceramic, 500 V
J=Input connector, shielded, for high-impedance magnetic phono pickup (10 mV output, approx.)
R_1=Value depends on type of magnetic pickup used. Follow pickup manufacturer's recommendations
R_2, R_7=2700 ohms, 0.5 watt
R_3, R_5=0.1 megohm, 0.5 watt
R_4=39000 ohms, 0.5 watt
R_6=0.47 megohm, 0.5 watt
R_8=0.68 megohm, 0.5 watt
R_9=15000 ohms, 1 watt
R_{10}=22000 ohms, 0.5 watt

Circuit Description

This two-stage audio preamplifier is intended for use with high-fidelity magnetic phonograph pickups. The two amplifier stages provide an overall circuit gain of approximately 150. The 7025 twin triode used in the circuit features exceptionally low hum and noise and is designed especially for use in high-fidelity circuits that operate at low signal levels. The preamplifier is ideally suited for use as the low-level input stage for audio power amplifiers such as the 50-watt unit, circuit 26-11. For use with audio power amplifiers such as the 15- and 30-watt units, circuits 26-9 and 26-10, which require higher input signals, another low-level amplifier (e.g., the tone-control amplifier, circuit 26-20) must be inserted between the preamplifier and the power amplifier to obtain the full rated output. The heater and dc operating power required for the preamplifier can usually be obtained from the power-supply circuit for the power amplifier.

The audio signal from the phonograph pickup is applied to J and coupled through a length of shielded cable to the control grid of the input stage of the preamplifier. The interstage coupling between the two amplifier sections of the preamplifier includes an RIAA equalization network (R_{10} and C_6). This network compensates for the Orthophonic recording characteristic* introduced into a record disc by the manufacturer. The output from the preamplifier is coupled from the plate of the second stage by output coupling capacitor

26-16 PREAMPLIFIER FOR MAGNETIC PHONOGRAPH PICKUP (Cont'd)

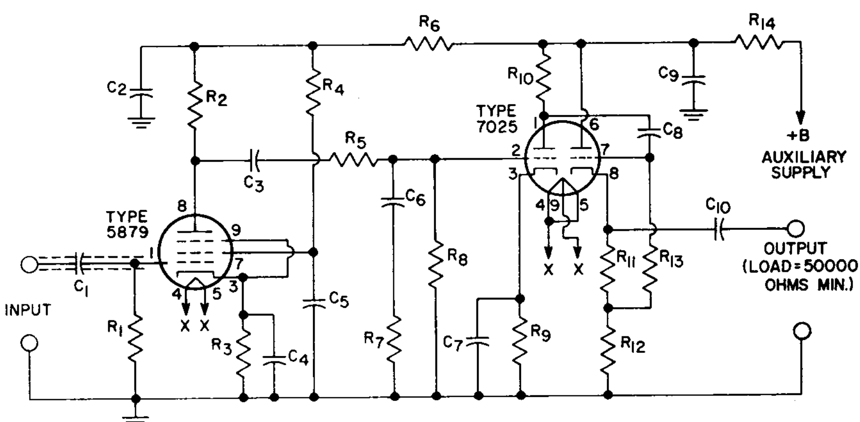
Circuit Description (Cont'd)

C_9 to the input of a tone-control amplifier (if used) or directly to the input of the power amplifier. Because of its relatively high output impedance, the preamplifier is recommended for use in systems in which the preamplifier is mounted on the same chassis as the power amplifier and/or tone-control amplifier. The preamplifier may be used at distances up to 6 feet from the following amplifier provided that the capacitance of capacitor C_8 is reduced approximately 30 picofarads for each foot of shielded cable used for the audio-frequency connection between the preamplifier and the following amplifier.

* To achieve wide frequency and dynamic ranges, manufacturers of commercial recordings use equipment which introduces a non-uniform relationship between amplitude and frequency. This relationship is known as a "recording characteristic." To assure proper reproduction of a high-fidelity recording, therefore, some part of the reproducing system must have a frequency-response characteristic which is the inverse of the recording characteristic. Most manufacturers of high-fidelity recordings use the RCA "New Orthophonic" (RIAA) characteristic for discs and the NARTB characteristic for magnetic tape.

26-17 HIGH-FIDELITY PREAMPLIFIER FOR TAPE-HEAD PICKUP

With NARTB Equalization



Sensitivity=3 millivolts rms input for output of 0.55 volt at frequency of 1000 Hz.

Parts List

C_1=0.047 μF, ceramic, 400 V
C_2=40 μF, electrolytic, 450 V
C_3=0.1 μF, ceramic, 400 V
C_4=25 μF, electrolytic, 25 V
C_5=0.22 μF, ceramic, 400 V
C_6=0.015 μF, ceramic, 400 V
C_7=25 μF, electrolytic, 25 V
C_8=0.22 μF, ceramic, 400 V
C_9=40 μF, electrolytic, 450 V
C_{10}=0.47 μF, ceramic, 400 V
R_1=1 megohm, 0.5 watt
R_2=0.1 megohm, 0.5 watt
R_3=1000 ohms, 0.5 watt
R_4=0.47 megohm, 0.5 watt
R_5=0.22 megohm, 0.5 watt
R_6=22000 ohms, 0.5 watt
R_7=3300 ohms, 0.5 watt
R_8=3.3 megohms, 0.5 watt
R_9=1500 ohms, 0.5 watt
R_{10}=0.1 megohm, 0.5 watt
R_{11}=1500 ohms, 0.5 watt
R_{12}=15000 ohms, 0.5 watt
R_{13}=0.47 megohm, 0.5 watt
R_{14}=4700 ohms, 0.5 watt

26-17 HIGH-FIDELITY PREAMPLIFIER FOR TAPE-HEAD PICKUP (Cont'd)

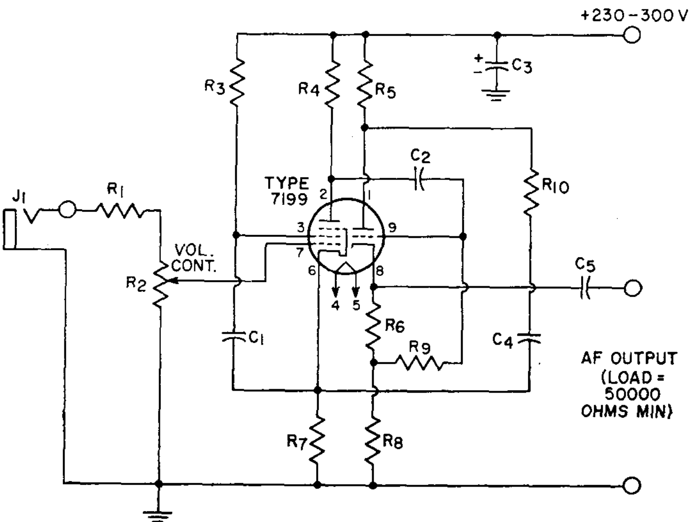
Circuit Description

This three-stage preamplifier provides the amplification necessary to increase the output from a tape-head pickup to the level required to drive an audio power amplifier. The circuit uses a 5879 low-noise sharp-cutoff pentode in a high-gain input voltage amplifier, one section of a 7025 twin triode in a second voltage amplifier, and the other section of the 7025 in a cathode-follower output stage. Because of the low-impedance cathode-follower output circuit, the preamplifier may be installed at distances up to 50 feet from the following stage (tone-control or power amplifier) without adverse effect upon its frequency-response characteristics. The preamplifier is intended for use as the low-level input stages for an audio power amplifier, such as the 50-watt unit (circuit 26-11) or, when followed by another low-level amplifier (e.g., the tone-control amplifier, circuit 26-20), the 15- or 30-watt unit (circuit 26-9 or 26-10). The heater and dc operating power for the preamplifier can usually be obtained from the power supply for the power amplifier.

The preamplifier provides an over-all circuit gain of 180. An input of 3 millivolts rms at the input terminals, is amplified by the pentode and triode voltage amplifiers to develop an output of approximately 0.55 volt rms at the cathode of the cathode-follower output stage. The interstage coupling between the pentode and triode voltage amplifiers equalizes the playback frequency response of the preamplifier to compensate for the NARTB recording characteristic introduced into the magnetic tape by the manufacturer. (See footnote for circuit 26-16.) The output of the preamplifier is coupled by capacitor C_{10} to the input of the audio power amplifier or to the input of an intermediate tone-control amplifier.

26-18 PREAMPLIFIER FOR CERAMIC PHONOGRAPH PICKUP

Cathode Follower (Low-Impedance) Output



26-18 PREAMPLIFIER FOR CERAMIC PHONOGRAPH PICKUP (Cont'd)

Parts List

C_1=0.1 μF, paper, 400 V
C_2=0.01 μF, paper, 400 V
C_3=20 μF, electrolytic, 400 V
C_4=0.25 μF, paper, 400 V
C_5=0.22 μF, paper, 600 V
J_1=Input connector, shielded, for high-impedance ceramic phono pickup (0.5-volt output)
R_1=1.8 megohms, 0.5 watt
R_2=Volume control, potentiometer, 0.5 megohm, audio taper
R_3=0.82 megohm, 0.5 watt
R_4=0.22 megohm, 0.5 watt
R_5, R_8=47000 ohms, 0.5 watt
R_6=4700 ohms, 0.5 watt
R_7=1000 ohms, 0.5 watt
R_9=1 megohm, 0.5 watt
R_{10}=1800 ohms, 0.5 watt

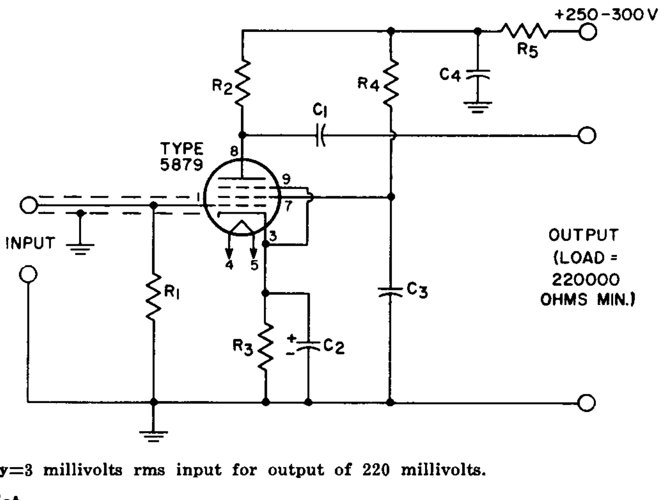
Circuit Description

This two-stage preamplifier is intended for use with a high-impedance ceramic phonograph pickup. The circuit features a cathode-follower (low-impedance) output which makes it possible to install the preamplifier at distances up to 50 feet from the succeeding stage (tone-control or power amplifier). The preamplifier operates from a dc supply of 230 to 300 volts and a heater supply of 6.3 volts. These voltages can usually be obtained from the power supply for the power amplifier in the audio system.

The preamplifier uses a 7199 triode-pentode in a high-gain pentode input stage and a triode cathode-follower output stage. These stages provide the amplification necessary to increase the output from a crystal phonograph pickup, applied at J_1, to the level required to drive an audio power amplifier. The output of the preamplifier, coupled from the cathode of the 7199 triode section, may be applied directly to the power amplifier, or to an intermediate tone-control amplifier.

26-19 LOW-DISTORTION PREAMPLIFIER

For Low-Output, High-Impedance Microphones



Sensitivity=3 millivolts rms input for output of 220 millivolts.

Parts List

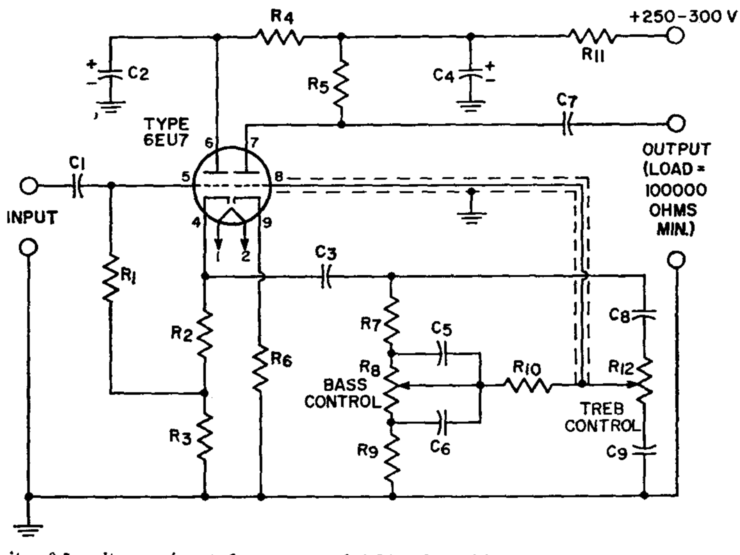
C_1=0.047 μF, paper, 400 V
C_2=25 μF, electrolytic, 25 V
C_3=0.22 μF, paper, 400 V
C_4=40 μF, electrolytic, 450 V
R_1=2.2 megohms, 0.5 watt
R_2=0.1 megohm, 0.5 watt
R_3=1000 ohms, 0.5 watt
R_4=0.47 megohm, 0.5 watt
R_5=22000 ohms, 0.5 watt

26-19 LOW-DISTORTION PREAMPLIFIER (Cont'd)

Circuit Description

This single-stage preamplifier is intended for use with a high-fidelity, high-impedance crystal or dynamic microphone. The circuit uses a 5879 low-noise sharp-cutoff pentode in a conventional amplifier circuit that has a high-impedance output, a voltage gain of approximately 70, and a flat frequency response over the audio range. Because of its high output impedance, the preamplifier should be mounted on the same chassis as the power amplifier and tone-control amplifier (if used). Heater and dc power for the circuit can be obtained from the power supply for the audio power amplifier.

26-20 BASS AND TREBLE TONE-CONTROL AMPLIFIER



Sensitivity=0.5 volt rms input for output of 1.25 volts with controls set for flat response.

Parts List

C_1=0.047 μF, paper, 400 V
C_2, C_4=20 μF, electrolytic, 450 V
C_3=0.1 μF, paper, 400 V
C_5=0.0022 μF, paper, 400 V
C_6=0.022 μF, paper, 400 V
C_7=0.22 μF, paper, 400 V
C_8=220 pF, ceramic or mica, 500 V
C_9=0.0022 μF, paper, 400 V
R_1=0.47 megohm, 0.5 watt
R_2=1500 ohms, 0.5 watt
R_3, R_{11}=15000 ohms, 0.5 watt
R_4=22000 ohms, 0.5 watt
R_5, R_7, R_{10}=0.1 megohm, 0.5 watt
R_6=1000 ohms, 0.5 watt
R_8=Bass control, potentiometer, 1 megohm, audio taper
R_9=10000 ohms, 0.5 watt
R_{12}=Treble control, potentiometer, 1 megohm, audio taper

Circuit Description

This high-fidelity tone-control amplifier uses a 6EU7 low-noise twin triode in a two-stage amplifier cascade that consists of an input cathode follower connected to a triode voltage amplifier through a frequency-sensitive (tone-control) interstage coupling network. The bass and treble controls in the coupling network can be adjusted to provide up to 16 dB of boost or attenuation (cut) at 30 Hz and at 15 kHz. With the bass and treble controls set at the mid-range positions, the amplifier provides an

26-20 BASS AND TREBLE TONE-CONTROL AMPLIFIER (Cont'd)

Circuit Description (Cont'd)

over-all voltage gain of approximately 2.5, and its frequency response is flat within ±1dB from 30 Hz to 15 kHz.

The tone-control amplifier is designed for use immediately ahead of an audio power amplifier, such as the 15-, 30-, or 50-watt unit (circuit 26-9, 26-10, or 26-11, respectively). Operating power for the tone-control circuit can usually be obtained from the power supply for the power amplifier. For operating convenience, the volume control on the power amplifier may be physically located on the tone-control chassis. In this case, it is advisable to insert a 1-megohm potentiometer in place of the volume control on the power amplifier. If partial compensation for the reduced high- and low-frequency sensitivity of the ear at low volume levels is desired, the volume-control potentiometer may be replaced by a loudness-control potentiometer.

26-21 SINE- SQUARE-WAVE AUDIO SIGNAL GENERATOR

Circuit Description

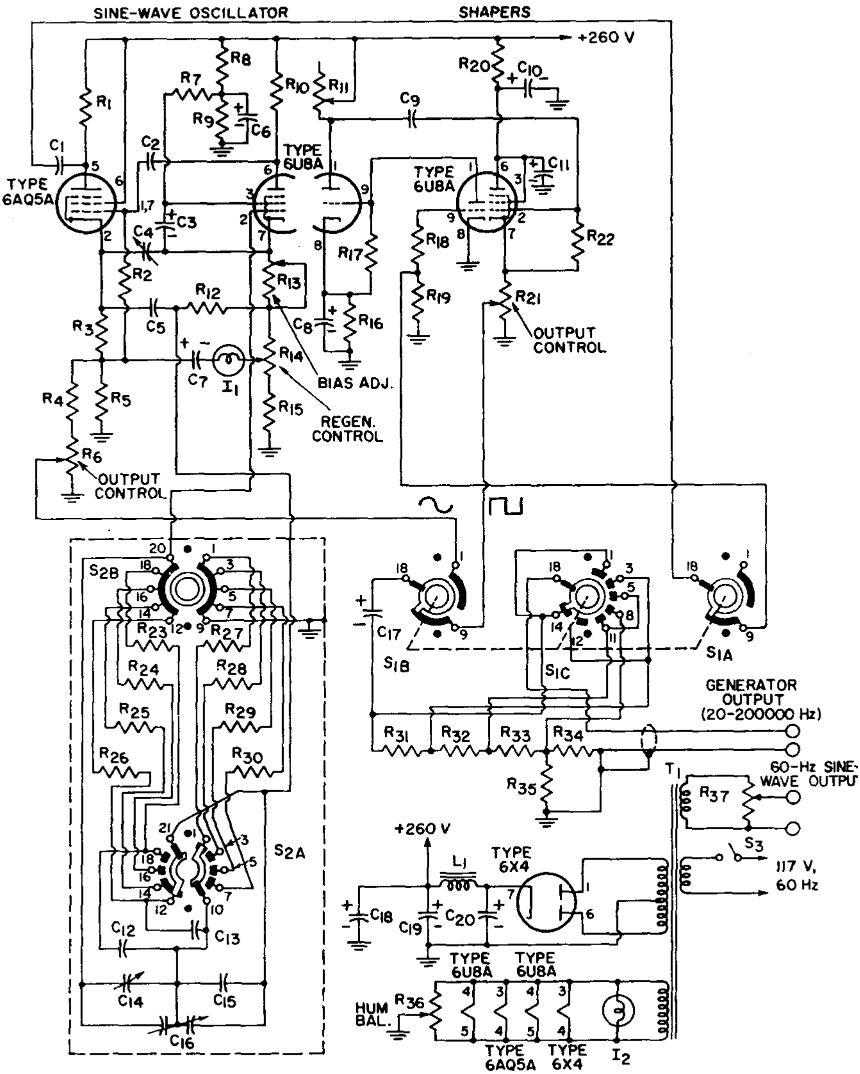
This audio-signal generator provides sine-wave or square-wave outputs at frequencies from 20 Hz to 200 kHz. The sine-wave outputs are adjustable from 0 to 10 volts rms, and the square-wave outputs are adjustable from 0 to 10 volts peak. The generator also provides a fixed-frequency (60-Hz) sine-wave output that is variable in amplitude from 0 to 6 volts rms. The 117-volt, 60-Hz ac input power to the generator is converted to dc operating power for the various circuit stages by a 6X4 full-wave rectifier circuit. Power for the tube heaters is supplied by a 6.3-volt winding of power transformer T_1. A panel lamp I_2 connected across this secondary winding lights when ON-OFF switch S_3 is closed to indicate the application of ac input power to the generator. A second 6.3-volt secondary winding of transformer T_1 provides the fixed-frequency sine-wave output. This 60-Hz signal is coupled from the wiper arm of the output voltage control R_{37} connected across the 6.3-volt winding.

The basic excitation in the main signal channel of the generator is provided by a variable-frequency bridged-T type of sine-wave oscillator in which the required amplification and switching are provided by the pentode section of a 6U8A triode-pentode. The Frequency-Range selector S_2, a four-position, two-section rotary switch, connects the proper combination of resistors into the bridged-T network to establish the desired frequency range for the oscillator—20 to 200 Hz (X1 position), 200 to 2000 Hz (X10 position), 2 to 20 kHz (X100 position), or 20 to 200 kHz (X1000 position). A two-gang variable (split-stator) capacitor C_{16} provides a vernier control of the oscillator frequency on each range. Capacitors C_4 and C_{14} are trimmer adjustments for the oscillator.

The sine-wave signal developed in the plate circuit of the oscillator stage is coupled to the control grid of a 6AQ5 pentode amplifier stage that provides both plate and cathode signals. The cathode signal is the sine-wave output of the generator. The plate signal is used to derive the square-wave output of the generator. The setting of the SINE-SQUARE attenuation selector S_1, an eight-position three-section rotary switch, determines whether the generator provides sine-wave or square-wave outputs. In addition, the selector provides four levels of attenuation for each type of output, as shown in the switch-position chart.

With the attenuation selector set to any one of the four square-wave positions, the sine-wave signal from the plate of the 6AQ5 stage is

26-21 SINE- SQUARE-WAVE AUDIO-SIGNAL GENERATOR (Cont'd)



SWITCH POSITIONS

S₂ Frequency Range	S₁ Sine/Square	
1—X1	1—Sine X10	5—Square X0.01
2—X10	2—Sine X1	6—Square X0.1
3—X100	3—Sine X0.1	7—Square X1
4—X1000	4—Sine X0.01	8—Square X10

This audio generator is similar to the RCA type WA-44C.

26-21 SINE- SQUARE-WAVE AUDIO-SIGNAL GENERATOR (Cont'd)

PERFORMANCE SPECIFICATIONS

Sine-Wave Output: 0 to 8 volts rms.
Square-Wave Output: 0 to 10 volts, peak.
Frequency Ranges: 20 to 200 Hz; 200 to 2000 Hz; 2000 to 20000 Hz; 20000 to 200000 Hz.
Notes: 1. "Sine-Square Attenuator" S_1 shown in "X10" position.
2. "Freq. Range" selector, S_2, shown in "X1" position.

Parts List

C_1=0.1 μF, ceramic, 400 V
C_2, C_9=0.25 μF, ceramic, 400 V
C_3, C_6, C_7=20 μF, electrolytic, 350 V
C_4=5—80 pF, trimmer
C_5=1 μF, paper, 200 V
C_8, C_{17}=40 μF, electrolytic, 150 V
C_{10}=100 μF, electrolytic, 150 V
C_{11}, C_{19}, C_{20}=3-section electrolytic; 20 μF, 250 V; 60 μF, 450 V; 20 μF, 450 V
C_{12}=2.2 pF, ceramic
C_{13}=3.3 pF, ceramic, 500 V
C_{14}=7.5—8 pF, trimmer
C_{15}=27 pF, ceramic, 600 V
C_{16}=Variable, 2 gang; RCA stock No. 220226 or equiv.
C_{18}=50 μF, electrolytic, 250 V
I_1=Lamp, 3 watts, 120 V
I_2=Pilot lamp, No. 47
L_1=Reactor, RCA stock No. 220215 or equiv.
R_1=3900 ohms, 2 watts
R_2, R_{12}, R_{22}=1 megohm, 0.5 watt
R_3=470 ohms, 1 watt
R_4=3900 ohms, 1 watt
R_5=12000 ohms, 1 watt
R_6=Potentiometer, 12000 ohms
R_7=3300 ohms, 0.5 watt
R_8, R_9=22000 ohms, 1 watt
R_{10}=56000 ohms, 0.5 watt
R_{11}=Potentiometer, 250 ohms, 0.5 watt
R_{13}, R_{14}=Potentiometer, 5000 ohms
R_{15}=8200 ohms, 0.5 watt
R_{16}=12000 ohms, 0.5 watt
R_{17}=4700 ohms, 1 watt
R_{18}=0.47 megohm, 0.5 watt
R_{19}=0.27 megohm, 0.5 watt
R_{20}=15000 ohms, 2 watts
R_{21}=Potentiometer, 750 ohms
R_{23}=36000 ohms, 0.5 watt
R_{24}=0.36 megohm, 0.5 watt
R_{25}=3.6 megohms, 0.5 watt
R_{26}=36 megohms, 1 watt
R_{27}=8 megohms, 1 watt
R_{28}=0.8 megohm, 0.5 watt
R_{29}=80000 ohms, 0.5 watt
R_{30}=8000 ohms, 0.5 watt
R_{31}=8200 ohms, 0.5 watt
R_{32}=820 ohms, 0.5 watt
R_{33}=82 ohms, 0.5 watt
R_{34}, R_{35}=18 ohms, 0.5 watt
R_{36}=Potentiometer, 100 ohms
R_{37}=Potentiometer, 100 ohms, part of assembly with switch S_3
S_1=Rotary switch, function selector, 8 position, 3 wafer, RCA stock No. 220216 or equiv.
S_2=Rotary switch, range selector, 4 position, 2 wafer, RCA stock No. 220217 or equiv.
S_3=ON-OFF switch, part of assembly with R_{37}
T_1=Power transformer, 117 volts rms, 60 Hz, RCA stock No. 220214 or equiv.

Circuit Description (Cont'd)

coupled through the S_{1A} section of the selector to the shaping amplifiers. The shaping amplifiers consist of two triode limiters and a pentode cathode-follower output stage in cascade. The triode limiters, each of which uses the triode section of a 6U8A triode-pentode, clip the positive and negative peaks of the sine-wave input to produce a square-wave signal. This signal is applied to the control grid of the pentode section of the 6U8A triode-pentode used in the cathode-follower output stage. The resulting square-wave signal developed across the square-wave output control R_{21} is coupled from the wiper arm of the control through the S_{1B} section of the SINE-SQUARE attenuation selector to the output attenuation network. If the attenuation selector is set to one of the four sine-wave positions, no square wave is developed, and the sine-wave signal from the wiper arm of the sine-wave output control R_6 is coupled through the S_{1B} section of the attenuation selector to the output network.

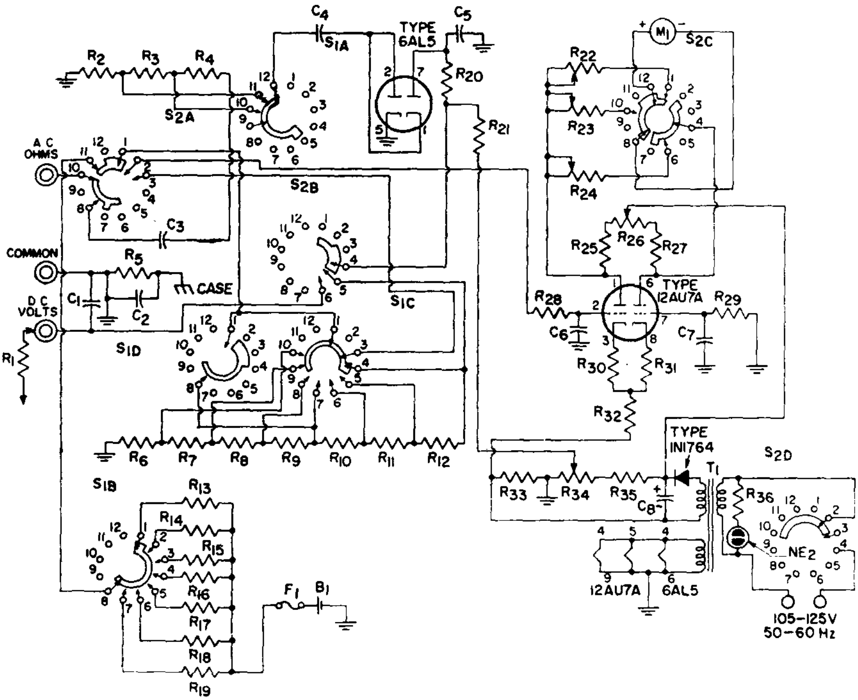
The output attenuation network is a tapped resistive voltage divider that provides four output levels with the three lower levels successively decreased to one-tenth of the next higher one. The S_{1C} section of the attenuation selector determines the tap on the voltage divider from which the sine-wave or square-wave output is obtained. The sine-wave or square-wave OUTPUT control provides continuous adjustment of the output level for any attenuation setting.

26-22 ELECTRONIC VOLT-OHM METER

Circuit Description

This electronic volt-ohm meter can be used to measure rms values of ac sine-wave voltages from 0.1 to 1500 volts, dc voltages from 0.2 to 1500 volts, peak-to-peak voltages from 0.2 to 4000 volts, and resistances

26-22 ELECTRONIC VOLT-OHM METER (Cont'd)



SWITCH POSITIONS

Position	Range Selector, S_1			Function Selector, S_2
1	1.5V	Rx1	4V	OFF
2	5V	Rx10	14V	AC VOLT
3	15V	Rx100	40V	−DC VOLTS
4	50V	Rx1000	140V	+DC VOLTS
5	150V	Rx10,000	400V	OHMS
6	500V	Rx100,000	1400V	
7	1500V	Rx1M	4000V	

Notes:
1. Switches are shown in their maximum counterclockwise position (S_1=1.5 V, R X 1; S_2="OFF").
2. For home construction of this or a similar circuit, the complete Kit-WV-77E(K) or RCA-WV-98C(K) is recommended because of the large number of special components used.
3. The accuracy of the volt-ohm meter depends upon the accuracy of the multiplier resistors.

Parts List

B_1=Battery, 1.5 V
C_1=470 pF, ceramic disc, 1600 V
C_2=0.001 μF, ceramic disc, 500 V
C_3=0.47 μF, tubular, 400 V
C_4, C_5=0.02 μF, ceramic disc, 400 V
C_6, C_7=0.005 μF, ceramic disc, 200 V
C_8=10 μF, electrolytic, 400 V
F_1=Fuse, 0.5 ampere
M_1=Meter, dc, 0-200 μA
NE_2=Neon lamp
R_1=DC-voltage probe isolating resistor, 1 megohm, 0.25 watt
R_2=138000 ohms, 0.25 watt
R_3=320000 ohms, 0.5 watt
R_4=0.9 megohm, 1 watt
R_5, R_{18}=1 megohm, 0.25 watt
R_6, R_{16}, R_{25}, R_{27}=10000 ohms, 0.5 watt
R_7=20000 ohms, 0.25 watt
R_8=70000 ohms, 0.25 watt
R_9=0.2 megohm, 0.25 watt
R_{10}=0.7 megohm, 0.25 watt
R_{11}=2 megohms, 0.25 watt

26-22 ELECTRONIC VOLT-OHM METER (Cont'd)

Parts List (Cont'd)

R_{12}=7 megohms, 0.25 watt
R_{13}=8.2 ohms, wire-wound, 0.5 watt
R_{14}=100 ohms, 0.25 watt
R_{15}=1000 ohms, 0.25 watt
R_{17}=0.1 megohm, 0.25 watt
R_{19}=10 megohms, 0.25 watt
R_{20}=20 megohms, 0.25 watt
R_{21}=91 megohms, 0.5 watt
R_{22}=10000 ohms, potentiometer ac calibration, 0.5 watt
R_{23}=10000 ohms, potentiometer dc calibration, 0.5 watt
R_{24}=15000 ohms, potentiometer, ohms adjustment, 0.25 watt
R_{26}=10000 ohms, potentiometer, zero adjustment, 0.25 watt
R_{28}=3.3 megohms, 0.5 watt
R_{29}=6.8 megohms, 0.5 watt
R_{30}, R_{31}=330 ohms, 0.5 watt
R_{32}=15000 ohms, 0.5 watt
R_{33}=27000 ohms, 0.5 watt
R_{34}=10000 ohms, potentiometer, ac balance, 0.5 watt
R_{35}=47000 ohms, 0.5 watt
R_{36}=0.22 megohm, 0.5 watt
S_1=Range selector switch, 7 position, RCA stock No. 217924 or equiv.
S_2=Function selector switch, 5 position, RCA stock No. 217923 or equiv.
T_1=Power transformer, 105-125 volts rms. 50-60 Hz, RCA stock No. 217921 or equiv.

Circuit Description (Cont'd)

from 0.2 ohms to 1000 megohms. Within these over-all limits, a Range Selector (S_1) can be used to select seven different measurement ranges for each measurement function, as shown in the switch-position chart. The mode of operation of the volt-ohm meter is determined by the setting of the five-position (OFF, AC, −DC, +DC, and OHMS) Function Selector (S_2). A section (S_{2D}) of the Function Selector is also used to control the application of the 117-volt, 60-Hz, input ac power. The ac input power is converted to dc power by the 1N1764 selenium rectifier and associated components. A 6.3-volt secondary winding of power transformer T_1 supplies power to the tube heaters. A neon lamp connected across the primary of power transformer T_1 lights when ac power is applied to the circuit.

A balanced push-pull dc amplifier, which includes a dc microammeter M_1 connected as part of a dc bridge network between the two plate sections of the stage, is used as the basic measuring circuit for each measurement function of the volt-ohm meter. This circuit has a linear response, excellent stability, and a very high input impedance. Calibration adjustments are provided for each mode of operation to assure that accurate measurements are obtained. If desired, the ZERO ADJ potentiometer R_{36} may be adjusted to provide a center-scale zero reading on the meter, which is useful in discriminator and bias voltage measurements.

For ac voltage measurements, Function Selector S_2 must be rotated to the AC position. The ac voltage to be measured, applied between the AC-OHMS and COMMON terminals, is coupled through contacts 10 and 9 of S_{1A} to the ac-voltmeter multipliers (R_2 through R_4). The ac voltage from one of the taps on the multiplier, as determined by the setting of the Range Selector (S_{1A} section), is rectified by the 6AL5 twin diode. The resultant dc voltage across the rectifier bleeder resistors R_{21} and R_{34} is proportional to the ac voltage from the multiplier network. This voltage is then coupled through contacts 4 and 5 of S_{2B}, through one of the contacts 4 through 10 (as determined by setting of Range Selector) and contact 1 of S_{1C}, and through contacts 1 and 2 of S_{2A} to the pin 2 control grid of the 12AU7A twin triode in the balanced dc amplifier. This input disturbs the balance of the amplifier and a current proportional to the ac input flows through the dc microammeter connected between the plates of the 12AU7. The pointer on the microammeter is then deflected to indicate the value of the voltage being measured.

With the Function Selector rotated to either −DC or +DC, a dc voltage being measured is coupled through the 1-megohm probe R_1, the DC VOLTS terminal, and contacts 6 and 5 of S_{2B} to the dc-voltmeter multipliers (R_6 through R_{12}). The 1-megohm resistance of the dc probe together with the resistance of the

26-22 ELECTRONIC VOLT-OHM METER (Cont'd)

Circuit Description (Cont'd)

multipliers results in an input resistance of 11 megohms for dc voltage measurements. The dc voltage from the appropriate tap on the multiplier network selected by the S_{1C} and S_{1D} sections of the Range Selector is coupled through contact 1 of these switch sections (or contact 3 of S_{1C}) and contacts 1 (or 3) and 2 of S_{2A} to the input of the balanced dc amplifier. The pointer of the microammeter in the balanced amplifier is then deflected to provide an indication of the value of the dc voltage being measured. The S_{2C} section of the Function Selector reverses the connections of the microammeter when the Function Selector is rotated from −DC to +DC so that current will flow through the microammeter in the same direction regardless of whether a negative or positive dc voltage is being measured.

For resistance measurements, the Function Selector is rotated to the OHMS position, and the external resistance to be measured is connected between the AC-OHMS and COMMON terminals of the volt-ohm meter. A 1.5-volt dry cell then causes current to flow through the external resistance, through contacts 10 and 11 of S_{2A}, and through one of the ohmmeter-section multiplier resistors (R_{13} through R_{19}), as determined by the setting of the Range Selector (S_{1B} section). Because the multiplier resistance is fixed for each range, the voltage developed across the external resistance provides an accurate indication of the value of this resistance. This voltage is coupled through contacts 10 and 2 of S_{2A} to the input of the balanced dc amplifier. The pointer of the microammeter is then deflected to indicate the value of the resistance being measured.

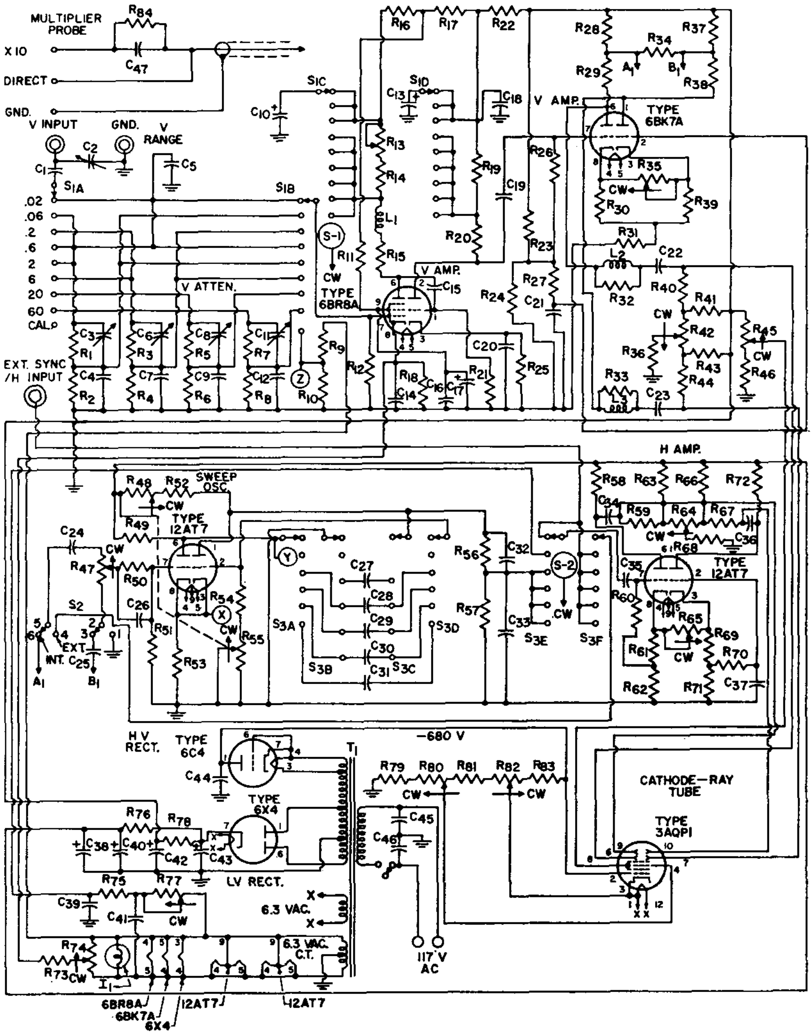
26-23 CATHODE-RAY OSCILLOSCOPE

Circuit Description

This oscilloscope provides a 3-inch cathode-ray-tube display of voltage waveforms at frequencies from 5.5 Hz to 5.5 MHz. It is very useful, therefore, for signal tracing and monitoring in the servicing of black-and-white and color television receivers, AM and FM radio receivers, high-fidelity audio systems, and other types of electronic equipment. The sensitivity of the oscilloscope is such that each 0.1 volt rms applied to the vertical-input terminal results in a 1-inch vertical deflection of the electron beam on the 3-inch cathode-ray-tube screen. The unit operates from a 117-volt, 60-Hz ac power line. A 6X4 transformer-coupled full-wave rectifier circuit converts the ac input power to the +320 volts used as the main dc supply voltage for the oscilloscope. A half-wave rectifier circuit that uses a 6C4 triode connected to operate as a diode converts the ac power developed across a high-voltage winding of power transformer T_1 to the −680 volts required for operation of the 3AQP1 cathode-ray tube. A 6.3-volt tap on the high-voltage winding of T_1 provides the heater power for the 6C4. A 6.3-volt secondary winding of T_1 provides the heater power for the 3AQP1 cathode-ray tube, and a center-tapped 12.6-volt winding supplies heater power for the remainder of the tubes in the oscilloscope.

A signal waveform applied to the vertical-input terminal is routed through contacts of the S_{1A} section of the Vertical Range selector to one of the input attenuation networks. The S_{1B} section of the Vertical Range selector couples the attenuated signal waveform from the appropriate input network to the input of the vertical amplifiers. The S_{1C} and S_{1D} sections of the Vertical Range selector automatically switch the vertical amplifiers from wide-band to narrow-band operation in the three highest-gain (lowest-attenuation) positions. With the Vertical Range selector in the CAL position, the

26-23 CATHODE-RAY OSCILLOSCOPE (Cont'd)



Parts List

C1, C3, C6, C8, C11=Trimmer capacitors, 4—40 pf, Arco No. 422 or equiv.
C2, C15, C19, C21, C35=0.1 μF, paper, 400 V
C4=64 pF ceramic disc, 500 V
C5=22 pF, ceramic disc, 500 V
C7=140 pF, ceramic disc, 500 V
C9=410 pF, ceramic disc, 500 V
C10, C13, C40, C43=20 μF, electrolytic, 450 V
C12=1500 pF, ceramic disc, 500 V
C14=1200 pF, ceramic disc, 500 V
C16, C24, C25=0.02 μF, ceramic disc, 600 V
C17, C38=10 μF, electrolytic, 450 V
C18, C42=40 μF, electrolytic, 450 V
C20=560 pF, ceramic disc, 500 V
C22=0.05 μF, ceramic disc, 200 V
C23=0.05 μF, paper, 200 V
C26=5 pF, ceramic disc, 150 V
C27=0.22 μF, paper, 400 V
C28=0.022 μF, paper, 400 V
C29=2200 pF, ceramic disc, 400 V
C30=220 pF, ceramic disc, 400 V
C31=15 pF, ceramic disc, 500 V

26-23 CATHODE-RAY OSCILLOSCOPE (Cont'd)

Parts List (Cont'd)

C_{32}=180 pF, ceramic disc, 200 V
C_{33}=150 pF, ceramic disc, 200 V
C_{34}, C_{36}, C_{37}, C_{41}=0.1 μF, paper, 200 V
C_{39}, C_{45}, C_{46}=0.01 μF, ceramic disc, 600 V
C_{44}=0.5 μF, paper, 1000 V
C_{47}=12 pF, tubular ceramic, 150 V
I_1=Pilot lamp, No. 47
L_1=Peaking coil, 20 μH
L_2, L_3=Peaking coil, 36 μH (wound on 10,000-ohm, 0.5-watt resistor)
R_1=0.68 megohm, 0.5 watt
R_2, R_{26}, R_{27}, R_{68}, R_{79}=0.47 megohm, 0.5 watt
R_3=0.91 megohm, 0.5 watt
R_4=0.11 megohm, 0.5 watt
R_5, R_7, R_{12}, R_{21}, R_{40}, R_{44}=1 megohm, 0.5 watt
R_6=3300 ohms, 0.5 watt
R_8, R_{32}, R_{33}, R_{34}=10000 ohms, 0.5 watt
R_9, R_{62}, R_{71}, R_{83}=15000 ohms, 0.5 watt
R_{10}=820 ohms, 0.5 watt
R_{11}=47000 ohms, 0.5 watt
R_{13}=Variable, wire-wound, 5000 ohms, 2 watts, Clarostat A43-5000 or equiv.
R_{14}=6800 ohms, 1 watt
R_{15}, R_{20}, R_{29}, R_{38}=1200 ohms, 0.5 watt
R_{16}=2200 ohms, 0.5 watt
R_{17}=Wire-wound, 2500 ohms, 5 watts, IRC Type PW5 or equiv.
R_{18}=100 ohms, 0.5 watt
R_{19}=4700 ohms, 1 watt
R_{22}=820 ohms, 1 watt
R_{23}=0.22 megohm, 0.5 watt
R_{24}=82000 ohms, 0.5 watt
R_{25}=120 ohms, 0.5 watt
R_{28}, R_{37}=1800 ohms, 1 watt
R_{30}, R_{39}=1000 ohms, 0.5 watt
R_{31}=Wire-wound, 2400 ohms, 5 watts, IRC Type PW5 or equiv.
R_{35}=5000 ohms, 0.5 watt
R_{36}=1.2 megohm, 0.5 watt
R_{41}, R_{43}, R_{63}, R_{66}=0.82 megohm, 0.5 watt
R_{42}, R_{48}=Variable, 1 megohm, 0.5 watt
R_{45}=Variable, 0.1 megohm, 0.25 watt
R_{46}=0.18 megohm, 0.5 watt
R_{47}, R_{77}=Variable, 0.25 megohm, 0.5 watt
R_{49}=0.1 megohm, 1 watt
R_{50}=68000 ohms, 0.5 watt
R_{51}=3300 ohms, 0.5 watt
R_{52}=0.27 megohm, 0.5 watt
R_{53}=680 ohms, 0.5 watt
R_{54}=39000 ohms, 0.5 watt
R_{55}=Variable, 5 megohms, 0.5 watt
R_{56}, R_{59}, R_{67}=2.7 megohms, 0.5 watt
R_{57}=3.3 megohms, 0.5 watt
R_{58}, R_{72}, R_{75}, R_{81}=0.12 megohm, 0.5 watt
R_{60}, R_{70}=10 megohms, 0.5 watt
R_{61}, R_{69}=2400 ohms, 0.5 watt
R_{64}=Variable, 2 megohms, 0.5 watt
R_{65}=Variable, 50000 ohms, 0.5 watt
R_{73}=0.1 megohm, 0.5 watt
R_{74}=Variable, 10000 ohms, 0.25 watt
R_{76}=4700 ohms, 0.5 watt
R_{78}=Wire-wound, 1500 ohms, 7 watts, IRC Type PW7 or equiv.
R_{80}=Variable, 0.5 megohm, 0.5 watt
R_{82}=Variable, 75000 ohms, 0.5 watt (includes ac switch)
S_1=Rotary switch, vertical range selector, 9 positions, 4 sections, RCA stock No. 219199 or equiv.
S_2=Switch, dpdt, sync, Stackpole Type SS-33 or equiv.
S_3=Rotary switch, horizontal sweep selector, 6 positions, 5 sections, RCA stock No. 219200 or equiv.
T_1=Power transformer, 117 volts, 60 Hz, RCA stock No. 218122 or equiv.
X, Y, Z,=Test points

For home construction of this circuit, the complete Kit RCA-WO-33A (K) is recommended because of the large number of special components used. This circuit is also available in wired form as the RCA-WO-33A.

Circuit Description (Cont'd)

vertical-input terminal and input attenuation networks are disconnected from the vertical amplifiers, and an internal calibrating (reference) voltage, obtained from the junction of voltage-divider resistors R_9 and R_{10}, is applied to the input of the vertical amplifiers. This calibrating voltage, the fact that the input attenuation networks are voltage calibrated, and the use of a graph screen scaled directly in volts make possible the use of the oscilloscope as a visual peak-to-peak voltmeter.

The signal waveform from the input attenuation network is amplified by a two-stage vertical-amplifier cascade that uses a 6BR8 in a high-gain pentode input stage and a triode voltage amplifier. The output of the triode amplifier drives a 6BK7 twin triode used in the vertical paraphase amplifier. The 6BK7 is operated in a push-pull differential-amplifier configuration to provide two equal-amplitude outputs (one from each plate section) that are 180 degrees out of phase. These signals are applied to opposite vertical deflection plates of the 3AQP1 cathode-ray tube to provide the push-pull vertical deflection of the electron beam that causes the horizontal sweep to track the signal waveform applied to the vertical-input terminal. The exceptionally high gain of the vertical-amplifier stages make the oscilloscope sensitive enough to provide useful displays of signals from low-level microphones, phonograph pickups, and other low-output sources. The VERT. CAL. control R_{35} in the cathode circuit of the vertical paraphase amplifier adjusts the sensitivity or calibrates the vertical amplifier to correspond with the position of the Vertical Range selector.

The circuits used to produce the

26-23 CATHODE-RAY OSCILLOSCOPE (Cont'd)

Circuit Description (Cont'd)

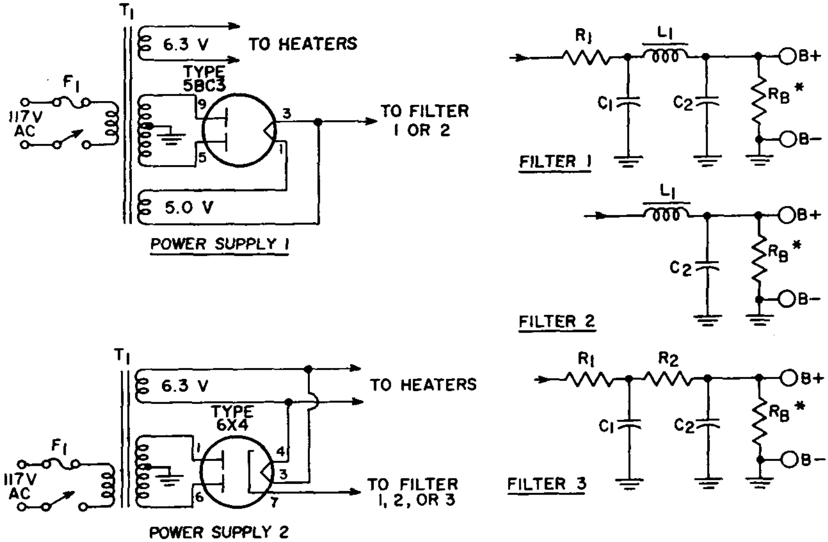
horizontal sweep on the oscilloscope screen include a horizontal oscillator (sawtooth generator) and a horizontal paraphase amplifier, each of which uses a 12AT7 twin triode. The oscillator generates sawtooth waveforms, at frequencies from 15 Hz to 75 kHz, in four basic ranges. The Sweep Selector S_3 connects the proper combination of capacitors into the stage for each range. The Sweep Vernier control (ganged potentiometers R_{48} and R_{56}), which overlaps the basic frequency ranges, provides exact adjustment of the sweep frequency. The oscillator exhibits excellent stability at high sweep rates, has a fast retrace, and provides adequate linearity throughout its overall frequency range. With the Sweep Selector set to any of the positions 3 through 6, the sawtooth waveform from the oscillator is applied to the pin 7 control grid of the 12AT7 twin triode used in the horizontal paraphase amplifier. The horizontal paraphase amplifier, which is essentially identical to the vertical paraphase amplifier except for significant differences in frequency-response characteristics, develops two equal-amplitude sawtooth waveforms that are 180 degrees out of phase. These waveforms are applied to opposite horizontal-deflection plates of the 3AQP1 cathode-ray tube to provide the push-pull deflection of the electron beam that results in a linear horizontal sweep on the oscilloscope screen.

The horizontal oscillator may be synchronized by either internal or external signals. With the Sync Selector S_2 set to INT, a signal from the vertical paraphase amplifier (points A_1 and B_1) synchronizes the oscillator to assure that the start of the horizontal sweep is coincident with the start of the signal applied at the vertical-input terminal. For internal-sync operation, the Sync/Phase control R_{47} at the input of the oscillator has its zero setting at the mid-range position and may be used to adjust both the amplitude and phase of the synchronizing voltage to lock the oscilloscope pattern to a stationary position. With the Sync Selector set to EXT, a signal from an external source, coupled through the EXT. SYNC/H INPUT terminal and contacts of S_{3F} (in positions 3 through 6) is used to synchronize the oscillator. For external-sync operation, the Sync/Phase control adjusts the amplitude of the external synchronizing voltage by normal clockwise rotation and the phase control feature is not provided (center position of control is not zero).

If desired, a signal from an external horizontal oscillator or the 60-Hz line voltage may be used to produce the sweep on the oscilloscope screen. With the Sweep Selector set to either HOR IN or to LINE, the horizontal oscillator is disconnected from the circuit, and the input to the horizontal amplifier is then obtained from either the EXT SYNC/H INPUT terminal or the center-tapped 12.6-volt heater winding of power transformer T_1.

The three-lead accessory probe shown with the circuit schematic facilitates the use of the oscilloscope. The ground lead of the probe is connected to the ground terminal of the oscilloscope, and the vertical input is then applied through the direct or the X10 attenuation lead. When the direct lead is used, the signal is applied directly to the vertical-input terminal. When the attenuation lead is used, a high-impedance network in the probe is connected in series with the test point and the vertical-input terminal of the oscilloscope. This high-impedance network presents an over-all input resistance of 10 megohms and an input capacitance of approximately 10 picofarads to the test circuit. This high impedance reduces circuit-loading effects and permits use of the oscilloscope in circuits which do not function properly if loaded by a conventional oscilloscope.

26-24 ALL-PURPOSE DC POWER SUPPLIES



POWER SUPPLY	TRANS-FORMER	CHOKE (L_1)	R_1	R_2	C_1, C_2	FIL-TER	OUTPUT VOLTS	OUTPUT MA
1 (5BC3)	Stancor PC or PM 8177 (300-0-300) or equiv.	140 mA, 7H, 165 ohms Stancor C1421 or equiv.	33 ohms 5W	—	40 μF 450 Vdc	1	360 340 320	60 80 120
						2	235 230 215	60 80 120
1 (5BC3)	Stancor PC or PM 8412 (400-0-400) or equiv.	200mA, 4H, 145 ohms Thordarson 20C54 or equiv.	56 ohms 10W	—	40 μF 600 Vdc	1	450 425 410	120 160 200
						2	310 300 280	120 160 200
2 (6X4)	Stancor P-6358 (300-0-300) or equiv.	80 mA, 12H, 375 ohms Thordarson 20C53 or equiv.	500 ohms 5W	500 ohms 3W	40 μF 450 Vdc	1	350 300 260	20 40 60
						2	250 230 220	20 40 60
						3	345 300 250	20 40 60
2 (6X4)	Stancor PM or PC 8419 (240-0-240) or equiv.	80 mA, 12H, 375 ohms Thordarson 20C53 or equiv.	500 ohms 5W	500 ohms 3W	40 μF 450 Vdc	1	265 225 190	20 40 60
						2	200 180 170	20 40 60
						3	260 220 180	20 40 60

* Bleeder R_B can be omitted if an external load is permanently connected across the ouput terminals. Bleeder current should be approximatly 10 per cent of the load current.

26-24 ALL-PURPOSE DC POWER SUPPLIES (Cont'd)

Circuit Description

In these power-supply circuits, 5BC3 and 6X4 full-wave rectifier tubes are used to convert ac input power to dc output power in various combinations of output voltage and load current. The 5BC3 tube is a directly heated novar type intended for use in power supplies for radio equipment, television receivers, and other applications that have relatively high dc requirements. The 6X4 tube is an indirectly heated miniature type used primarily in power supplies for automobile and ac-operated radio receivers and other equipment that have moderate dc requirements.

In each rectifier circuit, the 117-volt ac input power is applied to the primary of a step-up power transformer T_1. The two plate sections of the rectifier tube are connected to opposite ends of the center-tapped secondary winding of transformer T_1. With respect to the grounded center tap, the voltage applied to each plate of the rectifier tube, therefore, is 180 degrees out of phase with that applied to the other plate. With an external load connected to the rectifier cathode, pulses of current flow alternately to one plate and then to the other plate for each half cycle of the ac input power. This 120-Hz pulsating current develops a positive dc voltage across the load circuit.

Removal of virtually all the 120-Hz ripple component from the dc output can be accomplished by connection of a suitable filter network between the rectifier output (cathode) and the load circuit. Either Filter 1 or Filter 2 provides adequate filtering for the 5BC3 circuit. Any one of the three filter networks is satisfactory for use with the 6X4 circuit. Filter 3 is not recommended for use with the 5BC3 circuit because the use of the two resistors R_1 and R_2 in series with the relatively high output results in excessive power loss.

The chart shown with the rectifier circuits lists a wide range of dc output voltage obtainable for various values of load current. Proper selection of power transformer T_1, of the type of filter network, and of the values of filter choke L_1 and resistors R_1 and R_2 results in the desired combination of output voltage and current.

BLACK-AND-WHITE TELEVISION RECEIVER

Circuits 26-25 through 26-29 are essentially identical to the corresponding circuits in the RCA-KCS-152 Television Receiver. These circuits comprise a complete intercarrier television receiver with the exception of the deflection coils and the picture tube. Portions of any television receiver, however, are required to operate over an extremely wide range of very high frequencies. The construction of such circuits requires more than ordinary skill and experience and the use of sophisticated test equipment (see general consideration for the construction of high-frequency and broadband circuits at the beginning of this section). Home construction of such circuits is not recommended unless the builder has had considerable experience in this type of work.

The chassis of circuits 26-25 through 26-29 are connected to one side of the ac line during operation. Servicing of these circuits should not be attempted by persons not familiar with the following precautions necessary when working on this type of equipment:

1. An isolation transformer should be inserted between the receiver and the ac line before any servicing is attempted.

BLACK-AND-WHITE TELEVISION RECEIVER (Cont'd)

2. If the receiver must be operated directly from the ac supply, the power plug should be inserted in the proper direction to connect the chassis to the ground side of the ac line. An ac voltmeter should be used to measure the voltage between the chassis and the power-source ground; no voltage reading should be obtained. If a reading is obtained, the power plug should be reversed and another check made for a zero reading.

26-25

VHF TUNER

For Black-and-White Television Receiver

Circuit Description

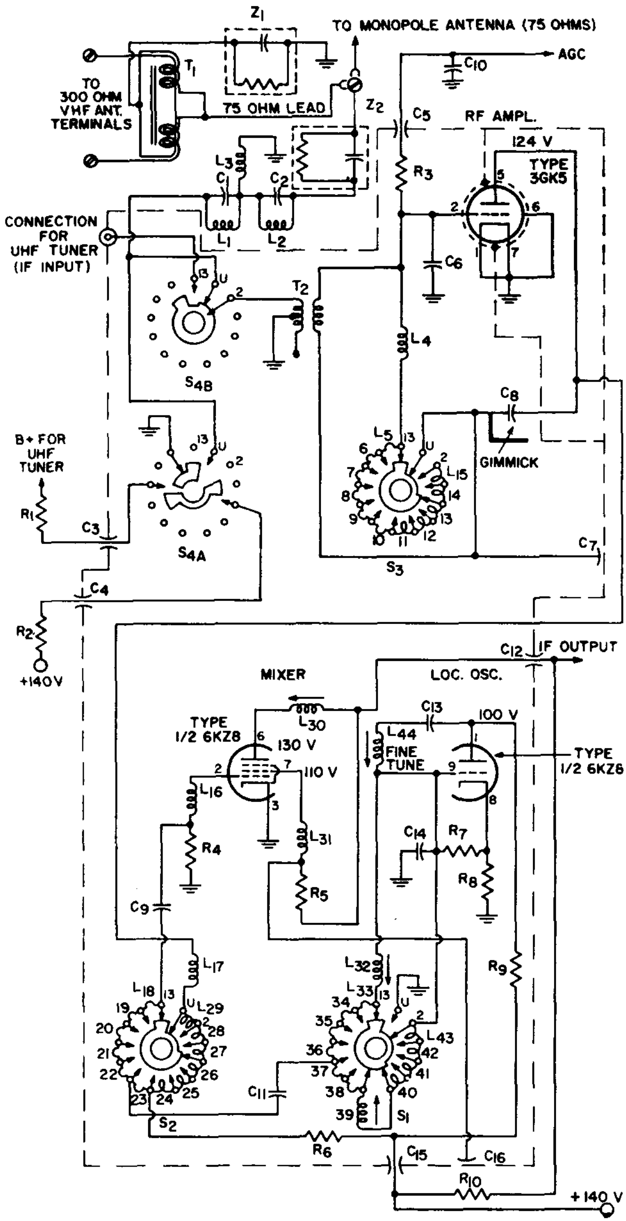
This vhf tuner selects the desired vhf frequency channel, amplifies composite video signals in the frequency channel selected, and converts the signal frequencies to the 45.75-MHz picture intermediate frequency and the 41.25-MHz sound intermediate frequency used in television receivers. When used with a uhf tuner, the vhf tuner is operated as a two-stage broadband rf amplifier tuned to 44 MHz (center frequency of the if band) and is essentially a pre-if amplifier for the television receiver. In each mode of operation, the tuner has a band pass that is broad enough to pass all the video information (including synchronizing and equalizing pulses) and the sound information superimposed on the video and sound carrier frequencies and has sufficient selectivity to assure adequate adjacent-channel and image-frequency rejection. The +140 volts used as the B+ supply for the vhf tuner is obtained from the low-voltage power supply of the receiver. The heaters of the tubes in the circuit are connected in series with those of other tubes in the receiver, and power for the series heater string is obtained directly from the input ac power line.

The antenna used with the vhf tuner may be either a 75-ohm monopole, as used with portable receivers, or a balanced 300-ohm antenna. A balanced 300-ohm antenna system can be matched to the unbalanced 75-ohm tuner input by means of the antenna-matching balun T_1. A 13-position channel selector, which consists of several wafer-switch sections (S_1 through S_4) mounted on a common shaft, establishes the operating frequency of the tuner for each of the vhf channels 2 through 13 or adapts the vhf tuner for operation with a uhf tuner. With S_4 set to any of the channel positions 2 through 13, the selected-channel signal from the vhf antenna is coupled through contacts U and 2 of S_{4B} and input transformer T_2 to the rf amplifier, and the input lead from the uhf tuner is not connected to the vhf circuit.

The vhf input signals are amplified by the 3GK5 high-mu frame-grid triode used in the rf amplifier stage. The S_3 section of the channel selector connects the appropriate combination of the inductors L_6 through L_{16} into the grid circuit of the rf amplifier to tune this stage to the desired frequency channel. The agc bias voltage applied to the control grid of the 3GK5 triode automatically controls the gain of the rf stage. The bias voltage, which varies directly with the amplitude of the received signal, is derived by a keyed agc amplifier in the television receiver.

The output of the rf amplifier is coupled through a resonant impedance network to the control grid of

VHF TUNER (Cont'd)



26-25 VHF TUNER (Cont'd)

Parts List

C_1, C_2=82 pF, ±5%, dual disc, ceramic, 500 V, N750
C_3, C_4, C_5, C_{15}, C_{16}=1000 pF, feedthrough, 500 V
C_6=12 pF, 5%, ceramic, 500 V, N750
C_7=20 pF, ±5%, feedthrough, 500 V, N470
C_8=0.56 pF, ±5%, headed lead, 500 V
C_9=100 pF, ceramic, 500 V, N1500
C_{10}=0.22 μF, ceramic, 500 V
C_{11}=0.82 pF, headed lead, 500 V
C_{12}=82 pF, ±5%, feedthrough, 500 V, N750
C_{13}=8 pF, ceramic, 500 V
C_{14}=10 pF, ±5%, radial leads, ceramic, 500 V, N330
GIMMICK=Trimmer-capacitor plate
L_1, L_2, L_3=RF coils; with two 82-picofarad capacitors, forms high-pass filter (antenna input network), RCA Stock No. 114458 or equiv.
L_4=RF amplifier grid coil, part of S_3 assembly
L_5 through L_{15}=RF-amplifier tuning coils, part of S_3 assembly
L_{16}=Mixer grid coil, part of S_2 assembly
L_{17}=Interstage coupling coil for rf amplifier and mixer, part of S_2 assembly
L_{18} through L_{29}=Mixer tuning coils, part of S_2 assembly
L_{30}=Variable rf coil; mixer plate tuning adjustment; RCA stock No. 112909 or equiv.
L_{31}=RF choke
L_{32}=Variable rf coil; local-oscillator tuning adjustment for channel 13
L_{33} through L_{43}=Local-oscillator tuning coils (variable coil L_{39} is tuning adjustment for channel 6), part of S_1 assembly
L_{44}=Variable rf coil; fine-tuning control; RCA Stock No. 113323, or equiv.
R_1=4700 ohms, 1 watt
R_2=5600 ohms, 0.5 watt
R_3=47000 ohms, 0.5 watt
R_4=0.1 megohm, 0.5 watt
R_5, R_7=10000 ohms, 0.5 watt
R_6, R_{10}=1000 ohms, 0.5 watt
R_8=2200 ohms, 0.5 watt
R_9=6800 ohms, 0.5 watt
S_1=Local-oscillator section of channel-selector switch; stator assembly, RCA Stock No. 114462 or equiv., includes local-oscillator tuning coils L_{33} through L_{43}
S_2=Mixer section of channel-selector switch; stator assembly, RCA Stock No. 114461 or equiv., includes mixer tuning coils L_5, L_6, and L_{18} through L_{29}
S_3=RF amplifier section of channel-selector switch; stator assembly, RCA Stock No. 114460 or equiv., includes rf-amplifier tuning coils L_4 and L_7 through L_{17}
S_4=VHF-UHF function selector; two-section switch ganged with channel selectors, S_1, S_2, and S_3; RCA Stock No. 114185 or equiv.
T_1=Antenna-matching balun; matches 300-ohm balanced antenna-lead line to 75-ohm unbalanced receiver-input line; RCA Stock No. 111973 or equiv.
T_2=Antenna transformer; RCA Stock No. 113195 or equiv.
Z_1, Z_2=Resistance-capacitance network (capristor), RCA Stock No. 109956 or equiv.

Notes:
1. All switches are ganged together on same shaft and are shown with shaft in channel 13 position.
2. Voltages shown are obtained with no signal input.
3. For dc voltage and heater supply, see circuit 26-29, page 616.
4. See additional notes on page 603.

Circuit Description (Cont'd)

the 6KZ8 pentode section used in the mixer stage. Section S_2 of the ganged channel selector selects the proper combination of the inductors L_{18} through L_{29} to tune the mixer input circuit to the same operating frequency as that of the rf amplifier. A signal from the plate of the 6KZ8 triode section used in the local-oscillator stage is also applied to the input circuit of the mixer. Section S_1 of the channel selector connects the right combination of the inductors L_{33} through L_{43} into the oscillator resonant circuit to maintain the operating frequency of the oscillator at 45.75 MHz above the video carrier frequency (41.25 MHz above the sound carrier frequency) of the vhf channel selected by the tuner. Inductor L_{44} in the series-resonant feedback circuit of the oscillator is the fine-tuning adjustment for the vhf tuner. This adjustment assures that the oscillator frequency accurately tracks the input tuning in each channel.

The signals from the rf amplifier and the local oscillator are heterodyned in the mixer stage to produce the 45.75-MHz amplitude-modulated and 41.25-MHz frequency-modulated difference frequencies used as the picture and sound intermediate frequencies, respectively, in the television receiver. The picture and sound if signals are coupled from the plate of the mixer to the if stages of the receiver.

When the multiple-section channel selector is rotated to the U position (for uhf operation), a connection from the B+ line of the vhf tuner through a 5600-ohm dropping resistor R_2, contacts 4 and 10 of S_{4A},

26-25 VHF TUNER (Cont'd)

Circuit Description (Cont'd)

and a 4700-ohm dropping resistor R_1 provides the B+ voltage for the uhf tuner. In addition, transformer T_2, which provides the input to the rf amplifier, is connected through contacts 2 and 13 of S_{4B} to the output of the uhf tuner, and the signal from the vhf antenna is shorted to ground through contacts U and 12 of S_{4A}. The input to the rf amplifier is then the amplitude-modulated 45.75-MHz picture if and frequency-modulated 41.25-MHz sound if signals from the uhf tuner.

In the U positions, switch sections S_3 and S_2 select the tuning inductors required for operation of the rf amplifier and mixer stages as broadband 44-MHz amplifiers, and section S_1 disables the oscillator stage by connection of the oscillator control grid directly to ground through switch contacts 2 and U. With these changes, the vhf tuner essentially becomes a broadband 44-MHz amplifier which provides two stages of amplification of the picture and sound if signals ahead of the receiver main if strip.

26-26 VIDEO-IF AMPLIFIERS AND SOUND-CHANNEL CIRCUITS

For Black-and-White Television Receiver

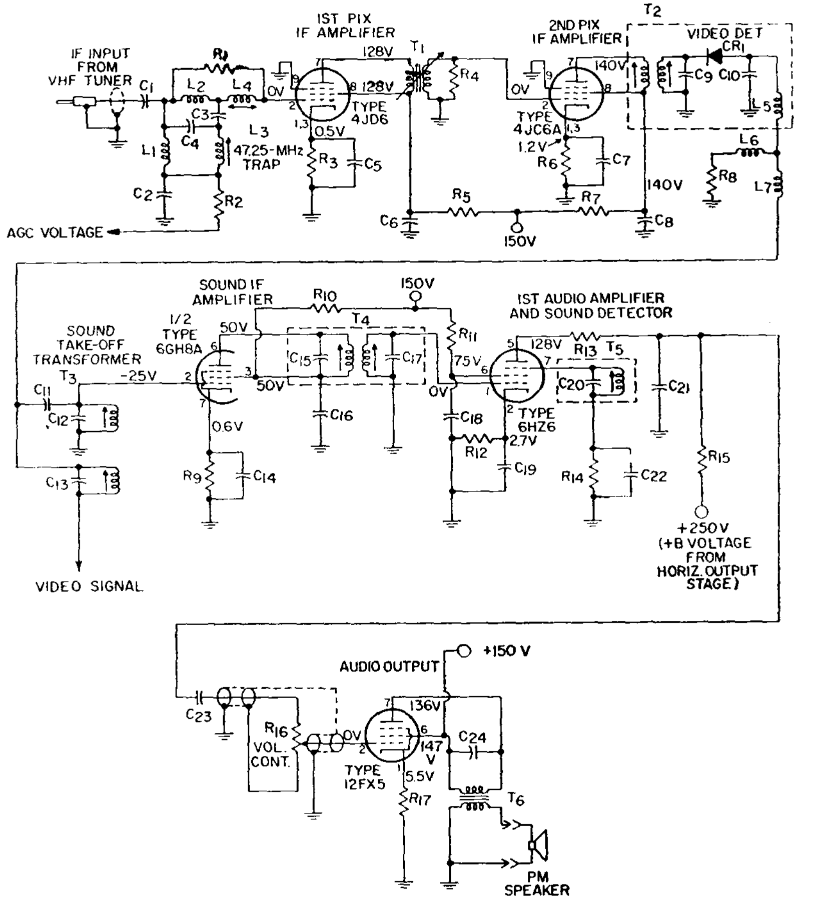
Circuit Description

These circuit stages are typical of those used in the if and audio channels of any intercarrier type of black-and-white television receiver. The over-all circuit operates from a dc supply of +150 volts obtained from the receiver low-voltage (B+) dc power supply. The heaters of the tubes in the circuit are connected in series with those of tubes in other sections of the receiver. Operating power for the series heater string is obtained directly from the 117-volt ac power line.

The input from the vhf tuner consists of amplitude-modulated 45.75-MHz picture if signals and frequency-modulated 41.25-MHz sound if signals. This composite input is coupled by a broadly tuned bandpass filter network to the control grid of the 4JD6 remote-cutoff pentode used in the first picture if amplifier. A dc bias voltage proportional to the input signal from the agc amplifier is also applied to the control-grid circuit to provide automatic gain control of this stage. The output of the first picture if amplifier is coupled by the single-tuned transformer T_1 to the control grid of the 4JC6A pentode used in the second picture if amplifier. The double-tuned transformer T_2 couples the output of this stage to the video detector (CR_1 and associated components). The input filter network and picture if transformers T_1 and T_2 are stagger tuned to obtain the broad response for the if amplifiers required to assure adequate passage of both the 45.75-MHz video and 41.25-MHz sound if signals.

The video detector demodulates the 45.75-MHz picture if signal, and the resultant video signal is coupled through inductors L_6 and L_7 and the lower winding of transformer T_3 to the video amplifier (shown in circuit 26-27). The video detector also operates as a second mixer circuit. The 45.75-MHz picture if signal and the 41.25 sound if signal are heterodyned to produce a second sound if carrier of 4.5 MHz. This 4.5-MHz second sound if carrier is still frequency-modulated by the audio components contained in the original rf signal input at the receiver antenna. The sound-takeoff transformer T_3, which forms a selective load impedance for the detector circuit at 4.5 MHz, couples the 4.5-MHz sound if signal to the control grid of the pentode section of a 6GH8A triode pentode used in the sound if ampli-

26-26 VIDEO-IF AMPLIFIERS AND SOUND-CHANNEL CIRCUITS (Cont'd)



Parts List

C_1, C_6=470 pF, ceramic, 500 V
C_2, C_7=0.001 μF, ceramic 500 V
C_3=7 pF, ceramic, 500 V, N150
C_4=2 pF, ceramic, 500 V, NPO
C_5=56 pF, ±5%, ceramic, 500 V, N750
C_8=560 pF, ceramic, 500 V
C_9=18 pF, 5%, ceramic, 500 V, N220
C_{10}=6 pF, ceramic, 500 V
C_{11}=10 pF, ceramic, 500 V, NPO
C_{12}=39 pF, ceramic, 500 V, N150
C_{13}=68 pF, ceramic, 500 V, N750
C_{14}, C_{19}=0.01 μF, ceramic, 500 V
C_{15}, C_{17}=12 pF, part of T_4
C_{16}, C_{18}=0.0022 μF, ceramic, 500 V
C_{20}=10 pF, part of T_5
C_{21}=680 pF, ceramic, 500 V
C_{22}=0.047 μF, paper, 200 V
C_{23}=0.01 μF, ceramic, 500 V
C_{24}=0.0068 μF, ceramic, 500 V
CR_1=Video detector, crystal diode, RCA Stock No. 112524 or equiv.
L_1=RF coil, RCA Stock No. 114315 or equiv.
L_2=RF coil, RCA Stock No. 114314 or equiv.
L_3=RF coil, 47.25-MHz trap RCA Stock No. 113097 or equiv.
L_4=RF coil, RCA Stock No. 113097 or equiv.
L_5=Video-detector peaking coil, 36 μH, RCA Stock No. 109758 or equiv.
L_7=Filter choke (reactor), 2.7 μH, RCA Stock No. 107463 or equiv.
R_1=3300 ohms, 0.5 watt
R_2=1000 ohms, 0.5 watt
R_3=39 ohms, ±5%, 0.5 watt
R_4=4700 ohms, ±5%, 0.5 watt
R_5=1500 ohms, 1 watt
R_6=100 ohms, 0.5 watt
R_7=470 ohms, 0.5 watt
R_8=3000 ohms, ±5%, 0.5 watt

VIDEO-IF AMPLIFIERS AND SOUND-CHANNEL CIRCUITS (Cont'd)

Parts List (Cont'd)

R_9=820 ohms, 0.5 watt
R_{10}=82000 ohms, 0.5 watt
R_{11}=15000 ohms, 1 watt
R_{12}=560 ohms, 0.5 watt
R_{13}=470 ohms, 0.5 watt
R_{14}=0.47 megohm, 0.5 watt
R_{15}=0.39 megohm, 0.5 watt
R_{16}=Volume control, potentiometer, 1 megohm
R_{17}=180 ohms, 0.5 watt
T_1=First pix if transformer, RCA Stock No. 109158 or equiv.
T_2=Second pix if transformer, RCA Stock No. 114317 or equiv.
T_3=Sound take-off transformer, 4.5-MHz, RCA Stock No. 114489 or equiv.
T_4=Sound if transformer (includes primary and secondary capacitors), RCA Stock No. 104137 or equiv.
T_5=Sound detector resonant circuit (includes 10-pF capacitor), RCA Stock No. 109948 or equiv.
T_6=Audio output transformer, matches speaker voice-coil impedance to tube plate load, RCA Stock No. 114490 or equiv.

Notes: 1. Voltages shown are obtained with no signal input.
2. For dc voltage and heater supply, see circuit 26-29, page 616.
3. See additional notes on page 603.

Circuit Description (Cont'd)

fier. The amplified if signal from this stage is coupled by the doubled-tuned 4.5-MHz transformer T_4 to the 6HZ6 audio detector-amplifier stage. This stage demodulates the 4.5-MHz sound if signal and amplifies the resultant audio signal voltage. The +250 volts used as the plate supply for the 6HZ6 is obtained from the horizontal output stage (shown in circuit 26-28) of the receiver.

The audio-signal power required to drive the speaker is developed by a 12FX5 pentode used in a single-ended audio output stage. The audio-signal voltage from the plate of the audio detector-amplifier is amplified by the 12FX5 and coupled by transformer T_6 to the voice coil of the speaker. The volume-control potentiometer R_{16} in the input circuit of the output stage provides manual adjustment of the sound level from the speaker.

26-27 VIDEO, AGC, AND SYNC AMPLIFIERS

For Black-and-White TV Receiver

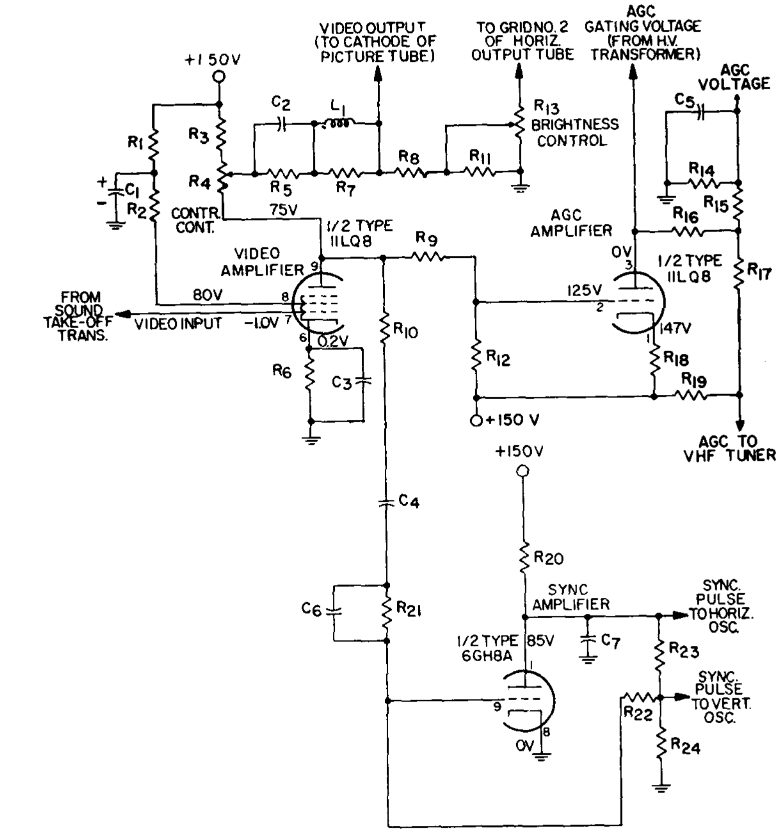
Circuit Description

This circuit shows video, agc, and sync amplifiers for a black-and-white television receiver. The video and sync amplifiers operate from a plate supply (B+) voltage of 150 volts obtained from the receiver low-voltage power supply. The plate supply voltage for the agc amplifier is a positive keying pulse from the high-voltage transformer in the receiver. The heaters of the three tubes are connected in series with those of tubes in other sections of the receiver. Operating power for the series heater string is obtained directly from the ac power line.

In the video amplifier, the pentode section of an 11LQ8 triode-pentode provides the required amplification of the video signal. The video signal is coupled from the video detector to the control grid of the video amplifier. The output from the voltage divider in the plate circuit of this stage is applied to the cathode of the picture tube to intensity-modulate the electron beam during its vertical and horizontal scanning of the picture-tube screen. The contrast control adjusts both the amplitude of the video output and the dc potential at the cathode of the picture tube to control picture contrast. The voltage-divider network in the plate circuit of the video amplifier is interconnected with another voltage-divider network. This second network includes the brightness control and the width control in the screen-grid circuit of the receiver horizontal-output tube (shown in circuit 25-29). The brightness control adjusts the cathode bias on the picture tube to control the intensity of the screen display.

An output from the video amplifier is also applied to the control grid of the 11LQ8 triode section used

26-27 VIDEO, AGC, AND SYNC AMPLIFIERS (Cont'd)



Notes: 1. Voltages shown are obtained with no signal input.
2. For dc voltage and heater supply, see circuit 26-29, page 616.
3. See additional notes on page 603.

Parts List

C_1=5 μF, electrolytic, 150 V
C_2=0.15 μF, paper, 200 V
C_3=0.033, paper, 200 V
C_4=0.0047, ceramic, 500 V
C_5=0.1 μF, paper, 200 V
C_6=470 pF, ceramic, 500 V
C_7=100 pF, ceramic, 500 V, N1500
L_1=Video-amplifier peaking coil, 18 μH, RCA Stock No. 109946 or equiv.
R_1=18000 ohms, 0.5 watt
R_2=330 ohms, 0.5 watt
R_3=1500 ohms, 0.5 watt
R_4=Contrast control, potentiometer, 4000 ohms, 3 watts
R_5=1 megohm, 0.5 watt
R_6=10 ohms, ±5%, 0.5 watt
R_7=22000 ohms, 0.5 watt
R_8=0.27 megohm, 0.5 watt
R_9, R_{10}, R_{20}=27000 ohms, 0.5 watt
R_{11}=27000 ohms, 1 watt
R_{12}=18000 ohms, 0.5 watt
R_{13}=Brightness control, potentiometer, 0.1 megohm
R_{14}, R_{17}=0.82 megohm, 0.5 watt
R_{15}=1 megohm, 0.5 watt
R_{16}, R_{21}=0.68 megohm, 0.5 watt
R_{18}=3300 ohms, 0.5 watt
R_{19}=8.2 megohms, 0.5 watt
R_{22}=5.2 megohms, 0.5 watt
R_{23}=33000 ohms, 0.5 watt
R_{24}=15000 ohms, 0.5 watt

Circuit Description (Cont'd)

in a keyed-agc amplifier stage. The operation of the agc amplifier is gated (keyed) by a positive pulse from the high-voltage power transformer (shown in circuit 26-28). This 450-volt keying pulse, which is synchronized with the video signal, overcomes the bias provided by the 150 volts applied to the cathode circuit and serves as the plate supply

26-27 VIDEO, AGC, AND SYNC AMPLIFIERS (Cont'd)

Circuit Description (Cont'd)

voltage for the agc amplifier. Portions of the video signal that occur coincident with the keying pulse are amplified by the agc stage. A 0.1-microfarad capacitor C_5 and a 0.82-megohm resistor R_{14} in the plate circuit of this stage filter out the pulsating components to obtain a negative dc voltage proportional to the video signal and thus to the rf input at the receiver antenna. The negative voltage developed in the plate circuit of the stage is applied as agc bias to the first picture if amplifier and to the rf amplifier in the vhf tuner.

Synchronizing pulses are included in the video signals transmitted by a television broadcast station to provide timing information required for synchronization of the transmitter and receiver scanning systems. The sync amplifier, or separator, separates and amplifies the synchronizing pulses contained in the composite video signal it receives from the plate circuit of the video amplifier. The circuit uses the triode section of a 6GH8A triode-pentode to develop the synchronizing pulses for the vertical- and horizontal-deflection circuits of the receiver. The sync amplifier is basically a class C limiter stage. With the video signal applied, the stage is biased beyond cutoff by the grid-leak bias network formed by the 470-picofarad capacitor C_6 and the 0.68-megohm resistor R_{21} in the control-grid circuit. Only the sync pulses in the composite video signal have sufficient amplitude to drive the sync amplifier into conduction. The resultant pulses developed across the output voltage-divider network are used as the synchronizing inputs to the horizontal- and vertical-deflection circuits.

26-28 VERTICAL- AND HORIZONTAL-DEFLECTION CIRCUITS AND HIGH-VOLTAGE RECTIFIER

For Black-and-White Television Receiver

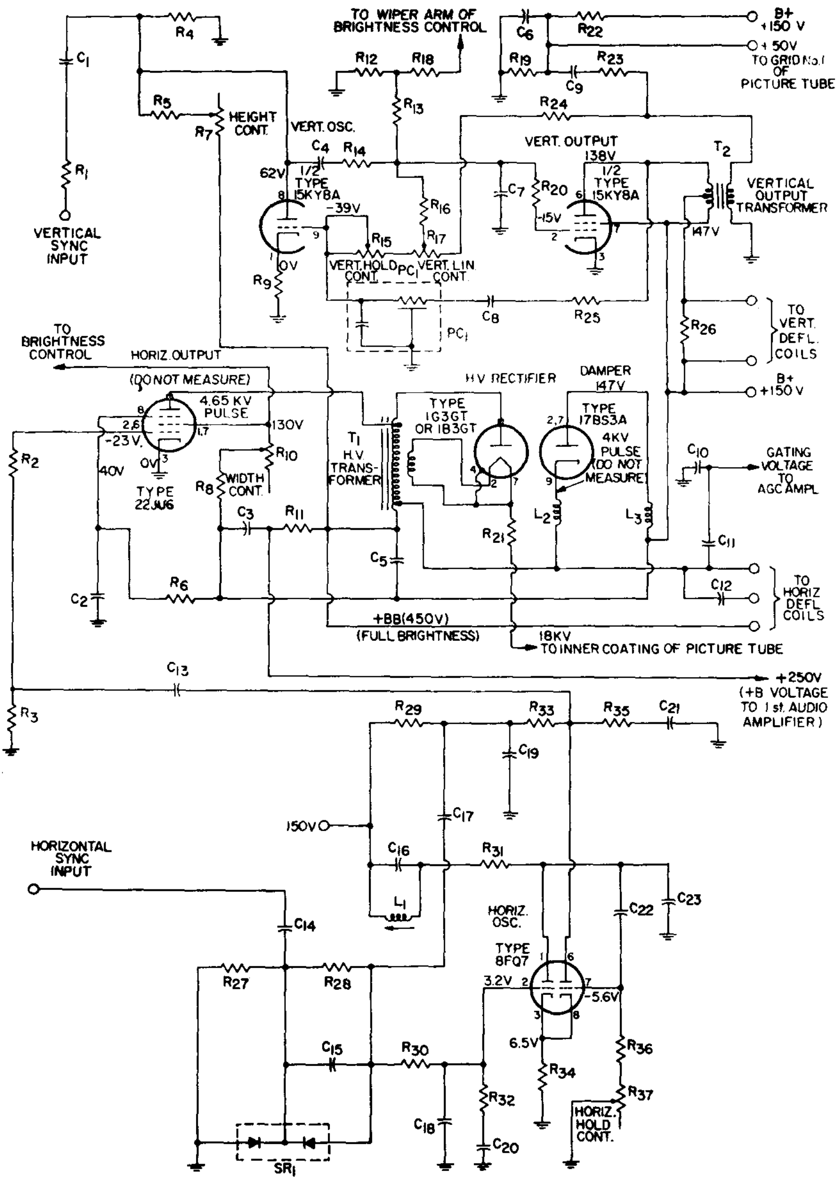
Circuit Description

These circuits develop the vertical and horizontal scanning signals and the dc operating potentials for the picture tube (RCA Type 16BGP4) used in the black-and-white television receiver and the boosted B+ voltage (+250 volts) used in the audio detector-amplifier (part of circuit 26-27). The circuits operate from a dc supply of 150 volts. With the exception of the 1G3GT (or 1B3GT) high-voltage rectifier tube, the heaters of the various tubes are connected in series with those of tubes in other sections of the receiver and are supplied by the input ac power line. Heater power for the 1G3GT (or 1B3GT) is provided by a 1.25-volt winding of the high-voltage tarnsformer T_1.

The vertical- and horizontal-deflection circuits are synchronized by negative signals from the sync amplifier (separator) which include horizontal sync pulses, equalizing pulses, and vertical sync pulses. When the composite video signal is generated at the television broadcast station, the leading edge of each horizontal sync pulse, of alternate equalizing pulses, and of alternate serrations of the vertical sync pulses are correctly timed to initiate the horizontal-retrace period. It is necessary, therefore, to extract the leading-edge components from the combined sync waveform prior to application of the synchronizing input to the horizontal-deflection circuit. Similarly, the vertical sync pulses must be separated from the combined waveform before they can be used to synchronize the vertical-deflection circuit.

The combined sync waveform is differentiated at the input to the

26-28 VERTICAL- AND HORIZONTAL-DEFLECTION CIRCUITS AND HIGH-VOLTAGE RECTIFIER (Cont'd)



Notes: 1. Voltages shown are obtained with no signal input.
2. For dc voltage and heater supply, see circuit 26-29, page 616.
3. See additional notes on page 603.

26-28 VERTICAL- AND HORIZONTAL-DEFLECTION CIRCUITS AND HIGH-VOLTAGE RECTIFIER (Cont'd)

Parts List

C_1=0.0039 μF, ceramic, 500 V, N5600
C_2=0.01μF, ceramic, 500 V
C_3, C_9=0.047 μF, paper, 200 V
C_4=0.033 μF, paper, 200 V
C_5=0.027 μF, paper, 600 V
C_6=0.015 μF, tubular paper, 200 V
C_7=0.022 μF, paper, 200 V
C_8=0.0022 μF, paper, 1000 V
C_{10}=0.0012 μF, ±5%, ceramic, 500 V, N3300
C_{11}=180 pF, ±5%, ceramic, 5000 V, N2200
C_{12}=47 pF, ceramic, 2500 V, N1500
C_{13}=0.0033 μF, ceramic, 500 V
C_{14}=68 pF, paper, 500 V, N1500
C_{15}=470 pF, ceramic, 500 V
C_{16}=0.0039 μF, mylar, 400 V
C_{17}=0.001 μF, ceramic, 500 V
C_{18}=0.0033 μF, ceramic, 500 V
C_{19}=0.001 μF, ceramic, 500 V
C_{20}=0.056 μF, paper, 200 V
C_{21}=150 pF, ceramic, 500 V
C_{22}=390 pF, mica, 500
C_{23}=68 pF, ceramic, 500 V, NPO
L_1=Oscillator coil, RCA Stock No. 114486 or equiv.
L_2, L_3=RF chokes (reactors), 8.2 μH, RCA Stock No. 107385 or equiv.
PC_1=Printed circuit (includes 0.001-μF and 0.0024-μF capacitors and 68000-ohm resistor), RCA Stock No. 114506 or equiv.
R_1=0.1 megohm, 0.5 watt
R_2=47 ohms, 0.5 watt
R_3, R_4=0.82 megohm, 0.5 watt
R_5=2.2 megohms, 0.5 watt
R_6=47000 ohms, 0.5 watt
R_7=Height control, potentiometer, 0.75 megohm
R_8=820 ohms, 1 watt
R_9=3300 ohms, 0.5 watt
R_{10}=Width control, potentiometer, 2000 ohms, 3 watts
R_{11}=0.68 megohm, 0.5 watt
R_{12}=47000 ohms, 0.5 watt
R_{13}=22 megohms, 0.5 watt
R_{14}=22000 ohms, 0.5 watt
R_{15}=Vertical-hold control, potentiometer, 0.75 megohm
R_{16}=1.8 megohms, 0.5 watt
R_{17}=Vertical-linearity control, potentiometer, 0.2 megohm
R_{18}=0.47 megohm, 0.5 watt
R_{19}, R_{25}=27000 ohms, 0.5 watt
R_{20}, R_{21}=1000 ohms, 0.5 watt
R_{22}=68000 ohms, 0.5 watt
R_{23}=10000 ohms, 0.5 watt
R_{24}=0.18 megohm, 0.5 watt
R_{26}=820 ohms, 0.5 watt
R_{27}=0.15 megohm, 0.5 watt
R_{28}=0.39 megohm, 0.5 watt
R_{29}=12000 ohms, 0.5 watt
R_{30}=1 megohm, 0.5 watt
R_{31}=15000 ohms, 0.5 watt
R_{32}=68000 ohms, 0.5 watt
R_{33}=33000 ohms, 0.5 watt
R_{34}=1500 ohms, ±5%, 0.5 watt
R_{35}=4700 ohms, 0.5 watt
R_{36}=47000 ohms, 0.5 watt
R_{37}=Horizontal-hold control, potentiometer, 70000 ohms.
SR_1=Selenium rectifier, RCA Stock No. 109474 or equiv.
T_1=High-voltage and horizontal-output transformer, RCA Stock No. 114498 or equiv.
T_2=Vertical-output transformer, RCA Stock No. 114502 or equiv.

Circuit Description (Cont'd)

horizontal-deflection circuit to obtain negative and positive voltage spikes which correspond to the leading and lagging edges, respectively, of the rectangular sync pulses. The amplitude of these voltage spikes is dependent upon only the peak value of the sync pulses and is not affected by the time durations of these pulses. The differentiating circuit, therefore, does not respond to the flat portions of the vertical sync pulses, and, with the exceptions of the serrations, the vertical sync pulses do not affect the operation of the horizontal-deflection circuits. The leading edge of alternate serrations, however, corresponds to the start of horizontal-retrace periods and thus may be considered as merely another horizontal sync signal.

The differentiated sync waveform is applied to the junction of the twin silicon diodes SR_1 used in a phase-discriminator network. The positive portion of the differentiated waveform has no effect on the discriminator network. The negative portion is compared with a feedback signal from the horizontal oscillator to derive the synchronizing voltage. The frequency of the horizontal oscillator and the repetition rate of the horizontal sync pulses should both be 15,750 Hz, the desired horizontal scanning rate for the picture tube. If the feedback signal from the oscillator does not occur coincident with the horizontal sync pulse, the phase discriminator develops a dc error voltage at the control grid of the input section of the 8FQ7 twin triode used in the oscillator stage. The resultant change in oscillator bias shifts the phase of the oscillator signal until it is locked in phase with the horizontal sync pulse.

The horizontal oscillator is basically a cathode-coupled multivibrator that free-runs, in asymmetrical half cycles, at a frequency of 15,750 Hz. A parallel LC circuit connected in series with the plate of the input section resonates at 15,750 Hz to provide frequency stabilization for the horizontal oscillator. The HOLD con-

26-28 VERTICAL- AND HORIZONTAL-DEFLECTION CIRCUITS AND HIGH-VOLTAGE RECTIFIER (Cont'd)

Circuit Description (Cont'd)

trol adjusts the basic multivibrator frequency to achieve an exact lock-in with the horizontal sync pulses. In a cathode-coupled multivibrator, one amplifier section conducts at saturation and the other section is cut off during one half-cycle of operation, and these states are automatically reversed for the next half cycle. Such circuits normally provide rectangular-wave outputs from each plate section that are 180 degrees out of phase and that switch between the saturation plate voltage and B+ (i.e., the cutoff plate voltage).

In the horizontal oscillator a series RC network is connected in parallel with the output tube section. Because of this network, the plate voltage does not immediately rise to the B+ value when the output tube section is cut off. Instead, there is a small immediate rise in plate voltage that results from the voltage drop across the resistor R_{35} in the output RC network produced by the initial charging current to the capacitor C_{21}. The plate voltage then rises gradually at a rate determined by the long-time-constant circuit through which the capacitor is charged. Before the capacitor can fully charge to the B+ voltage, the combination of the horizontal sync input and the feedback signal from the plate of the output section of the oscillator drives the grid of the input section below cutoff. The instantaneous rise in the plate voltage of the input section is coupled to the grid of the output section and causes this section to conduct. The capacitor C_{21} in the output RC network is then quickly discharged through the series resistor and the relatively low resistance of the output tube section. The output of the horizontal oscillator, therefore, is a trapezoidal voltage wave. The rising-slope portions of this wave (obtained when the ouput tube section is cut off) corresponds to the horizontal-trace period on the picture tube; the discharge portion of the trapezoidal wave corresponds to the retrace period. The time-constant coupling circuits between the input and output sections of the oscillator are designed so that the retrace period represents only about 5 to 10 per cent of the over-all oscillator cycle.

The trapezoidal voltage wave is coupled to the control grid of the 22JU6 pentode horizontal-output stage and causes a sawtooth current to flow through the high-voltage (flyback) transformer T_1 and through the horizontal-deflection coils of the picture tube. The gradually rising portion of the sawtooth current causes the horizontal scanning of the picture tube; the more rapid negative-slope portion of the current wave causes the retrace. During the retrace period, the picture-tube screen is blanked by a negative pulse applied to the control grid of the picture tube from the vertical-deflection circuits. The WIDTH control R_{10} in the screen grid of the horizontal-output stage adjusts the gain of this stage to control the width of horizontal scanning.

The vertical oscillator employs a 15KY8A triode-pentode in a basic plate-coupled multivibrator configuration. This free-running 60-Hz multivibrator is synchronized by the vertical sync pulses. The vertical pulses are separated from the combined sync waveform by integration of the combined waveform across the 0.022-microfarad capacitor C_7 in the control-grid circuit of the pentode output section of the multivibrator. The integrating network has negligible response for the narrow horizontal sync and equalizing pulses, but responds to the greater energy included in the much wider vertical sync pulses to develop a triangular voltage wave at the control grid of the pentode output section. The

26-28 VERTICAL- AND HORIZONTAL-DEFLECTION CIRCUITS AND HIGH-VOLTAGE RECTIFIER (Cont'd)

Circuit Description (Cont'd)

VERT LIN potentiometer R_{17} adjusts the charging period of the integrating capacitor to control vertical liearity. The VERT HOLD potentiometer R_{15} adjusts the frequency of the multivibrator to achieve an exact lock-in with the vertical sync pulses.

The voltage waveform at the control grid of the pentode output section results in a triangular wave of current through the vertical-output transformer T_2 and through the vertical-deflection coils of the picture tube. The rising portion of the triangular current wave produces the vertical scanning, and the decreasing portion of the wave provides the retrace. Blanking pulses to cut off the picture tube during vertical and horizontal retrace periods are coupled from the secondary of T_2 and from the VERT LIN potentiometer (combined sync waveform before integration) to the control grid of the picture tube.

The 1G3GT (or IB3GT) half-wave rectifier circuit develops the dc operating voltages for the picture tube. The ac input power to the rectifier is supplied by the horizontal-deflection circuits. The sudden cutoff of plate current in the horizontal-output stage at the beginning of the retrace period causes a very large, positive-going voltage pulse to be generated across the high-voltage transformer T_1. The rectifier converts this voltage pulse to a dc output voltage of approximately 18,000 volts, which is applied to the inner coating of the picture tube. Removal of negative overshoots that would be developed across the high-voltage transformer because of a flywheel effect is accomplished by connection of a 17BS3A rectifier (damper) tube across the horizontal-deflection coils which are in parallel with the lower tapped section of the high-voltage transformer. The polarity of the damper tube is such that the positive pulse developed across the high-voltage transformer causes no current flow through it. For negative pulses, however, the damper tube provides a low-impedance path for the current, and energy stored in the horizontal-deflection coils during the preceding half-cycle is dissipated as heat at the damper-tube plate to prevent oscillation in the coils. The current through the damper tube develops a dc voltage of 450 volts across the 0.027-microfarad capacitor C_5 in the cathode circuit. The 0.68-megohm dropping resistor R_{11} reduces this voltage to obtain the boosted B+ of 250 volts required for operation of the audio detector-amplifier (part of circuit 26-26).

26-29 LOW-VOLTAGE AND HEATER SUPPLY

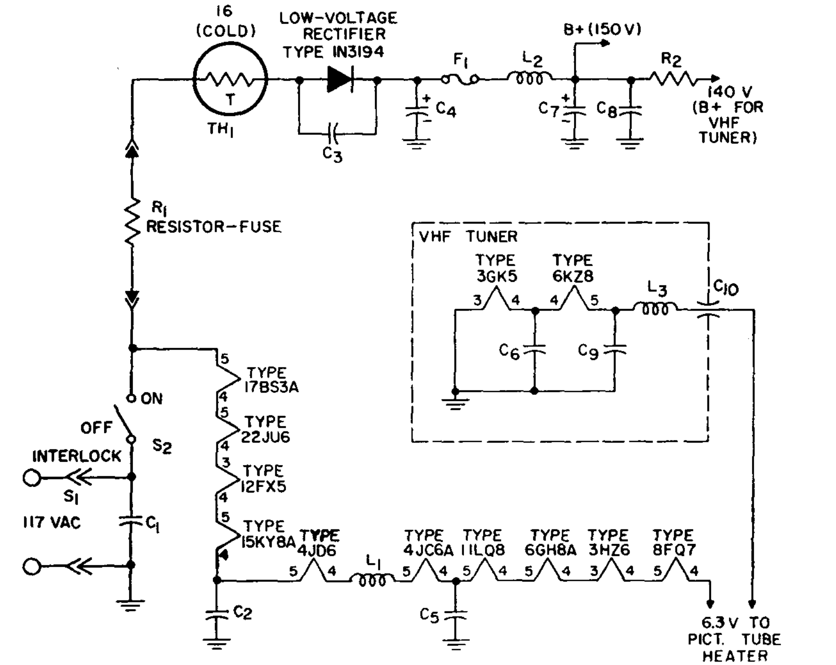
For Black-and-White TV Receiver

Circuit Description

This circuit includes the low-voltage (+150-volt) dc power supply and the series heater connections for circuits 26-25 through 26-28. As mentioned previously, the power supply and these four circuits comprise a complete black-and-white television receiver, with the exception of the picture tube and the vertical- and horizontal-deflection yokes.

The power supply is a half-wave type which uses a 1N3194 silicon rectifier. The 117-volt ac input is connected to the power supply through an interlock, S_1, which may be mounted on the back cover of the receiver. AC input power is then automatically disconnected from the receiver when the back cover is removed. ON-OFF switch S_2 controls the application of ac power to the power-supply circuit and to the tube heaters. With S_1 and S_2 both closed,

26-29 LOW-VOLTAGE AND HEATER SUPPLY (Cont'd)



Parts List

C_1=0.22 μF, paper, 600 V
C_2, C_5=0.001, ceramic, 500 V, part of assembly with L_1
C_3=680 pF, ceramic, 1000 V
C_4=250 μF, electrolytic, 200 V
C_6, C_9=680 pF, ceramic, 500 V
C_7=400 μF, electrolytic, 175 V
C_8=0.001 μF, ceramic, 500 V
C_{10}=1000 pF, feedthrough, 5000 V
F_1=Fuse, chemical, 0.45 ampere, RCA Stock No. 114446 or equiv.
L_1=RF choke, part of heater printed-circuit board, RCA Stock No. 114499 or equivalent (includes the two 0.001-μF capacitors C_2 and C_5)
L_2=Filter choke (reactor), RCA Stock No. 114501 or equiv.
L_3=RF choke for VHF tuner filament circuit
R_1=Resistor-fuse, 0.35 ohm, RCA Stock No. 114481 or equiv.
R_2=330 ohms, 1 watt
TH_1=Surge protection resistor (thermistor), 16 ohms (cold), RCA Stock No. 114480.

Circuit Description (Cont'd)

the 117-volt power from the ac power line is applied to the series heater network and to the 1N3194 rectifier circuit. Two 0.001-microfarad (C_2 and C_5) and two 680-picofarad (C_6 and C_9) bypass capacitors and rf chokes L_1 and L_3 are included in the heater circuit to filter out any stray high-frequency signals that may be coupled from the rf and if signal channels.

The 117-volt ac input is converted to pulsating dc by the 1N3194 silicon rectifier. A capacitor-input, pi-type LC filter network filters the rectifier output to obtain a smooth dc voltage that approaches the peak value of the input ac voltage. The 680-picofarad capacitor C_5 in parallel with the 1N3194 rectifier and the thermistor TH_1 in series with it provide surge-current protection for the rectifier. Initial surges of current that may result when power is first applied to the circuit (before a charge is developed across the input filter capacitor) are partially bypassed by the 680-picofarad capacitor and are limited in magnitude by the cold resistance of the thermistor. The

26-29 LOW-VOLTAGE AND HEATER SUPPLY (Cont'd)

Circuit Description (Cont'd)

thermistor has a negative temperature coefficient of resistance, and by the time the charge of the input capacitor C_4 builds up sufficiently to limit the current through the rectifier to a safe value, the resistance of the heated thermistor is small enough so that circuit power losses across this device are negligible. The resistor-fuse element R_1 in series with the 1N3194 rectifier provides protection against any continuous circuit overload. The +150-volt output from the power-supply filter network is used as the main B+ voltage for the television receiver. The 330-ohm, 1-watt dropping resistor R_2 at the output of the filter network reduces this voltage to the +140 volts required as the B+ voltage in the vhf tuner.

COLOR TELEVISION RECEIVER

Circuits 26-30 through 26-36 comprise a complete portable color television receiver. The brief signal-tracing analyses of these circuits assume that the reader has a basic knowledge of the purpose and operation of the various circuit sections of a color receiver. (The analyses can be more easily understood if the reader reviews the general discussions on television circuits given in the section on **Electron Tube Applications**, pages 57 through 65.) The receiver, which is essentially identical to the RCA Type CTC-22, features direct-line operation; the chassis of circuits 26-30 through 26-36, therefore, are connected to one side of the ac line during operation. Servicing of these circuits should not be attempted by persons not familiar with the precautions necessary when working on this type of equipment. (See notes 1 and 2 on page 603.)

Note: Circuits 26-30 through 26-36 are included in this manual primarily to illustrate applications of RCA electron tubes. Because of the exceptionally high voltages (up to 21,500 volts), high frequencies, and large bandwidths that are required and of the many special components that are used, home construction of these circuits is not recommended.

26-30 LOW-VOLTAGE POWER SUPPLY, DEGAUSSING COIL, AND HEATER CONNECTIONS

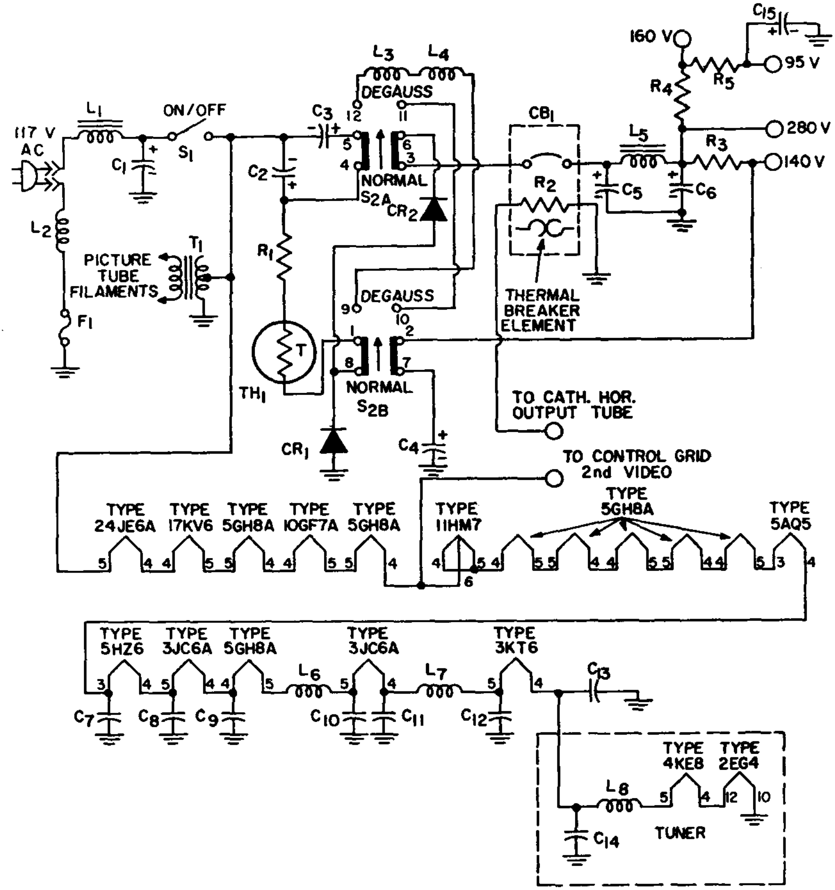
For Color Television Receiver

Circuit Description

This circuit includes the low-voltage (+280-volt) dc power supply, degaussing circuitry, and heater connections for a color television receiver. The tube heaters, with the exception of the color picture tube, are connected in series across the ac power line. Heater power for the picture tube is supplied by transformer T_1. With ON-OFF switch S_1 closed, the 117-volt power from the ac power line is applied to the series heater string and to the primary of transformer T_1. The 117-volt ac input power is stepped down by transformer T_1 to 6.3 volts at 0.8 amperes and applied to the heater of the 15LP22 color picture tube. Bypass capacitors and rf chokes are included in the series heater string to filter out any stray high-frequency signals that may be coupled from the rf and if signal channels of the receiver.

Two silicon rectifiers CR_1 and CR_2 are used in a voltage-doubler circuit to convert the 117-volt ac input power to the +280-volt B+ supply voltage for the receiver. This doubler circuit also provides a 160-volt output from the junction of resistors R_4 and R_5, a +140-volt output from the junction of resistor R_3 and capacitor C_4, and a 95-volt output from the junction of resistor R_6 and capacitor C_{15}. The dc voltage outputs

26-30 LOW-VOLTAGE POWER SUPPLY, DEGAUSSING COIL, AND HEATER CONNECTIONS (Cont'd)



Parts List

C_1=0.047 μF, paper, 600 V
C_2=250 μF, electrolytic, 175 V
C_3=50 μF, electrolytic, 250 V
C_4=100 μF, electrolytic, 300 V
C_5=150 μF, electrolytic, 350 V
C_6=100 μF, electrolytic, 350 V
C_7 through C_{14}=1000 pF, ceramic, 500 V
C_{15}=2 μF, electrolytic, 175 V
CB_1=Circuit breaker (includes R_2), RCA Stock No. 120784 or equiv.
CR_1, CR_2=Silicon rectifiers, RCA Stock No. 113998 or equiv.
F_1=Fuse, 7-ampere, 250-volt
L_1, L_2=Inductor, 60-Hz line filter
L_3, L_4=Degaussing coils, RCA Stock No. 120793 or equiv.
L_5=Filter choke, RCA Stock No. 120792 or equiv.
L_7, L_8=RF choke
R_1=2 ohms, wirewound, 7 watts
R_2=1.3 ohms, part of CB_1
R_3=3900 ohms, wirewound, 10 watts
R_4=47000 ohms, 1 watt
R_5=10000 ohms, 7 watts
S_1=ON-OFF switch, single-pole, single-throw
S_2=Degaussing switch, RCA Stock No. 120829 or equiv.
T_1=Filament transformer; primary, 117-volt; secondary, 6.3-volt, 8-ampere
TH_1=Thermistor; cold resistance, 120 ohms

See Note on page 617.

26-30 LOW-VOLTAGE POWER SUPPLY, DEGAUSSING COIL, AND HEATER CONNECTIONS (Cont'd)

Circuit Description (Cont'd)

are filtered by the pi-section filter network formed by L_5, C_5, and C_6.

The ac line is protected against any continuous circuit overload by a 7-ampere fuse, F_1, connected in series with one side of the line to ground. Surge protection is provided by a thermistor TH_1 connected in series with the B+ rectifiers (CR_1 and CR_2). The B+ circuit is protected by a special thermal reset circuit breaker CB_1. The circuit breaker opens the B+ line whenever the current demand on the low voltage power supply or the current through the horizontal output stage becomes excessive.

The circuit breaker has a resistive winding (approximately 1.3 ohms) that completes the ground return for the horizontal output tube. If the cathode current of the output tube becomes excessive, the resistive winding heats and causes the bi-metal strip in the circuit breaker to expand unequally. The resultant flexing of the bi-metal strip disconnects the breaker switch contacts and thereby opens the B+ line. The same action occurs when the B+ current demand becomes excessive.

Degaussing of the color receiver is initiated by depression of the spring-loaded switch S_2 to the DEGAUSS position. With S_2 in the NORMAL position, capacitors C_2 and C_3 are combined in parallel to provide the charging capacitance for the voltage-doubler circuit. For this condition, the parallel capacitors C_2 and C_3 are charged to approximately 142 volts and capacitor C_4 is charged to 140 volts to provide the +280-volt B+ voltage. When S_2 is depressed to the DEGAUSS position, capacitor C_2 is disconnected from the circuit, and degaussing coils L_3 and L_4 are connected in series with the power-supply rectifiers and capacitor C_8. When the line voltage swings positive, C_3 is charged through C_4, degaussing coils L_3 and L_4, and CR_2; when the line voltage is negative, C_3 is charged through CR_1 and the degaussing coils. This alternate cycling results in a symmetrical decaying wavetrain through the degaussing coils. The degaussing coils physically are looped about the receiver chassis in proximity to the color picture tube. The alternating magnetic fields developed by the decaying current wavetrain through these coils effectively demagnetizes the picture tube and adjacent chassis areas. The wavetrain decreases the zero when C_8 is charged to twice the peak value of the line voltage (approximately 330 volts dc). The degaussing action is completed in less than 1 second. It is only necessary, therefore, to momentarily depress switch S_2 to the DEGAUSS position. When the switch is released, it automatically returns to the NORMAL position.

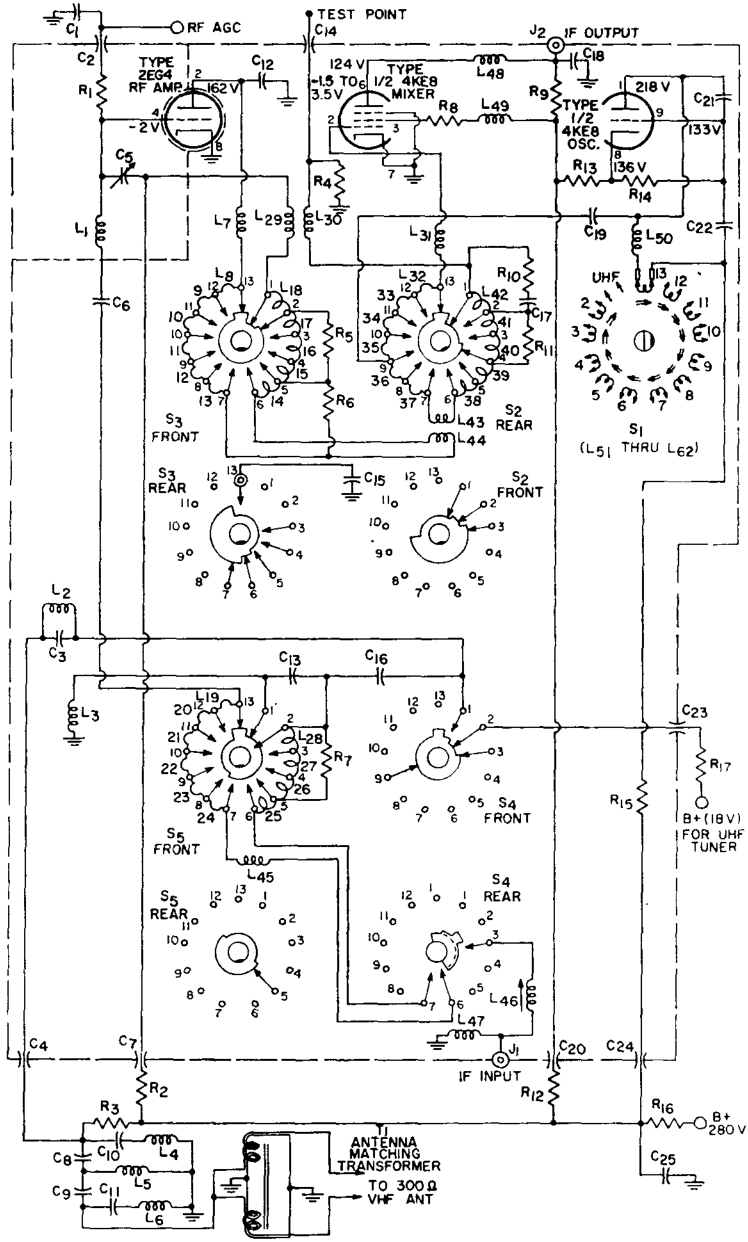
26-31 VHF TUNER

For Color Television Receiver

Circuit Description

This vhf tuner operates from a dc voltage of +280 volts obtained from the low-voltage power supply in the color television receiver. The tuner employs a 2EG4 nuvistor triode in the rf amplifier stage and uses a 4KE8 triode-pentode for the oscillator and mixer stages. The heaters of these tubes are connected in series with those of other tubes in the receiver; power for the series-heater string is obtained directly from the 117-volt ac power line. This tuner is very similar to

26-31 **VHF TUNER** (Cont'd)



Note: Switches S_1 through S_5 are ganged together on the same shaft and are shown in channel 13 position.

26-31 VHF TUNER (Cont'd)

Parts List

C_1=0.033 μF, paper, 200 V
C_2, C_{20}, C_{23}, C_{24}=1000 pF, feedthrough, 500 V
C_3=47 pF ±5%, ceramic, 500 V, N750
C_4=2 pF, feedthrough, RCA Stock No. 119595 or equiv.
C_5=Trimmer, 2 to 10 pF, RCA Stock No. 112038 or equiv.
C_6=27 pF ±5%, ceramic, 500 V, N750
C_7=47 pF, feedthrough, 500 V
C_8, C_9, C_{10}, C_{11}=27 pF ±5%, ceramic, 500 V, N470
C_{12}=2.7 pF, headed lead, 500 V
C_{13}=33 pF, ceramic, 500 V, N750
C_{14}=39 pF, feedthrough, 500 V
C_{15}=4.7 pF ±5%, headed lead, 500 V
C_{16}=680 pF, ceramic, 500 V
C_{17}=62 pF ±5%, ceramic, 1000 V, N1500
C_{18}=27 pF, ceramic, 500 V
C_{19}=2 pF, ceramic, 500 V, NPO
C_{21}=5.6 pF ±5%, ceramic, 500 V, N150
C_{22}=27 pF, ceramic, 500 V, NPO
C_{25}=0.047 μF, ceramic, 500 V
L_1=RF amplifier grid coil, part of S_3 assembly
L_2=UHF trap
L_3=RF amplifier grid-circuit coil, part of S_5 assembly
L_4, L_5, L_6=Filter coils for high-pass filter network, part of T_1 assembly
L_7=RF amplifier plate coil, part of S_3 assembly
L_8 through L_{18}=RF amplifier plate-circuit tuning coils, part of S_3 assembly
L_{19} through L_{28}=Antenna tuning coils, part of S_5 assembly
L_{29}, L_{30}=High-band coupling adjust coils
L_{31}=Mixer grid coil, part of S_2 assembly
L_{32} through L_{42}=Mixer tuning coils, part of S_2 assembly
L_{43}, L_{44}=Low-band coupling adjust
L_{45}=RF amplifier grid-circuit coil, part of S_5 assembly
L_{46}=IF input coil for signals from uhf tuner, RCA Stock No. 120782 or equiv.
L_{47}=RF coil, part of input circuit for signals from uhf tuner
L_{48}=Mixer plate coil, RCA Stock No. 112909 or equiv.
L_{49}=RF filter coil
L_{50}=Channel 13 range-centering coil
L_{51} through L_{62}=Local-oscillator tuning coils, part of S_1 assembly
J_1, J_2=Single-contact female connector, RCA Stock No. 104039 or equiv.
R_1=47000 ohms, 0.5 watt
R_2=16000 ohms, 3 watts
R_3=4700 ohms, 1 watt
R_4=82000 ohms, 0.5 watt
R_5=1500 ohms, 0.5 watt
R_6=10000 ohms, 0.5 watt
R_7=2200 ohms, 0.5 watt
R_8, R_{10}=10 ohms, 0.5 watt
R_9, R_{13}=1000 ohms, 0.5 watt
R_{11}=27000 ohms, 0.5 watt
R_{12}=68000 ohms, 1 watt
R_{14}=5600 ohms, 0.5 watt
R_{15}=6800 ohms, 0.5 watt
R_{16}=680 ohms, 1 watt
S_1=Local-oscillator section of channel-selector switch; stator assembly, RCA Stock No. 114837 or equiv., includes local-oscillator tuning coils L_{51} through L_{62}
S_2=Mixer section of channel-selector switch; stator assembly, RCA Stock No. 120084 or equiv., includes mixer tuning coils L_{31} through L_{42}
S_3=RF amplifier section of channel-selector switch; stator assembly, RCA Stock No. 120086 or equiv., includes rf amplifier plate tuning coils L_7 through L_{18}
S_4=UHF function switch assembly; part of channel-selector switch; stator assembly, RCA Stock No. 114807 or equiv.
S_5=Antenna section of channel-selector switch; stator assembly, RCA Stock No. 120087 or equiv., includes antenna tuning coils L_1, L_{45}, and L_{19} through L_{28}
T_1=Antenna matching transformer (includes coils L_4, L_5, and L_6 in high-pass filter network), RCA Stock No. 113968

See Note on page 617.

Circuit Description (Cont'd)

the tuner for a black-and-white television receiver (shown in circuit 26-25), and it operates equally well for either color or black-and-white transmissions.

The antenna used with the tuner is a balanced 300-ohm dipole type which is matched to the unbalanced tuner input circuit by the antenna matching transformer T_1. The ganged 5-section, 13-position channel selector, S_1 through S_5, establishes the operating frequency of the tuner for each of the vhf channels 2 through 13 or adapts the vhf tuner for operation with a uhf tuner. When used with a uhf tuner, the vhf tuner is operated as a two stage broadband rf amplifier and becomes essentially a pre-if amplifier for the color television receiver.

With the channel selector set to any of the channel positions 2 through 13, telecast signals, either color or black-and-white, from the selected channel are coupled from the antenna circuit through sections S_4 and S_5 of the channel selector to the control grid of the 2EG4 rf amplifier. For channel positions 2 through 13, the input lead (IF INPUT) from the uhf tuner is not connected to the vhf tuner.

The vhf input signals are amplified by the rf amplifier. The S_5 and S_3 sections of the channel selector connect the appropriate combinations of inductors into the grid and plate circuits of the rf amplifier to tune this stage to the desired frequency channel. An agc bias voltage, derived from the keyed agc amplifier

26-31 VHF TUNER (Cont'd)

Circuit Description (Cont'd)

in another section of the color receiver (circuit 26-33), is applied to the control grid of the 2EG4 to control the gain of the rf amplifier automatically.

The output of the rf amplifier is coupled through sections S_2 and S_3 of the channel selector to the control grid of the 4KE8 pentode section used in the mixer stage. Section S_3 of the ganged channel selector selects the proper combination of inductors to tune the mixer input circuit to the same operating frequency as that of the rf amplifier. A signal from the plate of the 4KE8 triode section used in the local-oscillator stage is also applied to the mixer. Section S_1 of the channel selector selects the required inductance so that the oscillator operates at a frequency 45.75 MHz above the video carrier frequency of the vhf channel selected by the tuner.

The signals from the rf amplifier and local oscillator are heterodyned in the mixer stage to produce the 45.75-MHz amplitude-modulated and 41.25-MHz frequency-modulated difference freqencies used as picture and sound intermediate frequencies, respectively. The composite color signal received at the antenna also includes a 3.58-MHz color subcarrier sideband. This subcarrier is also heterodyned with the local-oscillator frequency to produce a color-subcarrier intermediate frequency of 42.17 MHz. The picture, color-subcarrier, and sound if signals are coupled from the plate of the mixer through J_2 to the if stages of the receiver.

When the multiple-section channel selector is rotated to the UHF position, S_5 disconnects the vhf antenna circuit from the rf amplifier, and section S_4 completes a connection to the 280-volt B+ line through several voltage-dropping resistors to provide a dc voltage output of 18 volts for use as the B+ voltage for a uhf tuner. The video, sound and color-subcarrier if signals from a uhf tuner can then be applied through the IF INPUT jack J_1 and contacts of S_4 and S_6 to the control grid of the 2EG4 rf amplifier.

With the channel selector in the UHF position, switch section S_1 opens the B+ line to the local oscillator to disable this stage. In addition, sections S_2, S_3, and S_5 select the proper combination of components so that the rf amplifier and mixer stages operate as broadband 44-MHz amplifiers to provide two stages of amplification of the picture and sound if signals ahead of the receiver main if strip.

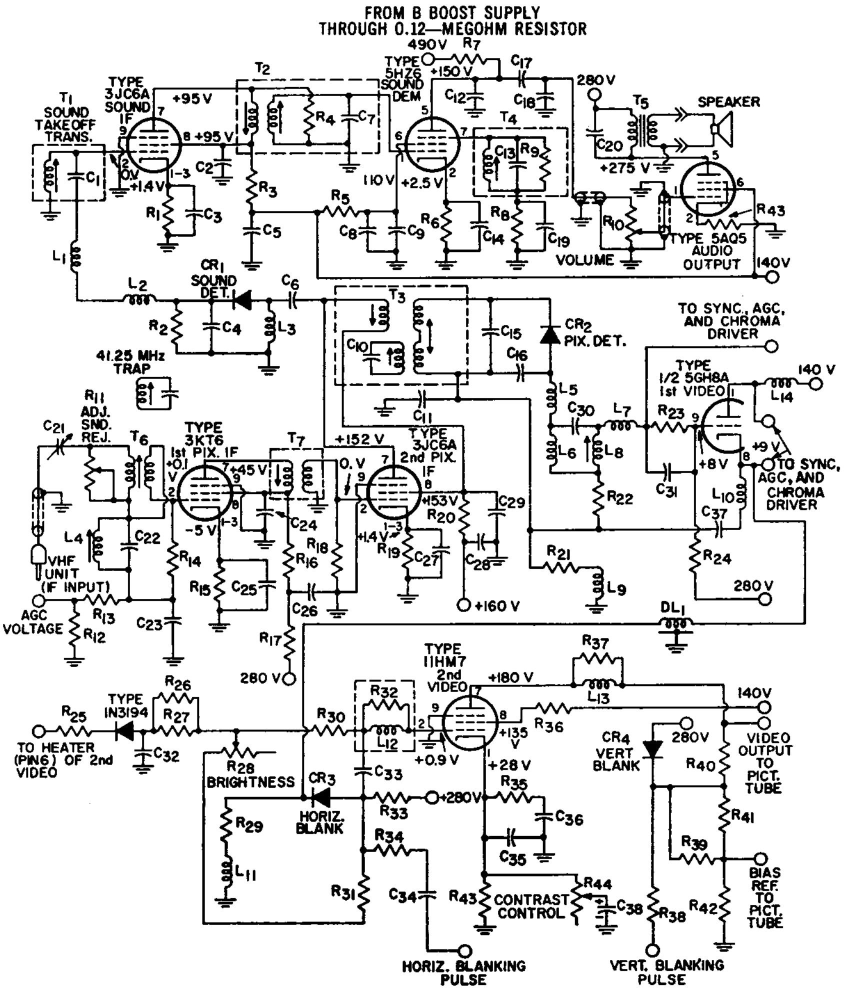
26-32 VIDEO-AND SOUND-CHANNEL CIRCUITS

For Color Television Receiver

These circuits form the video and sound channels for a color television receiver. The circuits operate from a dc supply voltage of 280 volts, obtained from the receiver low-voltage power supply. The tube heaters are included in the series-heater string for the over-all receiver. Operating power for the series-heater string is obtained directly from the 117-volt ac power line.

The picture if-amplifier circuit consists of two high-gain stages that use high-transconductance frame-grid tubes and double-tuned interstage coupling transformers. The composite if input from the vhf tuner which consists of amplitude-modulated 45.75-MHz picture signals 42.17-MHz color-subcarrier components, and frequency-modulated 41.25-MHz sound signals, are coupled by capacitor C_{21} and transformer T_6 to the control grid of the 3KT6 pentode used in the first picture if

26-32 VIDEO- AND SOUND-CHANNEL CIRCUITS (Cont'd)



Parts List

C_1=5 pF, part of T_1
C_2=1000 pF ±5%, ceramic, 500 V
C_3, C_5, C_9, C_{14}=0.01 μF, ceramic, 500 V
C_4=10 pF ±5%, ceramic, 500 V, NPO
C_6=1.5 pF, ceramic, 500 V, NPO
C_7=6 pF, part of T_2
C_8=47 pF, ceramic, 500 V, N750
C_{10}=150 pF, part of T_3
C_{11}=39 pF, ceramic, 500 V, N750
C_{12}=560 pF, ceramic, 500 V
C_{13}=10 pF, part of T_4
C_{15}=4 pF, ceramic, 500 V
C_{16}=10 pF, ceramic, 500 V, NPO
C_{17}=6800 pF, ceramic, 500 V
C_{18}=47 pF, ceramic, 500 V, N750
C_{19}=0.047 pF, ceramic, 500 V
C_{20}=0.0033 μF, paper, 1600 V
C_{21}=Trimmer, 3 to 15 pF, RCA Stock No. 116502 or equiv.
C_{22}=150 pF ±5%, mica, 500 V
C_{23}, C_{26}, C_{28}, C_{36}=1000 pF, ceramic, 500 V
C_{24}=330 pF, mica, 500 V
C_{25}=24 pF, ceramic, 500 V, NPO

26-32 VIDEO- AND SOUND-CHANNEL CIRCUITS (Cont'd)

Parts List (Cont'd)

C_{27}=4700 pF, ceramic, 500 V
C_{29}=430 pF ±5%, mica, 500 V
C_{30}=150 pF, mica, 500 V
C_{31}=0.047 μF, Mylar, 100 V
C_{32}=0.047 μF, ceramic, 100 V
C_{33}=0.1 μF, Mylar, 100 V
C_{34}=560 pF, ceramic, 500 V
C_{36}=680 pF, ceramic, 500 V
C_{37}=220 pF, ceramic, 500 V
C_{38}=50 μF, electrolytic, 50 V
CR_1, CR_2, CR_3=1N60 diode
CR_4=Vertical-blanking diode, RCA Stock No. 115867 or equiv.
DL_1=Delay line, RCA Stock No. 120786 or equiv.
L_1=RF choke, 3.9 μH, RCA Stock No. 116507 or equiv.
L_2, L_{10}=RF choke, 1.8 μH, RCA Stock No. 109248 or equiv.
L_3=RF choke, 12 μH, RCA Stock No. 120831
L_4=Inductor for 47.25-MHz trap, RCA Stock No. 121447 or equiv.
L_5=Video-detector filter coil, 5.6 μH, RCA Stock No. 109171 or equiv.
L_6, L_8=Part of 4.5-MHz trap, RCA Stock No. 121446 or equiv.
L_7=Video-detector filter coil, 36 μH, RCA Stock No. 16056 or equiv.
L_9=RF choke, 100 μH, RCA Stock No. 117380 or equiv.
L_{11}=Filter coil, 27 μH, RCA Stock No. 116511 or equiv.
L_{12}=Filter network (includes resistor R_{32}); RCA Stock No. 116499 or equiv.
L_{13}=Second-video plate coil, 330 μH, RCA Stock No. 118710 or equiv.
L_{14}=First-video plate coil, 1.8 μH, RCA Stock No. 78466 or equiv.
R_1, R_6, R_{35}, R_{43}=270 ohms, 0.5 watt
R_2, R_{25}=10000 ohms, 0.5 watt
R_3=8200 ohms, 0.5 watt
R_4=0.15 megohm, may be part of T_2
R_5=3300 ohms, 0.5 watt
R_7=0.68 megohm, 0.5 watt
R_8=0.47 megohm, 0.5 watt
R_9=68000 ohms, may be part of T_4
R_{10}=Potentiometer, volume control, 1 megohm, 0.5 watt
R_{11}=Potentiometer, sound-rejection adjustment, 7500 ohms, 0.5 watt
R_{12}=0.33 megohm, 0.5 watt
R_{13}, R_{30}=0.1 megohm, 0.5 watt
R_{14}=3900 ohms, ±5%, 0.5 watt
R_{15}=56 ohms, ±5%, 0.5 watt
R_{16}=1000 ohms, 0.5 watt
R_{17}=22000 ohms, 4 watts
R_{18}=6800 ohms, ±5%, 0.5 watt
R_{19}=150 ohms, ±5%, 0.5 watt
R_{20}=470 ohms, 0.5 watt
R_{21}=1200 ohms, 0.5 watt
R_{22}=4700 ohms, 0.5 watt
R_{23}=0.18 megohm, 0.5 watt
R_{24}=5.6 megohms, 0.5 watt
R_{26}=22 megohms, 0.5 watt
R_{27}=2.7 megohms, 0.5 watt
R_{28}=Potentiometer, brightness control, 0.25 megohm, RCA Stock No. 120775 or equiv.
R_{29}=680 ohms ±5%, 0.5 watt
R_{31}=0.22 megohm, 0.5 watt
R_{32}=2200 ohms, part of assembly with L_{12}
R_{33}=0.39 megohm, 0.5 watt
R_{34}=0.12 megohm, 0.5 watt
R_{36}=100 ohms, 0.5 watt
R_{37}=5600 ohms, 0.5 watt
R_{38}=560 ohms, 0.5 watt
R_{39}=22000 ohms, 3 watts
R_{40}=6800 ohms, 4 watts
R_{41}=10000 ohms, 3 watts
R_{42}=33000 ohms, 4 watts
T_1=Sound-takeoff transformer (includes C_1), RCA Stock No. 120824 or equiv.
T_2=4.5-MHz sound if transformer (includes C_7 and may include R_4), RCA Stock No. 120828 or equiv.
T_3=Pix if output transformer and 41.25-MHz trap, RCA Stock No. 120827 or equiv.
T_4=Sound-demodulator quadrature network (includes C_{13} and may include R_9), RCA Stock No. 120825 or equiv.
T_5=Audio output transformer, matches 5000-ohm tube-plate impedance to 3.2-ohm speaker voice coil, RCA Stock No. 120822 or equiv.
T_6=IF input transformer and 41.25-MHz trap, RCA Stock No. 116560 or equiv.
T_7=Pix if transformer, RCA Stock No. 120826 or equiv.
See Note on page 617.

Circuit Description (Cont'd)

amplifier. The 3KT6 tube has good remote-cutoff characteristics, and the automatic-gain-control (agc) bias voltage from the receiver agc amplifier (shown in circuit 26-33) is also applied to the control-grid circuit of this tube. The output of the first picture if amplifier is coupled by transformer T_7 to the control grid of the 3JC6A pentode used in the second picture if amplifier. Capacitor C_6 couples the output of the second picture if amplifier to the sound detector, and transformer T_3 couples the output to the video (pix) detector. Transformers T_6, T_7, and T_3 are stagger-tuned to obtain the wide band pass required for the if amplifiers to pass both the 45.75-MHz video AM signals and the 41.25-MHz sound FM signals, as well as the intermediate 42.17 color subcarrier.

The sound detector (CR_1 and associated components) is essentially a second mixer circuit. The 45.75-MHz picture if signal and the 41.25 sound if signal are heterodyned to produce a second sound if carrier of 4.5 MHz. This 4.5-MHz sound if carrier is still frequency-modulated by the audio components contained in the original rf signal input at the receiver antenna. The sound-takeoff transformer T_1 forms a selective load impedance for the 4.5-MHz if signal derived in the sound detector circuit.

The 4.5 MHz signal developed across sound-takeoff transformer T_1 is applied to the control grid of the 3JC6A sound if amplifier. The ampli-

26-32 VIDEO- AND SOUND-CHANNEL CIRCUITS (Cont'd)

Circuit Description (Cont'd)

fied 4.5 MHz FM if signal from this stage is then coupled by the double-tuned transformer T_2 to the control grid of the 5HZ6 sound demodulator. This stage demodulates the 4.5-MHz sound if signal and amplifies the resultant audio signal voltage. The +490 volts used as the plate supply for the 5HZ6 demodulator tube is derived from the 700-volt B Boost supply in the horizontal-output stage (shown in circuit 26-34) of the receiver.

The tuned secondary circuit of transformer T_3 selects the 45.75-MHz amplitude-modulated picture and 42.17-MHz color sideband signals from the composite if signal and applies this picture signal to the video detector (CR_2 and associated components). The detected video signal developed across the dector-circuit filter network (L_5, L_6, L_7, L_8, and C_{30}) is then coupled through C_{31} and R_{23} to the control grid of the 5GH8A triode section used in the first video amplifier (the pentode section of the 5GH8A tube is used in the sync-agc-and-chroma driver, shown in circuit 26-33). The first video amplifier supplies the input signals to the sync-agc-and-chroma driver and to the second video amplifier.

The second video stage performs many functions. The input circuit of the 11HM7 pentode used in this stage is the insertion point for horizontal blanking pulses (for eventual application to the cathodes of the color picture tube). The horizontal blanking diode CR_3 is placed in the conducting mode by a small positive voltage applied to its anode through the dropping resistor R_{33} from the 280-volt B+ source. During active video scanning time, diode CR_3 is forward-biased (conducting), and the video signal is coupled by capacitor C_{33}, to the control grid of the video amplifier. During horizontal blanking time, a negative pulse from the horizontal-output transformer (T_1 in circuit 26-34) is applied through C_{34} and R_{34} to the diode. This negative pulse is sufficient to cut off the diode during horizontal retrace time. The pulse is applied to the control grid of the second video amplifier and drives the grid more negative (than would the normal horizontal sync pulse). The negative signal at the grid is inverted at the plate; the added positive level coupled to the cathodes of the color picture tube is sufficient to provide blanking of horizontal retrace lines.

The brightness control for the color receiver is also located in the control-grid circuit of the second video amplifier. Negative dc grid bias for the 11HM7 second video tube is derived from the ac voltage obtained from the heater, pin 6, of the second video tube. The 11HM7 heater is in the approximate center of the series heater string (refer to circuit 26-30); at this point, approximately 60 volts of ac voltage is available. The negative dc voltage (about −75 volts) is developed across C_{32} by the IN3194 rectifier circuit. Adjustment of the brightness control, R_{28} alters the grid bias by "tapping" the positive voltage applied to the top of the control. This unique circuit arrangement provides automatic brightness compensation with changes in power-line voltage. If line voltage increases, the negative voltage across C_{32} increases; the increased bias that is then applied to the 11HM7 decreases the conduction of this tube. The opposite action occurs with a decrease in line voltage.

The cathode of the second video amplifier is returned to the contrast control R_{44}. Brightness stability is obtained by use of a fixed 150-ohm, 5-per cent resistor, R_{43}, for dc cathode bias. Adjustment of the contrast control does not change the dc characteristics of the cathode; only the ac signal gain of the stage is altered when the control is adjusted.

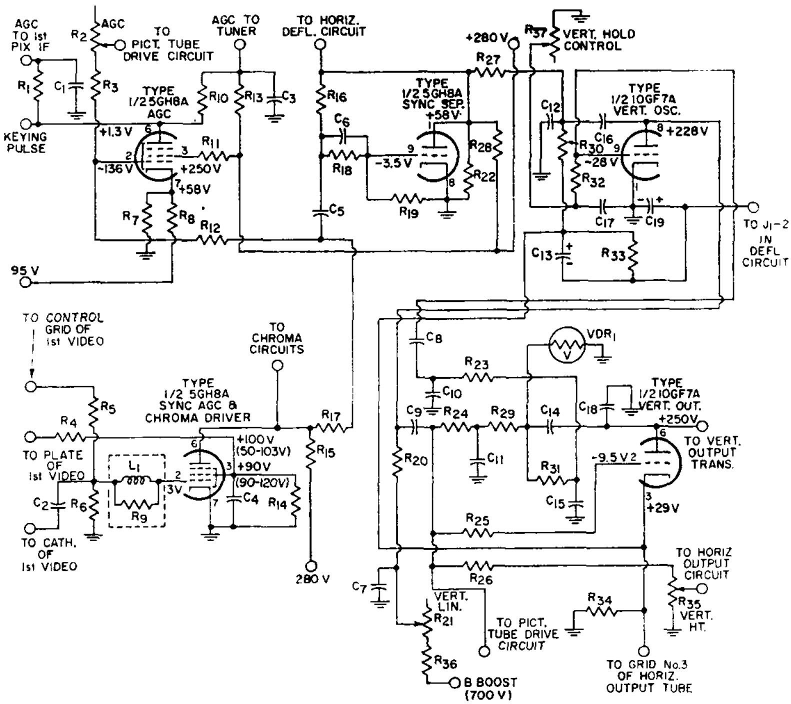
26-32 VIDEO- AND SOUND-CHANNEL CIRCUITS (Cont'd)

Circuit Description (Cont'd)

Vertical-retrace blanking is accomplished in the plate circuit of the second video amplifier. During active scan periods, the vertical-blanking diode CR_4 is forward-biased (conducts); during vertical retrace periods, however, a positive (blanking) pulse from the vertical-output transformer (T_2 in circuit 26-34) is applied through R_{38} to the cathode of the diode. This 60-volt positive pulse is large enough to bias the diode into cutoff. During the blanking interval, the positive voltage pulse is added to the plate voltage of the 11HM7 second-video tube and applied to the cathode circuits of the color picture tube. As a result of the increased positive potential at the cathode, the picture tube is cut off during vertical retrace periods.

26-33 SYNC, AGC, AND VERTICAL-DEFLECTION CIRCUITS

For Color Television Receiver



Parts List

C_1=0.18 μF, Mylar, 200 V
C_2=24 pF, ceramic, 500 V, NPO
C_3, C_{17}=0.01 μF, ceramic, 500 V
C_4=1000 pF, ceramic, 500 V
C_5=3300 pF, ceramic, 500 V
C_6=470 pF, ceramic, 500 V
C_7=0.1 μF, paper, 600 V
C_8=0.0056 μF, Mylar, 400 V
C_9=0.01 μF, Mylar, 600 V
C_{10}, C_{15}=680 pF, ceramic, 500 V
C_{11}=0.047 μF, Mylar, 100 V
C_{12}=1500 pF, ceramic, 500 V
C_{13}=50 μF, electrolytic, 75 V
C_{14}=0.0082 μF, paper, 1000 V
C_{16}=0.033 μF, Mylar, 600 V
C_{18}=0.001 μF, ceramic, 3000 V
C_{19}=25 μF, electrolytic, 25 V
L_1=RF choke, 120 μH, part of assembly with R_9, RCA Stock No. 120795 or equiv.

26-33 SYNC, AGC, AND VERTICAL-DEFLECTION CIRCUITS (Cont'd)

Parts List (Cont'd)

R_1, R_{18}=0.15 megohm, 0.5 watt
R_2=Potentiometer, agc adjustment, 50000 ohms, 0.5 watt, RCA Stock No. 120804 or equiv.
R_3=27000 ohms, 0.5 watt
R_4=3300 ohms, 0.5 watt
R_5, R_{17}, R_{28}=10000 ohms, 0.5 watt
R_6=6800 ohms, 0.5 watt
R_7=27000 ohms, 1 watt
R_8=1500 ohms, 0.5 watt
R_9=68000 ohms, 0.5 watt, part of assembly with L_1
R_{10}=0.56 ohms, 0.5 watt
R_{11}=1800 ohms, 0.5 watt
R_{12}, R_{27}, R_{31}=0.12 megohm, 0.5 watt
R_{13}=10 megohms, 0.5 watt
R_{14}=22000 ohms, 0.5 watt
R_{15}=10000 ohms, 3 watts
R_{16}=22 megohms, 0.5 watt
R_{19}=3.3 megohms, 0.5 watt
R_{20}=1.5 megohms, 0.5 watt
R_{21}=Potentiometer, vertical-linearity control, 3.4 megohms, 0.5 watt, RCA Stock No. 120807 or equiv.
R_{22}=56000 ohms, 0.5 watt
R_{23}=47000 ohms, 0.5 watt
R_{24}=4.7 megohms, 0.5 watt
R_{25}=1000 ohms, 0.5 watt
R_{26}=1.5 megohms, 0.5 watt
R_{29}=0.47 megohm, 0.5 watt
R_{30}=33000 ohms $\pm5\%$, 0.5 watt
R_{32}=0.22 megohm, 0.5 watt
R_{33}=3300 ohms, 1 watt
R_{34}=1500 ohms, wirewound, 3 watts
R_{35}=Potentiometer, vertical-height control, 1 megohm, 0.5 watt, RCA Stock No. 120805 or equiv.
R_{36}=0.1 megohm, 1 watt
R_{37}=Potentiometer, vertical-hold control, 0.75 megohm, 0.5 watt
VDR_1=Voltage-dependent resistor (varistor); 870 volts at 1 mA; RCA Stock No. 112876 or equiv.
See Note on page 617.

Circuit Description

This circuit shows the sync-agc-and-chroma driver, agc amplifier, sync separator, and vertical deflection circuit for a color television receiver. The sync-agc-and-chroma driver, the sync separator, and the vertical output tube operate from a plate supply (B+) voltage of 280-volts obtained from the receiver low-voltage power supply. The plate supply voltage for the agc amplifier is a positive keying pulse from the horizontal-output transformer, and the plate voltage for the vertical oscillator is obtained from the 700-volt B Boost supply in the horizontal output circuit. The tube heaters are connected into the series-heater string for the over-all color receiver; operating power for the heater string is obtained directly from the ac power line.

The drive signal for the sync and agc circuits is obtained from the cathode of the first video amplifier (shown in circuit 26-32). This signal is coupled by capacitor C_2 and the parallel LR network L_2 and R_9 to the control grid of the 5GH8A pentode section used in the sync-agc-and-chroma driver. (The triode section of the 5GH8A tube is used in the first video amplifier). The screen-grid and control-grid bias voltages for the driver pentode are also obtained from the first video amplifier. The output of the driver stage is applied to the control grids of the agc amplifier and the sync separator and to the chroma circuits (shown in circuit 26-35).

The agc amplifier uses the pentode section of a 5GH8A triode-pentode; the triode section of this tube is used in the sync separator. The operation of the agc amplifier is gated by a positive keying pulse from the horizontal-output transformer (shown in circuit 26-34). This pulse, which is synchronized with the video signal, overcomes the bias provided by the 95 volts (obtained from the receiver low-voltage power supply, circuit 26-30) applied to the cathode circuit of the agc amplifier. Portions of the video signal that occur coincident with the keying pulse (i.e. during the horizontal blanking interval) are amplified by the agc stage. Resistor R_1 and capacitor C_1, together with other filtering elements in the control-grid circuit of the first picture if amplifier, filter out the pulsating components in the video signal to obtain a negative dc voltage proportional the video signal and thus to the rf input at the receiver antenna. Similarly, an agc bias voltage for the vhf tuner is developed across the filter capacitor C_3.

Synchronizing pulses are included in the composite rf signals transmitted by a television broadcast station to provide timing information required for synchronization of the transmitter and receiver scan-

26-33 SYNC, AGC, AND VERTICAL-DEFLECTION CIRCUITS (Cont'd)

Circuit Description (Cont'd)

ning systems. The sync separator separates and amplifies the synchronizing pulses contained in the composite video signal it receives from the sync-agc-and-chroma driver. The 5GH8A triode section used in this stage is operated basically as a class C limiter. When the video signal is applied, the stage is biased beyond cutoff by the negative voltage developed by the grid-leak bias network formed by C_6 and R_{18}. Only the sync pulses in the composite video signal have sufficient amplitude to drive the sync amplifier into conduction. The resultant negative pulses developed in the plate circuit of the 5GH8A triode section are applied as the synchronizing inputs to the vertical and horizontal deflection circuits.

The vertical-deflection circuit employs one section of a 10GF7A dual triode in a vertical oscillator stage and a vertical output stage. These two stages form a basic plate-coupled 60-Hz free-running multivibrator that is synchronized by negative vertical sync pulses from the sync separator stage. The negative-pulse output from the zync separator, however, includes horizontal sync pulses and equalizing pulses in addition to the vertical sync pulses. The vertical sync pulses must be separated from the composite sync-separator output prior to the application of the synchronizing input to the vertical-deflection circuits. This separation is accomplished by integration of the composite sync-separator output across capacitor C_{12}. The integrating network (R_{27} and C_{12}) has negligible response for the narrow horizontal-sync and equalizing pulses, but responds to the greater energy contained in the much wider vertical-sync pulses to develop a triangular voltage waveform, coupled by C_{16}, C_9, and R_{28} to the control grid of the vertical-output triode section, that synchronizes the operation of the multivibrator. The combination of the triangular wave input to the grid of the output section and the square-wave multivibrator signal results in a trapezoidal voltage waveform at the plate of the output section. This trapezoidal voltage wave produces a triangular wave of current through the vertical-output transformer (T_2 in circuit 26-34) and through the vertical deflection coils of the picture tube (shown in circuit 26-36). The rising portion of the triangular current waveform produces the vertical scanning, and the decreasing portion of the waveform provides the retrace.

26-34 HORIZONTAL-DEFLECTION CIRCUIT AND HIGH VOLTAGE POWER SUPPLY

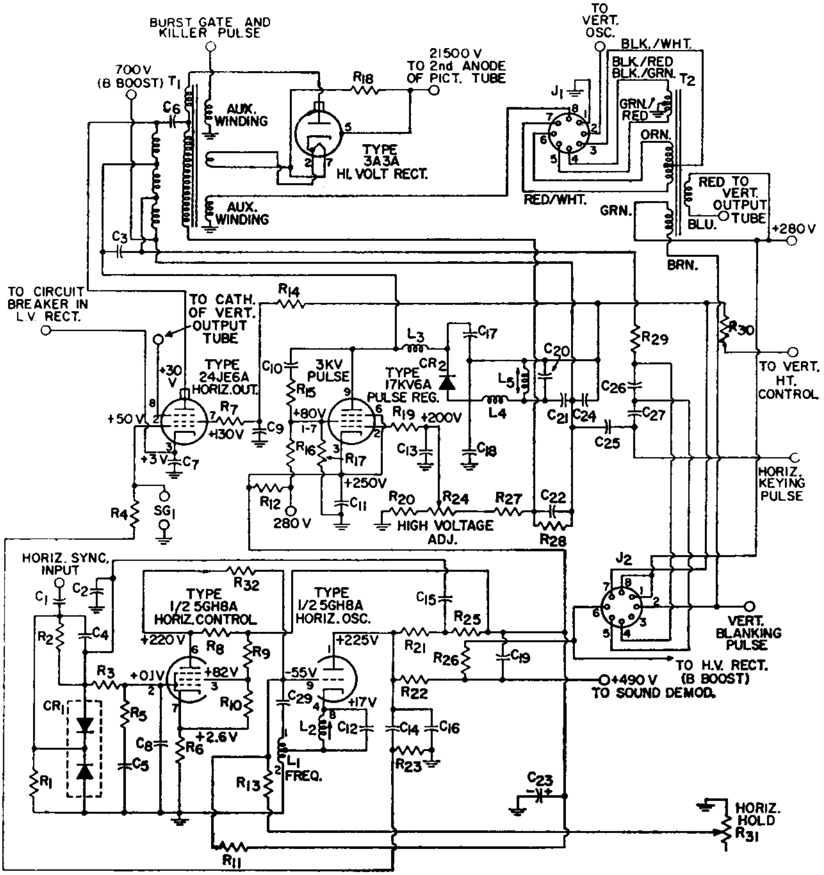
For Color Television Receiver

Circuit Description

These circuits develop the horizontal scanning signals and the dc operating voltage (21,500 volts) for the color picture tube (RCA Type 15LP22) and the receiver B Boost voltage (700 volts). The circuits operate from the receiver low-voltage (280-volt) supply. The heaters of the 5GH8A, 24JE6A, and 17KV6A tubes used in these circuits are included in the series-heater string for the over-all receiver; operating power for these heaters is obtained directly from the 117-volt ac power line. Heater power for the 3A3A high-voltage rectifier tube is obtained from a 3-volt secondary winding on the high-voltage transformer.

A blocking oscillator in which the transformer coil is located in the cathode circuit is used to obtain a large-amplitude horizontal-drive

26-34 HORIZONTAL-DEFLECTION CIRCUIT AND HIGH-VOLTAGE POWER SUPPLY (Cont'd)



Parts List

C_1=82 pF ±1 pF, ceramic, 500 V, NPO
C_2=1200 pF, ceramic, 500 V
C_3=0.0018 μF, paper, 1000 V
C_4=150 pF, ceramic, 500 V, NPO
C_5=0.15 μF, Mylar, 75 V
C_6=0.01 μF, Mylar, 600 V
C_7=0.01 μF, Mylar, 75 V
C_8, C_{15}=1200 pF, ceramic, 500 V
C_9, C_{19}=0.1 μF, Mylar, 400 V
C_{10}=15 pF, ceramic, 5000 V, N750
C_{11}, C_{13}=1000 pF, ceramic, 500 V
C_{12}, C_{14}=0.01 μF, Mylar, 400 V
C_{16}=270 pF ±5%, mica, 500 V
C_{17}=100 pF, ceramic, 5000 V, N1500
C_{18}=22 pF, ceramic, 1000 V, N750
C_{20}=0.1, Mylar, 200 V
C_{21}=0.033 μF, Mylar, 600 V
C_{22}=0.01 μF, Mylar, 600 V
C_{23}=40 μF, electrolytic, 350 V
C_{24}=0.047 μF, Mylar, 600 V
C_{25}=150 pF, ceramic, 2000 V, N1500
C_{26}=270 pF, ceramic, 2500 V, N1500
C_{27}=150 pF, ceramic, 2000 V, N1500
CR_1=AFC diodes, RCA Stock No. 109474 or equiv.
CR_2=Damper diode, RCA Stock No. 120818 or equiv.
J_1=Octal socket, convergence-circuit input jack, RCA Stock No. 77645 or equiv. (mates with P_1 on circuit 26-36)
J_2=Octal socket, deflection-yoke input jack, RCA Stock No. 102787 or equiv. (mates with P_2 on circuit 26-36)
L_1, L_2=Horizontal-oscillator dual-coil assembly, RCA Stock No. 109947 or equiv.
L_3, L_4=RF choke, 4.7 μH, RCA Stock No. 120839 or equiv.
L_5=Variable inductor, horizontal efficiency adjustment, RCA Stock No. 120794 or equiv.
R_1, R_{22}=0.22 megohm, 0.5 watt
R_2, R_{20}=0.39 megohm, 0.5 watt

26-34 HORIZONTAL-DEFLECTION CIRCUIT AND HIGH-VOLTAGE POWER SUPPLY (Cont'd)

Parts List (Cont'd)

R_3=0.27 megohm, 0.5 watt
R_4=100 ohms, 0.5 watt
R_5=15000 ohms, 0.5 watt
R_6=1200 ohms, 0.5 watt
R_7=47 ohms, 0.5 watt
R_8, R_{26}=0.12 megohm, 0.5 watt
R_9=0.15 megohm, 0.5 watt
R_{10}=82000 ohms, 0.5 watt
R_{11}=8.2 megohms, 0.5 watt
R_{12}=680 ohms, 2 watts
R_{13}=82000 ohms ±2%, 0.5 watt
R_{14}=82000 ohms ±5%, 4 watts
R_{15}=100 ohms, 0.5 watt
R_{16}=68000 ohms, 1 watt
R_{17}=33000 ohms, 0.5 watt
R_{18}=1000 ohms, 2 watts
R_{19}=10000 ohms, 0.5 watt
R_{21}=27000 ohms, 0.5 watt
R_{23}=10 megohms, 0.5 watt
R_{24}=Potentiometer, high-voltage adjustment, 0.5 megohm, 0.5 watt
R_{25}=33000 ohms, 0.5 watt
R_{27}=0.56 megohm, 0.5 watt
R_{28}=0.27 megohm, 1 watt
R_{29}=120 ohms, 0.5 watt
R_{30}=2.2 megohms, 0.5 watt
R_{31}=Potentiometer, horizontal-hold control, 50000 ohms, 0.5 watt
SG_1=Spark-gap capacitor, 0.5 pF, 1000 V, RCA Stock No. 120819 or equiv.
T_1=Horizontal-output (flyback) transformer, RCA Stock No. 120820 or equiv.
T_2=Vertical-output transformer, RCA Stock No. 120821 or equiv.

See Note on page 617.

Circuit Description (Cont'd)

waveform. A control stage establishes the bias for the oscillator and, in this way, controls the firing of the oscillator stage. The 5GH8A triode-pentode is used in these stages. The triode section is used as the oscilltor tube; the pentode section is used as a high-gain, low-drift control tube.

When the composite video signal is generated at the television broadcast station, the leading edge of each horizontal sync pulse, of alternate equalizing pulses, and of alternate serrations of the vertical sync pulses are correctly timed to initiate the horizontal retrace period. These leading-edge components are extracted from the composite output from the sync separator (shown in circuit 26-32) and are used to synchronize the operation of the horizontal oscillator.

The sync waveform is differentiated by the RC network (C_1 and R_2) at the input to the horizontal deflection circuit to obtain negative and positive voltage spikes that correspond to the leading and lagging edges, respectively, of the rectangular sync puluses. The amplitude of these voltage spikes is dependent upon only the peak value of the sync pulses and is not affected by the time durations of these pulses. The differentiating circuit, therefore, does not respond to the flat portions of the vertical sync pulses; as a result, with the exception of the serrations, the vertical sync pulses do not affect the operation of the horizontal-deflection circuits. The leading edge of alternate serrations, however, correspond to the start of horizontal-retrace periods and thus may be considered as merely another horizontal sync signal.

The differentiated sync waveform is applied to the junction of the twin silicon diode CR_1 used in a phase-discriminator type of afc network. The positive voltage spikes in the differentiated waveform have no effect on the discriminator network. The negative-voltage spikes are compared with pulses fedback from the horizontal oscillator to derive the synchronizing voltage. The frequency of the horizontal oscillator and the repetition rate of the horizontal sync pulses should both be 15,750 Hz, the desired horizontal scanning rate for the picture tube. If the pulses from the oscillator are not coincident with the horizonate sync pulses, the phase discriminator develops an error voltage at the control grid of the control tube. The control tube then varies the bias and, thus, the firing point of the oscillator until it is locked in phase with the horizontal sync pulses.

The parallel LC network (L_2 and C_{12}) in the cathode circuit of the oscillator resonates at 15,750 Hz to provide frequency stabilization for the oscillator. The HOLD control R_{31} adusts the frequency of the oscil-

26-34 HORIZONTAL-DEFLECTION CIRCUIT AND HIGH-VOLTAGE POWER SUPPLY (Cont'd)

Circuit Description (Cont'd)

lator to achieve an exact lock-in with the horizontal sync pulses. The output of the blocking oscillator is coupled through C_{14} and R_4 to the control grid of the 24JE6A power pentode used in the horizontal-output stage. This tube drives the high-voltage flyback transformer T_1 that develops the scanning voltage for the horizontal deflection coils (shown in circuit 26-36).

The sudden cutoff of plate current in the horizontal output stage at the end of the trace period causes a very large, positive-going voltage pulse to be generated across the high-voltage transformer T_1. The 3A3A half-wave rectifier circuit converts this pulse to a positive dc of 21,500 volts which is applied to the second anode of the color picture tube.

Regulation of the high voltage is achieved by use of a 17KV6A pulse-regulator stage connected in shunt with a section of the primary of the high-voltage flyback transformer. The regulator stage acts as a variable load on the flyback pulse source and, in this way, maintains an essentially constant pulse amplitude in the primary winding of the high-voltage transformer with changing loads on the high-voltage supply. This action assures that a constant-amplitude, stepped-up pulse is applied to the 3A3A rectifier. The rectifier output delivered to the picture tube, therefore, is maintained at a constant value of 21,500 volts.

Removal of negative overshoots that would be developed across the high-voltage transformer because of a flywheel effect is accomplished by the damper diode CR_2. This diode is shaped like a fuse and snaps into clips that can be mounted on the same circuit board with the horizontal deflection circuits and is readily replaced during servicing. The polarity of the damper diode is such that the positive pulse developed across the high-voltage transformer causes no current flow through it. For negative pulses, however, the damper diode provides a low impedance path for the current, and energy stored in the horizontal output transformer (and the horizontal deflection coils) is dissipated in the damper circuit. The rectified current through the damper diode develops the boosted B+ voltage of +700 volts across capacitor C_{21} in the damper anode circuit.

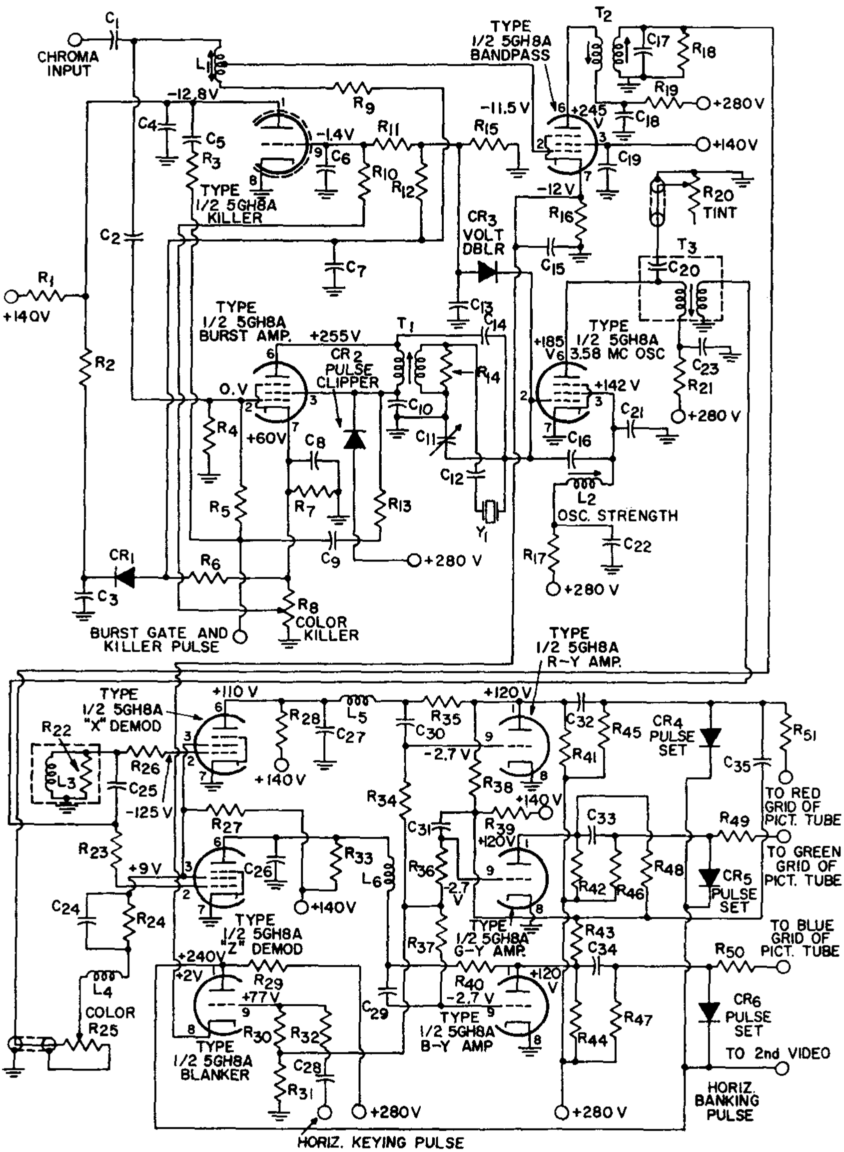
The two auxiliary windings on the high-voltage transformer supply supplementary pulse voltages. The upper winding supplies gating pulses to the burst-gate and the color-killer amplifiers (shown in circuit 26-35). The convergence pulse is developed across the lower auxiliary winding. Keying pulses for the agc amplifier and the horizontal blanking diode are derived from the capacitor network (junction of C_{26} and C_{27}) in the primary circuit of the high-voltage transformer.

Transformer T_2 shown in the circuit diagram is the vertical output transformer. The drive signal from the vertical output stage (shown in circuit 26-33) is developed across the primary of this transformer and coupled by the secondary winding through jack J_2 to the vertical deflection coils (shown in circuit 26-36). An auxiliary winding on transformer T_2 develops the keying pulse for the vertical blanking diode. The horizontal scanning signal from the high-voltage (horizontal-output) transformer are also coupled through jack J_1 to the horizontal deflection coils. The horizontal and vertical signals to the convergence board are routed through jack J_2. (Jacks J_1 and J_2 mate with plugs P_2 and P_1, respectively, on circuit 26-36.)

26-35

CHROMA CIRCUITS

For Color Television Receiver



See Note on page 617.

26-35 CHROMA CIRCUITS (Cont'd)

Parts List

C_1=27 pF, ceramic, 500 V, NPO
C_2=68 pF, ceramic, 500 V, N750
C_3, C_5, C_6, C_8, C_9, C_{22}, C_{23}, C_{29} through C_{34}=0.01 pF, ceramic, 500 V
C_4=390 pF, ceramic, 500 V
C_7=0.047 μF, Mylar, 100 V
C_{10}, C_{18}=1000 pF, ceramic, 500 V
C_{11}=Trimmer, 2 to 10 pF, RCA Stock No. 116501 or equiv.
C_{12}=220 pF, ceramic, 500 V
C_{13}=10 pF, ceramic, 500 V, N150
C_{14}, C_{16}=0.82 pF $\pm$5%, headed lead, 500 V
C_{15}=820 pF, ceramic, 500 V
C_{17}=390 pF $\pm$5%, Mylar, 500 V
C_{19}, C_{26}, C_{27}=33 pF, ceramic, 500 V, N150
C_{21}=10 pF $\pm$5%, ceramic, 500 V, NPO
C_{24}=0.027 pF, Mylar, 100 V
C_{25}=430 pF $\pm$5%, mica, 500 V
C_{28}=150 pF, ceramic, 500 V
C_{35}=1.2 pF, ceramic, 500 V
CR_1, CR_4, CR_5, CR_6=Silicon diode, RCA Stock No. 119596 or equiv.
CR_2=Diode, pulse clipper, RCA Stock No. 113998
CR_3=Diode, type 1N60
L_1=Variable inductor, chroma-takeoff coil, RCA Stock No. 120797 or equiv.
L_2=Variable inductor, oscillator strength adjustment, RCA Stock No. 120798 or equiv.
L_3=Phase-shift coil, 3.9 μH, part of quadrature assembly (RCA Stock No. 120830 or equiv.) with R_{22}
L_4=RF coil, 3.9 μH, RCA Stock No. 116510 or equiv.
L_5, L_6=RF choke, 620 μH, RCA Stock No. 109257 or equiv.
R_1=3.9 megohms, 0.5 watt
R_2=0.15 megohm, 0.5 watt
R_3, R_4, R_7=47000 ohms, 0.5 watt
R_5=82000 ohms, 0.5 watt
R_6, R_{10}=10 megohms, 0.5 watt
R_8=Potentiometer, color-killer adjustment, 1 megohm, 0.5 watt, RCA Stock No. 120805 or equiv.
R_9=82 ohms, 0.5 watt
R_{11}=2.7 megohms, 0.5 watt
R_{12}=2.2 megohms, 0.5 watt
R_{13}=3900 ohms, 0.5 watt
R_{14}, R_{16}=390 ohms, 0.5 watt
R_{15}=82000 ohms, 0.5 watt
R_{17}=47000 ohms, 1 watt
R_{18}=560 ohms, 0.5 watt
R_{19}=1500 ohms, 0.5 watt
R_{20}=Potentiometer, tint control, 10000 ohms, 0.5 watt, RCA Stock No. 120774 or equiv.
R_{21}=6800 ohms, 1 watt
R_{22}=120 ohms $\pm$5%, 1 watt, part of quadrature assembly with L_3
R_{23}, R_{26}=470 ohms, 0.5 watt
R_{24}=1500 ohms, 0.5 watt
R_{25}=Potentiometer, color control, 500 ohms, 0.5 watt, RCA Stock No. 120776 or equiv.
R_{27}=0.1 megohm, 0.5 watt
R_{28}, R_{39}=6800 ohms $\pm$5%, fixed film, 0.5 watt
R_{29}=4700 ohms $\pm$5%, 1 watt
R_{30}=0.22 megohm, 0.5 watt
R_{31}=8200 ohms, 0.5 watt
R_{32}=68000 ohms, 0.5 watt
R_{33}=8200 ohms $\pm$5%, fixed film, 0.5 watt
R_{34}, R_{36}, R_{37}=1 megohm, 0.5 watt
R_{35}, R_{40}=0.18 megohm, 0.5 watt
R_{38}=0.33 megohm, 0.5 watt
R_{41}, R_{42}, R_{44}=39000 ohms $\pm$5%, 1 watt
R_{43}=0.56 megohm, 0.5 watt
R_{45}, R_{46}, R_{47}=2.2 megohms, 0.5 watt
R_{48}=0.39 megohm, 0.5 watt
R_{49}, R_{50}, R_{51}=1000 ohms, 0.5 watt
T_1=Burst transformer, RCA Stock No. 120816 or equiv.
T_2=3.58-MHz oscillator transformer, RCA Stock No. 120815 or equiv.
Y_1=3.58-MHz oscillator crystal

Circuit Description

These circuits extract the color information from the 3.58-MHz chrominance sidebands included in the composite color video signal. The color information is included in the chrominance sidebands in the form of two difference-frequency components that have a phase difference of 90 degrees and that are derived in the color television transmitter by subtraction of the luminance (Y) signal from the red (R) and blue (B) color signals. [The green color-difference (G — Y) components are not transmitted, but instead, are derived in the color receiver by addition of complements (negative values) of the R — Y and B — Y signals.] To accomplish the demodulation function, the chroma circuits are required to develop two continuous-wave 3.58-MHz signals that have a phase difference of 90 degrees, each of which much be added vectorially to the chrominance sidebands. In other words, the 3.58MHz color subcarrier suppressed during transmission must be reinserted by the chroma circuits before the R — Y and B — Y color-difference information contained in the chrominance sidebands can be detected.

The chroma circuits operate from the color receiver low-voltage (280-volt) power supply. Five 5GH8A triode-pentodes fulfill the electron-tube requirements for the ten chroma stages. The heaters of these tubes are connected in series with those of other tubes in the receiver; operating power for the series-heater string is obtained directly from the 117-volt ac power line.

The input to the chroma circuits is the composite video signal after it has been amplified by the first video amplifier and the sync-agc-

26-35 CHROMA CIRCUITS (Cont'd)

Circuit Description (Cont'd)

and-chroma driver (shown on circuits 26-33 and 26-34, respectively). In addition to the chrominance sidebands, this composite signal includes the luminance signal (equivalent to the monochrome picture signal in black-and-white transmissions), the conventional horizontal and vertical sync pulses, and the color burst synchronizing signal. The color "burst" is a 3.58 MHz reference signal of approximately 8 cycles that occurs during the horizontal retrace blanking interval immediately following the horizontal sync pulse (refer to Fig. 96, page 73).

The chroma input is applied simultaneously to the chroma bandpass and burst amplifiers. When no burst signal is included in the chroma input (i.e., for black-and-white transmissions), the color-killer stage develops, by means of the current through diode CR_1, a negative dc voltage across capacitor C_7 that biases the chroma bandpass amplifier beyond cutoff; as a result the chroma input is not applied to the color demodulators.

The operation of the burst amplifier is controlled by a gating signal (burst-gate and killer pulse) from an auxiliary winding on the horizontal-output transformer (T_1 in circuit 26-34). This gating pulse is generated at the same time and has the same time duration as the horizontal blanking pulse used to blank out the horizontal retrace on the color picture tube. This interval corresponds to the period of the horizontal sync pulse and the 3.58MHz burst synchronizing signal that immediately follows the sync pulse. The burst amplifier, therefore, only amplifies this portion of the chroma input. The primary of transformer T_1 in the plate circuit of the burst amplifier, however, is tuned to 3.58 MHz so that only the 3.58-MHz burst signal is coupled from the plate of the burst amplifier.

The separated burst is coupled by transformer T_1 to the control-grid circuit of a 3.58-MHz injection-locked oscillator circuit. The oscillator, therefore, is forced to operate in step (with respect to both frequency and phase) with the incoming burst signal. The 3.58-MHz crystal Y_1 is used to assure excellent frequency stability in the oscillator circuit. The oscillator develops the continuous-wave 3.58-MHz reference signal applied to the control grids of the Z and X demodulators. The quadrature network (L_3 and R_{23}) causes a 90-degree phase shift in the 3.58-MHz signal applied to the control grid of the X demodulator. The 3.58-MHz chrominance sidebands must also be applied to the X and Z demodulators before these stages can derive the color difference signals. These sideband signals are obtained from the chroma bandpass amplifier.

The dc bias voltage developed in the grid circuit of the oscillator stage is used to control color-killer action and to derive an agc voltage for the chroma bandpass amplifier. The cathode-to-grid section of the oscillator triode, diode CR_3, and associated components from a two-diode voltage-doubler circuit. Any dc voltage developed in the oscillator grid circuit is approximately doubled at the voltage-doubler output (anode circuit of diode CR_3). When no color signal is received (i.e., no burst signal applied to the oscillator), the dc voltage at the grid of the oscillator is approximately -5 volts. The -10 volts developed across C_{13} and R_{15} in the anode circuit of voltage-doubler diode CR_3 is reduced to approximately -1.4 volts at the control grid of the color-killer stage. For this low level of bias, the color killer stage conducts and develops a cutoff bias for the chroma bandpass amplifier.

When color signals are being

26-35 CHROMA CIRCUITS (Cont'd)

Circuit Description (Cont'd)

received, the burst signals applied to the oscillator causes the oscillator grid bias voltage to increase to approximately −8 volts, depending on the amplitude of the burst signal. The dc voltage at the anode of the voltage-doubler diode then rises to approximately —16 volts, and the bias on the color-killer stage is increased to about −4 volts. For this bias level, no current flows through the color-killer stage, and the cutoff bias for the chroma bandpass amplifier provided b the color-killer stage is removed. The grid bias for the bandpass amplifier is then derived from the dc voltage at the grid of the 3.58-MHz oscillator. Because this voltage varies with the amplitude of the burst signal, it provides automatic-gain control for the bandpass amplifier.

With the removal of the cutoff bias provided by the color killer, the bandpass amplifier is allowed to amplify and and pass the 3.58-MHz chrominance sidebands contained in the chroma input (video signal). The single-tuned transformer T_2 in the plate circuit of the bandpass amplifier forms a selective load to the 3.58-MHz chrominance sidebands. The output of the bandpass amplifier, therefore, is a 3.58-MHz signal that contains the R− Y and B − Y color-difference information. The instantaneous phase difference of the 3.58-MHz color-difference components with respect to the burst synchronizing signal defines the color information being transmitted, as indicated by the chart on page 73 in the section **Electron Tube Applications.**

The 3.58-MHz color-difference signals from the bandpass amplifier are coupled by transformer T_2 to the screen grids of the X and Z color demodulators where they are mixed with the continuous-wave 3.58-MHz signal from the oscillator. The color demodulators are essentially synchronous detectors. These types of detectors are phase sensitive, and their output is determined not only by the amplitudes of the two input signals, but also by the phase relationship of these inputs. If the amplitudes of the chrominance and continuous wave inputs to the demodulators are considered to be constant, the input of the demodulators is affected by the phase relationship of the two input signals as follows: When the chrominance and the continuous signals are in phase, the output of the demodulators is maximum in the negative direction. When the two signals are 180 degrees out of phase, the output is maximum in the positive direction. A phase difference of 90 or 270 degrees results in a zero output from the demodulators.

The X and Z color demodulators are biased so that the plate current of each demodulator tube is small during the zero-signal condition. The continuous-wave signal applied to the control grid gates the tube into conduction for the full positive half cycle. During most of the negative half cycle, the tube is cut off. With no chrominance signal applied to the screen grid, the plate current of the demodulator tube consists essentially of 3.58-MHz pulses. A low-pass filter in the plate circuit of the demodulator removes the 3.58-MHz component so that the dc plate voltage decreases below the level obtained when there is no input to either the control or screen grid. The dc level obtained when only the continuous-wave reference signal is applied represents the zero output of the color demodulators; only changes in the average plate voltage above and below this level will be passed by the output coupling capacitor to the succeeding stages.

When the chrominance signal applied to the screen grid is in phase with the continuous-wave reference signal applied to the control grid,

26-35 CHROMA CIRCUITS (Cont'd)

Circuit Description (Cont'd)

the demodulator tube conducts more heavily during the periods that the reference signal permits conduction. The plate voltage of the demodulator then decreases below the zero level, and the output coupling capacitor couples the negative change to the next stage. Conversely, if the two signals are 180 degrees out of phase, the average plate current decreases. The attendant rise in average plate voltage causes a positive change to be coupled to the next stage. For 90- or 270-degree phase differences, the two signals tend to add together at certain times and to cancel each other times so that the average plate current is essentially unchanged.

In the development of the color-difference signals at the transmitter, the phase of the R — Y signal is shifted 90 degrees with respect to the burst reference signal and the B — Y signal is in phase with the reference signal. The B — Y component of the chrominance sidebands, therefore, is in phase with the reference signal applied to the Z demodulator, and the R — Y component is in phase with the phase-shifted reference signal applied to the X demodulator. The output of the Z demodulator then is the detected G — Y signal, and the output of the X demodulator is the detected R — Y signal. These signals are coupled to the B — Y and R — Y difference-signal amplifiers, respectively.

If strict consideration is given to signal phase relationships, the outputs of the X and Z demodulators are — (R — Y) and — (B — Y) signals. The positive versions of these color-diffrence signals results from the inversions provided by the R — Y and B — Y color-difference amplifiers. The G — Y color-difference signal is synthesized by addition of portions of the R — Y and B — Y signals from the plates of the R — Y and B — Y difference amplifiers in the resistor matrix network at the input to the G — Y color-difference amplifier. The vector sum of these quantities results in a —(G — Y) signal. This signal is amplified and inverted by the G — Y amplifier to obtain the G — Y signal.

The color difference amplifiers all operate in the grounded-cathode mode with the grid bias taken from the blanker circuit, and only capacitance coupling is used from the outputs of these amplifiers to the picture tube. The dc reference level for the three color grids of the picture tube are established by a clamp diode circuit in the output of each difference amplifier. The outputs of the R — Y, G — Y, and B — Y color-difference amplifier are coupled to the red, green, and blue grids, respectively, of the color picture tube.

26-36 PICTURE TUBE AND ASSOCIATED CIRCUITS

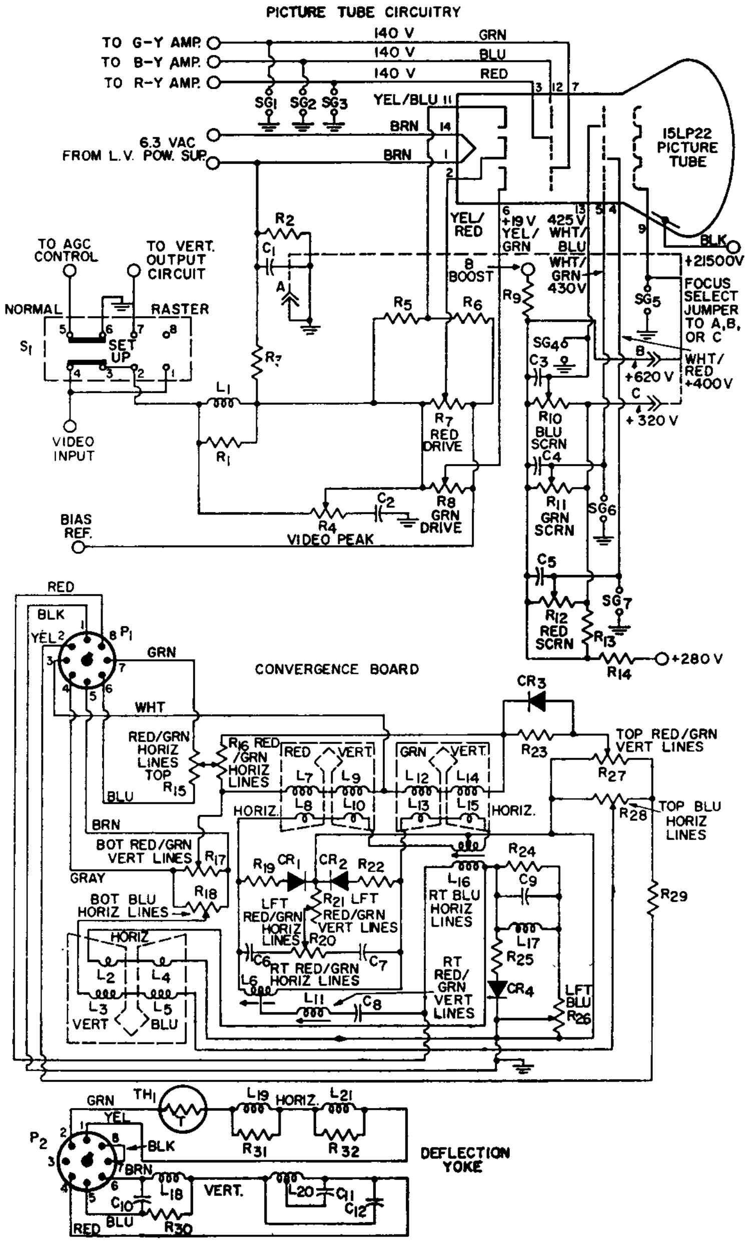
For Color Television Receiver

Circuit Description

These circuits include the picture tube and associated input-coupling and biasing networks, the convergence board, and the horizontal and vertical deflection coils for a color television receiver. The dc operating potentials for the picture tube are derived from the receiver low-voltage (280-volt) power supply, the B Boost (700-volt) voltage developed by the horizontal-output circuit, and the high-voltage (21,500-volt) rectifier circuit. The 6.3 volt heater power for the picture tube is obtained from a transformer (T_1 in circuit 26-30) connected across the 117-volt ac power line.

The 15LP22 color picture tube has a number of unique features. The phosphor-dot screen uses a rare-earth, red-emitting phosphor and improved blue and green phosphors.

26-36 PICTURE TUBE AND ASSOCIATED CIRCUITS (Cont'd)



26-36 PICTURE TUBE AND ASSOCIATED CIRCUITS (Cont'd)

Parts List

C_1=0.1 μF, Mylar, 400 V
C_2=47 pF, ceramic, 500 V, N750
C_3, C_4, C_5=1000 pF, ceramic, 500 V
C_6, C_7=0.15 μF, Mylar, 75 V (part of convergence-board assembly)
C_8=0.082 μF, Mylar, 100 V (part of convergence-board assembly)
C_9=0.27 μF, Mylar, 75 V (part of convergence-board assembly)
C_{10}=180 pF, 250 V, part of deflection-yoke assembly
C_{11}=3900 pF, part of deflection-yoke assembly
C_{12}=82 pF, 3000 V, part of deflection-yoke assembly
CR_1, CR_2, CR_3, CR_4=Selenium rectifier assembly, RCA Stock No. 120058 or equiv.
Convergence board=RCA Stock No. 120052 or equiv.
Deflection yoke=RCA Stock No. 120890 or equiv.
L_1=820 μH, part of network assembly (RCA Stock No. 120796 or equiv.) with R_1
(L_2—L_4, L_3—L_5) (L_7—L_9, L_8—L_{10}) (L_{12}—L_{14}, L_{13}—L_{15})=Convergence-coil assembly, RCA Stock No. 121343 or equiv., part of convergence-board assembly
L_6=Variable inductor, right red-green vertical lines adjustment, RCA Stock No. 120059 or equiv., part of convergence-board assembly
L_{11}=Variable inductor, right red/green vertical lines adjustment, RCA Stock No. 121443 or equiv., part of convergence-board assembly
L_{16}=Variable inductor, right blue horizontal lines adjustment, RCA Stock No. 120060 or equiv., part of convergence-board assembly
L_{17}=120 μH, RCA Stock No. 118245 or equiv., part of convergence-board assembly
L_{18}, L_{20}=Vertical-deflection coils, part of deflection-yoke assembly
L_{19}, L_{21}=Horizontal-deflection coils, part of deflection-yoke assembly
P_1=Connector for convergence board, 8-pin male type, RCA Stock No. 112728 or equiv. (mates with J_1 on circuit 26-34)
P_2=Connector for yoke assembly, 8-pin male type, RCA Stock No. 114767 or equiv. (mates with J_2 on circuit 26-34)
R_1=4700 ohms, 0.5 watt, part of network assembly with L_1
R_2=0.18 megohm, 0.5 watt
R_3=0.15 megohm, 0.5 watt
R_4=Potentiometer, video peak adjustment, 0.1 megohm, 0.5 watt, part of assembly with R_7 and R_8 (RCA Stock No. 120811 or equiv.)
R_5=5600 ohms, 0.5 watt
R_6=12000 ohms, 0.5 watt
R_7=Potentiometer, red drive adjustment, 6000 ohms, 0.5 watt, part of assembly with R_5 and R_8 (RCA Stock No. 120811 or equiv.)
R_8=Potentiometer, green drive adjustment, 6000 ohms, 0.5 watt, part of assembly with R_5 and R_7 (RCA Stock No. 120811 or equiv.)
R_9=33000 ohms ±5%, 0.5 watt
R_{10}, R_{11}, R_{12}=Three-section potentiometer; screen-grid adjustments for blue, green, and red electron guns, respectively; each section: 1.5 megohms, 0.5 watt; RCA Stock No. 120812 or equiv.
R_{13}=47000 ohms, 0.5 watt
R_{14}=1000 ohms, 0.5 watt
R_{15}=Potentiometer, top red/green horizontal lines adjustment, 120 ohms, 0.5 watt, RCA Stock No. 106320 or equiv. (part of convergence-board assembly)
R_{16}=Potentiometer, bottom red/green horizontal lines adjustment, 350 ohms, 0.5 watt, RCA Stock No. 116635 or equiv. (part of convergence-board assembly)
R_{17}=Potentiometer, bottom red/green vertical lines adjustment, 60 ohms, 0.5 watt, RCA Stock No. 105059 or equiv. (part of convergence-board assembly)
R_{18}=Potentiometer, bottom blue horizontal lines adjustment, 60 ohms, 0.5 watt, RCA Stock No. 105059 or equiv. (part of convergence-board assembly)
R_{19}, R_{22}=100 ohms, 1 watt, part of convergence-board assembly
R_{20}=Potentiometer, left red/green horizontal lines adjustment, 100 ohms, 0.5 watt, RCA Stock No. 120949 or equiv. (part of convergence-board assembly
R_{21}=Potentiometer, left red/green vertical lines adjustment, 100 ohms, 0.5 watt, RCA Stock No. 120949 or equiv. (part of convergence-board assembly
R_{23}=270 ohms, 0.5 watt (part of convergence-board assembly)
R_{24}=180 ohms, 1 watt (part of convergence-board assembly)
R_{25}=270 ohms, 1 watt (part of convergence-board assembly)
R_{26}=Potentiometer, left blue adjustment, 60 ohms, 3 watts, RCA Stock No. 114627 or equiv. (part of convergence-board assembly)
R_{27}=Potentiometer, top red/green vertical lines adjustment, 350 ohms, 0.5 watt, RCA Stock No. 116635 or equiv. (part of convergence-board assembly)
R_{28}=Potentiometer, top blue horizontal lines adjustment, 350 ohms, 0.5 watt, RCA Stock No. 116635 or equiv. (part of convergence-board assembly)
R_{29}=82 ohms, 0.5 watt (part of convergence-board assembly)
R_{30}=4700 ohms, 2 watts (part of deflection-yoke assembly)
R_{31}, R_{32}=220 ohms, 0.5 watt
S=Service switch, RCA Stock No. 120838 or equiv.
SG_1 through SG_7=Capacitor, spark-gap, 0.5 pF, 1000 V, RCA Stock No. 120819 or equiv.
TH_1=Thermistor; cold resistance, 1.3 ohms; RCA Stock No. 120891

See Note on page 617.

Circuit Description (Cont'd)

The new phosphors are more efficient and are capable of producing 38 per cent brighter highlights than previous color picture tubes. The directly viewed shadow-mask picture tube incorporates a screen with nearly straight sides and sharply rounded corners.

The 15LP22 is designed for operation with the blue gun down. The

26-36 PICTURE TUBE AND ASSOCIATED CIRCUITS (Cont'd)

Circuit Description (Cont'd)

anode bulb contact for high voltage connection is still located in the top section of the tube. Operation in the blue-down orientation, with respect to the viewing screen, provides optimum compromise of pincushion distortion at the top and bottom of the screen. The tube is equipped with an integral filter glass protective window, sealed to the base plate of the tube with a clear resin. An external magnetic shield is not required on the 15LP22. Another main feature of the color picture tube is an einzel-lens focus system. This system is relatively insensitive to variations of the high voltage so that the tube maintains good focus even with variations in picture brightness.

The focus system for the color picture tube is very similar to that used in instruments equipped with a black-and-white picture tube. Normally, the 15LP22 will have optimum focus when connected to ground potential. However, provisions to change the focus potential are facilitated by a pin connector from pin 9 of the picture tube. The focus selected jumper can be connected to 620 volts, 320 volts, or ground merely by relocating the slip-on connector to the proper stake extending from the circuit board.

A three-position service switch S_1 is incorporated into the picture-tube circuitry to facilitate receiver setup and adjustment. The NORMAL position of the switch, of course, permits normal receiver operation. With the switch in the SETUP or RASTER position, the video input is disconnected from the picture tube, and the ground return for the agc circuit is opened. Raster height and width and color and background levels can then be more easily adjusted.

The output of the color difference amplifiers are applied to the respective grids of the tricolor picture tube. The luminance signal from the second video amplifier is applied to the three cathodes of the color picture tube. These signals combine to intensity modulate the three electron beams to produce the color image on the picture-tube screen.

The horizontal and vertical deflection coils in a yoke on the neck of the picture tube deflect the electron beams, in response to signals received from the horizontal and vertical output stages, to produce the horizontal and vertical scanning required to trace the image on the picture-tube screen. (These coils are connected in shunt with the respective horizontal and vertical output transformer.)

The horizontal output circuit provides a sawtooth current waveform at a frequecy of 15,750 Hz to the horizontal-deflection coils, and the vertical output circuit provides a 60-Hz sawtooth current wave to the vertical-deflection coils. The picture tube electron beams are simultaneously deflected horizontally across the screen at a rate of 15,750 Hz and vertically at a rate of 60 Hz.

At the completion of each horizontal trace (end of rising portion of sawtooth current wave), the beam is deflected back to the left side of the screen (retrace) to start another trace period. A positive blanking pulse (included in the video signal) applied to the cathodes of the picture tubes cuts off the picture tube during this period so that the retrace lines do not appear on the tube screen. The picture tube is similarly blanked at the end of each vertical-trace period.

Correct color reproduction requires that the three beams of the color picture tubes meet, or converge, at the shadow mask and excite color dots of the same trios. The three electron guns of the color picture tube are mechanically tilted toward the center axis of the tube so that virtual convergence is ob-

26-36 PICTURE TUBE AND ASSOCIATED CIRCUITS (Cont'd)

Circuit Description (Cont'd)

tained with no external converging force applied. Slight bending of one or more of the beams may be required for exact convergence. The convergence circuit performs this function.

The components on the convergence board shown in the circuit diagram are mounted on a disk-shaped circuit board with a center hole that permits it to be fitted directly on the neck of the color picture tube. These components are interconnected in a dynamic type of convergence system. In this system, sine wave currents are used to provide horizontal convergence, and parabolic current waves are used to provide vertical convergence.

The sine waves of current used to provide horizontal convergence are derived from a voltage pulse developed across an auxiliary winding of the high-voltage transformer (T_1 in circuit 26-34) and applied through pin 8 of the convergence-board input connector P_1. The current through each of the three sets of horizontal convergence coils (L_2 and L_4, L_8 and L_{10}, and L_{13} and L_{15}) is individually adjustable in both amplitude and phase. The phase of the convergence current is adjusted b the Horizontal Shape control L_6, which resonates with the two 0.15-mcrofarad capacitors C_6 and C_7 at the line frequency (15,750 Hz). The sine-wave convergence current is produced by ringing this resonant circuit with the pulse obtained from the high-voltage transformer. Potentiometers R_{15}, R_{16}, R_{18}, R_{20}, and R_{28} adjust the amplitude of the sine-wave convergence current.

Vertical-frequency (60-Hz) sawtooth voltages obtained from secondary windings of the vertical-output transformer (T_2 in circuit 26-34), applied through pins 4 and 5 and pins 6 and 7 of connector P_1, are used to derive the vertical convergence-current waveform. Because of the integrating action of the convergence coils, this sawtooth voltage results in a parabolic current wave through the convergence coils. Potentiometer R_{21} adjusts the amplitude of the vertical voltage parabola applied to the three sets of vertical convergence coils (L_3 and L_5, L_7 and L_9, and L_{12} and L_{14}).

A vertical-frequency sawtooth voltage from a secondary winding of the vertical-output transformer, is applied across potentiometer R_{17}. The sawtooth voltage is obtained from center tapped transformers; the voltage at the center of potentiometer R_{17} therefore, is approximately zero with respect to circuit ground. Adjustment of this potentiometer mixes either positive or negative sawtooth voltages with the parabolic convergence voltage and, in this way, controls the shape of the convergence signal applied to the convergence coils.

INDEX

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on Electron Tubes, Semiconductor Products, and Batteries

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• PRODUCT GUIDE FOR RCA POWER TUBES—PWR-506B—40 pages. Contains tabulated data on all RCA power tubes in order of type designation within each general class of service. Includes maximum ratings, temperature ratings, heater or filament requirements, outline drawings, and basing diagrams. Price 30 cents.*

• RCA INDUSTRIAL TUBES PRODUCT GUIDE—TPG-200C (10⅞" x 8⅜")—28 pages. Covers all RCA industrial-tube product lines. Gives a brief description of each product line together with quick-selection data. Single copy free on request.

• RCA STORAGE TUBES AND CATHODE-RAY TUBES—STC-900B—16 pages. Contains technical information on RCA storage tubes, special-purpose kinescopes and oscillograph-type cathode-ray tubes including display-storage tubes, radechons, scan-conversion tubes, flying-spot tubes, monitor, projection, transcriber, and view-finder kinescopes; as well as data on fluorescent screens. Price 20 cents.*

• RCA TRAVELING-WAVE TUBE CLASSIFICATION CHARTS—MWD-101C—4 pages. Contains catalog-type data. Single copy free on request.

• RCA PENCIL TUBE CLASSIFICATION CHARTS—MWD-102B—4 pages. Contains catalog-type data. Single copy free on request.

• RCA CAMERA TUBES—CAM-600A—26 pages. Contains classification charts, defining data and typical characteristic curves for RCA image orthicons and vidicons. Camera tubes recommended for new equipment design are highlighted. Price 50 cents.*

• VIDICONS—CAM-700—16 pages. Supplies tube selection guidance and data on RCA vidicons for commercial, educational, industrial, and military service. Also included are tube replacement information and typical vidicon characteristic curves. The information contained in this publication supersedes the vidicon section of the booklet CAM-600A. Price 30 cents.*

● TECHNICAL BULLETINS—Authorized information on RCA receiving tubes, transmitting tubes, and other tubes for communications and industry. Be sure to mention tube-type bulletin desired. Single-copy on any type free on request.

Semiconductor Products

● RCA SEMICONDUCTOR PRODUCTS DATABOOK—SPD-100. Two loose-leaf binders for standard 8½" x 11" data booklets with more than 900 pages of data and curves on RCA semiconductor devices such as transistors, silicon rectifiers, and semiconductor diodes. Available on a subscription basis. Price $15.00* including service for first year. Also available with RCA Electron Tube Handbook HB-3 at special combination price of $30.00.*

● RCA SILICON CONTROLLED RECTIFIER EXPERIMENTER'S MANUAL—KM-71 (8⅜" x 5⅜")—136 pages. Contains 24 practical and interesting control circuits that can be built with a complement of active devices available in kit form. Includes photographs, schematic diagrams, and descriptive writeups. Also includes brief descriptions of solid-state components used (rectifiers, transistors, SCR's) and short section on troubleshooting. Price 95 cents.*†

● RCA SILICON POWER CIRCUITS MANUAL—SP-50 (8¼" x 5¼")—416 pages. Contains design information for a broad range of power circuits using RCA silicon transistor, rectifiers, and thyristors (triacs and SCR's). Gives design criteria and procedures for applications involving rectification, supply filtering, power conversion and regulation, ac line-voltage controls, rf power amplifiers, and control and low-frequency amplifiers. Shows design and practical circuits. Price $2.00.*†

● RCA TRANSISTOR MANUAL—SC-13 (8⅜" x 5⅜")—544 pages. Contains up-to-date definitive data on over 770 semiconductor devices including tunnel diodes, silicon controlled rectifiers, varactor diodes, conventional rectifiers, and many classes of transistors. Features easy-to-understand text chapters, as well as tabular data on RCA discontinued transistors. Contains over 40 practical circuits, complete with parts lists, highlighting semiconductor-device applications. Price $2.00.*†

● RCA TUNNEL DIODE MANUAL—TD-30 (8⅜" x 5⅜")—160 pages. Describes the microwave and switching capabilities of tunnel diodes. Contains information on theory and characteristics, and on tunnel-diode applications in switching circuits and in microwave oscillator, converter, and amplifier circuits. Includes data for over 40 RCA germanium and gallium arsenide tunnel diodes and tunnel rectifiers. Price $1.50.*†

● RCA SEMICONDUCTOR PRODUCTS GUIDE—SPG-201D (10⅞" x 8⅜")—44 pages. Contains classification chart, index, and ratings and characteristics on RCA's line of transistors, silicon rectifiers, semiconductor diodes, and photocells. Price 75 cents.*

● RCA DIFFUSED-JUNCTION SILICON RECTIFIER STACKS AND BRIDGES—SRS-300—10 pages. Contains technical data on RCA's diffused-junction silicon rectifier stacks and bridges. Characteristics of basic rectifier circuits are also given to assist in selection of proper RCA rectifier device. Price 20 cents.*

● RCA SMALL-SIGNAL SILICON N-P-N TRANSISTORS—SST-210—8 pages. Contains technical data on 2N2102 family of silicon transistors including high-voltage types, very-high voltage types, linear-beta types, and general types. Also includes quick-reference guide. Price 20 cents.*

● DESIGN OF TRANSISTOR SWITCHING CIRCUITS FOR DATA-PROCESSING EQUIPMENT—CTG-161—42 pages. Gives design considerations for a variety of transistor switching circuits for data-processing equipment such as logic gates, flip-flops, and memory drivers. It includes a review of switching theory, design procedures, methods of specifying characteristics and ratings for computer switching transistors; examples of design procedures; typical circuits using RCA transistors; and a complete listing of RCA Computer Transistors with ratings, characteristics, and performance data. Price 75 cents.*

● RCA MOS FIELD-EFFECT TRANSISTORS PRODUCT GUIDE—MOS-160—20 pages. Includes comprehensive data on RCA dual insulated-gate and single insulated-gate MOS FET's in easy-to-find format plus background information on MOS construction and application. Price 20 cents.*

● HEAT-SINK GUIDANCE FOR RCA THYRISTORS USING TO-5 AND "MODIFIED TO-5" PACKAGES—SCR-501—6 pages. Application guide on heat-sink methods for RCA thyristors. Single copy free on request.

● RCA HOMETAXIAL BASE SILICON POWER TRANSISTORS—HBT-400A—18 pages. Contains data, dimensional outlines and theoretical information on hometaxial-base silicon power transistors. Price 30 cents.*

● RCA LOW-NOISE COMMUNICATION-TYPE TRANSISTORS—CTG-165—Contains quick-selection graphs and charts and capsule data for RCA Bipolar Transistors and MOS Field-Effect Transistors for Low-Noise VHF and UHF Communication and Industrial Instrumentation Applications. Includes special characteristics curves showing quick-selection chart containing curves, G_p (dB) and NF (db) vs. f (30 to 1000 MHz) for each listed transistor type. Single copy free on request.

● MOUNTING HARDWARE FOR RCA INDUSTRIAL SEMICONDUCTOR DEVICES—MHI-300—4 pages. Contains mounting information for RCA industrial transistors, thyristors, and rectifiers. Single copy free on request.

● RCA RF POWER TRANSISTORS—RFT-700B—6 pages. Contains data, selection guide, and a quick-selection graph on RCA "overlay" transistors. Single copy free on request.

● RCA PHOTOCELLS—SOLID-STATE PHOTOSENSITIVE DEVICES—CSS-800A—32 pages. Contains detailed and updated information on RCA cadmium-sulfide and cadmium-sulfo-selenide photoconductive-cell characteristics, an extended section on photoelectric measurements, a new section describing design, new circuits, and an extension replacement guide. Price 35 cents.*

● RCA PHOTOCONDUCTIVE CELLS—File No. 312—8 pages. Contains descriptive material, characteristic curves, and classification charts on RCA cadmium-sulfide and cadmium-sulfo-selenide brood-area photoconductive cells. Single copy free on request.

● RCA SILICON POWER TRANSISTOR APPLICATION GUIDE—1CE-215—28 pages. For designers of industrial and military equipment. Discusses ratings, stability conditions, parameters and equivalent circuits. Includes design procedures and specific design equations for several transistor circuits. Price 50 cents.*

● SILICON VHF TRANSISTORS APPLICATION GUIDE—1CE-228—20 pages. For designers of industrial and military equipment. This guide describes the capabilities of RCA silicon vhf transistors for application at frequencies up to 300 MHz. Includes typical circuits for the 2NF1491 family of silicon vhf transistors. Maximum ratings and characteristics are included. Price 50 cents.*

● RCA THYRISTORS (SCR's AND TRIACS)—SCR-500A—22 pages. Contains tabulated data, classification charts and dimensional outlines for all-diffused silicon thyristors. Price 40 cents.*

● RCA TOP-OF-THE-LINE SOLID-STATE REPLACEMENT GUIDE—SPG-202-E—48 pages. Lists 31 RCA "Top-of-the-Line" SK-Series replacement semiconductor devices which can replace more than 9600 types of transistors, integrated circuits, and rectifiers used in entertainment electronic equipment, including U.S.A. industry-standard (EIA) types, foreign types, and types identified only by device-manufacturers' part numbers. Price 15 cents.*

● TRANSISTORIZED VOLTAGE REGULATOR APPLICATION GUIDE—1CE-254—12 pages. Discusses transistorized voltage regulators of the series and shunt types. Includes design considerations, step-by-step design procedures, and the solutions to sample design problems. Price 20 cents.*

Integrated Circuits

● RCA LINEAR INTEGRATED CIRCUIT FUNDAMENTALS—(8¼" x 5⅜") 240 pages. Contains basic principals involved in design and application of linear integrated circuits—includes description of silicon monolithic fabrication process—derivation of design equations and performance criteria—schematic diagrams, operating characteristics, and performance data for RCA (multiple-function silicon integrated circuits for a variety of linear applications. Price 2.00*†

Batteries

● RCA BATTERY MANUAL—BDG-111 (10⅞" x 8¾")—68 pages. Contains information on dry cells and batteries carbon zinc, mercury, and alkaline types. Includes battery theory and applications, detailed electrical and mechanical characteristics, a classification chart, dimensional outlines, and terminal connections on each battery type. Price 50 cents.*†

● RCA BATTERIES—BAT-134H (10⅞" x 8⅜"—36 pages. Technical data on 146 carbon-zinc, alkaline, and mercury batteries for consumer and industrial applications. Includes replacement information for 4000 portable radios, and cross-references 860 domestic battery types to their RCA replacements. Price 35 cents.*†

Test and Measuring Equipment

● INSTRUCTION BOOKLETS — Illustrated instruction booklets are available for all RCA test instruments at the prices indicated below.

WA-44A	(Audio Signal Generator)	$0.50*
WA-44C	(Audio Signal Generator)	1.00*
WO-33A	(Super Portable Oscilloscope)	1.00*
WO-88A	(5-in. Oscilloscope)	0.75*
WO-91A	(5-in. Oscilloscope)	1.00*
WO-91B	(5-in. Oscilloscope)	1.00*
WR-36A	(Dot-Bar Generator)	.0.50*
WR-46A	(Video Dot/Crosshatch Generator)	1.00*
WR-49A	(RF Signal Generator)	0.50*
WR-49B	(RF Signal Generator)	1.00*
WR-50A	(RF Signal Generator)	1.00*
WR-51A	(Stereo FM Signal Simulator)	1.00*
WR-52A	(Stereo FM Signal Simulator)	1.00*
WR-61B	(Color-Bar Generator)	1.00*
WR-64A	(Color Bar/Dot/Crosshatch Generator)	1.00*
WR-64B	(Color/Bar/Dot/Crosshatch Generator)	1.00*
WR-67A	(Test-Oscillator)	0.25*
WR-69A	(Television/FM Sweep Generator)	1.00*
WR-70A	(RF-IF-VF Marker Adder)	0.75*
WR-86A	(UHF Sweep Generator)	0.50*
WR-99A	(Marker Calibrator)	1.00*
WT-100A	(Electron-Tube Micro Mho Meter)	1.75*
WT-100A	(Electron-Tube Micro Mho Meter, Ser. No. 1001 and over)	2.00*
WT-100A	(Tube Chart 1CE-163)	3.00*
WT-110A	(Automatic Electron-Tube Tester)	0.75*
WT-110A	(1CE-174 Card Punch Data)	0.25*
WT-110A	(1CE-234 Card Punch Data)	1.00*

WT-115A (Color Picture Tube Tester) 0.50*
WV-37A (Radio Battery Tester) 0.25*
WV-37B (Radio Battery Tester) 0.25*
WV-38A (Volt-Ohm-Milliammeter) 0.50*
WV-65A (VoltOhmyst†) 0.25*
WV-74A (High Sensitivity AC VTVM) 0.75*
WV-75A (VoltOhmyst†) 0.25*
WV-76A (High Sensitivity AC VTVM) 0.75*
WV-77A (VoltOhmyst†) 0.25*
WV-77B (VoltOhmyst†) 0.25*
WV-77E (VoltOhmyst†) 1.00*
WV-84C (Ultra-Sensitive DC Microammeter) 0.75*
WV-95A (Master VoltOhmyst†) 0.25*
WV-97A (Senior VoltOhmyst†) 0.75*
WV-98A (Senior VoltOhmyst†) 1.00*
WV-98B (Senior VoltOhmyst†) 1.00*
WV-98C (Senior VoltOhmyst†) 0.50*
195-A (VoltOhmyst†) 0.25*

† Trade Mark Reg. U.S. Pat. Off.

* Prices shown apply in U.S.A. and are subject to change without notice.

† Suggested price.

Reading List

This list includes references of both elementary and advanced character. Obviously, the list is not inclusive, but it will guide the reader to other references

ALBERT, A. L. *Electrons and Electron Devices.* The Macmillan Co.

BECK, A. H. W. *Thermionic Valves.* Cambridge University Press

CHUTE, G. M. *Electronics in Industry.* McGraw-Hill Book Co., Inc.

DOME, R. B. *Television Principles.* McGraw-Hill Book Co., Inc.

DOW, W. G. *Fundamentals of Engineering Electronics.* John Wiley and Sons, Inc.

EASTMAN, A. V. *Fundamentals of Vacuum Tubes.* McGraw-Hill Book Co., Inc

EDSON, W. A. *Vacuum Tube Oscillators.* John Wiley and Sons, Inc.

FINK, D. G. *Television Engineering.* McGraw-Hill Book Co., Inc.

GHIRARDI, A. A. *Radio and Television Receiver Circuitry and Operation.* Rinehart and Co., Inc.

GRAY, T. S. *Applied Electronics.* John Wiley and Sons, Inc.

GROB, B. *Basic Television.* McGraw-Hill Book Co., Inc.

HENNEY, KEITH. *Radio Engineering Handbook.* McGraw-Hill Book Co., Inc.

HOAG, J. B. *Basic Radio.* D. Van Nostrand Co., Inc.

KOLLER, L. R. *Physics of Electron Tubes.* McGraw-Hill Book Co., Inc.

MAEDEL, G. F. *Basic Mathematics for Television and Radio.* Prentice-Hall, Inc.

MARCUS, A. *Elements of Radio.* Prentice-Hall, Inc.

MARKUS AND ZELUFF. *Handbook of Industrial Electronic Circuits.* McGraw-Hill Book Co., Inc.

MILLMAN AND SEELY. *Electronics.* McGraw-Hill Book Co., Inc.

MOYER AND WOSTREL. *Radio Receiving and Television Tubes.* McGraw-Hill Book Co., Inc.

PENDER, DELMAR, AND MCILWAIN. *Handbook for Electrical Engineers—Communications and Electronics.* John Wiley and Sons, Inc.

PREISMAN, A. *Graphical Constructions for Vacuum Tube Circuits.* McGraw-Hill Book Co., Inc.

HICKEY, H. V., and VILLINES, JR., W. M. *Elements of Electronics.* McGraw-Hill Book Co., Inc.

RCA TECHNICAL BOOK SERIES. *Electron Tubes, Vol. I and Vol. II.* RCA Review.

REICH, H. J. *Theory and Applications of Electron Tubes.* McGraw-Hill Book Co., Inc.

RICHTER, WALTHER. *Fundamentals of Industrial Electronic Ciruits.* Mc-Graw-Hill Book Co., Inc.

SEELY, S. *Electron Tube Circuits.* McGraw-Hill Book Co., Inc.

SPANGENBERG, K. R. *Vacuum Tubes.* McGraw-Hill Book Co., Inc.

STURLEY, K. R. *Radio Receiver Design.* Chapman and Hall, Ltd.

TERMAN, F. E. *Fundamentals of Radio.* McGraw-Hill Book Co., Inc.

TERMAN, F. E. *Radio Engineers Handbook.* McGraw-Hill Book Co., Inc.

The Radio Amateurs Handbook. American Radio Relay League.

ZWORYKIN AND MORTON. *Television: The Electronics of Image Transmission.* John Wiley and Sons, Inc.

KEY: BASING DIAGRAMS (Bottom Views)

•	Gas-Type Tube
BC	Base Sleeve
BS	Base shell
C	External Conductive Coating
CL	Collector
DJ	Deflecting Electrode
ES	External Shield
F	Filament
F+	Filament (positive only)
F−	Filament (negative only)
F_M	Filament Tap
G	Grid
H	Heater
H_L	Heater Tap for Panel Lamp
H_M	Heater Tap
IC	Do Not Use
IS	Internal Shield
K	Cathode
LC	Do Not Use, Except As Specified in Data
NC	No Internal Connection—May Be Used As Tie Point
P	Plate (Anode)
RCJ	Ray-Control Electrode
S	Shell
TA	Fluorescent Target

Subscript for multi-unit types:

- B = beam power unit
- D = *diode unit*
- HP = heptode unit
- HX = hexode unit
- P = pentode unit
- T = *triode unit*
- TR = tetrode unit

RCA Electronic Components, Harrison, N. J. 07029

ADV Plans, LLC