# Popular Electronics 

## How to Design Power Supplies Easy-to-build LED Projects

 Microprocessor Microcourse, Part II16-PAGE CB RADIO SUPPLEMENT, INCLUDING Buyers Guide to Super CB Mobile Transceivers


6L 7n O Ofid 26098080 dWJ 969089
Sharp RT-3388 Computer-controlled Cassette Deck
In This Wharfedale E-50 High-efficiency Speaker System Issue

Introducing the mobile that can move you out of the world of the ordinary and into the world of the serious CB'er The Cobra 138XLR Single Sideband. Sidebanding puts you in your own private world. A world where there's less congestion More privacy. More time to talk


It's all possible because instead of 40 channels you get your choice of 120 channels. Both AM and SSB. And instead of 4 watts of legal power you get 12 watts of legal power. So you get almost double the range of AM
With the 138XLR Single Sideband there's less background noise and less interference. So there's cleaner, clearer reception. Because like all Cobras, the 138XLR SSB is engineered to punch through loud and clear. Even in crowded metropolitan areas

And like all Cobras it comes equipped with such standard features as an easy-to-read LED channel indicator. Switchable noise blanking and limiting An RF/signal strength meter. And Cobra'sexclusive DynaMike gain control You'll find the 138XLR SSB wherever Cobras are sold. Which is almost everywhere. Because Cobra's got a nationwide network of dealers and Authorized Service Centers offering sales, installation, service and advice. So come on in And move on up.

Punches through loud and clear.
Cobra Communications Products DYNASCAN CORPORATION
6460 W Cortland St., Chicago, Illinois 60635

Write for color brochure

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# If you think this is merely the world's best omni-directional CB base antenna, THINK, TWICE. 

STARDUSTER II. Like its famous father, outperforms any $5 / \mathrm{s}$ wave omni-directional CB antenna ever designed. Period. But that's only half the story. We also designed it to work simultaneously on 88-108 MHz FM. It receives FM broadcasts right alongside your CB rig so you get superb double duty from one great antenna. FM reception is vertically polarized for a perfect omni pattern. Tuned traps and phasing coil for excellent performance across the entire band. And in case you're wondering, you still get the STARDUSTER's fantastic CB performance.
Tremendous 5dB gain. Our original full aperture design
that punches your signal farther out on the horizon. Plus a 500 to 1 power safety factor. Installation? STARDUSTER II's so super lightweight and compact you can carry it up a ladder and still hold onto the ladder. It's made of aircraft-grade seamless aluminum tubing, weighs just $\mathbf{4 . 0}$ pounds. You can assemble and put up this rugged beauty in minutes, and it'll stay up. Resists $\mathbf{1 0 0} \mathbf{~ m p h}$ winds.

Naturally, we pack in everything you need, even the coax splitter and 300 ohm patch cable.

STARDUSTER II. A great answer to two antenna Model M-800
before you buy any other CB antenna?
the antenna specialists co.

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## FREQOUI

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## Specifications．

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## Coming Next <br> Month

- MAGNETOMETERS FOR INVESTIGATING UFO'S
- MICROPROCESSOR MICROCOURSE, PART 3
- ADDING TRIGGER SWEEP TO AN OSCILLOSCOPE

TEST REPORTS:
Dynaco Stereo 416
Power Amplifier
Philips AH-673
AM/Stereo FM Tuner
Heath HW-2036 2-meter
AM Transceiver
Sencore CB41
CB Performance Tester

Cover Art by George Kelvin

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## Feature Articles

THE POWER RACE—WHEN WILL IT END / Julian Hirsch HOW TO DESIGN \& BUILD POWER SUPPLIES/ Joseph Carr

Part 1: Basics of transformers, rectifiers, filters, regulators and protection circuits.
MICROPROCESSOR MICROCOURSE / Forrest M. Mims
Part 2: Basic digital logic.
HOW ELECTRONIC MUSIC SYNTHESIZERS WORK / John S. Simonton, Jr. Noise generators, instrument dynamics and voltage control.

## Construction Articles

BUILD A COMPUTER MUSIC BOX PERIPHERAL / Martin Smaha
Low-cost, 12 -tone, 4 -octave music generator also produces test signals.
FOUR EASY-TO-BUILD LED PROJECTS / Ray Wilkins
A blinker. flasher, binary counter, or wheel of fortune.

## Special Focus on CB Radio

CB TODAY \& WHERE IT'S GOING
BUYERS GUIDE TO SUPER MOBILE TRANSCEIVERS / Ivan Berger
Specifications and necessary buying information for SSB and AM rigs
ABOUT THIS MONTH'S COVER
CHOOSING A MOBILE CB ANTENNA/ John J. McVeigh
HOW TO INSTALL / Ivan Berger
Mobile CB transceivers and antennas.

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EXPERIMENTER'S CORNER / Forrest M. Mims
Getting Acquainted with CMOS
HOBBY SCENE O\&A / John McVeigh
DX LISTENING / Glenn Hauser
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COMPUTER BITS / Leslie Solomon High-Resolution Graphics

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Editorial

## ELECTRONICS AT WINTER CES

CB radio was big at the giant Winter Consumer Electronics Show in Las Vegas. Many CB manufacturers unveiled their plans for the coming year: high-technology CB transceivers. This means that there will be a greater proportion of SSB models than before, and more models with microprocessor-based circuitry that makes possible keyboard entry functions, channel memory storage and scanning.

Furthermore, there were large exhibits of mobile underdash models, remotely controlled models, and indash models of every description. The latter included combined CB/AM/stereo FM models, many with a choice of cassette or 8-track tape players, adjustable shafts, and universal installation kits. Of particular note was the rising number of models with a channel-9 priority function.

A host of sophisticated base stations was displayed, too, extending in price to about $\$ 1000$. These included microprocessor-based types with keyboard entry; models with three large, independent meters; SSB models with a VFO for receive that include a separate digital frequency readout; a base station with a large, blue fluorescent/digital display; and so on.

CB antenna manufacturers were not content to simply lead with dressed-up versions of last year's line, either. For example, antenna specialists showed a mobile antenna with a solid-state circuit that displayed the traditional loading coil; GTE Sylvania displayed an electronic self-tuning trunk-mount antenna that's said to automatically maintain minimum VSWR; Avanti's new "Saturn" selectable vertical/horizontal base station omni floated high in the exhibit area; Shakespeare displayed its fiberglass "City Stick" for indoor use; American Antenna's exhibit showed its K-40 models mounted in a variety of locations on an actual automobile; and two manufacturers which were formerly OEM-only makers-Harada and True Temper-entered the consumer area.

Other personal communications highlights at the show included Prime Electronic's PR-1000 variable audio filter with $40-3000-\mathrm{Hz}$ Peak, Notch, Low-Pass positions; Stoner's matching-cabinet console system that adds a ham radio and an AM CB adapter to its SSB-only base station; Wilson's WR-500 rotor, with a disctype brake; and Panasonic's $49-\mathrm{MHz}$ "Walkie Talkie" entry.

Rumors were rife, too. They included development of a computer interface adapter so that CB'ers can communicate computer-to-computer over the air (there's a legality question on this, though); an FCC proposal to permit ASCII as well as Baudot; and illegal use of 27.505 MHz for Morse code practice (as if the FCC didn't have enough trouble with other violations).

In other product categories, highlights were TV receivers with dual viewing channels; a raft of high-quality auto-sound products which included power boosters and graphic equalizers for car use, scanners with microprocessors, an $\mathrm{X} / \mathrm{K}$ band radar detector from Convoy that uses a parabolic antenna in a standard fog-lamp housing; no-solder BNC connectors from Cambridge; innumerable projection TV systems and VCR's; calculators with new twists such as Toshiba's LCD-display pocket notebook that stores names, addresses, and phone numbers; Cannon's mini desktop with dual displays to show memory content or calculating process figures; Casio's pocket calculator/alarm clock/timer; and Sinclair's $\$ 34.95$ programmable with a library of 290 programs. Computers and games essentially joined microprocessors at the show. This will be noted next month in our "Computer Bits."

In short, this was a most exciting show, with some 550 exhibitors who used more than $300,000 \mathrm{sq} \mathrm{ft}$ of space. (About one third of the show was devoted to hi-fi equipment, covered by Ralph Hodges in this month's Stereo Scene.) So it looks like a good year coming up!


POPULAR ELECTRONICS

# Pocket CB 

## New integrated circuit technology and a major electronic breakthrough brings you the world's smallest citizens band transceiver.

Scientists have produced a personal communications system so small that it can easily fit in your pocket. It's called the PocketCom. and it replaces larger units that cost considerably more.

## MANY PERSONAL USES

An executive can now talk with anybody in his office, his factory, or job site. The housewife can find her children at a busy shopping center. The motorist can signal for help in an emergency. The salesman, the construction foreman, the traveler, the sportsman, the hobbyist - everybody can use the PocketCom.

## LONG RANGE COMMUNICATIONS

The PocketCom's range is limited only by its 100 milliwatt power and the number of metal objects between units. Its power reaches from a few blocks in the city to several miles on a lake. Its receiver is so sensitve that signals sent from stronger citizen band stations several miles away can be picked up.

## VERY SIMPLE OPERATION

To use the PocketCom: simply turn it on, extend the antenna, press a button to transmit, and release it to listen. And no FCC license is required to operate it. The PocketCom has two Channels-channel 14 and an optional second channel. Plug in one of the other 22 citizen band crystals, and slide the channel selector to the second position. Crystals for the second channel cost $\$ 7.95$ and can only be ordered after receipt of your unit.


The PocketCom components are equivalent to 112 transistors whereas most comparable units contain only twelve.

## A MAJOR BREAKTHROUGH

The PocketCom's small size results from a breakthrough in the solid-state device that made the pocket calculator a reality. Scientists took 112 iransistors, integrated them on a micro-silicon wafer, and produced the world's first transceiver linear integrated circuit. This major breakthrough not only reduced the size of radio components but improved their dependability and performance.

## BEEP-TONE PAGING SYSTEM

You can page another PocketCom user, within close range, by simply pressing the PocketCom's call A beep tone sounds on the other unit if it has been left in the standby mode. In the standby mode, the unit is silent and can be kept on for weeks without draining the batteries.

## SUPERIOR FEATURES

Just check the advanced features now possible because of this new circuit breakthrough: 1) Incoming signals are amplified several million times compared to only 100,000 times on comparable conventional systems. 2) Even with a 60 decibel difference in signal strength. the unit's automatic gain control will bring up each incoming signal to a maximum uniform level. 3) A high squelch sensitivity ( 0.7 microvolts) permits noiseless operation without squelching weak signals.


## EXTRA LONG BATTERY LIFE

The PocketCom has a light-emitting diode low-battery indicator that tells you when your ' $N$ ' cell batteries require replacement. The integrated circuit requires such low power that the two batteries, with average use, will last weeks without running down.


The PocketCom can be used as a pager, an intercom, a telephone or even a security device.

## MULTIPLEX INTERCOM

Many businesses can use the PockerCom as a multiplex intercom. Each employee carries a unit tuned to a different channel. A citizen band base station with 23 channels is used to page each PocketCom. The results: an inexpensive and flexible multiplex intercom system for large construction sites, factories, offices, or farms

## NATIONAL SERVICE

The PocketCom is manufactured exclusively for JS\&A and is the unit currently used on the hit TV show, Charlie's Angels. JS\&A is America's largest supplier of space-age productsfurther assurance that your modest investment is well protected. The PocketCom should give you years of trouble-free service. However, should service ever be required, simply slip your 5 -ounce PocketCom into its handy mailer and send it to our prompt national service-by-mail center.


The PocketCom measures approximately $3 / 4$ $\times 11 / 2^{\prime \prime} \times 512^{\prime \prime}$ and easily fits into vour shirt pocket. The unit can be used as a personal communications link for business or pleasure.

GIVE IT A REAL WORKOUT
Remember the first time you saw a pocket calculator? It probably seemed unbelievable. The PocketCom may also seem unbelievable, so we give you the opportunity to personally examine one without obligation. Order only two units on a trial basis. Then really test them. Test the range, the sensitivity, the convenience. Test them under your everyday conditions, and compare the PocketCom with larger units

After you are absolutely convinced that the PocketCom is indeed an advanced product breakthrough, order additional units, crystals, or accessories on a priority basis as one of our established customers. If, however, the PocketCom does not suit your particular requirements, then return your units within ten days after receipt for a prompt and courteous refund. You cannot lose. Here is your opportunity to test an advanced space-age product at absolutely no risk.

## A COMPLETE PACKAGE

Each PocketCom comes complete with batteries, high-performance Channel 14 crystals, complete instructions, and a 90-day parts and labor warranty. Send $\$ 19.95$ per unit (or $\$ 39.90$ for two) plus $\$ 2.50$ per order for postage, insurance, and handling, (Illinois residents add $5 \%$ sales tax.) or credit card buyers may call our toll-free number. But don't delay.

Personal communications is the future of communications. Join the revolution. Order your PocketComs at no obligation today.

## NEW LOW PRICE!

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## Letters

## CHEERS FOR "EXPERIMENTER'S CORNER"

Until I read the February, March, November. and December 1977 "Experimenter's Corner," I was in the dark about the digital IC's discussed. Now, I see the light. Thanks popular Electronics and Forrest Mims. Let's have more, more. -C. Britton, Scarborough, Canada.

## INTERNATIONAL LAW ON LICENSING

I take exception to the comments made by Gil Duddies in the "Letters" column in the October 1977 issue. It is international law-not the choice of the American citizen-that gives hams license to operate on the airwaves. The US is just one of the hundreds of member countries of the International Telecommunications Union (ITU) that make up the rules and regulations by which radio oper-
ations are governed. The suggestion that a technically competent person who does not demonstrate ability in Morse Code should be given a license is covered by Article 41 Section 3 Paragraph 1 of the ITU regulations: "Any person operating the apparatus of an amateur station shall have proved that he is able to send correctly by hand and to receive correctly by ear texts in Morse code signals. Administrations may, however, waive this requirement for stations making use exclusively of frequencies above 144 MHz .'
Again quoting from the ITU regulations, Article 41 Section 3 Paragraph 2: "Administrations shall take measures as they judge necessary to verify the technical qualifications of any person operating the apparatus of an amateur station.'
So according to the ITU, an applicant for an amateur license must prove technical competence and ability to send and receive code before a license can be granted. -Dr. Jerrold L. Patz, K1PKT, Wrentham, MA

## LIKES SWL ARTICLES

I was very pleased to discover the SWL features in the November 1977 issue of Popular Electronics. Mr. Woods' SWL schedules are accurate and extremely useful. Mr. Hauser's article is also most interesting. It is gratifying to see the emphasis placed on the program material instead of the search for rare DX. -Arthur Crookshank, New York, NY

## ELF ADDITIONS

I would like to add my name to what I am sure is a long list of satisfied subscribers who have built the "COSMAC Elf Microcomputer" (August 1976). Being an experimenter, I would like to show off my addition to the original Elf. The photos shown here are representative programs that demonstrate my 32-character oscilloscope display. The circuit I am using for this display is an adaptation of the "Scopewriter" featured in the August 1974 issue of Popular Electronics.

Photo 1 is self-explanatory. An ASCII keyboard was the input device for this line write/ edit program. Photo 2 is the word guessing game "Hangman" in progress. The secret word is "COSMAC," the player has de-

pressed " $G$," which is incorrect, and another letter has been added to HANGMAN. Photo 3 illustrates what I call a memory-map program. The eight bytes represent what is in memory, starting at the displayed address. This program allows inspection and correction of any part of memory. --Lynn R. Clock, Lompoc, CA.

## RADAR DETECTOR ADVOCATE

I was quite distressed to read your Editorial comments in the November 1977 issue of Popular Electronics. There is one important point that you didn't mention, legally referred to as probable cause. Your home cannot be searched or your telephone tapped without a search warrant. A warrant is issued only if the police can show probable cause that a crime has been committed. When a "Smokey" is beaming his MR7 moving radar at you on the road with it on all the time and does not wait for probable cause, he technically violates your right to privacy.

As used by most states, radar is for revenue purposes. The $55-\mathrm{mph}$ law has not been effective in saving lives or resources, a fact documented by the Comptroller of the United States and various state agencies. If our police were instructed to go hard on drunk drivers who account for $40 \%$ of all fatal accidents, reckless drivers, and "junk" cars, the roads would be safer, especially with a return to previous speed limits. --Bence D. Boelcskevy, Vice President, BMW Car Club of America, Inc., Cambridge, MA

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## New Products

Additional information on new products covered in this section is available from the manufacturers. Either circle the item's code number on the Reader Service Card inside the back cover or write to the manufacturer at the address given.

## Aiwa Stereo Cassette Deck

Aiwa has developed a mechanism that measures the reel rotation speed of the cassette and displays the remaining time for the tape. This device is incorporated into the company's Model AD-6550 stereo

cassette deck with built-in Dolby noisereduction system. The time is shown in minutes to the left of the meters. The deck's ratings include a wow and flutter of $0.05 \%$ weighted rms and a $\mathrm{S} / \mathrm{N}$ ratio of 65 dB. Its specified frequency responses are: 30 to $13,000 \mathrm{~Hz}$ for LH and 30 to 15,000 Hz for $\mathrm{CrO}_{2}$ and FeCr tape. A bias fine adjustor is provided for $\mathrm{LH} /$ normal tape. A two-step peak-indicating LED system is designed to activate at +3 and +7 dB . Additionally, there is a memory rewind system. $\$ 450$.

CiRCLE NO 92 ON fREE Information caro

## Avanti CB BaseStation Antenna

The Avanti Saturn CB base-station antenna is a coinductive system that combines four antennas that operate on two polarities in what is claimed to be the most unique antenna yet devised. The vertical radiator section is $22^{\prime}(6.7 \mathrm{~m})$ in height, giving it the desirable radiation of a 5/4wave design. While using vertical polarity, the horizontal section automatically serves as a ground plane for the radiator. The horizontal section consists of three half-wave dipoles that are electronically co-phased to

yield an omni-directional radiation pattern. The feed system is arranged so that equal power is applied to each dipole, via separate multipurpose balun coils. The method by which the power is equally divided and transferred to the dipole radiators is the heart of the Saturn's design.
ctrcle no 93 on free information card

## Ballantine $80-\mathrm{MHz}$ Frequency Counter

Ballantine's Model 5720A frequency counter covers a direct-count range of 10 Hz to more than 80 MHz . It features an audio tone multiplier circuit that provides a claimed resolution of 0.01 Hz in only a 1 second measurement time. Frequency and ratio measurements are read out from an eight-digit jumbo LED display. Direct readings are in megahertz, kilohertz, or

hertz as selected by a front-panel switch. The POWER and range switches are the only controls required for operation of the counter. The instrument's 1-megohm input impedance is shunted by 25 pF of capacitance. The input is fully protected against over-voltage to 250 volts rms from 10 to 1000 Hz , decreasing to 10 volts rms beyond 10 MHz . Sensitivity over the entire range is specified at $50 \mathrm{mV} . \$ 195$.

CIRCLE NO 94 ON FREE INFORMATION CARD

## Dual Direct-Drive Turntable

Dual's Model 604 semiautomatic turntable is direct-driven by a dc electronic motor whose speed is electronically regulated by a digital reference circuit. The tonearm is mounted in a four-point gyroscopic-type
gimbal suspension. Two mechanical antiresonance filters are housed in the counterbalance to attenuate acoustic feedback and parasitic resonances. The straight-line tubular tonearm offers what is claimed to be maximum rigidity with minimum mass. Other features include: mechanical sensor to locate lead-in grooves on $12^{\prime \prime}$ and $7^{\prime \prime}$ ( 30.5 and 17.8 cm ) discs; damped cueing system; automatic tonearm return and

shutoff at end of play; $10 \%$ pitch control and illuminated strobe; and die-cast platter. Rumble is rated at greater than 70 dB and wow and flutter at less than 0.03\%. Supplied with low-profile base and dust cover. $\$ 250$.

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## Philips Portable Oscilloscope

A portable $15-\mathrm{MHz} / 2-\mathrm{mV}$ oscilloscope, designated the Model PM 3211, is available from Philips Test \& Measuring Instruments, Inc. The scope features comprehensive display and triggering facilities and a double-insulated power supply that requires no grounding to eliminate hum and spurious signals. Triggering can be in auto or level set and multi-sourced, eliminating the need to change probes. Channel $B$ can be used as an $X$ input to facilitate

$X-Y$ displays with calibrated attenuation of both $X$ and $Y$ inputs. It can also be inverted and, with the scope's ADD function, can display $A+B$. An 18-speed timebase has a vernier control for simplified phase and timing measurements. $\$ 875$, including two probes

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## Genesis <br> Speaker System

The Model 3 speaker system from Genesis Physics Corp. employs three specialized drivers, each of which is claimed to be engineered for extremely long linear excursion relative to cone size. This approach is said to provide the large power handling capability of larger-diameter drivers with the sonic superiority of small, lightweight drivers with respect to transient response, suppression of resonance distortion and

coloration, and full frequency dispersion. A passive radiator extends the system's low-frequency limits. The midrange driver is housed in an acoustically isolated enclosure, and the tweeter is damped by a vis cous magnetic ferrofluid. A specially designed crossover network is said to preserve a high degree of phase integrity. $\$ 299.50$.

CIRCLE NO 97 ON FREE INFORMATION CARD

## SJE Morse Code Typewriter


S.J. Engineering's Model SMCT simulated Morse-code typewriter lets an aspiring ham radio operator learn and send Morse code simply by touching a built-in probe to selected letters and numerals. Once a
character is touched with the probe, the typewriter self-produces and completes the equivalent Morse-code output for that character. Speed is adjustable from 5 to 60 wpm. Also adjustable are the volume level and tone. The typewriter features two outputs: audio for learn/practice and reed relay for making on-the-air contacts. Power for the CMOS circuitry is from two 9 -volt batteries (not supplied). \$99.95, wired and tested; \$69.95, kit

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## Heath Electronic Cruise Control



According to the Heath Company, its new Model CS-1048 electronic cruise control

can provide increased fuel mileage and promote driving safety by reducing fatigue on long trips. The device is completely electronically controlled. Accuracy is reported to be within 2 mph of the preset speed regardless of load and other variables. A "resume" memory holds the preset speed. In addition, the cruise control is said to be capable of operation with both manual and automatic transmissions. It can also be removed from one vehicle and reinstalled in another vehicle with a minimum of changes. The device comes as-


[^0]sembled, ready for installation in vehicles having open driveshafts, as most cars do. Catalog price is $\$ 79.95$.

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## Pioneer AM/Stereo FM Receiver

The Model SX-1280 AM/sterec FM receiver from Pioneer is rated at 185 watts/channel minimum into 8 ohms from 20 to

Does the equalizer you're considering offer full ten band control with symmetrical "mirrorimage" boost and cut responses centered on ISO preferred octave bands? Does it have permanently-lubricated 60 mm metal-cased sliders with metal shafts and center click detents? Do the specs tell you what to expect at all settings . . . or only at the "flat" setting. where the critical tuned networks are bypassed? Does it employ advanced hum $\mathcal{E}$ saturation-free "gyrator" simulated inductors on all low and mid-frequency bands? How about truly differential balanced and unbalanced inputs and outputs for use in any audio system, amateur or pro? And "fit anywhere" packaging designed for 19" rack, in-wall, inconsole or optional wood cabinet mounting? What about truly flat response (both amplitude and phase angle) at the center reference setting?
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After careful inspection of the kit(s), you may return any or all items in their original unassembled condition for a full immediate refund if you are not totally satisfied ... (no questions asked). And, if you decide to keep and build the kit(s), our normal guarantee on the specs and parts still applies . . . if your properly assembled kit(s) fail to operate as stated, we will exchange any defective parts free for the first 90 days.

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$20,000 \mathrm{~Hz}$ at no more than $0.03 \%$ THD. Its power amplifier section employs a directcoupled OCL design, while separate power supplies are used for the two channels The AM/FM tuner uses a MOSFET in the $r-f$ mixing stage and a JFET in the mixing buffer stage. FM sensitivity is 9.8 dBf ; capture ratio, 1.0 dB . Two of the four meters in the receiver are used for indicating power, separately for each channel. They respond to output powers from 0.01 to 370 watts without requiring sensitivity switching. The volume control is a 32 -step attenuator type. Built-in are high and low filters, audible multipath switch, $25-\mu \mathrm{S}$ FM deemphasis switch. stereo adapter in/out terminals and switch. $\$ 900$

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## CPI AM/SSB CB Base Station

The Ultra CP-2000B "limited-edition" CB base station from Communications Power Inc., employs a modulator that is said to provide an unusually high level of modulation without exceeding FCC specifications. Spurious response is rated at -70 dB . The transmitter is said to have an infinite SWR mismatch tolerance. Specifications for the single-conversion receiver include -70 dB i-f rejection. -80 dB adjacent-channel rejection, and -85 dB intermodulation characteristics. Built in are a switchable speech

compressor, switchable anl/noise blanker, $r$ - $f$ gain control with automatic override, and microphone gain control. Separate r-f output power/S and SWR calibration meters are provided. An optional Model BC-2000 base-station console provides: 7 digit frequency counter; 6-digit 12/24-hour clock; FET receive preamplifier; dual antenna tuner; dual antenna switch; and -50-dB TVI filter. \$600.

[^1]
# MORE base justa 



Realistic ${ }^{( }$invents the base CB that's also a control center. Perfect for home, office, small warehouse or dispatcher-type operations. The push-to-talk phone-type handset is great for private listening or noisy areas. Add a remote speaker - the CB-Fone becomes a paging/PA amplifier that lets you monitor incoming CB calls simultaneously.
Emergency? Push the red priority button and you're instantly on Channel 9. Channel selection is electronic. Two buttons - count-up and count-down - replace rotary switches. Push once to change channels, hold down for rapid movement through all 40 channels. Big LED readout displays the channel number and indicates "PA" in public address mode. Flip-switches for delta-tune, blanker, handset or handset-plus-console speakers, mode selection. RF gain control. Lighted SWR and S/RF meters. On-the-air and modulation lights. AC and 12VDC versatility, positive or negative ground. U.L. listed. The CB-Fone - just 249.95* buys you a station that's something else!

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## New Literature

## TELEX CB MIKE BOOKLET

"The CB Power Mike Fact Book" explains and discusses basic points and facts regarding CB power microphones. One important feature discussed is how a CB noise-cancelling mike keeps unwanted sound out. Other basic points covered include "Why the Microphone is a Most Important Part of Your Transmitter"; "Will a Power Mike Work With My 40 Channel Radio?"; and "The Special Advantages of CB Power Mike Headsets." The booklet, with drawings and photos, may be obtained by writing to Telex Communications, Inc., 9600 Aldrich Avenue South, Minneapolis, MN 55420

## JENSEN TOOL CATALOG

A new 144-page catalog of tools for electronic and mechanical assembly is offered by Jensen Tools and Alloys. It includes over 3,000 tools in categories such as micro-tools,
test equipment, soldering equipment, tweezers, screwdrivers, cutters, drafting supplies, and power tools. Another section features tool kits and cases. Address: Jensen Tools and Alloys, 1230 South Priest Drive, Tempe, AZ 85281.

## SIGNAL TRANSFORMER CATALOG

"Transformers," a 16-page publication from Signal Transformer Co., Inc., combines a short-form catalog and an application guide. It presents specifications and mechanical data on the company's line of transformers. Schematics are included. Among the components featured are the conventional power transformers, filter chokes, rectifier, high-current, step-down auto, step-up or step-down power-isolation transformers and low-cost Split/TranR and Flathead ${ }^{R}$ printed-circuit board transformers. Address: Signal Transformer Co., Inc., 500 Bayview Ave, Inwood, NY 11696.

## MOTOROLA TWO-WAY RADIO TEST EQUIPMENT CATALOG

Thirty-one additions to Motorola's line of twoway radio test equipment appear in the company's new catalog. Featured are the models R-1200A service monitor, $R-1010 A$ signal generator, and S-1338A FM station monitor. Other test instruments for the maintenance of FM two-way radio communications equip-

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ment are included. Address: Motorola Literature Distribution Center, 1301 E. Algonquin Road, Schaumburg, IL 60196.

## NATIONAL SEMICONDUCTOR LED CATALOG

National Semiconductor has prepared a catalog describing its line of opto-electronic products. The catalog contains photographs, outline drawings, and specifications of red, yellow and green light emitting diode (LED) lamps, multi-digit numeric displays, small cal-culator-type numeric arrays and watch display die. An application section is included on mounting techniques for numeric displays. The catalog also contains a list of National's LED segment drivers and digit drivers, with specifications and ratings. Address: National Semiconductor Corp. 2900 Semiconductor Drive, Santa Clara, CA 95051.

## MALLORY GENERAL CATALOG

A new general catalog lists over 8,000 electronic components and related products made by P.R. Mallory \& Co., Inc. Information on lines of capacitors, controls, switches, semiconductors, audible signal devices, security products, and cassette recording tape is provided. Address: Mallory Distributor Products Co., Box 1284, Indianapolis, IN 46206.

## OEI COMPONENT CATALOG

The 1978 catalog of Optical Electronics, Inc., contains 41 data sheets on 18 operational amplifiers, 22 fast analog function modules, and a microcomputer. A selection guide indexes op amp modules by slewing rate, by gain-bandwidth at $\times 100$, and by settling time to $0.1 \%$. The selection guide for function modules is arranged by model number and function: logarithmic amplifiers, sample and hold, peak sense and hold, V-F-V, D-A-D, power supplies and others. Address: Optical Electronics, Inc., Box 11140, Tucson, AR 85734

## AUDIO-TECHNICA MICROPHONE GUIDE

Available from Audio-Technica, "A Brief Guide to Microphones," is an instructional booklet applicable to all brands of microphones. The 15 -page publication explains microphones through 8 basic terms: dynamic, condenser, omnidirectional, unidirectional (or cardioid), proximity effect, feedback, impedance and sensitivity. Address: AudioTechnica U.S., Inc., 33 Shiawassee Ave., Fairlawn, OH 44313

## INTERFERENCE INFORMATION

Interference and power-line surge damage common to audio equipment and FM receivers are described in a new flyer from Electronic Specialists. Problems such as lightning protection, ac power-line hash, and loudspeaker interference pickup are included. Cures for these problems are offered. Address: Electronic Specialists, Box 122, Natick, MA 01760

# Understanding Digital Electronics New teach-yourself courses 



Design of Digital Systems is written for the engineer seeking to learn more about digital electronics. Its six volumes - each $11-1 / 2^{\prime \prime} \times$ 8 -1/4" are packed with information, diagrams and questions designed to lead you step-by-step through number systems and Boolean algebra to memories, counters and simple arithmetic circuits, and finally to a complete understanding of the design and operation of calculators and computers.

## The contents of Design of Digital Systems include:

Book 1 Octal, hexadecimal and binary number systems; conversion between number systems; representation of negative numbers; complementary systems; binary multiplication and division.
Book 2 OR and AND functions; logic gates; NOT, exclusive-OR, NAND, NOR and exclusive-NOR functions; multiple input gates; truth tables; De Morgans Laws; canonical forms; logic conventions; Karnaugh mapping; three-state and wired logic.
Book 3 Half adders and full adders; subtractors; serial and parallel adders; processors and arithmetic logic units (ALUs); multiplication and division systems.
Book 4 Flip flops; shift registers; asynchronous and synchronous counters; ring, Johnson and exclusive-OR feedback counters; random access memories (RAMs) and read only memories (ROMs).
Book 5 Structure of calculators; keyboard encoding; decoding display data; register systems; control unit; program ROM; address decoding; instruction sets; instruction decoding; control program structure.
Book 6 Central processing unit (CPU); memory organization; character representation; program storage; address modes; input / output systems; program interrupts; interrupt priorities; programming; assemblers; computers; executive programs; operating systems and time sharing.


Digital Computer Logic and Electronics is designed for the beginner. No mathematical knowledge other than simple arithmetic is assumed, though the student should have an aptitude for logical thought. It consists of four volumes - each 11-1/2" $\times 8-1 / 4^{\prime \prime}-$ and serves as an introduction to the subject of digital electronics. Everyone can learn from it - designer, executive, scientist, student, engineer.

Contents include: Binary, octal and decimal number systems; conversion between number systems; AND, OR, NOR and NAND gates and inverters; Boolean algebra and truth tables; De Morgans Laws; design of logic circuits using NOR gates; R-S and J-K flip flops; binary counters, shift registers and half adders.

In the years ahead the products of digital electronics technology will play an important part in your life. Calculators and digital watches are already commonplace. Tomorrow a digital display could show your automobile speed and gas consumption; you could be calling people by entering their name into a telephone which would automatically look up their number and dial it for you.

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## WITH THE HIGH ROLLERS IN VEGAS

FOR SOME TIME, the Winter Consumer Electronics Show (WCES) has been overshadowed by its much larger summertime brother. If this past January's installment was any different, it is probably because a change of venue from frigid Chicago to the comparatively benign climate of Las Vegas piqued enthusiasm, curiosity, and deeply harbored desires for a brief vacation after a grueling Christmas. It did not seem like much of a vacation from this side of the footlights, however

The new WCES (it will return to Las Vegas next year) proved considerably enlarged in attendance and square yardage, particularly since it sprawled well beyond the official Convention Center headquarters and into a host of neighboring hotels and motels in true summer CES fashion. This raised certain tactical problems. For security reasons, the hotels refused to divulge room numbers, so that my search for some exhibitors simply had to be abandoned for lack of sufficient time or a competent wilderness guide. Next year I shall be more familiar with the territory.

As usual, the show offered no challenge to its summer counterpart in sheer numbers of new audio product introductions, but several of the debuts that did take place were provocative. A sampling follows.

Musclemen. Thanks to Pioneer's new SX-1980, receivers are now up to 270 watts per channel and very probably still climbing. Happily, the newer super receivers are not growing much in size over the largest models of yesteryear. Instead, the problems involved with cooling units of this capacity are being solved on the engineering level. Pioneer's solution is a power transistor with a new body style that makes more surface area available to contact the heat sink. Combined with an internal "chimney" that encourages efficient flowthrough of air, the transistors evidently can throw off heat rapidly enough to ob-
viate the need for forced-air cooling- an impressive achievement at a time when pundits have only half-humorously predicted that the helium-cooled receiver is just around the corner. Clearly, the su-per-power receiver is well on its way to becoming commonplace. Even Sanyo, a company heretofore associated with middle-fi equpiment and clock radios, has broken new ground with the 120 -watt-per-channel JCX2900K.
In another accelerating trend, receivers, having been threatened somewhat by the growing popularity of separates, are now in the process of becoming separates themselves. Rotel introduced a receiver with a "docking" power-amplifier section a year or so ago, and Mitsubishi applied the same concept to an integrated amplifier. At WCES, Sansui
joined in with the G22000, the top model of the $G$ Series at 220 watts per channel. The forerunners of the G Series appeared at last June's show, and this new addition continues the interesting styling that distinguishes the line
Many of the new amplifier introductions clustered around the mid-power level, an example being the Marantz 300DC power amplifier at 150 watts per channel. The Marantz unit is also indicative of a growing concern about slew rate and transient intermodulation distortion; the manufacturer lays heavy emphasis on the efforts made to control these performance parameters. Besides being augmented by a new 500 -watt-per channel (!) stereo power amplifier (the Dual 500) and an analog time-delay/ reverberation unit (Model 6000), the Phase Linear line has been completely restyled. The restyling included circuit modifications to virtually all the existing Phase Linear models, with FET inputs being miuch in evidence

Hitachi has employed MOSFET output devices in two new power amplifiers, the HMA 7500 and HMA 9500 , with 75 and 100 watts per channel, respectively. The $\$ 500$ price of the HMA 7500 would appear to be a bargain for FET amplifiers. The company also introduced a matching preamplifier, two new three-
(Continued on page 20)



# The1980 Kenwoods. 

No. We're not kidding. By 1980, the kind of performance these new Kenwoods deliver will be considered commonplace. Here's a summary:

1. The KA-7100 is an integrated DC amplifier with dual power supplies delivering 60 watts per channel, minimum RMS at 8 ohms from $20-20 \mathrm{kHz}$, with no more than $0.02 \%$ total harmonic distortion. Not only is that the lowest THD of any integrated amp, the KA-7100 is the lowest priced DC integrated amp on the market. ( $\$ 300$ *)
2. The KT-7500 marks the next plateau for FM tuners. For optimum reception under any condition it has two independent IF bands: the narrow band virtually eliminating interference when stations are close together, the wide band for lower distortion and maximizing stereo separation. In addition, we've developed new circuitry which eliminates the high
frequency beat distortion (that is, swishing noises) thought to be inherent in stereo FM broadcast. Even we're impressed that it costs only $\$ 275$ *

This combination of separate amp and tuner not only gives you performance unheard of in other separate components, it gives you performance that will remain elusive in receivers for quite a while. The Kenwood KA-7100 and KT-7500. Solid evidence that the breakthroughs occurred ahead of schedule, and available to you now for a truly remarkable price $\$ 575^{*}$. for the pair.
*Nationally advertised value Actual prices are established by Kenwood dealers Handles optional

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head cassette decks, and a novel directdrive turntable. Sony's pulse-width modulation amplifier has now been designated the TA-N88 and is rated at 160 watts per channel. And among SAE's new introductions is a combined integrated amplifier and parametric equalizer, the Model 2922.

Loudspeakers. Helium-cooled amplifiers we may not have, but heliumcooled speakers may finally be here. A company called Plasmatronics exhibited a mid- and high-frequency driver consisting of a small quartz cell with a small square opening in its frontal surface. A continuous stream of helium is bled into the cell in the presence of a high polarizing voltage. An eery violet glow with tinges of green and yellow promptly appears; and when an audio signal is applied (from an internal vacuum-tube amplifier), what comes out is sound.

The designer, Alan Hill, is not yet able to be totally specific about the operating principle (patent considerations are involved, I imagine); but this is clearly an ionized-gas loudspeaker with an unusually wide frequency response (down to 700 Hz ) and dynamic range. Speculation about the purpose of the helium continues. Some believe that it assists in high-frequency propagation (ever heard a man speaking in a helium-oxygen atmosphere?) while others suspect it is merely a coolant (the cell radiates an impressive amount of heat). In any case, a tank of the gas lasts an estimated 300 hours, after which you must return for a refill. The Plasmatronics speaker has a conventional dynamic woofer and some very impressive frequency-response and impulse-response specifications. The associated preamplifier offered by the company is highly recommended; it has a switch to halt the helium flow.

Acoustic Research has a new top-of-the-line model, the AR9, standing over 4 feet tall and equipped with dual side-firing woofers ( 12 inches each) placed for optimum acoustic coupling with the nearest room boundaries. It is a fourway design with what AR says is the flattest response and highest power-handling capability it has ever achieved.

Infinity's latest speaker system, the Quantum Reference Standard, is likewise a top-of-the-line model, offered on a custom-order basis at a projected price of $\$ 6500$ the pair. They make use of vertical arrays of the EMIT film-diaphragm tweeters, to which the manufacturer is now fully committed; and they have a multi-faceted cabinet that is cer-
tainly one of the largest ever introduced to the consumer market.

JBL's new products consisted of the L50, the lowest-price three-way design in the JBL catalog, while KLH launched several additions to its Baron series. In addition, ESS presented quite a few modifications to many of its existing products and unveiled a productionready version of the Transar loudspeaker with the unique Heil woofer.

And Elsewhere. Kenwood's new top-of-the-line turntable, the KD-750, follows a number of acclaimed successes in turntable design by that manufacturer, and also incorporates a 20 -pole dc motor. (Multi-element motors are becoming increasingly favored by designers of di-rect-drive systems as the fight to eliminate any trace of cogging continues.) The Harman-Kardon/Rabco ST-8 is a thoughtful redesign of the ST-7, using the same radial tracking principle but benefitting from various refinements. AR has a rather special turntable in the works as well, but in the meantime the company offers the AR77-XB, the latest evolution of its classic AR turntable.
Listeners to older $78-\mathrm{rpm}$ recordswhich in many cases were not quite 78 rpm-have long bemoaned the lack of turntables that are continuously variable in speed. They will be glad to learn that the elusive Lenco record players are now back and firmly entrenched with Neosonic Corporation of America acting as importer and distributor. The three Lenco variable-speed models operate at anything from 30 to 86 rpm , and can be bought with and without bases.
Turntables, tuners, and cassette decks are the areas in which automation and computer techniques have found their happiest consumer application so far, and this tradition continues with a new cassette deck from Sharp/ Optonica. Incorporating all of the pro-gram-search features of previous models from this manufacturer, the new machine also has a liquid-crystal display that indicates the operating function and the time of day, as determined by a 24hour quartz-crystal clock with a sevensegment digital readout. Perhaps needless to say, the clock also functions as a timer. Therefore, the deck can be set to record any program at any time and then turn itself off, all without any operator's attention.
Another interesting cassette development is the re-emergence, of the cassette changer/player. Lenco's RAC-10 machine, capable of handling up to ten
cassettes at a time, has returned to the U.S. It is not a device for everyone, but it appears to be the single survivor of a possible trend that showed promise some years ago.

Accessories is a product category that has grown from an afterthought to a major center of consumer attention, and few accessories were as popular at this show as racks-equipment-holding racks, that is. Kenwood, Sansui, Mitsubishi, and Harman-Kardon are a few of the major companies that have followed the lead of Pioneer and Nakamichi in introducing vertically tiered equipment cabinets, some with casters, that will readily accept the 19 -inch front panels of much current equipment. Whether they will accept the varying depths and heights of different manufacturers' amplifiers, preamplifiers, and tuners is still an unanswered question, as are such matters as ventilation and provision for interconnecting cables. Optonica showed a variety of other types of cabinets to house hi-fi components. They feature push-to-open glass doors for the LP record storing section.

In the meantime, record-care accessories are proliferating beyond any hope of keeping track of all of them; even tape companies like Memorex have devised extensive lines of stylus-force gauges, record-tracking dust removers, and hand-held record decontaminators. So far, none of these new products conspicuously demonstrates any new thinking about the problem of dirt on record surfaces, but neither does any of them seem ill-suited to its purpose.

Many are calling the new ADS 10 Acoustic Dimension Synthesizer the best of the available reverberation devices for the home user. It is an entirely digital processor with a frequency response (at -20 dB ) of 30 to $13,000 \mathrm{~Hz}$ and a continuously variable decay time of up to 1.6 seconds. ADS has built in a 100-watt-per-channel stereo amplifier, and includes a pair of L10 speaker syslems for rear-channel use with the package. The operation of the A-to-D/D-to-A converter is still classified, but it was revealed that a special circuit is built-in to limit the reverberation enhancement afforded to mono signals when desired. Consequently, FM programs can be listened to without the announcer sounding as if he were speaking from a cathedral floor.

All of these products should be getting some attention in these pages in months to come. And from first appearances, they will deserve it.

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SAE has long been involved in the field of tone equalization. From cur pioneering efforts in variable turn over tone controls to our more recent advancements in graphic equalizers, we have continually searched for and developed more fexible and responsive tone networks. From these efforts comes a new powerful tool in tone equalization the Parametric Equalizer. Now you have the power of precise cont=ol.
Our 2800 Dual Four-Band and 1800 Dual Two-Band Parametrics offer you controls that not only cut and boost, but also vary the bandwidh and tune the center frequency of any segment of the audio range.

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# Julian Hirsch 

THE POWER RACE-WHEN WILL IT END?

IT IS EASY to be carried away by the ever-increasing power ratings of today's receivers and amplifiers. Not long ago, a basic power amplifier that could deliver 100 watts/channel rated the name "super power." (Such power levels were unheard of in receivers.) Then amplifiers began to have ratings of 200 to 250 watts, and receivers crept up to $120,150,160$, and 180 watts. Last year, the leader in the receiver power race was a giant from Marantz, rated at 250 watts/channel, but a 270 -watt receiver from Pioneer now threatens its position. In basic power amplifiers, a 500-watt/channel Rotel amplifier currently leads the race, but similarly rated amplifiers from Phase Linear and possibly others are in the offing. Apparently the end of the race is not yet in sight.

The justification for this race to greater and greater power is multifaceted. Partly, it is merely because technology now makes it possible, at reasonable cost, to generate such huge power outputs. Some of today's power transistors can dissipate hundreds of watts per device, yet are priced low enough for use in consumer products. Manufacturers can hardly be blamed for attempting to convert some of this new technology into saleable hardware. And the prices of some of these products are not unreasonable in view of their performance, though some may find the size and weight excessive for their homes.

Another side to the question of "how much power" is one of need, or at least, utility. No one will claim that a power level of hundreds of watts per channel is always necessary for good reproduction. However, there are convincing arguments that prodigious power can be used to advantage under some circumstances. If it can be utilized and if it can be generated without prohibitive expense, why not have it available in our deluxe amplifiers and receivers?

One of the two questions most often posed to me is what amount of power is needed for a good hi-fi system? Even after I narrow down the scope of the question by establishing what type of speaker systems will be used in what size room, etc., the answer is still necessarily vague and unsatisfying. What it boils down to is that you need as much power as is necessary to play music as loud as you like to hear it. While this is the literal truth, it is of little help in selecting system components. Under most home conditions, reasonably high acoustic levels can be generated
with only 1 or 2 watts of amplifier power. It is, therefore, difficult for a layman, or even an experienced audiophile, to understand how a reserve of hundreds of watts can be beneficial or justifiable.
The explanation for this "need" for high power lies in two different, though related, aspects of sound. Most natural sounds, either music or speech, are nonsinusoidal. They consist of a multitude of frequencies whose combined waveform usually has a high crest factor. The maximum instantaneous peak amplitude is many times the rms value of the waveform, computed through a complete cycle of its fundamental frequency. (In contrast, the instantaneous peak value of a sine wave is only 1.41 times its rms value.) The perceived loudness of a sound is roughly related to its average or rms value, the two usually being of the same order of magnitude. Brief peaks that exceed the rms level by a factor of 10 or 100 times are not heard as being louder. The effect is familiar to everyone who has heard the greatly increased apparent loudness of TV or radio commercials, as compared to the average program levels. The peak modulation levels are unchanged (any transmitted increase would result in illegal overmodulation), but the average level of the commercial is raised by compression so that it is heard at a much higher volume.

The second factor is the intermittent nature of music and speech. Regardless of the wave shape of the sound, it is characterized by large variations in level during the program. Musical passages may be very soft, or there may be periods of complete silence, followed by periods of very high program levels. The average level, over the entire duration of the program, is normally very much less than the maximum level attained during that interval. Thus, to reproduce either music or speech in a natural manner, the amplifier must have the capability of delivering peak outputs many times the average power.
The amount of reserve "headroom" needed in an amplifier varies widely with the program material, since all recorded and broadcast programs are limited to keep their peak levels within the limits set by the transmission medium. A headroom of only 10 dB may be adequate in many cases, but for most realistic reproduction, 20 dB is preferable.
The $20-\mathrm{dB}$ headroom is a power ratio of 100 times. If the average power during a musical passage is

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about 1 watt (it could range from a few milliwatts to several watts), the amplifier will be called upon to handle an occasional peak of as much as 100 watts. "Occasional" here means as infrequently as a few milliseconds out of a half hour of playing, or perhaps as often as $10 \%$ of the time, but more likely the former. If the amplifier cannot deliver those peaks, the result will be clipping, which may be audible.

Most of us manage to enjoy our music systems without undue distress from amplifier peak clipping, even with considerably less than 100 watts of power available to us. Recorded programs rarely have more than a $10-\mathrm{dB}$ peak-to-average ratio, since they are designed for playback on mass-produced record players with very limited power capabilities. Some speaker systems are so efficient that they require only a tenth as much power as others for the same volume level.
With the foregoing in mind, why do we need "su-per-power" amplifiers, and what possible justification can there be for receivers or amplifiers that can deliver hundreds of watts per channel? The answer, or part of it, lies in the logarithmic nature of human hearing. To make a program sound twice as loud, we must increase the power ten-fold. Another doubling of apparent loudness, and we require 100 times the original average power level (and as much as 100 times that level to handle the peaks).
It might appear that there is no practical solution to the power "shortage." If our hypothetical 1-watt average program, heard at a comfortable level (and requiring from 10 to 100 watts of amplifier power in reserve just to avoid peak clipping) is to be played "just a little bit louder," we can easily find ourselves running out of power, even with a $200-$ or 300 -watt amplifier. This is not as far-fetched as it may seem. I find it very easy to clip the outputs of an amplifier rated at 200 watts/channel or more, at levels far below those that exist at a discotheque or in the vicinity of a small instrumental combo when using speakers of average efficiency.

Not surprisingly, the equipment manufacturers'
answer to the clipping problem is to make more powerful amplifiers. Unfortunately, this is no answer at all, since a 500 -watt amplifier will play only a barely detectable 3 dB louder than a 250 -watt amplifier. It is also much more likely to blow out one's speaker systems if used carelessly. Speaker systems can not handle unlimited power.

The reason for public acceptance of less-than-ideal program dynamic range in home music reproduction, in my opinion, lies in the ready availability of a knob, usually labelled volume or loudness, and found on every receiver and amplifier. We turn this knob clockwise to increase the loudness. Perhaps, if we are lucky, the program reaches what we consider a natural level and is free of unpleasant sounds. Well and good, since this is what hi-fi is all about. Suppose that before this level is reached, however, we hear obvious and unpleasant distortion. Rather than try for higher levels, most people react by turning down the volume until the distortion becomes acceptable.

In turning down the volume, we have adapted the reproduced program level to the limitations of our equipment, and are willing to accept a less-thannatural listening level. In a very short time, most of us no longer find such levels unnatural, which is very fortunate, since it is manifestly impractical to generate concert hall levels in one's home unless the neighbors are distant or deaf or both.
The foregoing is why hundreds of thousands of people are very satisfied with 20 -watt receivers, thousands more find the 60 to 80 watts of a good middlepriced receiver more than satisfactory, and only a few can justify the considerable expense of the superpower variety. Personally, I appreciate having 200 watts or so on tap when needed, but recognize that the listening benefits of even an increase to 300 or more watts will often be judged too subtle for the added cost. Of course, that will not stop manufacturers from creating such amplifiers or receivers, or dedicated audiophiles from buying them. I can only say, more power to them!
$\longrightarrow$


## HARMAN-KARDON MODEL 730 AM/STEREO FM RECEIVER

Medium-power receiver features unusual tuning meter.


Transient performance has long been an important part of HarmanKardon's design approach to its high-fidelity products. To this end, the company has espoused amplifiers with very wide-band, low phase-shift circuits, and, as in the latest crop of H-K receivers, separate power supplies for the two channels. The Model 730 AM/stereo FM receiver that heads the current H-K re-
ceiver line embodies these design aims, although its 45 -watt output power is barely moderate by present standards.

The styling of the Model 730 is distinctive, with the front panel presenting an uncluttered, utilitarian appearance. The most unusual external feature of the receiver is its tuning meter that, on FM, is actually a quieting indicator. It indicates the reverse of the usual signal-strength meter, with its pointer deflecting downscale as signal-strength increases and noise level decreases. In effect, the me-
fer indicates $S / N$ in a qualitative, rather than a quantitative, manner. When the pointer enters a small circle at the lower limit of its travel, the receiver is fully quieted. For AM reception, the meter's pointer swings up-scale with increasing signal strength, as is usually the case with most of the conventional signalstrength meters.

Dimensions of the Model 730 are $51 / 2^{\prime \prime}$ $H \times 17^{\prime \prime} W \times 141^{\prime \prime} D(14 \times 43 \times 37$ cm ). Suggested manufacturer's price for the Model 730 is $\$ 399.95$.


Inset shows FM quieting indicator also used for AM signal strength.

## Product Focus

Traditionally, the tuning meters in FM tuners and receivers indicated re-ceived-signal strength and/or centerchannel tuning. Harman-Kardon has made a change from tradition by having its Model 730 stereo receiver's meter indicate relative signal quality so that, when nulled, tuning is as near perfect as possible for maximum S/N.

The signal-quality meter system employs a discrete transistor amplifier and two tuned resonant circuits. Each resonant circuit is tuned to about 100 kHz . The signal that drives this filter/ amplifier combination is derived from an amplifier stage that follows the tuner's discriminator.

The amplifier/fitter combination amplifies the noise that appears at 100 kHz at the discriminator's output. The $100-\mathrm{kHz}$ frequency was chosen to minimize the effects of audio modulation on the meter's reading. (It prevents the meter pointer wobble common to multipath meters.)

Since it is tuned to 100 kHz and "looks" only at that frequency, the metering system is quite selective and is thus able to give a pretty good indication of S/N. (If the frequency selected were more than 100 kHz , the discriminator would roll off the response.)

The system is not a true signal minus noise meter. Rather, it indicates noise itself. The circuit employs an agc system to prevent the meter from becoming crowded at one end, which linearizes the meter scale.

General Description. During the past year or so, we have seen several amplifiers and receivers with separate power supplies for the two audio channels, some of which utilized a single common power transformer and others with separate transformers. Doing this is supposed to eliminate interaction between channels that results when a common power supply impedance is used, especially at very low frequencies. The benefits of such isolation are difficult to demonstrate by objective or subjective means. Judging from the size of the power transformers and filter capacitors used in the Model 730, there is probably little difference in material cost between the Harman-Kardon approach and simply making a single, heavy duty supply with sufficient output capacitance to provide adequate isolation. Of course, with separate power supplies there is absolutely no difference in the performance of one channel whether or not the other channel is driven. The FTCmandated preconditioning period of one hour at one third power, with both chan-
nels driven, is still necessary to bring the amplifier to a fully heated condition.

Harman-Kardon's wide-band philosophy is illustrated principally in the low-frequency response of the Model 730 , which extends to 4 Hz and below. Clearly, most of the amplifier stages are direct-coupled. In a brochure furnished with the receiver, a $20-\mathrm{Hz}$ square-wave response with only a $3 \%$ tilt is displayed to emphasize the receiver's extended low-frequency response.

On the rear apron of the receiver, there are pushbutton reset circuit breakers that protect the speaker outputs and a control shaft that permits adjustment of the FM muting threshold over a wide range. The ventilating slots at the rear of the top cover run across the width of the receiver. The output transistors and their heat sinks extend from front to rear, at the center of the receiver's chassis. Thus, the ventilating holes have very little direct cooling effect. Even so, despite the relatively large size of the receiver (for its power rating), it runs notably cool at all times


Noise and sensitivity curves for $F M$ section of receiver.

# WHAT THE EXPERTS CALLED THE BEST LASTYEAR WASN'TGOOD ENOUGH FOR US. "IT CANNOT BE FAULTED." 

SA9500 - Stereo Review
"AS NEAR TO PERFECT AS WE'VE ENCOUNTERED."

TX9500 - Popular Electronics

## "CERTAINLY ONE OF THE BEST... AT ANY PRICE."

TX9500 - Modern Hi Fi


Last year, the experts paid Pioneer's integrated amps and tuners some of the highest compliments ever.

The challenge was obvious: to build even better amps and tuners. Amps and tuners that would not only surpass anything wéd ever built before, but anything anyone ever built betore. Here's how we did it.

## THE NEW PIONEER TX9500II TUNER: EVEN CLOSER TO PERFECT.

When Popular Electronics said our TX9500 tuner was "as close to perfect" as they'd encountered, they obviously hadn't encountered our TX9500II. It features technology so advanced, some of it wasn't even perfected until this year.

Our front end, for example, features three newlydeveloped field effect transistors that work to let you pull in beautiful FM reception no matter how far you live from the transmitter. And no matter how much interference there is in your neighborhood.

Where most tuners give you one band for all FM stations, the TX9500II gives you two. A wide band with a new surface acoustic wave tilter to take advantage of strong stations, and a narrow band with five ceramic filters to remove the noise and interference from weaker ones.

And where conventional multiplex circuits accidently cut out frequencies that add depth and presence to music, the multiplex circuit in the TX9500II doesn't. It features a Pioneer-developed integrated circuit that's far more accurate than anything else around. So the music begins to sound as if it's coming live from your living room, instead of from some radio station miles away.

## THE NEW SA9500II AMPLIFIER:

 HOW TO GET THE MOST OUT OF THE BEST.After building one of the world's best tuners, we had no choice but to create an amplifier that could match it.

The result is the new SA95001I. An $80^{*}$ watt integrated amp that was designed to let you get every-
thing out of your tuner. Perfectly.
Our output stage, for example, features a new parallel push-pull circuit that reduces total harmonic distortion to less than $0.1 \%$. Well below the threshold of human hearing.

To all but eliminate cross-talk, the SA9500II comes with a separate power transformer for each channel, instead of the usual single transformer for both.

And where some amps give you two, or three tone controls, the SA9500II gives you four. Two for regular treble and bass, and two for extended treble and bass. They're calibrated in 2 dB click stops, which means you have a virtually endless variety of ways to get the most out of your music.

Obviously, both the SA9500II and the TX950011 are very sophisticated pieces of equipment. But all of the engineering skill that went into making them has gone into every tuner and amplifier in our new series II. No matter what the price, no matter what the specifications.

And that's something you don't have to be an expert to appreciate.

## SA9500II-TX9500II

| POWER MIN. RMS, 20 TO $20,000 \mathrm{~Hz}$ | 80 | SIGNAL TO NOISE RATIO | Mono 82dB <br> Stereo 77dB |
| :---: | :---: | :---: | :---: |
| TOTAL HARMONIC DISTORTION | 0.1\% | FM SENSITIVITY <br> (IHF'58) | 1.5 uV |
| PHONO OVERLOAD IEVEL | 300 mV | SELECTIVITY | (wide) 35dB (narrow) 85dB |
| INPUT: PHONO/AUX/ TAPE | 2/1/2 | CAPTURE RATIO | (wide) 0.8 dB (narrow) 2.0dB |

*Minimum RMS continuous power output at 8 ohms, from 20 to $20,000 \mathrm{~Hz}$, with no more than $0.1 \%$ total harmonic distortion.
(N) NE BRING IT BACK ALIVE.
U.S. Pioneer Electronics Corp., 85 Oxford Drive, Moonachie, New Jersey 07074 OUS PIONEER HFCTRONICS COKP 1977


TX9500II
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Made of specially treated sheeting

| PERFORMANCE SPECIFICATIONS |  |  |  |
| :---: | :---: | :---: | :---: |
| Specification | Rating | Measured | Comment |
| Power Output <br> ( 8 ohms, $20-20,000 \mathrm{~Hz}$ ) | 40 W (\% 0.1\%THD | $40 \mathrm{~W} .0 .05 \%$ THD |  |
| Clipping Power $(1000 \mathrm{~Hz})$ | $N / A$ | 51 W (8 ohms) <br> $64.8 \mathrm{~W}(4 \mathrm{ohms})$ <br> 33.6 W (16 ohms) |  |
| Frequency Response <br> (1-watt output) | $\begin{aligned} & 4 \mathrm{~Hz}-140 \mathrm{kHz} \\ & \pm 0.5 \mathrm{~dB} \end{aligned}$ | $\begin{aligned} & 5 \mathrm{~Hz}-220 \mathrm{kHz} \\ & \pm 0.5 \mathrm{~dB} \end{aligned}$ | 5 Hz is lower test limit |
| Rise Time | $1.5 \mu \mathrm{~s}$ | $1.5 \mu \mathrm{~s}$ |  |
| Slew Rate | $N / A$ | $9 \mathrm{~V} / \mu \mathrm{s}$ |  |
| IM Distortion $(60 / 7000 \mathrm{~Hz}, 4: 1)$ | $\begin{aligned} & 0.12 \% \text { (1) } 40 \mathrm{~W} \\ & 0.15 \% \text { (a) } 1 \mathrm{~W} \end{aligned}$ | $\begin{aligned} & 0.056 \% \text { ( ( } 140 \mathrm{~W} \\ & 0.065 \% \text { ( a } 1 \mathrm{~W} \end{aligned}$ |  |
| Hum \& Noise (unweighted) | -60 dB re 40 W | $\begin{aligned} & -83.6 \mathrm{~dB}(\mathrm{AUX}) \\ & -81 \mathrm{~dB}(\mathrm{PHONO}) \end{aligned}$ |  |
| Input Sensitivity (for 40-watt output) | $\begin{aligned} & 150 \mathrm{mV}(\mathrm{AUX}) \\ & 2.5 \mathrm{mV}(\mathrm{PHONO}) \end{aligned}$ | $\begin{aligned} & 136 \mathrm{mV}(\mathrm{AUX}) \\ & 2.2 \mathrm{mV} \text { (PHONO) } \end{aligned}$ | 1000 Hz |
| Phono Overload $(1000 \mathrm{~Hz})$ | 95 mV | 105 mV |  |
| RIAA Equalization Accuracy | $\pm 1 \mathrm{~dB}$ | $\pm 1 \mathrm{~dB}$ | $30-15,000 \mathrm{~Hz}$ |
| FM Sensitivity: |  |  |  |
| IHF MONO | $1.9 \mu \mathrm{~V}$ | $1.85 \mu \mathrm{~V}(10.6 \mathrm{dBf})$ |  |
| -50 dB MONO | $3.5 \mu \mathrm{~V}$ | $2.4 \mu \mathrm{~V}(13 \mathrm{dBf})$ |  |
| - 50 dB STEREO | $35 \mu \mathrm{~V}$ | $30 \mu \mathrm{~V}(35 \mathrm{dBf})$ |  |
| Ultimate $\mathrm{S} / \mathrm{N}$ | 70 dB | $\begin{aligned} & 68.5 \mathrm{~dB} \text { (MONO) } \\ & 68 \mathrm{~dB} \text { (STEREO) } \end{aligned}$ |  |
| Capture Ratio | 2 dB | 1.76 dB |  |
| Image Rejection | 80 dB | 85 dB |  |
| AM Rejection | 60 dB | 70 dB |  |
| Alternate Channel Select. | 80 dB | 71.4 dB |  |
| Adjacent Channel Select. | N/A | 7.8 dB |  |
| Stereo Separation $(1000 \mathrm{~Hz})$ | 40 dE | 47 dB |  |
| FM Distortion $(1000 \mathrm{~Hz}$ ) | $0.2 \%$ MONO <br> $0.3 \%$ STEREO | $0.11 \%$ MONO <br> $0.29 \%$ STEREO | $65 \mathrm{~dB}{ }^{\text {f }}$ |
| Pilot Carrier Leakage | $N / A$ | $-75 \mathrm{~dB}$ |  |

laminated to a curved frame reflecting almost all its incident light back to the viewer.

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User Comment. The RIAA phono equalization is not only very accurate over the defined range of 30 to 15,000 Hz (it falls off somewhat at the lowest frequencies), but it is totally unaffected by cartridge inductance. The audio distortion of the receiver was quite insignificant, and the clipping power of about 50 watts should be adequate for most
home installations, with speakers of normal efficiency
Although the tone controls have a satisfactory range and choice of turnover frequencies, the high-cut filter is essentially useless. (It virtually duplicates the response of the treble tone control near its minimum setting.) A 6-dB/octave filter with a $-3-\mathrm{dB}$ response at 3000 Hz is


SINE-WAVF POWER OUTPUT PER CHANNEL IN WATTS
Total harmonic distortion and 60/7000-Hz distortion.
really not worth the control space allotted to it on the panel. The low-frequency filter, although not ideal, is considerably better; it has a $12-\mathrm{dB}$ /octave slope and a -3 -dB frequency of 50 Hz . The loudness compensation boosts only low frequencies, to a moderate degree, yet manages to sound undesirably heavy because it affects almost the entire range below 1000 Hz .
The FM tuner section was excellent, handily meeting or surpassing almost every specification. The quieting meter proved to be a very accurate tuning indicator. Although this point is not mentioned in the instruction manual we noted that the meter's pointer fluctuated with multipath distortion and gave a steady indication on signals free of that effect. This represents the first triple function FM tuning meter we have used (center channel, as shown by a definite minimum reading; relative signal strength; multipath distortion). Its only drawback, compared to conventional meters, is that one must tune quite slowly through a station to establish the minimum reading, and then tune back to that point-again, slowly-since the me-
ter seems to respond with a lag. In contrast, a typical center-channel meter allows one to stop tuning as soon as the pointer reaches the indicated center.
Although the FM tuner had a ruler-flat frequency response to $15,000 \mathrm{~Hz}$, the $19-\mathrm{kHz}$ pilot carrier leakage was a very low -75 dB . This speaks well for the design of the multiplex filters in the tuner section. The stereo channel-separation characteristic was unusual, with a very wide separation of more than 60 dB at 30 Hz , which decreased smoothly to 24 $d B$ (still good) at $15,000 \mathrm{~Hz}$. As a whole, the tuner's FM performance was excellent. The AM tuner was as limited in its frequency response as most we have used, but was exceptionally quiet and free of the buzzes and noises that usually accompany a scan of the AM broadcast band.

In listening tests, the Model 730 did a first rate job. Like most good receivers, it sounded as good or as bad as the program material allowed. When we transmitted a high-quality signal to it through our laboratory signal generator, it left nothing to be desired. The muting circuit has a slight delay that allows the dial to


Harmonic distortion at three power levels.
be twirled from one end of the FM band to the other in perfect silence; about a second after a station is tuned in, its modulation appears. In slow tuning through a station, we heard only a faint click as the muting operated.

One of our two criticisms of the Model 730 (both relate to its FM tuner section) concerns a basic design factor and the other is certainly due to a fault in our test sample. The pointer of the tuning meter is not illuminated or even colored white for contrast against the black and dark green background of the meter scale. As a result, it can hardly be seen, except at point-blank range. On our test unit, the FM dial calibration was poor, showing the result of a misalignment. It was accurate below 94 MHz but had an error of 300 kHz at 96 MHz , and 700 kHz at 106 MHz . As a result, tuning was approximate over much of the band.
The Harman-Kardon Model 730 is convincing evidence that a mediumpriced receiver can deliver top-quality sound, combined with all the operating flexibility most people will ever need and an attractive appearance.

CIRCLE NO 101 DN fREE INFORMATION CARD

## SHARP MODEL RT-3388 FRONT-LOADING CASSETTE DECK

Built-in "computer control" system and LCD display are featured.


The Sharp Model RT-3388 is a basic cassette deck of conventional design and good quality. Where it differs from other cassette decks is in its built-in "computer control" system that gives it an operating flexibility and convenience previously unavailable in a consumer deck. The cassette deck proper is a two-head, sin-gle-motor transport with a servo-con-
trolled dc motor. The front-loading mechanism provides excellent visibility of the entire cassette through its clear plastic window that makes up almost the entire cassette compartment door. The door can be removed by loosening two thumbscrews to provide complete ac* cess to the tape heads for cleaning and adjustment. The operating controls are mechanical "piano-key" levers that are solenoid assisted for certain functions.

The cassette deck measures $173 / 8^{\prime} \mathrm{W}$
$\times 127 / 8^{\prime \prime} \mathrm{D} \times 53 / 8^{\prime \prime} \mathrm{H}(44.1 \times 32.7 \times 13.7$ cm ) and weighs $20 \mathrm{lb}(9.1 \mathrm{~kg})$. Its nationally advertised retail value is approximately \$350.

General Description. The deck has separately switchable bias (HIGH/LOW) and equalization ( $70 / 120 \mu \mathrm{~s}$ ). The manual that accompanies the deck lists recommended settings for most popular tape brands and formulations. Unlike the case with most recorders, the manual

also identifies the specific tapes for which the deck was adjusted at the factory. They are Maxell UD (normal), Maxell UD-XL II (HIGH/70 $\mu \mathrm{s}$ ), and Sony Ferrichrome (LOW/70 $\mu \mathrm{S}$ ).
The recorder has built-in Dolby noise reduction, separate microphone and line inputs that can be mixed, and large illuminated level meters supplemented by a PEAK LED indicator. While this is conventional, the rest of the deck is far from conventional, as is immediately obvious.
A "computer" keyboard and display panel dominate the center of the front panel. The "computer" that controls the operation of the deck provides memory functions for stopping the tape at a preselected point, either in normal or fast speeds. It also supplies the tape index counting function usually performed by a belt-driven mechanical counter. It controls Sharp's Automatic Program Locator Device (APLD) that can be set to skip any number of separately recorded selections (up to 19 in all) on a tape, stopping and playing a preselected piece that occurs later in the tape. The APLD contains a quartz controlled digital clock that is constantly in operation. When the machine is plugged into the power line, the clock is powered whether or not the recorder is turned on. Internal batteries can power the clock for up to a year without recourse to the ac power line. The clock can turn the deck (and anything plugged into its single switched outlet) on and off at preset times with split-second accuracy and time the running of a tape in minutes and seconds. It also provides the usual index-counter function. As a timepiece, it can be set for a 12 - or a 24 -hour format and automatically provides AM and PM identification in the 12 -hour mode.

All the display functions appear on an LCD (liquid-crystal display) panel located above the keyboard. When the recorder is turned off, the display automa-
tically reverts to its clock function. Seconds are indicated by the blinking of the colon that separates the hours and minutes digits. When the deck is turned on,

| SPECIFICATION | N RATING | MEASURED | REMARKS |
| :---: | :---: | :---: | :---: |
| Wow \& Flutter | 0.06\% wrms | 0.065\% unweighted rms |  |
| Frequency | $30-13,000 \mathrm{~Hz} \pm 3 \mathrm{~dB}$ | $25-14,200 \mathrm{~Hz} \pm 1 \mathrm{~dB}$ |  |
| Response | (normal) | UD-XLI (LOW/120 $\mu \mathrm{s}$ ) |  |
|  | $30-15,000 \mathrm{~Hz} \pm 3 \mathrm{~dB}$ | $24-14,600 \mathrm{~Hz} \pm 1 \mathrm{~dB}$ |  |
|  | (UD-XL II) | UD-XLII ( $\mathrm{HIGH} / 70 \mu \mathrm{~S}$ ) |  |
|  | $30-16.000 \mathrm{~Hz} \pm 3 \mathrm{~dB}$ | $22-15,000 \mathrm{~Hz} \pm 3 \mathrm{~dB}$ |  |
|  | (FeCr) | Sony FeCr (HIGH/120 $\mu \mathrm{s}$ ) |  |
| Playback | N/A | $40-12,500 \mathrm{~Hz} \pm 2 \mathrm{~dB}$ |  |
| Response |  | TDK AC-337 (120 $\mu \mathrm{s}$ ) |  |
|  |  | $40-10,000 \mathrm{~Hz} \pm 0.5 \mathrm{~dB}$ |  |
|  |  | Teac 116SP ( $70 \mu \mathrm{~s}$ ) |  |
| S/N Ratio | 55 dB (normal) | 47.4 dB UD-XL । | 3\% playback |
|  |  | 45.3 dB UD-XL II | distortion, |
|  |  | 44.7 dB Sony FeCr | unweighted |
|  |  | 56.0 dB UD-XL | CCIR/ARM |
|  |  | 58.8 dB UD-XL II | weighted |
|  |  | 56.0 dB Sony FeCr |  |
|  |  | 65.6 dB UD-XLI | CCIR/ARM |
|  |  | 66.5 dB UD-XL II | weighted, |
|  |  | 64.4 dB Sony FeCr | with Dolby |
| Noise Increase |  | +6.5 dB | mic input at |
|  |  |  | maximum gain |
| Sensitivity | 50 mV (LINE) | 54 mV |  |
|  | 0.2 mV (MIC) | 0.2 mV |  |
| Output Level | 775 mV at 0 dB (LINE) | Varies with tape; | Phones not |
|  | 89 dB (8-Ohm PHONES) | $700 \text { to } 800 \mathrm{mV}$ | measured |
| Playback | N/A | . $0.45 \%$ UD-XL I | $1000-\mathrm{Hz}$ test |
| Distortion (0- |  | $0.25 \% \text { UD-XL II }$ | tone |
| dB record level) |  | 0.80\% Sony FeCr |  |
| $\begin{gathered} \text { Crosstalk ( } L-R \\ \text { at } 1000 \mathrm{~Hz} \text { ) } \end{gathered}$ | $N / A$ | -51 dB TDK AC-352 |  |
| PEAK indicator On | N/A | +6.5dB |  |
| Meter Ballistics | N/A | 10\% overshoot | $0.3-\mathrm{s}, 1000-\mathrm{Hz}$ <br> tone bursts |
| Meter Calibration | N/A | within 0.5 dB |  |
| Fast Forward and Rewind Time | N/A | - 94 s | C-60 cassette |

Note: Manual specifies Low bias and $70-\mu$ s equalization for FeCr tape. These settings yielded sharply rising high-frequency response; flattest response was obtained with HIGH bias and $120-\mu$ s equalization.

# Better stereo records are the result of better playback pick-ups 



# Enter the New Professional Calibration Standard,Stanton's 8815 



Mike Reese of the famous Mastering Lab in Los Angeles says: "While maintaining the Calibration Standard, the 881 S sels new levels for tracking and high frequency response. It's an audible improvement. We use the 881 S exclusively ior callibration and evaluation in our operation'

The recording engineer can only produce a product as good as his ability to analyze it. Such analysis is best accomplished through the use of a playback pick-up. Hence, better records are the result of better playback pick-up. Naturally, a calibrated pick-up is essential.

There is an additional dimension to Stanton's new Professional Calibration Standard cartridges. They are designed for maximum record protection. This requires a brand new tip shape, the Stereohedron*, which was developed for not only better sound characteristics but also the gentlest possible treatment of the record groove. This cartridge possesses a revolutionary new magnet made of an exotic rare earth compound which, because of its enormous power, is far smaller than ordinary magnets.

Stanton guarantees each 881S to meet the specifications within exacting limits. The most meaningful warranty possible, individual calibration test results, come packed with each unit.
Whether your usage involves recording, broadcasting or home entertainment, your choice should be the choice of the professionals . . the STANTON 881 S .


the display becomes a tape counter, although any of its other functions can be selected at that time by pressing the proper buttons. Arrows in the display indicate direction of tape motion and tape speed (the latter by blinking on and off when a fast speed is selected). A Dolby trademark symbol appears in the display when the DOLBY switch is on. Short bars above each column of keys indicate which function has been selected, and a letter $M$ appears above a bar when specific memory data has been entered for a given function.

The computer-controlled functions of the Model RT-3388 are so diverse that most of the very comprehensive instruction manual is devoted to them, with the aid of copious illustrations and photographs. Only nine of the 36 pages in the manual deal with normal tape recording functions. In general, the desired information, whether it be an index counter reading or a time, is fed into the computer by sequential operation of the keys, followed by a touch of the s (set) key. Previously set functions, which are not incompatible with the new one, remain unchanged and in use. Thus, one can switch between index counter, seconds counter, and actual time display at will while the tape is running without miscounting or affecting the operation of the machine. When the recorder is turned on, the LCD panel is back lit for full visibility. With the power switch off, a separate switch can be used to activate only the display lights so that the clock can be used at any time.

The "computer" section is physically very similar to a hand-held calculator in its size and configuration of keyboard and display. It extends only about $1 / 2^{\prime \prime}$ $(12.7 \mathrm{~mm})$ behind the panel and cannot be seen without extensive disassembly. According to the manual, a single LSI chip supplies all computer functions.

Most of the remaining electronic circuits are on a single large circuit board that is very clearly labelled with individual circuit reference symbols for all components and the function of each section of the board (Dolby, Preamplifier, etc.). A second large board contains the power supply and control circuitry. The tape transport occupies only a small fraction of the internal space of the recorder.

User Comment. This is a most impressive product, largely because it offers so much for such a moderate price without appearing to have sacrificed any significant aspect of its performance. The tape saturation characteristics at a o-dB recording level, compared to the -20 dB response, suggest that the heads are of only ordinary efficiency, neither better nor worse than one would expect on a machine in the $\$ 250$ to $\$ 300$ price range. On the other hand, the measured wow and flutter were about as low as we have measured on any cassette recorder and are better than we would expect to find at this price.

The recorder delivered excellent sound quality, both from prerecorded tapes and from recordings we made off the air and from records. We also recorded FM tuner interstation hiss at a -10-dB level and compared the playback to the original sound. This test simultaneously checks frequency response and tape saturation effects and reveals even slight deviations from an accurate recording. With Maxell UD-XL I tape, the results were good, but not quite perfect; we could hear a slight dulling of the highest frequencies. The "chrome equivalent" Maxell UD-XL II, on the other hand, yielded perfect reproduction of the highest frequencies. In the case of ferrichrome tape, we are not certain how to judge the machine's performance. With the HIGH bias and $120-\mu \mathrm{s}$

## Product Focus

The unique feature of the Sharp Model RT-3388 cassette deck is its one-chip microcomputer/liquid crystal display ( $\angle C D$ ). Commands and numbers for the memory are activated by a 24-key matrix.

The microprocessor chip that controls the deck is a single, square 60 pin CMOS LSI device. It has a built-in clock, 54 commands, 2268 bytes of ROM, and 96 words of RAM. A crystal oscillator supplies energy for the clock, timer, and counter that indicates tape time in seconds.

The timer serves as an alarm clock and sleep-time indicator. It also permits unattended recording and can switch on and off the tape deck and a device plugged into the deck's accessory outlet. The timer features twostage operation: one for independent timer stop; the other for switching from timer start to timer stop and vice versa.

When the tape counter is set, the deck plays to the present point and automatically shuts off. Also in fast forward and rewind, the deck automatically stops at the preset point.

The tape counter keeps track of the tape passing the heads but cannot accurately register absolute values. A second counter is provided for more accurate indications.
Setting the number of a selection on the tape puts the deck into fast forward or rewind to locate the desired selection. This automatic locating device can be keyed for as many as 19 program steps ahead of or behind the desired selection.
equalization that gave flattest frequency response, we could hear a slight difference between the input and output of the recorder. Using the recommended settings of Low bias and $70-\mu$ s equalization, the results were audibly perfect, as good as with UD-XL II, in spite of the measured low-level frequency response being far from flat. Obviously, some user experimentation would be in order if FeCr tape is to be used. We also obtained completely acceptable results
with a number of comparable tapes of all kinds, using the appropriate switch settings. The tapes chosen for testing were those specified by Sharp; they gave the flattest response in our tests.

The computer functions were fascinating to use. Space does not permit a full account of what this machine can do (many pages of the manual are devoted to that), but we were continually impressed by the accuracy with which it responded upon reaching a preset point on a tape, whether in a memory mode or
on the APLD function. The latter counts the silent intervals between recorded selections to locate the desired point, so it cannot be "foolproof" in its operation (when making one's own recordings, an EDITOR switch cuts off the program while a short blank section is recorded between program segments). All in all, we found that the APLD worked correctly in the vast majority of cases, even with commercially recorded tapes. To fully utilize the capabilities of the deck, one must only study the manual carefully
and practice extensively with its controls to become familiar with it.

In sum, the Sharp Model RT-3388 is an above-average tape recorder for its price in all basic performance aspects and is unique at this time in its operating features. It is as much fun to use and to look at as it is for listening. Once you have been exposed to the comprehensive LCD panel, mechanical counter and function display indicators on other recorders appear old fashioned.

CIRCLE NO 102 ON FREE INFORMATION CARD

## WHARFEDALE MODEL E50 SPEAKER SYSTEM

High-efficiency system produces crisp sound and tight bass.


Wharfedale has re-entered the U.S. hi-fi market after an absence of several years with a new line of highly efficient speaker systems. The British Company's new Model E50 speaker system resembles some of the recent products from Japan with its drivers surrounded by machined aluminum rings set against a flat black speaker board. The grille is an openmesh plastic that appears to be equally transparent to light and sound. Cutouts provide access to the two level control switches, labelled HIGH and LOW, even with the grille in place.

The Model E50 speaker system measures $32^{\prime \prime} \mathrm{H} \times 14^{\prime \prime} \mathrm{D} \times 13^{1 / 2} 2^{\prime \prime} \mathrm{W}(81.5$
$\times 36 \times 43.2 \mathrm{~cm}$ ) and weighs $70 \mathrm{lb}(32$ kg ). Nationally advertised value is $\$ 390$.

General Description. The Model E50 (" $E$ " stands for efficiency) is designed to deliver high acoustic levels without dynamic compression or other distortion when driven at moderate power levels. Its $10^{\prime \prime}(25.4-\mathrm{cm})$ woofer operates in a vented enclosure, the $4^{\prime \prime}$ (10.2$\mathrm{cm})$ port of which has been designed to operate as a fourth-order, maximally flat Butterworth system. At 800 Hz , there is a crossover to a $4^{\prime \prime}(10.2-\mathrm{cm})$ cone driver, while at 7000 Hz , there is another crossover to a horn-loaded compression driver that has a $1^{\prime \prime}(2.54-\mathrm{cm})$ diaphragm. The frequency response is rated at 55 to $18,000 \mathrm{~Hz} \pm 3 \mathrm{~dB}$.

The system's level controls are fiveposition switches, with their 0 (maximum) settings providing the flattest response. The frequency ranges affected by these switches do not correspond exactly to the range of any single driver. For example, the Low control, which should really be called mID, varies the output between 200 and 2000 Hz , with a maximum reduction of 5 dB . The HIGH switch produces a shelf in the response above 2000 Hz , with a maximum range of 5 dB . The system impedance is normally 8 ohms, and it has been designed to be greater than 6.8 ohms at all audible frequencies.

Laboratory Measurements. We measured the low-frequency response, up to about 300 Hz , with a closely spaced microphone. We separately recorded the response of the port and the
woofer cone and combined the two to obtain a single reading. The highs were measured in the reverberant field of the room, using a swept warble tone signal and averaging the curves obtained from the two speakers in a normal setting for stereo listening.

With due allowance for the very different measurement techniques used by Wharfedale and ourselves, it was clear that the speaker system easily met its response rating. A moderate high-frequency peak was the only part of the composite response curve that exceeded the $\pm 3-\mathrm{dB}$ limits. From 45 to 8000 Hz , the response was smooth and uniform, within $\pm 3 \mathrm{~dB}$. It rose to about +5 $d B$ at 11,000 or $12,000 \mathrm{~Hz}$. The woofer's response, from 50 to 300 Hz , was exceptionally flat, attesting to the success of the computer design employed.

The low-frequency distortion at a 1 watt input was a mere $0.5 \%$ or so down to 60 Hz . It rose to only $5 \%$ at 30 Hz . At a 10-watt drive level, the distortion was less than $2 \%$ down to 50 Hz , rising to $6.7 \%$ at 35 Hz . When the input was adjusted to maintain a 90 dB SPL at 1 meter with changes of frequency, the result was similar to the 10-watt measurement except that the distortion rose more rapidly below 50 Hz .

The level controls had approximately the specified effects. (Typical response curves as well as fairly complete system specifications are on a nameplate affixed to the rear of the cabinet.) The HIGH control began to take effect at about 1000 Hz , overlapping the Low control, which modified the response between 150 and 1500 Hz . The system

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Course 4 covers basic amplifiers, typical amplifiers, operational amplifiers, power supplies, oscillators, pulse circuits, modulation and demodulation with emphasis on integrated circuits. Included are texts, records and more than 110 electronic components for 18 different experiments. The ET-3100 Experimenter/Trainer Kit is also available as an option. Course 4 requires the completion of Courses 1 through 3 or equivalent knowledge. The average completion time for Course 4 is 40 hours.
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Tone burst responses at frequencies (left to right) of 100,1000 , and 7000 Hz .
impedance varied considerably with frequency, with a maximum of about 40 ohms at 22 Hz and a minimum of about 6 ohms at 200 and 1500 Hz .

The tone-burst response was exceptionally good, with nearly perfect burst shapes at most frequencies. The high efficiency of Wharfedale's computerdesigned speaker system was dramatically illustrated by the Model E50's ability to deliver a 95-dB SPL at 1 meter when driven by 1 watt of random noise in the octave centered at 1000 Hz .

The simulated "live-versus-recorded" listening test confirmed that the measured smoothness and range of the system's frequency response was quite real. The chief difference between the sound of our "live" source (a reference speaker system that reproduced a specially taped program) and the sound of the Model E50 attempting to imitate it was an added brilliance in the sound of the latter. This could have been inferred from its slightly rising high-frequency response characteristic. Its effect was to add "sizzle" or "bite" to the sound of wide-range program material that contained appreciable high-frequency energy. The most accurate reproduction was with the HIGH switch at its minimum setting, but the highs were still "hot."

User Comment. The sound of the Wharfedale Model E50 might appeal more to the pop or rock music listener than to the classical enthusiast, especially since it can produce prodigious levels of very clean sound when driven by a modestly priced receiver or amplifier. It lacks the deep bass response fa-
vored by some people (although it is by no means shy of bass). However, it is outstandingly flat and true over most of the audio range.

The speaker system seemed to us to have a razor-sharp, almost clinical quality. This may be related to its notably good transient response, as confirmed by our tone-burst tests, as well as to its accentuated top-end response. As a rule, we prefer to set a speaker system's balance controls only once, upon installation. In this case, we found it desirable to readjust the HIGH level control according to the program content. With most FM broadcasts and records, we found that one or two steps of high-frequency reduction gave the best results. With programs containing exceptionally strong high-end content, such as the Sheffield direct-disc recordings played with a good moving-coil cartridge, we had to cut the speaker system's highs all the way down. (The Low switch was left at maximum.)

We found the Model E50 to be an exceptionally clean and easy-to-listen-to speaker system. Since its emphasis was principally at frequencies above 10,000 Hz , its high-frequency response was never audible as stridency or even as "presence." Instead, it gave a crispness or edge to the sound that seemed well matched to the very tight and nonboomy bass reproduction of the speaker system. The ability to operate effectively at any usable listening level from an amplifier rated at, perhaps, 20 watts output is a major advantage of this speaker system over most of its competitors.
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WHEN YOU design or build projects, particularly large-scale designs in which many transistors and/or IC's are used, most of the effort goes into creating the final circuit and the printed circuit board. All too often, the power supply is just an afterthought. This is unfortunate because even a welldesigned and assembled project may operate borderline if the power supply is not delivering the correct voltage at the required current. This problem is compounded when the supply must deliver large amounts of current, as in multi-IC digital circuits, especially microcomputers. Hence, the power supply deserves special attention, since it is often critical to the success of an electronic project.
In this first of a two-part article we will discuss power supply basics, some design concepts, etc. By the end of Part II, you should be able to design low-voltage, high-current power supplies that can perform as required for just about any project.

Transformers. The transformer is generally a voltage converter. It reduces the standard 117 -volt ac power line potential to the lower voltages required in solid-state electronics. Most discrete circuits operate with potentials in the 1.5 -to-28-volt range; linear IC systems operate in the range from $\pm 4.5$ to $\pm$ 18 volts; CMOS circuits require between 4 and 18 volts; and TTL requires the use of a tightly regulated 5 -volt supply line.
Because a transformer is very efficient, stepping the line potential down in the secondary winding increases the current available for any given voltage level. The primary VA (volts times amperes) rating is very nearly equal to the VA rating of the secondary. Simply stated, $E_{\text {pri }} \times I_{\text {pri }}=E_{\text {sec }} \times I_{\text {sec }}$, where $E_{\text {pri }}$ is the potential in the primary winding ( 117 volts ac); lpri is the primary current; $\mathrm{E}_{\text {sec }}$ is the voltage in the secondary; and $\mathrm{I}_{\text {sec }}$ is the secondary current.

Rectifier. The rectifier converts the alternating current from the transformer's secondary into pulsating direct current (dc). The simplest of rectifiers is the halfwave circuit shown in Fig. 1.

All rectifiers operate on the same principles, whether they are solid-state or vacuum-tube types. They conduct current in only one direction. When an ac sine wave is applied to the input of this circuit, current passes through the rectifier only when its anode is more positive than its cathode, as in Fig. 1A. On the other half of the ac cycle, the rectifier

## BY JOSEPH CARR



## PART 1

Basics of transformers, rectifiers, filters, voltage regulators
and protection circuits.


Fig. 1. Forward-biased diode (A) conducts current while reverse-biased diode (B) does not. In (C), upper trace is ac input, lower trace is pulsating dc across load.


Fig. 2. At (A) is fullwave rectifier. Ac in primary of T1 is upper trace in (B), pulsating dc is at bo'tom.


Fig. 3. A full-wave bridge rectifier. Broken lines show current flow during each half of the ac input cycle.
is reverse-biased (Fig. 1B), thus preventing the flow of current through the external load, $\mathrm{R}_{\mathrm{L}}$.

The waveforms associated with the
half-wave rectifier are shown in Fig. 1C. The top waveform is that of the ac sine wave applied to the input, while the bottom waveform shows the rectified pul-
sating dc output across $R_{L}$. Note that the pulsating dc output exists only when the input waveform is in its positive alternation. Because half of the input waveform is not used, the half-wave rectifier is very wasteful of electrical energy. And half-wave rectification presents difficulties in filtering the output to pure dc with no ripple component.
The half-wave rectifier has an average output potential of approximately 0.45 times the applied rms potential and its ripple amounts to $120 \%$. To add to the problems of this design, the transformer used must have a primary VA rating $40 \%$ greater than is required if fullwave rectification were used.

A basic full-wave rectifier using a cen-ter-tapped transformer is illustrated in Fig. 2A. At any given ac peak, one end of the transformer's secondary is positive, while the other end is negative. The center tap is at a potential that is half that across the entire secondary. Therefore, if the center tap is used as the common reference, equal and opposite polarity potentials will be found at either end of the secondary with respect to the center tap.

Let us consider the case when the top of the secondary is more positive than the bottom. Current flows from the common center tap through $R_{L}$ and for-ward-biased rectifier D1 (whose anode is more positive than its cathode) and then back to the transformer. During this period, $D 2$ is reverse-biased due to the negative potential at its anode so that no current can flow through it.

On the alternate half-cycle, D1 becomes reverse-biased and D2 conducts. Current then flows from the center tap through $R_{L}$ and forward-biased D2 and back to the secondary of the transformer. Note that, in both cases, the current flows through the load in the same direction. This produces the "doublehumped" waveform across $R_{L}$ shown in the lower trace of Fig. 2B. In essence, the negative-going portion of the applied ac sine wave has been "folded up" to produce the double-frequency waveform shown in the figure.

The bridge circuit shown in Fig. 3 is another type of full-wave rectifier. It employs a diode "ring" (D1 through D4) for rectification. The secondary of the transformer is not center tapped; the diode ring provides the negative (sometimes ground) reference point. The two "corners" of the bridge labelled "+" and "-" and go to the positive and negative sides of the filter capacitor.

Since the bridge rectifier circuit em-

ploys the entire secondary potential, it produces an output potential (pulsating dc) of twice that of the ordinary full-wave rectifier using the same transformer. There is one catch, however. The bridge rectifier supply delivers only half the current of the full-wave rectifier for a given primary VA rating. There are occasions when it is possible to exceed this and draw nearly the full rated current from the transformer's secondary without causing damage, but this is not dependable in all cases.
The average dc output potential in an unfiltered full-wave supply is approximately 0.9 times the applied rms potential, or about twice the voltage obtained with the half-wave rectifier. Both types of full-wave rectifiers have an output ripple component of about $48 \%$ and, thus, need filtering to produce the dc required by the electronic circuits. Also, the ripple frequency in the full-wave rectifier circuit is 120 Hz , which is twice the frequency of the line power.

Filters. The filter smoothes out the pulsating dc output from the rectifier to create the nearly pure dc required by the electronic circuitry load.

The half-wave rectifier produces one dc pulse for each ac cycle (Fig. 1C), while the full-wave supply produces two dc pulses per cycle (Fig. 2B). These waveforms illustrate the difference in ripple frequency- -60 Hz for the half-wave and $120-\mathrm{Hz}$ for the full-wave rectifiersand implies that the higher frequency of the half-wave rectifier's output is easier to filter.

The usual high-value capacitor found in power supplies is shown in Fig. 4A. In this circuit, the bridge rectifier is shown in block form, since it is most often a bridge-rectifier assembly rather than a set of four discrete rectifier diodes. Filter capacitor C1 is connected directly across the rectifier.

The value of Cl is critical to the performance of the power supply. It should be no less than $1000 \mu \mathrm{~F}$ per ampere of output current; many authorities claim that $2000 \mu \mathrm{~F}$ per ampere should be the minimum. In any event, it is good practice to use not less than $1000 \mu \mathrm{~F}$ in projects that draw 1 ampere or less current. A typical 5-volt, 4-ampere dc power supply for a small digital computer would require not less than 8000 microfarads for a good filtering.

The waveform shown in Fig. 4B illustrates how the filter capacitor reduces
the level of the pulsations in the rectified output waveform. Capacitor Cl charges up as long as the pulsating dc applied to it is rising. Once the peak potential has been reached and the rectified waveform begins to drop toward zero, the capacitor dumps its charge to fill up the spaces (shaded area in Fig. 4B) between the pulses. Obviously, the greater the charge dumped, the smoother will be the top of the output waveform from the filter. The five waveforms shown in Fig. 5 were obtained from a low-voltage, 5-ampere supply using different amounts of filter capacitance. The circuit employed was that shown in Fig. 4, using a transformer rated at 13 volts and 10 amperes.

The Fig. 5A waveform shows the unfiltered output across the load resistor. The base line represents the 0 -volt level, while the peak of the pulsating dc waveform is just short of 19 volts. The result of connecting a $150-\mu \mathrm{F}$ capacitor across the load is shown in Fig. 5B. Note that the ripple has been reduced and has taken the shape of the filtered output shown in Fig. 4B. A dc voltmeter connected across the load indicated approximately 13 volts when there was


Fig. 5. How capacitors smooth waveform. Lower trace is zero volts. (A) is full-wave dc without filter; (B) has $150-\mu F$ capacitor; (C) $2000 \mu F$; (D) $5000 \mu F$; (E) $18,000 \mu F$.


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no filtering. With the $150-\mu \mathrm{F}$ capacitor installed, it indicated 16.8 volts.

Connecting a $2000-\mu \mathrm{F}$ capacitor across the load produced the Fig. 5C waveform. The ripple is substantially reduced and the average dc potential has risen to about 18 volts. The situation is even better in Fig. 5D, where the capacitance is $5000 \mu \mathrm{~F}$. The ripple has lessened to the point of almost disappearing. The dc potential has risen only an additional 0.7 volt, to 18.7 volts. In the Fig. 5E waveform, an $18,000-\mu \mathrm{F}$ capacitor is across the load, which results in less ripple but no increase in the dc output potential. Bear in mind that this is for a 4 -ampere power supply in which the formula capacitance should have been 8000 picofarads.

Voltage Regulators. Circuits that maintain their output potential constant over a wide range of load variations are termed "voltage regulators." Most computers and all TTL circuits fare better on such regulated power supplies.

Voltage regulators for low-current levels are reasonably simple. Up to 5 amperes, conventional low-cost three-terminal IC regulators can be used. The circuit of a power supply in which a threeterminal regulator is used is shown in Fig. 6A.

Several different but essentially similar families of three-terminal IC regulators exist. Probably the most familiar is the LM309 series, the LM309H being a 100-mA device in a TO-5 package and the LM309K a 1 -ampere device in a TO-3 case.

There is also the LM340 series in which the output voltage is indicated by a number suffix added to the basic series number. For example, the LM340-5 is a 5 -volt regulator, while the LM340-12 is a 12 -volt device. These devices are available with outputs up to 24 volts.

They come in two package styles-the $K$ package for 1 -ampere and the $T$ package for $750-\mathrm{mA}$ capacity.

The LM320 devices are essentially the same as the LM340 devices, except that they are designed for negative output voltages. Note that the pinouts for the negative regulator shown in Fig. 6B are different than for the positive regulator. Failure to observe this fact can result in catastrophic damage when the power supply is turned on.

Another well-known regulator family is the 7800 (positive) and 7900 (negative)


Fig. 7. Bottom waveform shows how a regulator "limits" peaks on rectified pulsating dc.
series. The output potential is given by the last two digits in the type number ( 7805 for +5 volts and 7812 for +12 volts output, for example).

As shown in Fig. 6A, all three-terminal regulators should have noise bypass capacitors (C2 and C3) across their input and output terminals. Various manufacturers specify different values for these bypass capacitors, the most common being between 0.33 and $2 \mu \mathrm{~F}$. These noise filters should be wired as close to the terminals of the regulator as possible. If you use the lower value, the capacitor should be of ceramic disc construction. If the higher value is used, select a tantalum capacitor.

Capacitor C4 is optional but desirable, especially where output current demands are very dynamic. The capacitance usually specified is on the order of $100 \mu \mathrm{~F}$ per ampere, or about a tenth of the value of the main filter capacitor. This added capacitor is not specifically used for filtering but to provide a "hedge" against output voltage droop under transient load conditions.
It is necessary to use a filter capacitor before a regulator; and Fig. 7 reveals why. The upper trace is the pulsating dc obtained from the rectifier, while the lower trace is the output of the regulator when filter capacitor $C l$ is disconnected from the circuit. The unfiltered but regulated output waveform rises in each cycle until it reaches the cutoff point of the regulator, at which point it clips. Now, examine the waveform shown in Fig. 8. Although these waveforms appear to be similar to those shown in Fig. 5, they are different. In Fig. 5, the lower trace was used to indicate the 0 -volt base line, while in Fig. 8 they illustrate the input (upper) and output (lower) of a three-terminal regulator. These traces are accoupled to the oscilloscope so that the dc component is suppressed and to permit the $5000-\mu \mathrm{F}$ ripple component to be


Fig. 8. Waveform at top is before regulator. Lower trace shows how ripple is eliminuted.
shown on a larger scale. The preregulation waveform of the upper trace was taken using a 0.2 -volt/cm sensitivity in the scope's vertical channel, while a 0.01 -volt/cm sensitivity was used for the lower trace. Even at 20 times sensitivity, no apparent ripple appears in the output waveform on the scope.

Overvoltage Protection. Unfortunately, there are occasions when "something happens" in the regulator that permits the output voltage to rise above the required level. This potentially disastrous situation can be averted with an overvoltage protection circuit like that shown in Fig. 9. This circuit is called a


Fig. 9. Crowbar overvoltage protection circuit "fires" when tovolt line exceeds breakdown voltage of zener diode.
"crowbar" because it operates by shorting the output to ground in the same manner as a conducting metal crowbar would if il were actually connected across the supply.

Normally, the supply potential (in this case +5 valts) is too low to allow zener diode D1 to conduct. Consequently, the SCR presents a high impedance that makes it "invisible" to the dc line. When the potential on the supply line exceeds 5.6 volts, D1 conducts and generates a voltage across $R 2$. This voltage is then applied, via R1, to the gate of the SCR,
which triggers on. When this occurs, the short circuit that results causes fuse F1 to blow and shut down power. Although this circuit appears to be a little crude, it is extremely effective and can prevent damage to an expensive system connected to the power supply.

If you decide to use the crowbar protection circuit shown in Fig. 9, select an SCR that can handle about twice the current normally delivered by the power supply. Also, use a conventional fastblow fuse for F1.

Some of the circuits we will discuss in

Part II employ commercially available overvoltage protection devices, such as the OV-1 shown in Fig. 6A.

Current Limiting. This feature is usually found in supplies that employ more sophisticated voltage regulator circuits than those described above. Essentially, a small-value resistor is connected in series with the output lead of the regulator and the current drawn by the load generates a small voltage drop across this resistor. This voltage is applied to a comparator/amplifier that shuts down the power supply if excess current is drawn by the load.

Coming Up. In Part II of this article, we will discuss further design criteria. We will also present four construction projects: a +8 -volt, 15 -ampere power supply for Altair (S-100) bus microcomputers; +5 -volt, 4-ampere power supply; $\pm 12$-volt, 1 -ampere power supply; and a sophisticated 5 -volt, 10-ampere power supply with overvoltage protection and current-limiting shutdown. $\diamond$

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# PART 2. BASIC DIGITAL LOGIC 

The Basic Logic Gates. All digital logic circuits, from the simplest counter to the most sophisticated microprocessor, are made from interconnected combinations of simple building-block circuits called logic gates. There are four basic gates, and they are designated according to their function as YES, NOT, AND, and OR circuits. Each of these basic gates has one or more inputs, a single output, and a couple of power supply terminals.

Various combinations of the binary bits 0 and 1 can be applied to the inputs of a gate by allowing a low voltage to represent logic 0 and a high voltage logic 1 . This is called positive logic. In negative logic, the definitions are reversed.

The YES gate transmits the logic state ( 0 or 1 ) at its single input directly to its output. It's often used to interface logic circuits that are otherwise electronically incompatible. For this reason it's often called a buffer.

The NOT gate inverts or complements the logic state at its single input so it's often called an inverter. The NOT function is often indicated by a bar or vinculum over the symbol for an input or output that's been inverted. Thus if $A$ is 0 and $B$ is 1 , then $A=\bar{B}$. (The $\bar{B}$ is read and sometimes written "not B.")

The AND gate is a decision making circuit with two or more inputs. The output of the AND gate is logic 0 unless all the inputs (inputs A and B and $\mathrm{C} . .$. ) are logic 1 .

The OR gate is also a decision making circuit with two or more inputs. Its output is logic 0 unless any or all of its inputs (input A or B or $\mathrm{C} .$. ) are 1.

The operation of a gate can be de-
fined by a table that shows the combination of input bits that produces a particular output bit. Such a table is called a truth table. The truth tables and standard symbols for each of the four basic logic gates are shown in Fig. 1.

Compound Logic Circuits. Combining two or more of the basic gates into a compound logic circuit can provide some very important operating features. The two most important compound logic circuits are the AND-NOT and OR-NOT combinations. These are called the NAND and NOR gates and their symbols and truth tables are shown in Fig. 2.

As shown in Fig. 3, various combinations of NAND (or NOR) gates alone can simulate YES, NOT, and AND circuits. This is important, but the most fascinating characteristic of the NAND and NOR functions is their logic equivalence. Thanks to a rule known as DeMorgan's theorem, a positive logic NAND gate is equivalent to a negative logic NOR gate and vice versa.

You can prove this for yourself by writing the appropriate truth tables and finding that they are indeed identical. DeMorgan's theorem simplifies digital logic to the point where combinations of only NAND gates or NOR gates can implement any logic function. Figure 4, for example, shows how NAND gates alone can implement both the OR and NOR functions. Notice how NAND gates are used as inverters to change the inputs from positive to negative logic.

Complex Logic Systems. Simple and compound gates can be tied togeth-
er to implement a countless variety of logic functions. Some of the resulting logic systems contain only a handful of gates; others may use dozens or even hundreds of gates. All of these complex logic systems can be divided into two broad categories: combinational and sequential.

Combinational circuits are characterized by their fast acting operation. Exclusive of the brief time delay required for its gates to react to an incoming logic 0 or 1 (the propagation time), the output(s) of the most complex combinational circuit instantaneously reflects the pattern of 0's and 1's at its input(s).

Sequential circuits include storage or delay elements that permit the logic result of a previous input to directly influence a new input. This makes sequential circuits slower than combinational circuits. But it also makes possible important applications such as memory registers, counters, dividers, sequencers, and microprocessors.

## Combinational Logic Circuits.

 The simplest combinational logic circuit is the Exclusive-OR gate. The symbol and truth table for this circuit are shown in Fig. 5.Look at the Exclusive-OR truth table for a moment. The Exclusive-OR function is just that; it gives a logic 1 output only if one or the other of its two inputs is logic 1 . Otherwise the output is 0 . This is identical to the binary addition rules with the exception of the carry output needed for $1+1$.

It's easy to generate the carry output bit needed to use the Exclusive-OR circuit as a binary adder. Look at the logic


| A OUT |  |
| :---: | :---: |
| 0 | 0 |
| 1 | 1 |



| $\mathbf{A}$ | $\mathbf{B}$ | OUT |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 0 |
| 1 | 0 | 0 |
| 1 | 1 | 1 |



| A | OUT |
| :---: | :---: |
| 0 | 1 |
| 1 | 0 |



| $A$ | $B$ | OUT |
| :---: | :---: | :---: |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 1 |

Fig. 2. A NAND
 gate and a NOR gate, with their repective truth tables below.


AND


Fig. 1. The four basic logic gates: YES, NOT, AND, OR.


Fig. 3. Using NAND
gates to simulate other gates. At top, YES; middle, NOT; bottom, AND.


Fig. 4. Using NAND gates to prove NOR DeMorgan's theorem.

At top, OR; on the bottom, NOR.
circuit for an Exclusive-OR in Fig. 6. If you'll study the operation of this circuit, you'll find that the output of AND gate 1 provides the carry output we need. In the other circuit in Fig. 6, we use this carry output to form a circuit that can add any two binary bits. It's known as a half adder.

A half adder is useful, but it can only accept two input bits. To complete the binary addition rules, we need an adder circuit that will accept a carry bit as well. The circuit that accomplishes this goal is the full adder. As you can see in Fig. 7, a full adder can be made from two half adders and an OR gate.

It's possible to connect a string of adders together to form a binary adder capable of adding multiple-bit binary words. Figure 8, for instance, shows a 4bit adder that will sum two words applied to its inputs. Try adding $1101+0101$ using this adder to prove to yourself it really adds.

A binary adder forms part of a microprocessor's arithmetic-logic unit (ALU), a combinational circuit that performs addition, subtraction, and various logic operations upon two incoming words. The ALU is instructed what operations it is to perform by binary signals applied to its control inputs. We'll learn more about the ALU later in this course.

Encoders and Decoders. An encoder is a combinational network of OR gates that converts or encodes a nonbinary input into binary. For example, an octal-to-binary encoder has eight inputs (one for each octal digit) and three outputs (one for each binary bit). A logic 1 at one of the inputs produces the binary equivalent at the output

Encoders can provide other conversion operations, too. Keyboard encoders, for instance, convert individual key positions into their assigned binary words. An example is the ASCII (American Standard Code for Information Interchange) encoded keyboard, which generates the 7 -bit word 0100101 when the $\%$ key is pressed.

A decoder is a combinational circuit that converts a binary number at its inputs into a logic 1 at one or more of its outputs. In digital electronics it's often necessary to convert a binary number into some other format, and one common decoder application is the conversion of binary numbers into the format required to activate the appropriate segments in a 7 -segment decimal display.

Decoders are also used to decode binary instructions in a microprocessor,


Fig. 5. The combinational circuit at top providesan Exclusive-OR, as shown in the middle. The truth table is below.

Fig. 6. How an Exclusive-OR can be used to make a half-adder combinational circuit.


Fig. 7. Two half adders and an OR gate can be used to make a full adder circuit.

assist in the production of sequential timing signals for advanced logic circuits, and convert binary numbers into their octal, decimal, and hexadecimal counterparts. Figure 9 summarizes the operation of encoders and decoders.

Multiplexers and Demultiplex. ers. The multiplexer is the digital logic equivalent of a multiple-position rotary switch. A typical multiplexer is a combinational logic circuit that selects one of several input lines and applies any data on that line to a single output. A special set of address inputs determines which input line is selected.

One typical multiplexer has eight data inputs, three address inputs, and a single data output. When the address 101 is applied to the multiplexer, input 5 is connected to the output.

A common application for multiplexers is driving the readouts of pocket calculators to reduce the number of pin connections on the calculator's chip. The multiplexer lets all the digits in the readout share a common set of terminals. It activates each digit or one segment in all the digits in rapid succession to fool the eye into thinking the display is continually illuminated.
The demultiplexer transfers the binary data at its input onto one of two or more output lines. Like the multiplexer, an address input controls the output.
Demultiplexers are used with multiplexers to convert multiplexed data back to its original form. They can even function as decoders by applying a logic 1 to the single input and using the address inputs as data inputs. Figure 10 summarizes the operation of multiplexers and demultiplexers.

Sequential Logic Circuits. Unlike combinational logic circuits, sequential circuits have memory. Their output(s) can reflect the effect of an input that occurred seconds or even days earlier.
The simplest sequential circuit is the flip-flop. A microprocessor together with a read/write memory incorporates doz-ens-perhaps thousands-of flip-flops.
There are several different kinds of flip-flops, but all are capable of storing a single binary bit. This makes possible such applications as counters, dividers, memory registers, and others. Here are the four basic kinds of flip-flops.

The RS Flip-Flop. The simplest flipflop is made from two NAND or NOR gates with crisscrossed inputs and outputs as shown in Figure 11. This basic circuit is called a reset-set (RS) flip-flop


Fig. 9. An encoder is a combinational network that converts a nonbinary input to a binary output. A decoder does just the reverse.


Fig. 10. A multiplexer is the equivalent of a multiple-position switch. A demultiplexer converts multiplexed data back to original form.


Fig. 11. Simplest flip-flop is made from two NAND's or two NOR's with truth tables as shown.


| CLOCK | $\mathbf{s}$ | $\mathbf{R}$ | $\mathbf{a}$ | $\overline{\mathbf{Q}}$ |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 1 | NO CHANGE |  |
| 0 | 1 | 0 | NO CHANGE |  |
| 1 | 0 | 1 | 1 | 0 |
| 1 | 1 | 0 | 0 | 1 |

Fig. 13. A data, or
D, flip-flop is made by adding an inverter to input of one flip-flop.

Fig. 12. A clocked RS flip-flop is a sequential circuit with truth table as shown here.


D $\mathbf{a} \overline{\mathbf{a}}$
$\begin{array}{lll}0 & 0 & 1 \\ 1 & 1 & 0\end{array}$
or simply a latch. Figure 11 also shows the truth tables for NAND and NOR gate versions of the RS flip-flop.

Notice that the two outputs of the RS flip-flop complement one another. When $Q$ is logic 1 , the flip-flop is set. When $Q$ is logic 0 , the flip-flop is reset or cleared.

Clocked RS Flip-Flop. The basic RS flip-flop is asynchronous; it responds to inputs as soon as they occur. A way to synchronize the operation of the RS flipflop with other logic circuits is to gate its inputs so they can respond only when activated by a logic 1 from a clock. A clock is a sequential circuit that produces a stream of alternating 0 's and 1 's. Fig. 12 is a clocked RS flip-flop.

The Data or D Flip-Flop. The D flipflop is a further modification of the clocked RS flip-flop. As shown in Figure 13, an inverter is added to one of the two inputs of the flip-flop and the remaining input and the inverter's input are tied together. This guarantees that the inputs to the RS section of the flip-flop will always complement one another. And it insures that the logic state of the Q output will always correspond to the logic state of the D input.

The JK Flip-Flop. The JK flip-flop is a clocked RS flip-flop with a refinement that allows a logic 1 to be simultaneously applied to both inputs. Figure 14 shows the logic circuit and truth table for this flip-flop. The JK flip-flop can easily simulate any of the other kinds of flipflops, so it's commonly used in sequential logic circuits.

The JK flip-flop can be used to make a toggle or $T$ flip-flop. The $J$ and $K$ inputs are tied together and called the $T$ input. When a logic 1 is applied to $T$, the flipflop changes state or toggles each time a clock pulse arrives.

Storage Registers. A string of D flipflops called a register can be used to store a binary word. A register like this can be made far more useful by adding some combinational logic to simultaneously clear all the flip-flops to 0 when a logic 1 is applied to a clear input. A load input can also be added to force the register to ignore incoming data. When the load input is logic 1 , the input data will be accepted by the register when the next clock pulse arrives.

Data storage registers like this are sometimes called buffer registers. They're used in logic circuits and in microprocessor units to temporarily hold a data word.

Shift Registers. Considerably more


Fig. 14. The JK flip-flop is a clocked RS flip-flop that allows a logic 1 to be simultaneously applied to both inputs. Shown here is a NOF gate version with truth tables for $Q=0$ and $Q=1$.


Fig. 15. This shift register made from $D$ flip-flops accepts data a bit at a time and has a serial output as well as parallel outputs from each flip-flop.


Fig. 16. A four-bit counter made from T flip-flops that will count from 0000 to 1111 and then recycle.
versatile than the buffer register is the shift register shown in Fig. 15. This particular register accepts data a bit at a time (serial input) while making available the contents of all its flip-flops simultaneously (paraliel output). The data bits in the register are shifted right a bit at a time by clock pulses to make room for incoming bits.

Universal shift registers that can accept and output data as serial bits or parallel words as well as shift the data left or right are available. The various operations of a universal shift register are selected by applying logical 0's and 1's to an array of control inputs. Microprocessors incorporate at least one shift register to perform some of the data manipulation required to multiply and divide binary numbers.

Counters. Remember the toggle or $T$ flip-flop we discussed earlier? The $Q$ output of this flip-flop alternates between logic 0 and 1 for each incoming clock pulse: 0 ... 1 ... 0 .. 1 ... In other words, the Q output is logic 1 for half the incoming clock pulses. This means a single flip-flop can be used to divide an incoming stream of bits by two. The $Q$ output of a toggle flip-flop also counts! Thus, 0
1 ... is the same as counting from 0 to 1 in binary over and over again.
Higher capacity binary counters (and dividers) can be made from a string of $T$ flip-flops. Just connect the Q output of one flip-flop to the clock input of the next flip-flop. Figure 16, for instance, shows a 4-bit counter made from four T flip-flops. This counter will count from 0000 to 1111 and then recycle.

There are many different kinds of flipflop counters. The modulo of a counter specifies the maximum count it reaches before recycling. Modulo 10 counters are very popular because they recycle after the tenth input pulse and therefore provide a convenient way to count in decimal. They are often called BCD (binary coded decimal) counters. Their count sequence is $0000\left(0_{10}\right) \ldots 0001$ $\left(1_{10}\right) \ldots 0010 \quad\left(2_{10}\right) \ldots \ldots 1001$ $\left(9_{10}\right) \ldots 0000\left(0_{10}\right)$
Counters can have a variety of control inputs. A typical counter, for example, can count up or down. It may also have control inputs for clearing the count to all 0 's, presetting the count to any desired value, and enabling the counter to count. Finally, since counters store the accumulated count until the next clock pulse arrives, they can be considered storage registers.

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## Examines noise generators, instrument dynamics, and voltage control.

THE ELECTRONIC music synthesizer has changed the face of recorded music and live rock-band performances. This "instrument" can produce a myriad of unconventional, sometimes weird or eerie sounds. Yet it can also emulate the sounds of any conventiona! instrument.
There are two basic types of synthesizers: studio and performance. Modern studio synthesizers are made up of modular sections that consist of voltagecontrolied amplifiers, filters, oscillators and noise generators, modulators and other devices. These modules can be interconnected in virtually any order by plugging their inputs and outputs together with "patch" cords. The output of almost any given module can be used either as part of the tone you eventually hear, or as the control voltage for another module. It all depends on how one "patches" the elements together
Synthesizers designed for live performances do not use patch cords, as the time lost in changing patches during a performance would ruin the musical continuity. Here, the various modules are hard-wired together and the sounds are changed by a host of conveniently
located switches and potentiometers, used much as stops are on an organ. Since a player can handle only a limited number of controls efficiently, performance synthesizers can't be made as flexible as the studio type.

How It Works. A synthesizer's output waveforms and control signals can be considered as a vast kit of parts from which musicians can assemble any desired sounds. The different parameters of each note-pitch, overtone structure, attack time, duration, and decay--are, in conventional instruments, fixed within narrow limits. In the synthesizer these parameters are independently, and almost infinitely, variable

The modules that control those parameters are shown in Fig. 1. Pitch and overtones are controlled by the voltagecontrolled oscillator (vco) and by the voltage-controlled filter ( vcf ). The note's attack, sustain, decay and release-its "envelope" in time-are controlled by the voltage-controlled amplifier (vca), which in turn is controlled by the envelope generator. The latter is sometimes called ADSR, the initials of the
four parameters it governs-attack, delay, sustain and release. The musician's control input for these modules is usually a keyboard

Normally, pitch is governed by the voltage-controlled oscillator. As the name implies, its frequency varies with the control voltage fed to it. But the vco also has an effect on overtone structure, or timbre. Its output can, on most synthesizers, be a pure sine wave, with no overtones at all, or a ramp, triangle, pulse, or square wave (see illustrations in Fig. 2), each of which has a different mixture of overtones.

Conventional instruments have somewhat similar overtone structures. A violin note, for instance, begins as a ramp waveform, the bow grabbing the string and deflecting it until the string's tension overcomes the friction of the bow, and the string snaps back again. A saxophone reed's opening and closing makes the equivalent of a square wave.

Don't think that these waveforms fully represent the sound of the instrument. They don't. The actual sound produced depends on the instrument's resonances, which accentuate or attenuate



Fig. 2. Typical vco waveforms.


Fig. 3. Filter modifies timbre.

Individual harmonic components in the raw waveform. The sound of a saxophone may begin as a square wave, but after passing through the instrument, it appears rather like the "ringing" waveform shown in Fig. 3.

The synthesizer's voltage-controlled filter has a similar effect to that of the saxophone's bell or the violin's hollow body. Whereas the resonators of mechanical instruments are reasonably fixed, a synthesizer's equivalent module (its vcf) can be used as either a lowpass, band-pass or high-pass filter over a frequency range of many octaves, while its " $Q$ " (curve sharpness) is adjustable, too. The vcf can also give pitch to the output of an unpitched signal source, the 'noise generator.' At first it may seem odd to include such a module in a synthesizer, since so much attention is paid to eliminating noise in electronic systems. There are many applications, though, as illustrated in the next section.


Fig. 5. The combination of a vca and ADSR.

The Noise Generator. The noise generator is used in simulating such instruments as the snare drum and the high-hat cymbal, or such natural sounds as wind, surf, explosions and thunderclaps. Using the synthesizer's other resources, it's also possible to create sounds that a real drum or real surf could never make. Let's try, for example, to invent a new musical instrument, with a voice like the wind. With enough experimentation, I'm sure that would sound much like the wind-perhaps a pipe with a plugger of some kind in it. With enough practice, we might even learn how to play it so that we were on pitch (and so that it didn't break into oscillation, which would ruin the effect). But it would take some work.

With a synthesizer, this type of task is almost ridiculously simple. First, we substitute a noise generator for the vco (Fig. 4). Then we recover pitch information from the noise (which contains all possible pitches) by passing it through the filter. To play this new instrument, we apply the keyboard's output voltage to the vcf's control input. Now, each keystroke will shift the filter's frequency range, controlling the pitch of the note.

Since the wind usually builds and dies away slowly, we'll want our instrument to have a slow attack and decay, and to sustain as long as we hold down the key. This is where the attack, delay sustain and release, or envelope generator, comes in.

Instrument Dynamics. Every instrument's note varies in amplitude over time. This variation-the instrument's "dynamics"-is one of our chief clues to which instrument we're hearing. (Note how odd most instruments sound when recordings of them are played backwards, even though pitch and overtones are the same.) In some instruments, the sound of each note builds up (attacks) quickly, and dies away (decays) slowly. In others, the attack may be slow, and the decay rapid. Some instruments' outputs begin to die away as soon as a peak is reached; others "sustain" for a time before decay begins.

Though it may not be immediately apparent, all of these characteristics are functions of the way energy is added to each instrument's mechanical system. In instruments where the energy is added all at once (whether by hitting it with a stick or strumming it with a plectrum), the note's volume will be at a peak immediately after striking; and since all of the energy goes in at once, there will be


Fig. 6. By varying its ATTACK, DECAY, SUSTAIN and RELEASE, envelope generator can produce different output waveforms.


Fig. 7. An integrator adds glissando to a voltage controlled oscillator (vco).


Fig. 8. Summation of control voltages
results in special effects like vibrato.


Fig. 9. Varying control voltages from vco or ADSR produce unusual effects.
none left over to sustain the sound-it's downhill all the way until the next note is played. This quick attack and moderate-to-slow decay is typical of instruments in the percussion family: guitar, piano,
drums, xylophone, the heads of contestants on the "Gong Show," and such.

It would be natural to assume that we similarly "strike" the oscillator somehow to simulate this type of sound. But natural as this may seem, that's not how it's done. There's an easier way.

In a synthesizer, the oscillator runs all the time. Whether we hear it or not depends on whether the voltage-controlled amplifier (the last element in our "classic" patch) is on or off. And we control the dynamic shape of the note we're building by controlling the rate at which that vea turns on or off.

What controls the vca is the envelope generator or ADSR (Fig. 5). When it receives a trigger signal (usually from the keyboard), the ADSR's control voltage rises to a peak at a rate determined by the setting of the attack control.

After reaching this peak, the control voltage begins to decrease at a rate set by the DECAY control. It doesn't fall all
the way back to ZERO, though, just to a level set by a third control, sustain, where it holds for as long as the original triggering signal is present. Only when the trigger signal goes away (usually when the key on the keyboard is released) does the control voltage fall from the sustaining voltage to zero, and then at a rate set by the RELEASE control.

By adjusting these four controlsATTACK, SUSTAIN, DECAY and RELEASEwe can cause the ADSR to generate a control voltage which, when used to determine how much signal passes through the vca, simulates the dynamics of any natural instrument (Fig. 6). It can also produce dynamics that would be difficult to produce with a mechanical instrument. An example would be combining the percussive attack of a drum with the sustaining properties of an oboe.
A synthesizer's oscillators and filters operate over an impressively wider frequency range than their mechanical counterparts, and have more modes of operation. A single electronic unit can even simulate a number of properties that might be mutually inconsistent in a mechanical device. But that's only part of the story.

Voltage Control. The real story is voltage control. A synthesizer's oscillator and filter frequencies and amplifier gain are all functions of the control voltages applied. That's more significant than it may look at first.
The keyboards used with most synthesizers are nothing but switch-selectable voltage dividers. Press a key, and a voltage that represents that key appears at the keyboard's output. Press another key and the voltage instantaneously changes to a new level. If the voltage from the keyboard is being used to set the pitch of the oscillator, the output instantly steps from the first note to the second. There will be times, though, when it will be desirable to produce a sound that doesn't instantly change from one pitch to another, but rather glides (glissandos) between notes. Because of voltage control, a simple integrator (nothing more than a register, capacitor and buffer amplifier) placed in the key-board-to-oscillator control voltage path will produce this effect by slowing down the change (Fig. 7).

Other special effects can be added easily by summing control voltages from several sources. Vibrato, for example, which is a slow-speed modulation, is realized by summing a slowly varying ( $7-12 \mathrm{~Hz}$ ) control voltage into one of the
control voltage inputs of the vco (Fig. 8). Applying this same modulating voltage to the vca produces tremolo, a slow variation in the output amplitude.

Since the center or corner frequencies of the filters used are also functions of summed control voltages, several vastly different acoustical instruments can be simulated simply by changing the source of the control voltages that we apply to this element.

If the voltage comes from a fixed source, the output would simulate an instrument with a fixed resonator such as a guitar or piano. However, changing the control voltage allows one to instantly change the properties of the resonator (try that with a piano). Moreover, if the control voltage comes from the same source that is supplying pitch information to the oscillators, we can simulate the properties of instruments with variable resonators, such as reed and wind instruments (Fig. 9).

Filter control voltages that vary with time can vary independently of the voltage that controls the oscillator, allowing the filter to sweep the harmonic content of the oscillator's output cyclically (a cross between tremolo and vibrato that is rare among natural instruments). Or, the control voltage can be derived from an ADSR, for a "waa-waa" effect.

Getting back to our "wind" instrument, the desired sighing quality can be easily attained by simply adjusting the envelope generator for a slow attack and decay, a high sustain level, and a slow release. Tiring of that, one need only change the settings of the ADSR to fast attack, no sustain, and moderate decay, to set up an entirely new "instrument." This one will sound like a "chromatic high hat cymbal," one that is played from a keyboard. This would not be only for rhythm, but also as a lead voice in a composition. (Try that with your pipe and plugger!) The pitch and range of the filter can be changed, too. Bring an oscillator in, and begin simulating snare drums that can be played in pitch. There's no end to the possible variations achievable with patch-cord type synthesizers.

If you have access to a synthesizer, you'll have instrumental and other sound available for recording right at your fingertips. And if you think that you don't like synthesizer sounds, perhaps you should begin listening more carefully to modern pop, jazz and rock recordings. The next time you hear what appears to be a chorus of violins, read the album's liner notes-they might not be real violins at all!

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## Build a <br> COMPUTER MUSIC BOX PERIPHERAL

UDGING from the many commercial plug-ins available, computergenerated music appears to be the "in" thing today. If you have found the singlebit method is too limited and the digital/
analog converter approach too expensive, the low-cost (less than \$30) Music Box described here may be just for you.

The Music Box has a 12-note, fouroctave range. It can be used with any
computer that has a parallel output port. And to simplity its use, no strobes or other handshake signals are required.

The Music Box circuit is not limited to making music. It can easily be pro-


C1. C2-0.01- $\mu \mathrm{F}$ Mylar capacitor
DI through D12-1N914 diode
ICI-74154 4-line to 16 -line decoder
IC2-555 timer
IC3. IC4-7473 dual JK flip-flop
1C5-740) quad NAND gate
QI-2N3906 transistor
R1 through R13-2000-ohm linear-taper po trimmer potentiometer
The following are $1 / 4$-watt, $5 \%$-tolerance resistors:
R14-5600) ohms.
R15.R19. R20-1000 ohms
R16- $10.0 \times 0$ ohms
R17-22.(NO) ohms
R18. R21, R22, R23, R24-100,000 ohms
Misc.-IC sockets (optional): suitable prototyping berard: suitable enclosure; etc.

As bits 0 through 3 from computer change, the vco changes frequency. Other four bits ( 4 through 7) determine the octave of the audio output.
grammed to generate a mix of tones, up to a total of 16 , for use as test and re-mote-control signals.

Circuit Operation. The circuit (see schematic diagram) can be broken down for discussion purposes into three major subsections: note decoder/selector, voltage-controlled oscillator (vco), and octave decoder/selector.

The note decoder/selector consists of integrated circuit IC1, a 4 -line to 16 -line decoder. As the four control bits from the computer (bits 0, 1, 2, and 3) are entered into IC1, one of the 16 output lines is driven low. When the output line goes low, it allows its associated diode (D1 through D12) and series potentiometer (R1 through R12) to control the voltage and, hence, the frequency of the vco made up of IC2, Q1, and their associated components. Since only 12 tones per octave are used in music, output lines 13,14 , and 15 of IC1 (pins 15, 16, and 17) are not used. (These three lines can be used to control an external device, as we will discuss later.) When IC1's output 0 at pin 1 is low, the vco is cut off to provide a no-note condition.

Timer IC2 is configured as an oscillator, with transistor Q1 serving as a volt-age-controlled resistor that works in conjunction with frequency-determining capacitor C1. By varying the bias applied to the base of Q1, the output frequency of the vco system can be made to vary.

Resistor R18 determines the low- and R16 and R17, the high-frequency ends of the range. Capacitor $C 1$ can be

## TABLEITHE WELL-TEMPERED MUSICAL SCALE

| Control bit | Frequency <br> $\mathbf{7 6 5 4 3 2 1 0}$ <br> $(\mathbf{H z})$ | Note <br> 5th Octave |
| :---: | :---: | :---: |
| 10000000 | 0 | Off |
| 10000001 | 523.25 | C |
| 0010 | 554.37 | $\mathrm{C} \#$ |
| 0011 | 587.33 | D |
| 0100 | 622.25 | $\mathrm{D} \#$ |
| 0101 | 659.26 | E |
| 0110 | 698.46 | $\mathrm{E} \#$ |
| 0111 | 739.99 | F |
| 1000 | 783.99 | G |
| 1001 | 830.61 | $\mathrm{G} \#$ |
| 1010 | 880.00 | A |
| 1011 | 932.33 | A |
| 1100 | 987.77 | B |

changed to select the desired frequency range. The output of the oscillator at pin 3 is fed to the flip-flops in IC3 and IC4 for octave generation.

The four octaves of square waves generated by IC3 and IC4 are summed with the four octave-control bits (bits 4, 5,6 , and 7) by the four AND gates in IC5. The resulting selected octaves are mixed in R21 through R24 for application to an external audio system. Any combination of four octaves can be selected simply by changing the status of bits 4 through 7 . If all octave bits are low, no tone appears at the output. Note that no status signals are required.

Since the audio output consists of square waves, it is not difficult to introduce various types of filters to create different sounds.

Construction. The entire circuit can be assembled on any prototyping board that can be connected to the parallel output port of the computer in which the Music Box is to be used. The power for the Music Box can be taken from the +5 -volt and ground lines in the computer. Alternatively, you can use an external power supply rated at 100 mA minimum. In either case, a common ground must be used between the Music Box and computer.
You can use sockets for the IC's if you wish and small board-mounted trimmer potentiometers for R1 through R13.

Calibration. Although the Music Box was designed for use with a computer, it does not require a computer for calibration. All you need is a 5 -volt dc power source and an audio system. A frequency counter will simplify calibration but is not a necessity.

Before applying power to the Music Box, set R1 through R12 to their maximum series resistance and R13 to its center of rotation. If you have a frequency counter, connect it to the TEST POINT. Otherwise, connect the output of the Music Box to an amplifier/speaker combination so that the pitch of the output signal can be compared with the sound of a known musical instrument.

Using temporary jumpers to the +5 volt (1) and ground ( 0 ) lines, set the control bits to the values given in Table I and adjust the corresponding trimmer potentiometer (R1 through R12) to obtain the indicated frequency (or the correct tone when compared with the sound from a musical instrument). If the entire range cannot be obtained, readjust R13 and perform the above procedure again.

## TABLE II-TEST VALUES

| Note | Number value (n) |
| :--- | :---: |
| Off | 0 |
| C | 1 |
| C\# | 2 |
| D | 3 |
| D\# | 4 |
| E | 5 |
| E\# | 6 |
| F | 7 |
| G | 8 |
| G\# | 9 |
| A | 10 |
| A\# | 11 |
| B | 12 |
| Octave | $n+m b e r$ value |
| 5 | $n+64$ |
| 4 | $n+32$ |
| 3 | $n+16$ |

Note: $B_{5}$ is the highest note $(n=140)$
$C_{2}$ is the lowest note $(\mathrm{n}=17)$
$\mathrm{C}_{5}$ is middle $C$
$A_{4}$ is $A_{440}$

Operation and Use. Since there is no data latch, the Music Box tracks the data that appears at the parallel output port. Connect the common ground and eight data lines between the Music Box and the output port. To test the system, execute an output of the number value that corresponds to that note as given in Table II.

The software program you write will depend on the music requirements. Arrays can be used to store melody information and loops can be used to control the length of the note.

The four-octave range of the circuit can be shifted by halving the value of $C 1$ to raise the pitch one octave or it can be doubled to lower the pitch one octave.

Other Uses. The three decoded outputs from IC1 at pins 15, 16, and 17 can be used to trigger a percussive device (such as the "Cabonga" featured in the August 1977 issue of Popular ElecTRONICS) or to latch an external control device. These decoded output signals are TTL level. If music is not what you want, you can use the circuit to provide 16 preadjusted tones for use in testing or remote-control applications (see "Computer Bits," August 1977). To obtain all 16 tones, you must add diodes and potentiometers to the circuit as shown for the other outputs.


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EXAMINING a CB radio mobile handbook published in 1962 underscores how far this low-cost, two-way radio communications system has traveled . . . and how much the same it is today. One striking sentence published fully 16 years ago in this book (four years after the CB band was moved to 27 MHz from higher frequencies authorized in 1947) was: ". . . the Commission never envisioned the CB'er who owns one set and talks to other licensees!" That was the reason given then for the FCC's initiation of the famous five-minute conversation limit rule.

The Citizons Band Today? It's a more meaningful form of short-distance radio communications than ever for the general public. There are said to be some 25 -million CB transceivers spread around the U.S., creating a viable local network for emergency communications, informational assistance for motorists, and social conversations. And it's got a new name-the Citizens Band Radio Service, which is a subdesignation in the Personal Radio Service.
The new 17 CB channels added last year will most certainly be used more and more since the FCC banned sales of new or used 23-channel transceivers as of January 1, 1978 (excepting handheld types, which may be marketed without 40 -channel capability until August 1, 1978). Although there was an increase of 74 percent in channel availability that now stretches from 26.96 to 27.41 MHz , channel 9 remains the emergency calling channel.

A host of modified rules and decisions over the past year or two has changed the complexion of CB for the better. For example, one can get on the air legally without waiting for a station license by using Form 555-B packed with each new CB transceiver; subpart D of the FCC's Part 95 Rules and Regulations
has been broken out as a separate publication to be packaged with new transceivers; there is no charge for a CB license; new 40-channel transceivers must meet more stringent technical requirements; and, most recently, proceedings on creation of a Class $E$ (224-225 MHz) CB band have been terminated

Other decisions aftest to CB radio's growing importance in our lives. The U.S. Coast Guard, for instance, announced it would install CB gear at Coast Guard Search and Rescue stations throughout the U.S. in time for the 1978 recreational boating season. Federal funds are now issued to states with organized CB programs, with the National Highway Traffic Safety Administration stating that the "Citizens Band offers the only existing method convenient to the public by which the motorist can enter the emergency response system from his/her vehicle." As a consequence, it will be progressively easier to reach a highway patrolman for aid as time goes on. Right now, in fact, CB radio public safety programs have been established by 94 percent of the nation's state police organization's, with CB radios already installed in 48 percent of the police vehicles in 34 states. Supporting the federally funded NEAR (National Emergency Aid Radio) program, REACT, the largest voluntary CB emergency service organization, was recently awarded a contract to develop a CB channel-9 monitor training program for public safety officers and volunteers.

The foregoing does not mean that CB radio is cleansed of all its problems. There are still violators of FCC rules, the most important being: (1) Out-of-band communications; (2) Overpowered transmitters; (3) Indecent language; (4) Communication over 150 miles away (skip); and (5) Failure to identity by callsign. Furthermore, CB-caused TV and
audio interference is still a challenge.
The CB industry itself is in the process of stabilizing. The problems it had appear to have been precipitated by the FCC's announcement in July 1976 that 40-channel CB transceivers could be sold as of January 1977. While people waited for the introduction of new 40channel models, 23-channel inventories mounted. The result was a drastic reduction in prices of 23 -channel models that proved irresistible to the public, which, in turn, caused discounted prices on 40-channel models. The upshot was that most CB manufacturers lost money during this period, while the public was getting the best values in memory. With a more mature market, you can anticipate that selling prices will edge upward.

Where is CB Coing? In the short term, more and more single-sideband and increasingly sophisticated AM CB transceivers are likely to be purchased. The reasons are multifold. For one, advanced technology presents prospective users with a better rig, both in raw performance and in convenience features: electronic digital readouts, cal-culator-type control keypads, memory, automatic channel scan, separate VFO for receive-only with a second digital channel display, improved front-end overload circuitry, superior selectivity, precision SSB tuning, remotely controlled mobiles, and so on. Secondly, many millions of CB'ers are ripe for upgrading their present equipment, having tasted the benefits of two-way radio communication. (It's also expected that many CB'ers in this group will turn to amateur radio, expanding their communication horizons while maintaining CB rigs for emergencies.)

Technological developments do not necessarily mean exhobitant cost, either. For example, National Semiconductor's single-chip frequency synthe-
sizer/programmer for CB radio use is claimed to reduce component count by as much as $60-70$ percent. Features include two-speed-slew, up/down channel selection capability, busy or clear channel scan, channel memory, LED blanking control in the event an auto clock circuit is added, channel 9 selectability, and more. ECL and I2L fransistors are fabricated on the same chip.

You can expect standardization of selective calling signals somewhere down the road for CB radio. A technical committee of the EIA is already working on this for add-on to existing models and incorporation into future ones. There will be ongoing efforts to reduce interference on TV sets, too. CB radio manufacturers have already suggested that $C B$ harmonic emissions be limited to -75 dB as compared to the present -60 dB . Further, FCC Commissioner Lee told TV manufacturers to get ready for an FCC drive to upgrade interference capabilities of TV receivers.
With tew FCC enforcement personnel, it's expected that self-regulation efforts will be pursued to minimize communication violations by CB'ers. The

FCC may well utilize the services of volunteers in a program of first-level screening and offender identification, a proposal already being considered. Also, a new series of radio Public Service Announcements has been prepared to remind CB'ers on proper usage of personal communications equipment. The on-again/off-again ATIS (automatic transmitting identification signal) proposals appear to be shelved for now.

There are many spectrum alternatives being analyzed to cover future growth of personal radio services. None seem imminent at this time, but somewhere down the road such a decision would have to be made to avoid channel congestion. Right now, though, the new 17 channels are essentially clear and clean, providing users with fartherreaching signals than they got with the older channels.

At some distant time, one may also expect a CB rig to perform as a true mobile telephone. Interestingly, International Resource Development Inc., a market research company, does indeed project a future merger between CB and the telephone, as well as skyrocketing
growth in personal radio use after a short pause. Among other prospects is the use of repeaters in sparsely populated areas to extend a CB radio's range.

It would seem appropriate, too, in future planning to carve out SSB-only channels. To take full advantage of the spectrum-saving function of SSB, even at present, would require tighter technical specifications so that both lower (most popular) and upper sidebands on one channel can be used optimally

Clearly, CB radio is no longer a fad, The over-riding reason for purchasing a CB rig, in fact, is to handle distress situations, as borne out by a recent study by the International Trade Commission. This isn't surprising since more than 40 million emergency and assistance requests are reported to be handled annually by the CB Radio Service, Equally impressive, the former Superintendent of the Missouri State Highway Patrol ádvised that lapsed time between occurrence and notification of accidents was almost halved when reported by $C B$ ra dio as compared to conventional means.

See "Editorial" for report on CB at the 1978 Consumer Electronics Show. 厄

BY IVAN BERGER Senior Editor

ANEW breed of super mobile CB transceivers has been introduced to the marketplace this year. They feature higher performance capabilities than ever before, lower interference emission and more convenience features. Each CB manufacturer's best AM/SSB and AM model is listed in the buyer's guide on the following pages.

Specifications. All of the top-of-theline SSB and AM mobiles listed in our chart cover 40 channels and virtually all manufacturers claim $100 \%$ modulation and the maximum power that the law allows: 4 watts on AM, 12 watts PEP on SSB. All háve squelch, on-off-volume knobs and external speaker jacks; nearly all (except most of the control-in-themike remote-mount models) have $S / r-f$
metering, and virtually all have digital channel displays. Because of these similarites, you won't find those features in the table. Sensitivity figures for about $90 \%$ of these rigs are $0.5 \mu \mathrm{~V}$ for 10 dB ( $\mathrm{S}+\mathrm{N}$ )/N on AM and $0.25 \mu \mathrm{~V}$ for SSB, and most makers rate audio power output at the $10 \%$-distortion point. Exceptions to these specifications are noted under "Remarks."

Two specifications measure a CB rig's ability to reject unwanted signals. Adjacent Channel Rejection, as the name implies, tells how well the rig can reject signals of channels on either side of the tuned-in one. It's measured at $\pm 10 \mathrm{kHz}$, of course. The higher this number, the better. Selectivity here is the bandwidth or i-f "window" at which the receiver's response falls by 6 dB . The narrower
this frequency range, the greater the immunity to "splatter" from overmodulating stations on adjacent channels. If it's too narrow, however, voices will sound unnatural and that's not good either.

Image Rejection measures the receiver's ability to attenuate an undesired image signal generated by the converter stage. Spurious Response Rejection measures the transceiver's resistance to spurious signals created by the interaction within the tuner front end of strong external signals. The higher both figares, the better.

Most mobile sets will operate on both negative-ground cars (the most common typé) and positive-ground ones. Where the manufacturers supplied the information, we show the permissible ground polarities as either " $\pm$ " or "-".

## About This Month's Cover

AIthough the CB mobile transceiver illustrated on our front cover doesn't exist, most of its features are available in one or another of the new CB models available today. With more and more CB manufacturers limiting much of their development and manufacturing efforts to "high-end" models, it would not surprise us to find all the features in our imaginary model combined on one chassis at some future time.

Like virtually all of today's top CB transceivers, our dream model has electronic tuning and LED digital channel readout. This permitted us to include several tuning options that are currently available (though not all on the same model), plus a few practical approaches we haven't seen yet.

The UP and DOWN buttons (duplicated on the mike) let you scan manually to any desired channel, while the keyboard lets you jump to any channel directly. If you're trying to hold a conversation, but find the channel is crowded, press our dream rig's SCAN VACANT rocker switch. (It's one of the controls in the group labelled SCAN on our drawing-detailed switch designations aren't shown.) This will find and display the nearest vacant channel number. After you tell the party with whom you were modulating what channel is vacant, pressing the switch the other way, to TUNE VACANT will move
you directly to that channel. If you're looking for a conversation, the SCAN ACTIVE button will find it for you. SCAN MEMory looks for activity, too, but only among the stations you've programmed into either of the two memories: M1 for the channels you use within your normal driving area, or M2 for channels you use when driving elsewhere.
selective call facilities like our dream rig's should be available soon: one button, under the numerical keypad, programs in the selective-call number to be transmitted; the button beneath it is used to transmit that calling code. Two other keys work with the number pad to set your own call number and the channel or channels you expect to be called on. After that, any properly coded calls on any of those channels automatically take priority over the one you're using.

ON-OFF and VOLUME controls are separate, so you needn't reset the volume control whenever you turn the rig on. The VOLUME, SQUELCH and AM/SSB Selector settings can be determined by touch, for safety, and the two SSB con-trols-the sideband selector and the cla-rifier-are grouped together. squelch control setting can be determined by touch, as can that of the AM/SSB selector just below it. The CLARIFIER control is to the AM/SSB selector's left.

Grouped under qUALITY are: an ANL
switch; a two-position NOISE BLANKER (for a choice of blanking time), a whisTLE FILTER and a bROAD/NARROW i-f selector. To the left of the foregoing are RF GAIN and MIKE GAIN knobs; to the right is a TONE control. Still further right is the TALK TIMER It squawks when you've exceeded the statutory 5 -minute limit on talking time, and its surround ring blinks as you approach that limit.

The optional clock module counts down from 5 minutes whenever you press the talk switch on the mike. It also tells you the time of day, between transmissions; and has an elapsed-time mode, too. A separate hours reset button simplifies resetting when you drive from one time zone to another. An interlock button prevents accidental resets.

The meters, of course, are as large as practical. In addition to the switchselectable SWR metering, there's an alarm LED to warn you when the SWR approaches limits that could damage your transmitter-if the antenna is damaged, for example.
If our dream rig looks unusual, it's because we paid more attention to human engineering than to styling. Hence, the wide variety of controls' shapes and sizes (for easy touch identification) and the angled keypad (a more natural angle for the driver's hand with center buttons recognizable by touch.

Maximum current drawn by the rig is shown in the chart, too

Listed Features. In addition to the basic features found on virtually all CB transceivers, there are several fairly common extras noted here. (For rarer ones, see "Remarks".) The Automatic Noise Limiter (ANL) circuit clips noise pulses that ride in on the received sighal, but not noise which was part of the signal as it was transmitted. The Noise Blanker circuit momentarily squelches noise within the signal. It is more effective than ANL on AM, and quite a bit more effective on SSB. Where manufacturers have specified that their ANL or noise blanker controls are switchable, we've marked them with an " S ". Switching out the ANL will usually increase volume and decrease distortion slightly. Switching out the noise blanker should have no audible effect (save for an increase in noise.) All too often, however, this will increase the receiver's apparent
sensitivity, rather as if the squelch had been turned down a bit

Some rigs have R-F Gain controls to prevent strong-signal overload; an " S " here means a two-position switch (sometimes labelled Local/DISTANT) instead of a variable knob. Mike Gain controls assure maximum legal modulation even if you're speaking softly-on most sets, automatic level controls perform this function). Next comes a Tone control, followed by Up/Down Tuning (as in our dream rig).
Remote-mount means that the bulk of the transceiver circuitry is in a featureless, concealable box, with most or all controis on the microphone. (Some sets have a few controls or indicators on a separate speaker box.) PA switches are found on virtually all CB rigs; where the makers specified a Separate PA Speaker Output, that column is checked. A LED Dimmer brightens the channel numbers for daytime viewing and dims them at night to prevent glare.

Many sets today have various Priority Channels, allowing you to switch instantly to channel 9 for emergencies, to channel 19 for traffic advisories, or perhaps to some other channel or channels of your choice. (These are given only in the table for AM rigs.) Some will also scan the priority channels (" $S$ " in our chart). Switching to one when there's activity. Others will, like our dream rig, scan for active channels ("Sa") or for vacant ones (" Sv "). Since this feature is rare on SSB rigs, we've used Clarifier Range for that column in the SSB ta-ble-priority and scan features, if any are also under "Remarks."

Indicators for Percent Modulation and SWR may be either meters (" $M$ ") or LED's ("L'). Modulation meters differ from the usual "RF" meters primarily in being calibrated in percentages, rather than in arbitrary units. If LED's are used for SWR, they usually serve only to warn when SWR has become high enough to possibly damage the transmitter.


## SUPER AM MOBILE TRANSCEIVERS



Symbols and abbreviations: ${ }^{\text {a }}$ AM; L=LED indicator; mameter; S=switched; "SSB; Sa=scan for active channel; Sv= scan for vacant channel; Sescan(in Priority-Channel column); "Distortion unspecified; U\&L=Upper \& Lower sideband indicator 11ghts;
*See "Remarks" for distortion level;

## SUPER A H M MOBILE TRANSCEIVERS Continued



Symbols and abbreviations: ${ }^{\text {am }}$; L=LED indicator; M=meter; S=switched; ${ }^{s}$ SSB; Sa=scan for active channel; Sv= scan for vacant

*See "Remarks" for distortion level;

# Choosing a Mobile CB Antenna - How to Select and Install Mobile CB Antennas. 

BY JOHN J. McVEIGH, Asst Technical Editior

WHERE range is concerned, the only factor truly under the control of the CB'er is the antenna-its type and where it is installed.

Antenna Basics. All antennas employed in two-way communications perform two functions: (1) accept r-f power from the transmitter and radiate it into space, and (2) capture a portion of the radio signals passing by and present them as small voltages to be processed by the receiver. For maximum range, antennas should perform these two functions as efficiently as possible.

The elementary antenna from which most others are derived is the half-wave dipole. It is composed of two quarterwavelength conductors and is fed in the center by a transmission line. The dipole can be installed so that its conductors lie in the horizontal or vertical plane.

Although a horizontal dipole possesses some qualities desirable in CB mobile communications, such as its greater ability to reject ignition noise, it's size makes it impracticable. Only a small fraction of all the vehicles on the road could accommodate a dipcle 17.2' (5.2 $m$ ) long! The vertical dipole, shown in Fig. 1A, is a more realistic alternative.

One of the characteristics of a dipole is directivity. That is, the antenna works better in some directions than others. If it is mounted in free space, the dipole's radiation pattern is as shown in Fig. 1B. It is most effective at right angles to itself, and least effective off its ends.

To an observer in the same plane as the antenna (standing upright and looking directly at it), the antenna appears to be omnidirectional. For a given radiation angle (the angle at which the signal takes off from the antenna), the horizontal radiation pattern of the vertical.dipole
is circular-the antenna receives signals from or radiates them to all points of the compass with equal facility. In other words, it is omnidirectional. Keep in mind, however, the three-dimensional aspect of the pattern. The vertical dipole is most effective when signals strike or leave it at right angles.
The vertical dipole's omnidirectionality in the horizontal plane means that no loss in signal will occur if the mobile you're talking with up ahead makes a left or right turn or follows a curve in the


General arrangement of a vertical dipole is shown at (A); while $(B)$ is its radiation pattern in free space

road. If horizontal dipolés were employed, each would end up in the null of the other's radiation pattern. It is true that the described patterns are those of antennas in free space, and that proximity to the earth and metallic objects distort them. But even real-life horizontal dipoles display some directionality.

Another beneficial quality of the vertical dipole is its low angle of radiation, especially when mounted near ground. If it is mounted at right angles to the earth's surface, it is most sensitive to (and best radiates) ground-wave sig-nals-those travelling parallel to the surface of the earth. In line-of-sight communications, the ground wave predominates. The vertical dipole will therefore reject skip signals to some extent, and will be most efficient for ground-wave mobile-to-mobile and mobile-to-base communications.

There are, however, some disadvantages associated with vertical antennas. Because they are omnidirectional in the horizontal plane, they will bring in stations from all points of the compassboth wanted and unwanted. On transmit, signal power will be radiated in all directions, not only toward the stations with whom you are communicating, but also areas where there are no stations.

The major disadvantage associated with the vertical half-wave dipole is its size-about 17.2 feet ( 5.2 m ) at CB frequencies. A mobile antenna of that size would not only be physically unwieldy, but would also be highly vulnerable to damage from shocks inflicted by highway overpasses, overhanging tree branches, etc. Full-size, half-wave "coaxial" vertical antennas are commonly employed at base stations and on CB-equipped pleasure craft, but their dimensions rule out their use for mobiles.

Ground Planes provide a solution to the vertical dipole's height problem. If a quarter-wave vertical conductor is placed over a large horizontal conductive sheet, a "phantom" quarter-wave element is generated. The conductive sheet, called a ground plane, acts like a mirror and supplies the antenna element necessary to form a half-wave radiator. Ideally, the ground plane should be a perfectly conducting disc with a radius 'large as compared to the wavelength at the operating frequency:

Practicality imposes several limits on the structure and composition of the ground plane. Even at base stations, a metallic disc with a radius of only a quarter wavelength (about 8.6 feet or 2.6 m )

## Choosing Mobile Antennas contiriued

would be unwieldy. Four radial wires, each one-quarter wavelength, provide a ground plane at many CB base installations. The mirror image produced by the ground plane permits the use of an antenna only a quarter wavelength high.

In mobile applications, four radial wires are impractical. Instead, the metallic vehicle body is used. Although a far from ideal ground plane, the car body will work-to an extent-and is eminently practical. However, the vehicle body must be metallic

Loading. Even though the use of a ground plane reduces the height of a vertical antenna by one half, a quarterwave whip at CB frequencies (104" or 2.6 m ) is rather large. Some CB'ers install such antennas on their vehicles, but for practical reasons they can mount them only on a bumper. (That is not a very good mounting location, as will be developed later.)

When positioned over a ground plane, a vertical whip will resonate at the frequency for which its length is one-quarter wavelength. That is, its feedpoint impedance will be purely resistive. If the whip is too short to be a quarter wavelength at the operating frequency, its feedpoint impedance will contain some capacitive reactance. CB transceivers, however, operate optimally when looking into purely resistive 50 -ohm loads. By adding the right amount of inductive reactance-supplied by a loading coilto precisely cancel out the antenna's capacitive reactance, only the resistive component is left. The antenna's electrical length is such that it resonates at the operating frequency.

A loading coil, in effect, supplies the missing physical length required for resonance. For a given operating frequency, more and more inductance is required for resonance as the antenna is physically shortened Some CB mobile antennas are only $18^{\prime \prime}(45.7 \mathrm{~cm})$ longquite a reduction from 104" ( 2.6 m )! They require a lot of loading inductance. Most mobile whips designed for roof or trunk mounting use less loading and longer whips, keeping element length in the physically manageable 48-to-54inch (1.2-to-1.37-meter) range.

Loading is a compromise solution to the antenna height problem, not a perfect one. Two of its principal drawbacks are a reduction in antenna efficiency and a narrowing of bandwidth. The degrada-
tion of antenna efficiency is caused by the resistance of the wire which forms the coil. Some of the transceiver's r-f output will be wasted heating the coil, rather than being radiated by the antenna. For a given wire composition, more inductance means more turns of wire, hence more resistance and increased power loss. A mini-whip, requiring a large loading coil, will therefore be less efficient than a longer antenna with a smaller loading coil. Manufacturers try to keep this power loss low by using lowresistance wire in their loading coils.

Reduced bandwidth is a natural consequence of antenna loading because the introduction of inductive reactance raises the " $Q$ " of the antenna. As the operating frequency moves away from that of resonance, the feedpoint impedance of the antenna changes, producing a mismatch between the antenna and feed line. Standing waves then appear on the line and the transceiver sees a reactive load.

Some impedance mismatch is inevitable and is tolerable up to a point. A mismatch of impedances by a factor of two (by definition, a standing wave ratio or SWR of $2: 1$ ) is acceptable, but greater mismatches can cause problems to the transmitter and a loss of antenna efficiency. Just how much the antenna's impedance changes with frequency determines how far above or below resonance it can be used before the mismatch becomes intolerable.

Fig. 2. Radiation pattern of vertical whip ( $A$ ) is distorted when antenna sways backward (B) or forward (C).


A full-size quarter-wave whip working against a good ground plane will easily cover the 40 CB channels while maintaining an acceptable match. However, a very short miniwhip will have difficulty presenting a matched impedance over even the original 23 channels. Most midsize whips can be used on 40 channels if tuned for the best impedance match at the center of the band.

Apart from physical convenience, there are two compelling reasons for using inductively loaded, less-than-full size vertical whips in mobile CB communications. One deals with ground-plane effects, and will be considered later. The other results from the physical behavior of a large whip on a moving car. With the vehicle stationary and the antenna upright, a radiation pattern like that shown in Fig. 2A is produced.

If the whip is long and springy, onrushing air will deflect it once the car is moving, resulting in the radiation pattern shown in Fig. 2B. This pattern indicates that some of the radiated $r$-f will be sent up into the blue-at the expense of the ground wave, the real medium of communications! If the car hits a bump or sharply decelerates, the whip can pitch forward, again producing an undesirable radiation pattern (Fig. 2C). Of course, distortions in the pattern will also occur if the whip sways from side to side.

As the whip bounces around, the signal received by another CB'er will flutter in strength. The same effect will be experienced on receive by the operator using the tall whip. Also, the load impedance as seen by his transmitter output stage and the SWR on the line will vary. These undesirable effects can be minimized by either shortening the antenna and inductively loading it or constructing it so that it is rigid

Loading, within limits, is a good solution. Making the antenna rigid will cure the problem, but makes the antenna susceptible to damage from overhanging objects. Those manufacturers who produce stainless steel antennas use special alloys and element tapers to enhance the antenna's ability to remain upright while maintaining the flexibility necessary for absorbing shocks. Loaded fiberglass whips are more rigid than stainless steel and thus have lessflutter.

Fiberglass antennas áre also less susceptible to the buildup and discharge of static electricity, which can produce a hissing noise in the receiver. (Stainless steel whips usually have tip-mounted balls to inhibit static discharge.) The principal drawback to the use of a rigid


Fig. 3. Top-loaded antenna ( $A$ ) is shown at left; center loading (B) in middle; base loading (C), right.
fiberglass whip is vulnerability to impact from overhanging objects.

A stainless steel shock spring can be inserted at the base of a long whip to make it more shock absorbent. The spring must have a braid-shorting strap fastened internally between its two ends. Otherwise, the spring will act as a coil and upset the feedpoint impedance of the antenna.

Types of Loading. An antenna designer has three choices as to the location of the loading coil-at the base, in the center, or at the top of the whip. Each has advantages and disadvantages, and we will examine them in turn.

Base loading is shown in Fig. 3C. The plastic cylinder at the base of the antenna houses the loading coil, isolating it from the detuning and corrosive effects of the environment. Base loading re-
quires the least amount of inductance for a given antenna length, so coil resistance can be kept to a minimum. Also, base loading results in a physically sturdy structure because the weighty loading coil is at the bottom of the antenna. The light weight of the whip minimizes pendulum-like oscillations. An electrical benefit is also obtained.IKeeping the coil fixed in relation to the ground plane reduces variations in feedpoint impedance as the whip flutters.

Most antennas that employ base loading are at dc ground. That is, there is a direct short from a dc point of view between the whip and the ground plane. Dc grounding is introduced to bleed off the hiss-producing static charges that accumulate on a stainless steel whip before they can discharge into the atmosphere.

The major disadvantage associated with base loading is a distortion in the distribution of voltage and current along the radiating element. Compare Fig. 4 A , the $V$-I distribution along a full-size vertical radiator, with Fig. 4B, the distribution along a base-loaded antenna. This distortion in voltage and current distribution means that the base-loaded antenna has a fairly low radiation resistancelower, in fact, than that exhibited by cen-ter- or top-loaded whips. Radiation resistance, a concept developed by antenna theorists, is a fictitious resistance that accounts for the power radiated by the antenna. The higher the radiation resistance, the greater the portion of the r-f delivered to the antenna that is actually radiated into space. Base loading, therefore, is less efficient than other loading techniques.

Center loading (Fig. 3B) offers improved $V$-I distribution along the radiator and greater radiation resistance than base loading. However, for a given whip length, more loading inductance is required. This implies greater coil losses than those experienced with base loading. There is a compensatory factor: a base loading coil is situated at the current maximum, but a center loading coil is positioned at a point where there is less current in the radiator. Heat losses are determined by the familiar relationship $P=I^{2} R$, where $l$ is the rms current and $R$ the dc resistance of the coil. That less current flows in a center loading coil compensates, somewhat, for the greater resistance of the larger loading coil.

Top loading (Fig. 3A) improves the $V$-I profile and increases the radiation resistance even more. This loading technique requires the most inductance and
thus introduces the most coil resistance. However, the top of the radiator is where the least current flows, so coil losses are not as severe as you might first think. The trade-off of some coil losses for increased radiation efficiency is more than a break-even proposition.

The real problem that center or top loading introduces is whip sway. This adversely affects the radiation pattern and causes variations in feedpoint impedance: Whip sway is more likely with center or top loading because the coil housing increases the wind resistance of the whip. Impedance variations are more severe than with base loading because the relationship between the ground plane and the loading coil is not fixed but dependent on the deflection of the whip. As mentioned earlier, whip sway can be prevented by making the antenna rigid, but again, a stiff upright radiator is more susceptible to impact damage. However, if an antenna up to about $60^{\prime \prime}(1.2 \mathrm{~m})$ is mounted on an auto roof or trunk, the likelihood of its coming in contact with overhanging objects is not too great. Vans and trucks are a different story.
Another loading technique is continuous loading. No discrete loading coil is employed. Rather, the inductance is distributed along the entire radiator. The continuously loaded antenna is formed by helically winding one quarter-wave length of wire on an insulating pole (Fig. 5). Epoxy or a fiberglass sheath is used to secure the helix in place.

Continuous loading produces better voltage and current distribution than lumped constant (discrete coil) loading. if's feedpoint impedance is also a better match for 50 -ohm coax. If a tapered pitch is employed when the helix is wound, a very good match will be obtained. The radiation resistance of a continuously loaded antenna is comparable to that of a top-loaded whip. Because of these advantages, many of the fiberglass antennas on the market (except, of course, full-size whips) employ continuous loading.

Mounting the Antenna. We have already seen that a vehicle's body falls short of fulfilling a required ground plane. This has a significant effect on the performance of a mobile whip. Recall that the ground plane is analogous to a mirror. If it has high resistance, it will act like a dirty mirror, giving a faint reflection. If it is too small, the entire image will not fit in it. If the ground plane is not symmetrical, it will act like one of those

## Choosing Mobile Antennas continued

trick mirrors in an amusement park, producing a distorted reflection.

The size of the ground plane (the auto body) is fixed once you have acquired the vehicle. There is not much you can do about the resistance of the car body except to ensure that all its components are bonded together by low-resistance, metal-to-metal connections. The one ground plane characteristic you can determine, to some extent, is its symmetry. This is done by your choice of the antenna's mounting site.

The best place to mount a mobile whip is at the center point of the vehicle's body, midway between the front and back and equidistant between the two sides. This will usually be located on the roof of the vehicle. If the antenna is mounted here, it will have a polar radiation pattern like that shown in Fig. 6A. (The arrow points towards the front of the automobile.) Although this pattern is not the ideal circle, it is a fair approximation. Because a car is longer than it is wide, the antenna is more effective fore and aft than it is side to side.
Mounting the antenna on the trunk lid results in the pattern shown in Fig. 6B. The distortion introduced by shifting the antenna from its ideal mounting position is obvious. It is due to the lack of ground plane symmetry. However, the pattern is acceptible and its imperfect shape is considered by many CB'ers to be less objectionable than drilling a hole in the vehicle roof for an antenna mount.

The polar response of an antenna mounted on the left corner of the rear bumper is shown in Fig, 6C. The anten--na strongly favors the front right, and its overall performance is degraded. It is clear from these plots of field strength near the antenna that an unsymmetrical ground plane produces a polar response that is also unsymmetrical. If possible, this should be avoided.

Types of Mounts. CB'ers can choose either permanent or temporary antenna mounts. The final decision will be guided by performance, theft protection, and aesthetic considerations.

Permanent mounts include hardwaresecured roof, trunk-lid, and bumper installations. Roof mounts require a $3 / 8^{\prime \prime}$ -to-3/4" (9.5-to-19-mm) hole, but offers the best antenna performance. Trunk-lid mounts do not require drilling because they are held in place by two set screws which grip the inside edge of the trunk


Fig. 4. Voltage and current distribution along a full-size quarter-wave antenna (A) is distorted when a loading coil is introduced at base of antenna (B) and the whip length is reduced.
lid. As noted earlier, antenna performance is somewhat degraded as compared to roof mounts. Bumper mounts usually employ straps which wrap around the bumper. The new "safety bumpers," however, have a lip which allows a clamp mount to be used with no straps or chains. Bumper-mounted antennas will not perform as well as those on the trunk or roof. The same is true of mirror- and cowl-mounted whips.

These permaneent antenna mounts offer a solid connection to the ground plane, which is desirable, but they present security problems.

You can minimize calling attention to your CB gear by either removing the antenna and its mount each time you leave the car. (A quick-disconnect dévice is useful for this purpose) or by using antennas that are self-stowing. Using a multi-purpose antenna for AM and FM reception as well as for CB work is one way to do this. Such "disguise" antennas, if properly designed, look like a standard auto antenna. Electrically powered antennas that retract at the touch of a switch (some actuated by turning off the transceiver or ignition) similarly re-

Fig. 5. Continuously loaded antenna is made by winding one-quarter wavelength of wire on insulating pole.

duce the risk of loss while adding to convenience. Some of these, though, still leave a tell-tale $1 / 2^{\prime \prime}$ jutting through the opening. "Hideaway" trunk mounts, hinged to allow the antenna to be swung down into the trunk when not needed, also reduce antenna visibility.

Though these anti-theft measures reduce the chance of theft, they do degrade antenna performance somewhat. Because "disguise" antennas are mounted on a fender, and "hideaway" trunk mounts must be placed along one side of the trunk, they work against an unsymmetrical ground plane and have a skewed polar pattern.

Antennas with temporary mounts include those that clip on the raín gutter and those with magnetic mounts. There are several disadvantages associated with gutter mounts. First, gutters are not sturdy so the antenna must not have large wind resistance. This dictates the use of short whips and large loading coils, resulting in coil losses and decreased antenna efficiency. Second, rain gutters do not always have lowresistance connections to the rest of the car body (ground plane). Third, the ground plane is highly unsymmetrical.

A magnet-mounted antenna can be effective both as a radiator and an antitheft device. It can be tossed in the trunk when not needed, and placed on the trunk or roof in a few seconds. A good magnet-mount antenna meets the following requirements. It must resist being dislodged when the whip is deflected. It must not "walk" along the roof or trunk as the car body vibrates and the whip sways. The mount must display a relatively high capacitance to the car body. The last requirement is electrically important because a magnet mount, unlike the others that have been mentioned, does not offer a direct connection to the vehicle body. Although the braid of the coaxial cable is grounded at the transceiver, for best results it should also be grounded at the mount. If the magnet mount is properly designed, there will be sufficient capacitance between it and the vehicle body.

No matter what type of mount you are thinking of using, be sure that it will keep the antenna upright. For example, if you are mounting the antenna on a fastback or other angled surface, choose a mount that will permit you to compensate for the mounting angle and adjust the whip so it is upright.

Co-Phasing. If two vertical antennas are spaced one quarter wavelength or

Fig. 6. Polar radiation patterns for roof (black line), trunk (colored line), and bumper (solid color), mounted antennas. Arrow points toward the front of the vehicle.

more away from each, other, fed with properly phased feedlines, and there are no vertical conductors within a twowavelength radius, some gain and a fig-ure-eight radiation pattern (favoring fore and aft) will be obtained. CB'ers have attempted to take advantage of this by mounting twin whips on truck or camper
mirrors, car bumpers, etc. Unfortunately, it doesn't work! Except, perhaps, on some very large trucks, a whip spacing of $104^{\prime \prime}(2.6 \mathrm{~m})$ cannot be obtained. Also, there are enough intervening metallic elements within two wavelengths ( $72.8^{\prime}$ or 22.2 m ) to upset the carefully phased electromagetic fields.

The two mirror- or bumper-mounted antennas together will, however, have a better polar response than one by itself (recall Fig. 6C) because of the complementary phasing between them. But one antenna mounted on the vehicle roof or trunk displays a pattern superior to that of the phased twins, and will usually give superior performance!

Antenna Materials. Choosing a given type of loading, mounting position, etc. must be guided by factors that are unique to a given situation. Therefore, we can't recommend a specific antenna type. But we should say a word about antenna materials. If you want a metalic whip, be sure that it is "17-7 PH" type stainless steel. This alloy is very strong, does not corrode, and will flex but is resistant to permanent deformation. Shock springs and similar components should be triple-chrome plated.

Finally, mounting hardware should be heavily plated to resist corrosion.

In Conclusion. A properly installed and tuned antenna is your transceiver's best friend. Therefore, take care to mount it so thatits ground plane is symmetrical. Make sure that the antenna has provisions for adjusting its length so that it can be fine tuned for the CB channels. Tune the antenna in accordance with the manufacturer's instructions and you'll ensure that it and your transceiver are giving you their maximum performance capabilities.

# How to Install - Mobile CB Transceivers - Mobile CB Antennas 

look first for the best and easiest places to install them. For example, be sure that there is sufficient room under your dash to accommodate the transceiver's depth before you buy it!

For a typical mobile setup, we'd recommend that the transceiver be mounted on a quick-release slide bracket beneath the center of the dash. Also, the antenna should be mounted so it can be flipped down into the trunk for concealment when necessary. That combination offers easy access to the set's controls, reasonably good signal output and reception, straightforward installation, and excellent theft protection.

The reasons for the dashboard site are obvious: hext to a combo CB /stereo unit that's built into your dash (and such equipment is difficult to install), a centered, under-dash mount is easiest for both driver and passenger to reach and is least likely to get in anyone's way.

WANT to make $\$ 25$ or more, taxfree, for about an hour's easy work? Then install that new CB mobile radio in your car yourself, and you'll save at least that much. You don't need
special knowledge or unusual tools.
You can mount a CB transceiver almost anywhere within reach of the driver's seat and mount an antenna almost anywhere outside the car. But it pays to

Slide mounts come in two parts: a stationary section that attaches to the bottom part of the dashboard, and a sliding section that holds the transceiver. Ideally, you'll find a mounting spot for the stationary section where the transceiver's controls will be easy for every driver in your family to see and reach, where the rig can be easily and securely mounted, and where there will be space to install the rig with clearance for attachment of its plugs and fasteners.

Mounting Surfaces. With a little bit of luck, you may find some screws al-
themselves around your drill bit. Even foam padding, although less of a problem, must be treated cautiously.

If you can, it's best to punch your way through a padded metal underdash. An alternative approach is to cut an " $X$ " in the top covering where you want to make the hole, peel back the four flaps made by the cut, and carefully dig or cut out any underpadding. Then smear Vaseline on the bit to make it slippery and drill in short bursts, watching carefully so you can stop the drill at the first sign of cloth or padding wrapping itself around the bit. After installing anything


A typical 2-part slide mount tor easy removal of an under-dash CB transceiver. Note that copper finger contacts on part A (that attaches to transceiver) makes power and signal connection to part B (affixed to underside of dashboard) so that the user need not physically disconnect or connect wires to remove unit.
ready on the underside of the dash that can be used to hold the slide or the transceiver's mounting bracket (although you might have to replace those screws with slightly longer ones). If not, check the available surface to be sure it is strong enough to carry the transceiver's weight and to determine what techniques will be needed to drill through it.

Plain, painted metal is an ideal surface. It can support the radio securely and you can use short, self-tapping sheet-metal screws. With an electric drill, clean holes are easy to make. Just be sure to mark each hole with a centerpunch before you drill so the bit won't slip while you're working and scratch the paintwork-or your skin.

Metal with a padded covering requires a bit more care when drilling. Fabric coveis or wooly underpadding may wrap
on a padded surface, check and re-tighten the screws from time to time to compensate for any gradual compression of the padding.

Plastic and fiber panels are less secure mounting surfaces than metal, but sometimes they're all you have. Check behind such panels to see if there are metal structural supports that you can reach with longer bolts or screws. If not, the main problem is a tendency for the transceiver's weight to cause screw holes to enlarge until the screws pull out, CB rig and all. To prevent this, drill very carefully, with a sharp, fresh bit and gentle pressure to avoid cracking the panel. Then spread the weight over as much surface area as possible, using several, widely spaced screws and placing the largest possible washers under each nut. (Use lockwashers or Loctite to pre-
vent your mountings from vibrating loose.) if you can't get your fingers behind the panel to insert nuts and washers, use Molly screws. For screw holes near the edge of a thin panel, Tinnerman nuts can hold the mounting bolts, although larger washers will spread the load more.

If your mounting surface is an underdash, plastic parcel shelf, rest your rig atop the shelf instead of hanging it below; here, the screws only maintain the transceiver in position, rather than supporting its weight.

Before you drill the first hole, make a final position check: Have someone hold the slide and transceeiver in place and check for such often-overlooked details as cords that might get tangled in the gear-shift or pedals, side-mounted microphone plugs that poke the driver's leg, etc. Double-check behind the dash to make sure the drilk bit won't hit wires, puncture air-conditioning ducts, or hit ashtrays. Often you can move some of the obstacles out of the way before you drill. If they are not too close to the panel you're drilling, you can also protect them by slipping a drill stop (available at hardware stores) over the bit to limit the depth of its penetration, or improvise a stop with duct tape. Otherwise, you'll have to pick a new location.

Determine which half of the slide mount attaches to the car and use itnot the mounting bracket supplied with the transceiver-as your template. (The transceiver bracket then bolts to the other, sliding half of the mount.) Make sure the bolt heads don't protrude enough to prevent sliding the transceiver into place. After you have drilled the first hole, attach your bracket to the car and double check the position you've marked for the second one.

Other Mounting Spots. If your dashboard is not a suitable or convenient spot for your transceiver, you might consider mounting it on the car's transmission hump. It's best to use speaker/rigmount combinations in this situation. They're designed specifically for floor use, incorporating better speakers than those in mobile units. Moreover, since most mobile rigs have downward-facing speakers, this accessory avoids a sound-output problem. They usually disconnect easily for storage in the trunk when not in use, but they do take up some floor space.

Although the dash and the hump are the most poputar locations for a $C B$ rig, they are not the only ones. Some CB'ers
mount their sets, controls up, between front bucket seats; others mount them in their cars' center consoles. A small transceiver can even be installed in the car's open well that serves as a glove compartment. Then, too, there are temporary setups where the rig rests on the seat, lifted at the edge so as not to muffle speaker output. Power here is obtained from a lead plugged into the cigar lighter socket, while the antenna is either attached to a rain gutter or by a magnet mount in the middle of the roof.

Many new transceivers are built for concealment, with all controls in the microphone head and the rest of the circuitry in a featureless box that can be locked in the trunk or hidden elsewhere (mounted on the firewall, for example).

Extension Speakers. Most mobile sets have one or two extension speaker jacks on their rear panels. These allow the use of external speakers for better sound within the car, for PA use outside the car, or both. In rigs with two jacks, one feeds the stations you are listening to through an extension speaker in the car, while the other, when you switch the transceiver to its "PA" mode, feeds whatever you say into the mike through a speaker outside the car or under its hood. Don't use hi-fi car-stereo speakers since their frequency range is too wide for voice radio communications, which rarely exceed 3 kHz .

Slide-Out Bracket Details. Not all slide mounts are alike (although some models may show up under several brand names). Insist on a bracket that has a built-in coaxial socket and plug for the antenna, so you don't have to disconnect that separately when you slide out the transceiver. If you don't have this feature, you're sure, someday, to forget to re-attach the antenna and blow your power transistors.

The flimsy locks often found on slide mounts don't discourage thieves, who can often snap them in seconds. Even if the lock is a robust one, why invite a thief to try and rip up your car? So it's best to throw away the key and remember to remove and store your rig, perhaps in the trunk, when you leave the car unattended.

Most slide-mount sets have an extra pair of contacts for extension or PA speakers. At least one has two pair, so you can use both extension and PA speakers with your transceiver, or use the same mount for a car stereo (which needs connections for two speakers).


Installing a mobile $C B$ rig is not usually a major undertaking. It can be challenging, though, when an AM/FM electric-powered antenna is replaced by a CB/AM/FM one since auto radio must be removed to make power connections.

Power Connections. Your mobile transceiver will get its power from the car's electrical system. It's easiest to make these power connections while the transceiver is in the car, but before it's actually bolted in place. You may connect it either to a circuit that is always "live" or to one which only carries power when the ignition key is turned. The owner's manual for your car should tell you which circuits are which and
may-especially if it's a foreign carinclude a circuit diagram to help you find the wires you'll need.

Running the power leads to a switchcontrolled circuit ensures against accidentally draining the battery by leaving the car with the transceiver turned on. It also prevents unauthorized use of the radio while you're out of the car. (Remembering to remove and hide the rig each time you get out would also take

This external speaker-system/CB-rig mount is positioned on top of the car's transmission hump for enhanced sound. It is removabie and can be hidden in the trunk for safety.


## How To Install continued

care of these two problems.) If you want to listen or transmit with the engine off, most car's ignition switches have "accessory" positions to allow this.

The easiest place to connect the CB power leads is to one of the terminals or wires already in the dash. You can tap into an existing wire, to a terminal of the ignition switch (if that's not buried out of reach), or to a terminal on one of the other dashboard switches or controls. (If you do use a switch terminal, make sure you have the switch's "hot" side-you don't want a CB radio that only operates if the headlights are on.)

The best place to tap in is usually the car's fuse box-there's less chance of picking up interference there. Whether you get your power directly from the fuse box or from another power source under the dash, make sure the fuse involved has enough capacity to handle your transceiver plus other devices it's al-

ready powering. If not, select a circuit with some spare capacity-don't just install a larger-amperage fuse. And never connect your transceiver or other accessory to the same circuit as the headlights or other vital systems.

Which Side is Ground? Virtually all cars today have 12 -volt electrical systems, with the negative side of the circuit grounded to the car's frame. But some older ones and many trucks have posi-tive-ground systems. Most current transceivers will operate on negative-ground, 12 -volt systems, with many also capable of working on positive-ground systems. The instruction manual accompanying the CB transceiver will note which type you have. But if you don't have that information, you can use the ohmmeter function of a multimeter to determine it. Measure resistance between the transceiver's case and each of its two power

Photo at left shows a new type of 3 M connector that comes in handy for making power connections to your mobile CB rig. Below, is an Amphenol PL-250 plug to show how easily RG-58A/U antenna coax wire can be connected without soldering.

ple mount their antennas back there for a variety of reasons. One is that short, coil-loaded antennas that clamp onto the trunk lid are about the easiest to install, since no holes have to be drilled in the car's outer body. Also, an antenna mounted to the front, hinged edge of the trunk is in a fairly efficient location, especially if the car's a hatchback with its "trunk lid" hinged at roof level. Nine-foot whip antennas can only be mounted conveniently on a car's rear bumper.

Antennas serve to alert would-be thieves that there's probably CB equipment inside that is worth stealing. So more and more installations take this into account. Quick-disconnect attachments are often used for easy removal of the vertical antenna section, though this still leaves the antenna base. To completely hide an antenna, more and more CB'ers have turned to the electricpowered type or to an antenna mounting device that permits one to manually swing the antenna down into the trunk. Both are side-mounted types.

Flip-down mounts may clamp on, without drilling, or may require some small screw holes in the rain gutter that surrounds the trunk opening. Since signal radiation from a móbile antenna is greatest toward the farthest point of the car, mounting the antenna by the side of the trunk opening will send most of your transmitted power towards the car's opposite front corner. But there will still be substantial radiation in most directions, so that is not a serious problem-especially compared to the performance you get if your rig is stolen!

Whether you have a trunk mount, a bumper-mounted whip, or a flip-down, your biggest problem will be getting its lead out to the transceiver. You're most likely to run your antenna cable from the trunk toward the dash.

Getting the lead into the trunk should be easy. Bumper-mount whips may require a hole drilled in the trunk wall (remember to line it with a rubber grommet, both to protect the cable and keep rain out), although it is sometimes possible to bring the cable through a hole that now carries wire to your back-up or li-cense-plate lights. Antennas that mount on the edge of the trunk opening require only a little care to ensure that the cable wor't kink when the lid closes.

Coming from the trunk, your first obstacle will likely be the partition between the trunk and the passenger compartment. If there's a gap between the partition and the trunk's floor, or if the car is a wagon or hatchback whose seat folds


A fold-away trunk mount makes it easy to conceal a mobile CB antenna when leaving an automobile.
down, half the problem is already solved for you. You may also find a channel that carries tail-light and other wires, with enough space left to carry your antenna cable too. More often, though, you'll have to make a hole. If the partition is of metal, you'll have to drill. If it's of fiberboard or a similar substance, you may be able to punch a hole through it instead. It's probably best to drill, using gentle pressure. Again, be sure to line. the hole with a rubber grommet. Check out the area on the other side of the panel before locating the hole to be sure your drill won't chew into upholstery or encounter other problems where it comes through. If the rear seat cushion has to come out, it's generally best to do that first. (It may take two people to do this.) Most seats are held by catches, some by bolts.

Before drilling, you should also check the passenger compartment to see where it will be best to run the cable. The best route is usually along the side of the car. You can tuck it under the edges of side panels (you may have to loosen their mounting screws), run it under the sill-plates at each door (they're easy to remove, usually held by just a screw or two), or insert the cable under the edges of the carpets and floor mats. In some cases, it's more practical to lift the carpets and run the cable along the lower edge of the transmission hump and through the center console.

If you are unable to remove the rear seat cushion, you could try to snake the wire straight through the trunk, or fish it
through with a hook made from coathanger wire. If that fails, try raising the seat's front edge enough to get your hand under it. (You may want to slip a block beneath the seat's front edge.) If the seat doesn't lift up easily, check on how it's mounted; it may be bolted in place.

As you work, you should be using cable clamps or cable ties to keep the antenna lead out of the way. This is most important in the trunk where a loose cable could be snagged by luggage or other cargo, and when passing across the front of the car where it could tangle with the steering gear or pedals.

The antenna cable may not be exactly the right length, but that's no problem. If it's too long, you can cut the excess and use a new PL-259 connector. The solderless connector type for RG-58 A/U coaxial cable is best'for this purpose.

If the cable is too short, you can buy an extension. Buy one with plugs already installed and the shortest length that will do the job.

Other Antenna Sites. Because your signal's coverage increases as you raise your antenna, and because its pattern is most symmetrical when the antenna is at the center of the car, a roof mount is a most efficient choice. Unfortunately, a roof mount can be quite difficult to install. Also, most people hesitate to drill holes in an area that is so conspicuous.

For a roof antenna, you'll have to drill a hole in the roof and snake the antenna lead from there, under the car's cloth

## How To Install continued

headliner, then down a window pilfar to the dash or floor level. That may be easier if you have a dome light in the center of the car's roof-take out the light and you can usually drill directly through the roof. Just be sure there will still be room to put the light back, once the antenna is installed. If you have no center light, you'll have to take the headliner down to do the job neatly-and in most cars, that's not easy.

You can also use a magnetic-base antenna. It requires no drilling and installation (the cable could be passed through the rubber gasket around the car's door opening or through a slightly opened window). Furthermore, the antenna is easily hidden in the car when you park.

Antennas that clip temporarily to the car's roof gutter are available, too, as are permanent fender-mounted or cowlmounted types. The latter include elec-tric-powered antennas, which might also combine AM and FM broadcast radio.

Fine-Tuning Your Antenna. It's not enough to just install and connect your antenna. You also have to adjust it for minimum SWR-standing wave ratio. This is the ratio between the power that comes out of your transceiver and the losi power that bounces back from the antenna line instead of going out over
the air. (It will similarly affect the strength of a received signal.)
To do this, you'll need a modestly priced SWR meter, if your CB transceiver doesn't have one built in. This will reduce some of the savings you made by doing your own installation, but is a worthwhile investment for making periodic checks on your antenna system.
The meter plugs in between the transceiver and the antenina cable. If you have added an extension to your antenna's cable, you can connect the meier between the cable and extension. Otherwise, you'll need a short stub of cable with plugs on both ends to connect the meter to your transceiver; some meters include this cable, but not all do.

You can't tune an antenna for the same SWR on every channel. The farther you go in frequency from the channel for which the antenna has been tuned, the higher the SWR will be. So, unless you do almost all your talking on a single channel, you should tune the antenna to the center of the band for the best average efficiency on all channels.

For 40-channel transceivers, this would be channel 20 or 21. Listen for a break between conversations before you press the mike "talk" button. You don't want to interfere with other CB users' conversations! Also, you may legally transmit a silent carrier only when

Measuring and adjusting for minimum SWR is an important final step to ensure optimum antenna efficiency and avoid damaging a CB transceiver.

making adjustments like this and then for not more than one minute out of every five.

Your meter will probably have a switch marked FWD on one side and REV or SWR on the other. Switch it to the forward position and press your mike's talk switch. The needle should move up the scale. Adjust the meter's calibration knob until the pointer reaches a red line or other index mark. Then flick the switch back to SWR or REV and read the pointer again. It should have dropped back to a reading somewhere between 1.1 and 1.5 or so. Now release your PTT mike switch so you can make antenna adjustments that will produce the lowest SWR reading.

Theoretically, an SWR of $1: 1$ is perfect; in practice, it will be higher, say, 1.2:1 at mid-channel and 1.5:1 to 2:1 at end channels.
To lower an antenna's SWR, you must adjust its length. This requires either lengthening or shortening the antenna. To tell which, repeat your SWR checks on a moderately low channel (around channel 10) and a fairly high one (say, channel 30). If SWR is lower on channel 10 than channel 20 , you'll have to shorten the antenna; if it's lower on channel 30 than on 20 , you'll have to lengthen it. (If it is much higher than 2:1 on any channel, stop transmitting on that channel at once and check cable connections at the transceiver and the antenna ends.)

Most of the better antennas have length adjustments, either an adjustable tip (on top- or center-loaded antennas or on some fiberglass "continuously loaded" types), or an adjustment on the coil housing (of base-loaded types). To lengthen or shorten such an antenna, simply loosen the adjustment setscrew with the Allen wrench supplied, make a small height adjustment, and lock it in place while you re-check your SWR with the meter.

Less expensive antennas and some older types may have to be trimmed to obtain a proper match. Either a hacksaw or bolt cutter will do the trick. Cut no more than $1 / 8$ inch at a time.

Eventually, your SWR will get as low as it's going to be and your next adjustment will only serve to raise it a trifle. At that point, go back one step to where you got your lowest reading. Then, for a final check, measure your SWR on both channel 1 and channel 40 . If they're not quite equal, readjust until they are.

Now, at last, you're ready to go onithe air and ask for a radio check.


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## four <br> EASY-TO-BUILD

WITH THE prices of LED's and CMOS IC's continuing to drop, electronics experimenters should take advantage of the circumstances and build some of the many interesting projects that can be made using these devices. The four circuits described in this article are not only fun to build, they also teach the builder quite a bit about the devices and their uses.
The circuits take advantage of the fact that CMOS devices require very low power, so no power on/off switches are used. The quiescent current drawn by the CMOS chips (when the LED's are off), allows normal battery shelf life. Once the pushbutton switch on a project is operated, the circuit "does its thing," and then stops.

Blinker. As shown in Fig. 1, this circuit uses a single CMOS hex inverter to provide both timing and drive to make the two LED's blink alternately. Built with two small red LED's, the circuit makes


Fig. 1. Dual LED alternate blinker uses parallel gate output for more $L E D$ driving current.


FLASHER PARTS LIST
BI-9-voli battery
$\mathrm{Cl}-100-\mu \mathrm{F}, 10-\mathrm{V}$ electrolytic
C2- $4.7-\mu \mathrm{F}, 10-\mathrm{V}$ electrolytic
D1-1N914 diode
ICl-401] CMOS quad 2-input NAND gate
LEDI-Red light emitting diode
RI-10-megohm resistor
R2-100,000-ohm resistor
R3-470,000-ohm resistor
R4-10,000-ohm resistor
R5-3.3-megohm resistor
S1-Normally open pushbutton switch

Fig. 2. Single LED
flasher also uses
parallel gate output
for driving
the LED.
an ideal HO-gauge model railroad crossing blinker. With LED's of two different colors (in one package), it can be used to obtain other effects.
Resistors R2 and R3 and capacitor C2 determine the flash rate, while R1 and $C 1$ set the total display time. The component values shown here produce a blinking rate of two per second and an on time of about 20 seconds. To change the timing, change the values of the capacitors since decreasing the value of the resistors will increase the quiescent battery current drain.

Flasher. A simple variable-rate LED flasher is shown in Fig. 2. The voltage across C1 determines the flash rate. When the pushbutton switch is closed, capacitor C1 charges to 9 volts and the flasher blinks rapidly. As the voltage is discharged through R1, the flasher slows down until the charge on C1 reaches about 4.5 volts, at which point the oscillator stops and the LED stays off. The flash rate is set by the values of R2, R3, R4, R5, and C2. Capacitor C1 and bleeder resistor R1 create the slowdown period.

Binary Counter. A circuit that demonstrates the operation of a six-bit binary counter is shown in Fig. 3. When the

Fig. 3. Simple binary counter illustrates counting in the 1-2-4-8-16-32 mode.


BI--9-volt battery
C1-0.01- $\mu \mathrm{F}$ disc capacitor
$\mathrm{C} 2-0.022-\mu \mathrm{F}$ disc capacitor
IC 1-4011 CMOS quad 2-input NAND gate
1C2-4024 CMOS binary counter
LED1-LED6-Red light emitting diode
R1.R4-10.000-ohm resistor
R2-10-megohm resistor
R3-4.7-megohm resistor
SI-Normally open pushbutton switch
pushbutton switch is depressed, the circuit starts counting from zero (all LED's off) to 63 (all LED's lit). After reaching the full count, the circuit automatically resets to zero and shuts itself off. The six LED's come on in a binary ( $1,2,4,8$, 16,32 ) sequence which is typical of digital counters.

two things occur simultaneously. Counter IC2 is reset to zero by the signal on pin 2, thus placing all of the IC2 outputs at their low states ( 0 volts). Thus, none of the LED's can glow. The second action is an enable level signal ( +9 volts) at pin 13 of IC1. This action allows the oscillator (the middle two gates) to start, thus producing an input signal to the counter IC through the last gate of IC1.

The counter then counts until it is full, illuminating the LED's in the proper sequence. One count after full count is reached, pin 3 of $I C 2$ goes high. This signal is inverted by the first gate of $I C 1$, and its output goes low, thus disabling the oscillator. The circuit then remains in the "all LED's off" state until the pushbutton is depressed again. The value of C2 can be changed to increase or decrease the counting speed.

Wheel of Fortune. The circuit shown in Fig. 4 is a 10 -LED spinning wheel with audible 'clicks' as the wheel passes each point. The rotation starts fast, then gradually slows down to a random stop (with a click at each position). After the rotation ceases, the selected LED stays lit for about 10 seconds, then goes out. The cycle restarts by depressing the pushbutton switch.
The logic requires onty two IC's. Of these, IC1A, IC1B and IC1C form a vari-

Fig. 4. "Wheel of Fortune" sequentially lights one of 10
LED's and generates audible clicks.

B!-9-volt battery
$\mathrm{C} 1-0.01-\mu \mathrm{F}$ disc capacitor C2-200- $\mu \mathrm{F}, 10-V$ electrolytic $\mathrm{C} 3-1-\mu \mathrm{F}, 10-\mathrm{V}$ clectrolytic
C4-3.3- $\mu \mathrm{F} .10-\mathrm{V}$ electrolytic
DI-1N914 diode
ICI-4069 CMOS hex inverter 1C2-4017 CMOS decade counter decoder LED1-LED10-Red light emiṭting diode Q1, Q2-2N2222 transistor R1-100.000-ohm resistor R2-470.000-ohm resistor R3-3.3-megohm resistor R4, R6-10,000-ohm resistor R5-1-megohm resistor SI-Normally open puishbutton switch


Fig. 5. Modifying the Wheel of Fortune for use with conventional 6 volt lamps.


Fig. 6. Foil pattern and component installation. The four circuits are separated along the dotted lines.
Note: Pc board available from Ray Wilkins, Box 551, Hanover, NH 03755 for $\$ 4.50 \mathrm{ppd}$.

able frequency oscillator operating exactly like the oscillator in the Fig. 2 flasher circuit. Then IC2 is a combination decade counter, decoder and driver that powers 10 LED's in sequence, with the LED's arranged in a circular display. Each pulse from the oscillator advances the count by one.

The oscillator pulses are buffered by IC1D and amplified by transistor Q1 to drive a small loudspeaker. Capacitor C3 affects the speed of rotation, while C2 determines the total length of time that the display stays lit. The dc voltage APRIL 1978
across C2 is also applied to a pair of buffering inverters (IC1E and IC1F) with the output used to turn on switching transistor Q2. When this transistor is saturated, it allows the LED's to turn on. When the voltage across C2 drops, the output of inverter IC1F drops to zero, causing Q2 to cut off, thus turning off the LED's.

It is possible to substitute conventional 6 -volt, $40-\mathrm{mA}$ lamps in place of the LED's by using the circuit shown in Fig. 5. To operate these optional lamps, an extra 6 -volt battery is required.

Construction. Any type of construction can be used for any of the projects. If you want to use a printed circuit, you can use part or all of the foil pattern shown in Fig. 6. The four sections of the pattern can be separated at the dotted lines. Component layouts are also shown in Fig. 6. Install passive elements first, then the IC's. Be sure to observe the polarities of the electrolytic capacitors, diodes and IC's. Use a conventional 9 -volt battery clip and leads for the connections. The red lead is positive, and the black lead is negative.

By Forrest M. Mims

## GETTING ACQUAINTED WITH CMOS

THOSE OF YOU who have built some of the digital circuits presented in this column over the past several months have probably noticed a prob lem common to all the 7400 -series TTL integrated circuits. They are power hungry. The 7489 64-bit RAM, for example typically draws 80 milliamperes. Some 7489 chips require up to 120 milliamperes. That's one-fourth of the current demand of a 6 -volt lamp in a portable sealed-beam lantern!

One way around the TTL power problem is to use low-power Schottky TTL chips. These chips typically require only twenty percent of the power of conventional TTL. Low-power Schottky devices, which aren't as easy to find and cost more than conventional TTL, are designated with an "LS", such as: 74LS89, 74LS90, etc.


Fig. 1. Basic CMOS gate.
The best solution to the TTL power problem is to use CMOS IC's instead. In case you're not familiar with CMOS, it's a logic family which uses voltage-sensitive, metal-oxide semiconductor (MOS) field effect transistors, as opposed to current-sensitive, bipolar transistors.
CMOS has an ultra-low power requirement. The CMOS 74C89, for example, is functionally almost equivalent to the TTL 7489 64-bit RAM. The CMOS version, however, typically consumes only 0.050 microampere in operation!

This power saving feature makes CMOS ideal for battery-powered devices like digital watches, pocket calculators, and spacecraft.
Why does CMOS require so little power? The answer lies in the very structure of CMOS circuitry. A basic CMOS gate (an inverter) is shown in Fig. 1. Note that a complementary (the " C " in CMOS) pair of enhancement-mode MOSFET's comprise the inverter. (Although other CMOS logic elements are more complex, they all use complementary MOSFET's and the following description captures the essential characteristics of the CMOS logic family.) An enhancement MOSFET is normally off, displaying a high resistance between drain and source. To turn it on, you must apply a sufficiently large voltage between the gate and source.

In this circuit, Q1 is a p-channel MOSFET and Q2 is an n-channel device. The two form a series circuit between $V_{D D}$ and ground. If $V_{\text {IN }}$ is low, the p-channel MOSFET is on and the $n$-channel MOSFET is off. Thus, Q1 exhibits a relatively low resistance between drain and source and Q2's channel resistance is very high. The output terminal is therefore effectively connected to $+V_{D D}$ and isolated from ground, and $V_{\text {OUT }}$ is high. If $V_{I N}$ is high, there is no potential difference between the gate and source of Q1, so the p-channel MOSFET is off. However, $V_{G S}$ for Q2 is high, and the nchannel device turns on. This grounds the output terminal, making $V_{\text {OUt }}$ low.

Note that in either case ( $V_{\text {IN }^{\prime}}$ high or low), one MOSFET is on and the other is off. Because the two devices are connected in series, a high-impedance path exists between $+V_{D D}$ and ground for either input state. That's why CMOS requires so little supply current. In fact, the only time its current demand rises is during an input state transition ( $V_{\mathbb{N}}$ going towards $+V_{D D}$ or ground). During such a transition, the devices will have channel resistances between the two ex-
tremes and more current will be drawn from the source

Among the other advantages to using CMOS are the small chip area required for each gate, the very high noise immunity, the large fan out (the number of CMOS gate inputs that can be driven by one output), a wide range of permissible power supply voltages, and low output and high input impedances. There are a few drawbacks, however. A MOSFET's gate structure is very fragile, and the extremely high input impedance makes CMOS susceptible to damage from static electricity. Also, CMOS employs both p- and n-channel MOSFET's in close proximity to each other on the same chip. This makes CMOS more costly to manufacture than conventional TTL, which employs npn transistors exclusively. Finally, the structure of CMOS results in relatively large stray capacitances, which combined with high input impedances result in relatively slow logic. The typical maximum speed for CMOS logic is 1 to 5 MHz .
In many applications, high speed isn't required and CMOS is perfectly acceptable. The other major problem with CMOS, its vulnerability to static electricity, has been dealt with by diffusing protective zener diodes at the sensitive gate structures. The diodes shunt high voltages away from the gates, preventing their destruction. However, you may not know if a particular device is diodeprotected, and external appearance will not tell you. Unless you know for a fact


Fig. 2. Astable multivibrator. that a particular CMOS device is protected, play it safe and handle it carefully.

In recent years a wide range of CMOS chips has become available at prices attractive to the experimenter from many of the advertisers in the Electronics Marketplace pages of this magazine. One common family is a pin-for-pin equivalent of the traditional TTL 7400 series. These chips even use the same num-
bers, inserting a " C " after the 74 prefix. Thus, a 74 COO is the CMOS version of the TTL 7400. Another common CMOS family is the 4000 series. Let's use a chip from each family in test circuits.

CMOS Astable Multivibrator. A good way to become better acquainted with CMOS is to build a simple astable multivibrator from a few of the inverters in a 74C04 hex inverter. One possible circuit is shown in Fig. 2. This circuit will flash its LED at ample brightness while consuming only a few milliamperes from the four series-connected 1.5 -volt alkaline AA cells. Excepting current through the LED, the circuit consumes less than half a milliampere.

The circuit's repetition rate can be varied by adjusting the setting of RI. Naturally, current demand goes up when the pulse rate is increased.

Avoid touching the pins of the 74C04 when you build the circuit. CMOS chips that do not have diode protection are almost always sold with the pins inserted in conductive plastic foam (not styrofoam) or with the pins otherwise shorted together to prevent damage from static electricity. Grasp the ends of a CMOS DIP between your forefinger and thumb, and then insert it into a solderless breadboard. Use insulated connection wires,


Fig. 3. Pin outline of the 4017. Each activated output goes high for one clock cycle then returns to the low state.
and avoid touching exposed conductors. (You'll want to take more elaborate precautions when handling expensive CMOS chips such as microprocessors.)

Incidentally, CMOS chips can be operated from a power supply delivering from 3 to 15 volts, so feel free to use a higher supply voltage.

## CMOS Divide-by-10 Counter/

Decoder. The 4017 CMOS divide-by-10 counter/decoder is an exceptionally handy chip. It does the job of a 7490 TTL decade counter and a 7441 TTL 1 -
of-10 decoder. Figure 3 shows the pin outline for this versatile IC.

We can put the astable multivibrator we just built to work by using it as a source of clock pulses for the 4017 counter. Figure 4 shows one possible arrangement in which the 4017 successively flashes each of ten LED's. Only one LED series resistor is needed since only one LED is on at any given instant.


Fig. 4. Sequence generator.
Can you think of any applications for the 74C04 clock/4017 counter? If you adjust the clock so that it supplies one pulse each second, you can use the circuit as a handy darkroom timer. The circuit also makes an unusual light flasher or attention getter. Just arrange the LED's in a circle or in a random pattern and adjust for the best visual effect.

Another application for the circuit is as a sequence generator. I originally designed the circuit as a microinstruction sequencer for a homebrew digital controller made from a dozen or so TTL chips. The TTL drew so much current from my power supply that it was necessary to use CMOS for the sequencer circuitry. Since the controller was designed to operate at relatively slow operation (below 100 kHz ), CMOS was the logical choice in this case

Another excellent way to learn more about CMOS is to read Don Lancaster's CMOS Cookbook (Howard W. Sams \& Co., Inc., 1977). This 414-page book is filled with useful tips, applications and design possibilities.

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# Hobby Scene / <áy 

By John McVeigh


#### Abstract

Have a problem or question on circuitry, components, parts availability, etc? Send it to the Hobby Scene Editor, populah electronics, One Park Ave., New York, N.Y. 10016. Though all letters can't be answered individually, those with wide interest will be published.


"AC VOLTAGES"

## Q. I would like to know if commercial power is 110 volts peak or working voltage.-Paul Coelho, Mill Valley, CA.

A. An alternating voltage (terms like "ac voltage," "ac current," "dc voltage," etc. are really misnomers) can be expressed in several ways. The wave-form delivered by commercial power companies is sinusoidal, similar to the waveform shown in the figure. Such a wave can be described by its peak voltage $V_{P}$, the highest (or lowest) voltage the waveform attains during one cycle. It can also be described by its peak-to-peak voltage $V_{P-P}$, which, in the case of a symmetrical waveform, is twice the peak voltage. Finally, it can be described by its "working," "effective," or root mean square (rms) voltage, which is the square root of the mean value of the square of the instantaneous voltage. Strictly speaking, it is not the average value of the voltage, which for a sine wave is zero (the positive and negative portions of the signal cancel each other out in straight averaging). In non-mathematical terms, the rms value of a signal is the voltage required

## CW BANDPASS FILTER

Q. I need a sharp audio bandpass filter for CW reception. Do you have one that peaks around 800 Hz ?Dean Poeth, WB8TMD, Columbus, OH .
A. The best circuit I have on hand is the input stage of the Morse-A-Letter, the Morse decoder which appeared in the January 1977 issue. This circuit is not only an active bandpass filter, but has excellent agc characteristics as well. You could use a low-impedance earphone in place of the speaker at edge connector location A8. Of course, the digital circuitry can be omitted if you want to decode the Morse yourself!
to produce the same amount of energy (say, heat dissipated by a resistor) that a steady direct voltage would. For a "110volt" sine wave from a commercial power station, the rms voltage $V_{\text {RMS }}$ is 110 volts, the peak voltage $V_{p}$ is $\sqrt{2} \times 110$ volts or 155.6 volts, and the peak-topeak voltage $V_{P \text { p }}$ is 311.2 volts.

## HOW AN LED WORKS

Q. Why does a light emitting diode emit light, and why just one color?Lou D'Antuono, Queens, NY.
A. A LED is similar to a germanium or silicon diode in that it is composed of semiconductor material. On one side of the semiconductor junction, the material contains impurity atoms with extra electrons as compared to the majority material. These excess electrons are not bound into the crystal lattice structure, and will move with a little push. This side is called the $n$ side. On the other or $p$ side, impurity atoms are added that are deficient in electrons as compared to the majority atoms. Thus there are "slots" in the bounding structure called holes into which electrons can fall. The excess electrons occupy a "conduction band" and have higher average energies than the electrons in the "valence band" on the p side.

Forward biasing a pn junction pushes electrons from the $n$ side to the $p$ side, where they fall from the conduction band into the valence band (into holes). Any electron that falls into the valence band gives up energy in the form of heat or light. The process is shown in the figure. The wavelength of the emitted electromagnetic energy depends on the gap between the two bands. The wider the gap, the more energy is given up, and the higher the frequency (shorter wavelength) of the emitted radiation. If the gap is sufficiently wide, the radiation will be in the visible spectrum. The width of the gap depends on the material used to form the diode. That's why only one narrow group of wavelengths (color, in the case of visible light emitters) is radiated by a particular diode. Gallium Arsenide (GaAs) or Gallium Arsenide Phosphide (GaAsP) is used to form most LED's.


## SUPERSCOPE AIRCOMMAND AM CB TRANSCEIVER

Features LED meter readouts, channel-9 scan and SWR indicator.


THE Superscope Aircommand 40channel mobile AM CB transceiver departs from convention, as do a few others, by indicating relative signal strength, output power, etc., with a row of LED's instead of a meter. Another special feature is channel 9 scanning with individual squelch control and an audible alarm that sounds whenever a signal appears on channel 9 .

Aside from its special features, the Aircommand has a Delta tune control with center detent; individually switched automatic noise limiter (ANL) and noise blanker (NB); detachable microphone with a headphone connection at the mike plug; bottom-facing speaker; PA mode; external-speaker jacks; auxiliary input jack for feeding output from a portable radio or tape player through the transceiver's audio section; and electronic voltage regulation. It operates from a 12-to-16-volt dc, negative- or positive-ground source.

The transceiver measures 9 1/16" D $\times 79 / 16^{\prime \prime} \mathrm{W} \times 27 / 16^{\prime \prime} \mathrm{H}(23 \times 19.2 \times$ 6.2 cm ). It comes with mobile-mounting hardware and rubber feet for base station installations and carries a suggested retail price of $\$ 229.95$

Technical Details. Although a schematic diagram was not supplied with our
test transceiver, we were able to surmise the following details. The receiver employs double conversion to i-f's of 9785 and 455 kHz . A filter at the second $i-f$ provides the selectivity. The circuit lineup consists of the customary r -f amplifier, mixers, i-f amplifiers, detector, agc, squelch, audio ANL, and audio amplifier stages. (The last doubles as the modulator for the transmitter.)

The PLL system follows the usual pattern. It uses a $10,240-\mathrm{kHz}$ crystal-controlled oscillator from which the standard reference signal is derived. This oscillator is also used for the second conversion, using the difference between its frequency and that of the first $i-f$. The voltage-controlled oscillator (vco) at the first mixer operates at a frequency 9785 kHz below the CB signal to minimize the possibility of high receiver radiation above 28 MHz and to insure better image rejection.

The transmitter's carrier is derived from the vco, the signal from which is routed through the $r$-f amplifiers and the driver and power-output amplifiers. A multielement output network matches to 50 -ohm loads and minimizes spurious responses and harmonics that might otherwise cause TVI and other service interference. Amc is included in the transmitter to maintain high average
modulation without excessive overmodulation that could cause adjacentchannel splatter.

The rf gain and audio volume and the delta tune and sal (squelch) controls on the transceiver's front panel are arranged in concentric pairs. Although the SWR CAL control appears to be a concentric pair, it is actually a single control. The other functions are handled by lever-type switches located in the lower center of the panel. The switch at the left is for selecting between PA and $C B$ operation. The next three toggles are for switching in and out the noise blanker (NB), for Channel 9 sCan/hold/off selection, and for switching in and out the ANL. The last switch is for setting the LED display to indicate relative sWR in its up position and relative output power (PWR) in its down position and for calibrating the system in the center (CAL).

Just right of center on the front panel is a horizontal row of eight discrete red LED's. The one on the left is a power ON indicator. The next four LED's are labelled for relative power or SWR at 1.5, 3,5 , and 10. The sixth LED is labelled CAL for SWR and is additionally used with the last two LED's to indicate modulation level. Calibration labelling below the LED's is for relative signal strength starting with S4 and ending with S9 +20 dB . The lowest SWR that can be indicated is 1.5 ; if the low-end LED does not come on, it is assumed that the SWR is less than 1.5:1.

The red seven-segment numeric LED displays used for the channel indicator are located directly below the row of discrete LED's. The channel selector switch dominates the right end of the front panel.

With the channel-9 lever in the off position, normal channel selection is via the channel selector knob. With the switch in the sCAN position, channel 9 comes up only when a signal is being received on this channel, at which time a CH9 LED at the upper right of the panel comes on and a beeping tone sounds. By placing the switch in its HOLD position, channel 9 is kept open.

Laboratory Measurements. The sensitivity of the receiver measured a nominal $0.5 \mu \mathrm{~V}$ for $10 \mathrm{~dB}(\mathrm{~S}+\mathrm{N}) / \mathrm{N}$ with 1000 Hz at $30 \%$ modulation. The agc threshold range was 0.5 to $50 \mu \mathrm{~V}$. The agc held the audio output to within. 7 dB with a $20-\mathrm{dB}$ r-f signal change at 1 to 10 $\mu \mathrm{V}$ and to 10 dB with an 80-dB change at 1 to $10,000 \mu \mathrm{~V}$. A nominal $50-\mu \mathrm{V}$ input signal registered an $S 9$ reading.


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Adjacent-channel rejection and desensitization was 60 dB minimum. Image and i-f rejection were 75 and 80 dB , respectively. Other unwanted-signal rejection was 60 dB except in the area of 25 MHz , where the rejection was 40 dB .

The overall audio response at the -6dB points was 500 to 4000 Hz . The audio output was 5 watts with a $1000-\mathrm{Hz}$ sine-wave signal with an 8 -ohm output amplifier load. The THD at this output level was $4 \%$. The amplifier was also capable of putting out 6 watts of power at $10 \%$ THD and could also be driven into clipping. The PA output was about $10 \%$ lower in each test.

Voltage regulation is apparently provided for the transmitter, since the carrier output power maintained a constant 3.75-watt level at any power supply potential between 12 and 16 volts. Such regulation is quite unusual. In all cases, the power gradually drifted down to 3.5 watts after a short period of operation.

With a $1000-\mathrm{Hz}$ signal at microphone input levels 16 to 25 dB greater than that required for $50 \%$ modulation, the modulation held to nominally $90 \%$ at $2 \%$ THD. Under these conditions, the adjacentchannel splatter was down 45 to 50 dB . Using a $400-\mathrm{Hz}$ tone, the THD was $8 \%$. When we switched to a $2500-\mathrm{Hz}$ tone, the splatter was only 38 dB down but still within the FCC regulation.

During operation with maximum voice levels, the modulation slightly exceeded $100 \%$ on occasion on the negative and positive peaks. However, the splatter was down 50 to 55 dB . The overall audio response at the $-6-\mathrm{dB}$ points was 300 to 3500 Hz . The transmitter's frequency held to within -193 Hz on all channels.

User Comment. The transceiver's performance was startlingly good. Particularly noteworthy was the effectiveness of the ANL and noise blanker. Both performed excellently in our bench tests and in actual on-the-road tests in a noisy vehicle. The audio quality was unusually crisp on receive and of better-than-average quality on transmit. Moreover, the modulation was highly effective. In this respect, we observed that one must take care to avoid speaking too closely into the microphone to avoid "breathy" sounds on the receiving end.

The sophisticated channel 9 emergency scanning function, boasting independent squelch, worked beautifully. It's a most welcome feature.

Although the use of LED's to indicate operating parameters, as on this transceiver, provides only approximate val-
ues, the meihod can be of more use to the CB'er than the often supplied miniature analog meter movements. Also, the LED's add a colorful and flashy touch to the otherwise bland-looking transceiver:

About the only mincr criticism we can
make is that the black-finished transceiver's panel markings (though they are white) are difficult to see at night. In addition, the concentric rotary controls are located at the top of the panel, where they protrude far enough to obCIRCLE NO 104 ON FREE information caro
scure the labelling of the switches below them. The solution to this problem, of course, is to tilt the transceiver upward as is often done in mobile installations.

All in all, we feel this transceiver ranks among the best we have tested.

## HAZELTINE MODEL 1500 COMPUTER TERMINAL

Professional terminal in kit form.


AS ACTIVITY in the home microcomputer field has matured, several "big-narre" manufacturers have introduced products for use in this new market. One such manufacturer is the Hazeltine Corporation, which is making
available its well-known Model 1500 computer terminal in semikit form at a savings over the assembled version's price.

The Model 1500 terminal features an 80 -character/line $\times 24$-line format with upper- and lower-case characters. It can be interfaced with a microcomputer via either a $20-\mathrm{mA}$ current loop or an RS-232 system.

The terminal measures $20.6^{\prime \prime} \mathrm{D} \times$ $15.5^{\prime \prime} \mathrm{W} \times 13.5^{\prime \prime} \mathrm{H}(52.1 \times 40 \times 34.3$ cm ) and weighs $35 \mathrm{lb}(15.9 \mathrm{~kg})$. Contact your local store for price.

General Description. The Model 1500 is a classical high-quality terminal. lis character set is displayed in a $7 \times 10$
dot matrix. Standard and reverse video is provided for all 94 ASCII characters available. The refresh rate is 60 frames per second, noninterlaced, with the display on the built-in $12^{\prime \prime}(30.5-\mathrm{cm})$ diagonal glare-proof CRT screen. Dual intensity of any word or character is selectable, and all data is stored in an on-board $2048 \times 8$ bit RAM.

The terminal also has an on-board ROM that accepts a number of remote (computer-generated) commands.
These include cursor address, incremental cursor, read cursor address, clear screen, clear foreground, clear to end of screen, clear to end of line, home cursor, set high/low display intensity, audible alarm, backspace, keyboard lock/unlock, insert/delete line, and remote tab function.

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## Hazeltine's Model

 1500 computer terminal comes in semikit form as shown here and is easily assembled.rent loop or RS-232 system provides full or half-duplex operation, with compatibility with a 103A or 202 modem. Eight baud rates between 110 and 19,200 baud are individually selectable. Also provided is a choice of odd, even, or no parity and a choice of one or two stop bits. All selections are made via a set of DIP switches located under a small liftoff panel near the keyboard.

User Comment. It is difficult to consider the Model 1500 terminal as a kit
since its major electronic elements come assembled, tested, and guaranteed by Hazeltine. All that must be done is to follow the assembly instructions given in the well-written manual as the various elements of the terminal are mounted. Then the individual elements are simply interconnected with a solderless wiring harness.

Exercising care in installing the elements and making neat interconnections, we assembled the kit in about four hours, counting from the time we
opened the carton in which the kit was shipped. At the end of this time, our terminal was up and running with our computer. The job of assembly is very simple and straightforward; it can be accomplished by anyone who has a rudimentary knowledge of electronics and mechanics. The only hand tools required are a screwdriver and pliers.

Following assembly, the first test we made was to deposit a full screen of m's and w's to test the edge-to-edge clarity of the character display because of the wide bandwidth required to keep the characters from filling in. The full screen of characters also permits checking the linearity of both the horizontal and the vertical portions of the sweep. In our test sample, these were extremely linear and in sharp focus.

When we filled the screen with upperand lower-case characters, the terminal's character generator, in conjunction with the wide bandwidth, produced an extremely clear display. All keys and key combinations that generate ASCII codes operate in a "typamatic" mode. That is, depressing any of these keys initially produces the selected character onscreen; and, if the key is held down for

more than three-quarters of a second, the character is repeated at a rate of 15 characters per second. In all cases, the terminal displays its characters in a very professional manner, including scrolling. The alphanumeric keyboard and the numeric cluster have finger-tip contoured keytops and a good "touch."

While we were using the terminal, we tried changing the baud rate without changing that of the computer we were using. We discovered that if terminal and computer are not set to the same baud rate, the terminal initiates an insistent "beeping" via its internal alarm and the CRT screen displays a line of PE (parity-error) symbols.

Using the simple instructions provided in the manual that accompanied the kit, we made several changes to a couple of our BASIC programs to permit us to use some of the special remote commands to the terminal (such as screen clear and two-level brightness). Also, using the lock/unlock commands, we managed to defeat the keyboard to prevent anyone's toying with it while a special program was running.

Examining the voluminous and very well-written assembly and operating manual, we discovered that the Model 1500 is more than just a computer terminal. The system is controlled by an 8080A microprocessor and a complete set of support IC's, has internal ROM and RAM, and video display section. With its built-in keyboard, I/O porting, and high-resolution CRT monitor, this excellent terminal can certainly give homebrewers some interesting thoughts on expanding the system.

The manual provided with the terminal kit is by far the most complete we have seen this side of military gear manuals. The maintenance sections are complete, with oscilloscope waveforms, voltage measurements, logic guides, flow charts, etc. There is virtually nothing that can go wrong with the terminal that is not covered by the manual.

If you are a serious computer enthusiast and have spent time and money upgrading your computer system, you should give some thought to moving up to a Hazeltine Model 1500 "professional" computer terminal, especially at its reasonably moderate price. The same goes for small businesses and educational institutions on the lookout for a commercial terminal at a savings in cost.

After working with the Model 1500 for several weeks, we were very favorably impressed with its performance.


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## NON-ENGLISH BROADCASTS TO NORTH AMERICA

THOSE who are not initiated into the world of DX listening often assume one must be multilingual to get anything out of short-wave broadcasts. This is emphatically not the case. Yet, nearly every country that broadcasts in English to North America also broadcasts in its native language and a few use additional languages as well. They deserve your attention, and the native music can be enjoyed with no language barrier

All these stations send out free program schedules on request. If you don't hear an exact address announced on the air, simply use the station name, city and country. The stations get so much mail that little else is really necessary. Some are reticent about sending schedules in a certain language, unless you write to them in that language. Here's a survey of non-English broadcasts to North America (all times are GMT; frequencies are subject to change.)

Albania. R. Tirana's English broadcasts are so dogmatic and insulting that many SWL's prefer the incomprehensible Albanian. Try 9790, 7300 and 6200 kHz at $0000-0100$ and 0200-0230; or 11,985 and 9500 kHz at 1030-1100.

Argentina. RAE English is in three one-hour blocks, exceeded by Spanish at 0000-0300 and 0400-0600 on 9690 kHz weekdays. On weekends, all programs are in Spanish--before 2400 on $11,710 \mathrm{kHz}$, after 0000 on 9690 kHz .
Australia. Has French for the Pacific at 0000-0100 on 15,320 and 15,160 kHz . Quebec is just a bit farther in the same direction. In our summer, try again at $0500-0600$ on additional frequencies in the 16- and 19-Meter bands.
Austria. ORF has German at 2300-0100 and 0200-0330 on 6155 and 9770 kHz ; French at $0100-0130$; and German at 0400-0430 and 0500-0600 on 6015 kHz .

Belgium. BRT-4 precedes its 0015 English with a half hour of Dutch on the same frequency.
Bulgaria. R. Sofia's 0000 English is
flanked by Bulgarian at 2330-2400 and $0100-0130$ on the same frequency.

Canada. Among RCl's many French broadcasts audible in North America, these are intended for us: 0100-0127 on 5960 kHz ; 0200-0227 on 9605 and 5960 kHz ; 0330-0357 on 9535 and 5960 kHz . Listen for a DX show and mailbag on Sundays (GMT Mondays). Spanish: 0030-0057 on 9535 kHz ; 0130-0157 on 9535, 6185, and 5960 $\mathrm{kHz} ; 0230-0257$ on the same channels as 0300 English.

China. R. Peking in Chinese generally uses frequencies that at other hours are in English. Cantonese is at 0100 and 0300; Standard Chinese at 0200 and 0400.

Cuba. R. Habana Cuba's extensive Spanish schedule is designated for North, Central and South America. If you can't find this, your receiver is turned off! Also, French at 0300-0330 on 11,760 kHz.

Czechoslovakia. R. Prague's "Interprogramme" at 2300-0100 on 9630 and 6055 kHz includes Czech \& Slovak at 2300, German at 2330, and French at 0000. Regular broadcasts include Czech \& Slovak in the half hour preceding the 0100 English on most of the same channels.
Chile. The Voice of 'Free' Chile uses the same antennas and frequencies for non-English as for English: French at 0000,0130 , and 0300 and Spanish at 0100 and 0230.
Denmark. R. Denmark is stymied by a law that forbids it to broadcast in English! Danish to North America is on $15,165 \mathrm{kHz}$; most transmissions before 1805 GMT are either for Greenland or North America in general.
Dominican Republic. R. Clarin is in Spanish whenever it isn't in English. Check for a mailbag show at 2230; Sundays after 1500 GMT, hear results of the national lottery-"90 pesos!"-being shouted several times an hour.

Ecuador. HCJB—or the evangelists
buying time on it-have the curious idea that it's worthwhile to broadcast to North America in Russian at 0130-0200 in the 19- and 25-Meter bands. Last year there was even Czech. French follows Russian at 0200-0230 on the same frequencies. Spanish comes this way at 1230-1530 on $11,910 \mathrm{kHz}$; 1630-2100 on $15,160 \mathrm{kHz} ; 2230-0500$ on 11,960 $\mathrm{kHz} ; 0100-0500$ on 6050 kHz . Japanese to Japan crosses North America on the way at $1130-1225$ on 9715 kHz .

Egypt. R. Cairo follows its English program on 9475 kHz with an hour in Arabic at 0330.

Finland. R. Finland has been ordered to cut its external broadcasts 2 hours and 45 minutes per day. Will this come out of English or Finnish? The latter was scheduled in the hour preceding the 1300 English on the same frequency (exc. Swedish Wed. 1230-1255), and again at 1500-1600 (exc. Wed. 1500-1525 Swedish).
France. Does not use English for North America or any language in our evenings. But at 1200-1710 it's a powerhouse in French on 15,440 and $17,780 \mathrm{kHz}$. When France is on summer time, this shifts to 1100-1610.

Germany, East. R. Berlin International inserts 45 minutes of German between the 0100 and 0230 English on the generally inaudible 9730 kHz . On the same frequencies as the 0330 English, German follows at 0415.
Germany, West. Almost every night, Deutsche Welle jokes about its "enormous" 20-minute English program. For true enormity, try its $230-\mathrm{min}$ ute German program, aired twice at 2200 and 0200 on many hard-to-miss frequencies from Germany, Rwanda, Malta, Canada, Antigua and Montserrat. For this, DW publishes a day-by-day, hour-by-hour schedule similar to BBC's "London Calling." DW also provides morning news in German at 1300-1320 and 1330-1350.

Greece. On the same frequencies as English, Voice of Greece is in Greek at 0000-0015, 0045-0215, 0230-0350, 1200-1215, 1230-1250, 1500-1515, 1530-1550. And in French at 0030-0045.

Hungary. R. Budapest broadcasts more Hungarian to us than English, and on the same frequencies, at $0130-0200$, 0230-0300, 0330-0400. On GMT Mon days this expands to 0130-0300 and 0330-0500.

Iran. Voice of Iran is slow to beam English our way, but there's plenty of music and Farsi after 2000 on 15,084
kHz . During the summer this is audible well into the night.

Israel. Each Israel Radio English program is accompanied by one in French on the same channels: 0515-0530, 1230-1300, 2030-2055, and 2200-2230. Hebrew news follows English on same frequencies at 2300. The Hebrew home service is relayed our way on mostly weaker SW transmit-ters-at 0400-0610 on 7465 and 5882.5 kHz ; 1740-2315 on 9355 and 12,077 kHz . At times, a channel above 15,500 kHz can also be heard.

Italy. English broadcast to Japan is longer than the one to North America! But Italian is something else: 2230-0100 on $11,905,9710,9630,9575$, and 6010 kHz . The variety of music and information on this service puts their leaden English $1 / 3$ hour to shame. It's followed by French at 0120-0140.

Japan. R. Japan has a 15-minute Japanese newscast every hour of the day at 15 past (except for half-hours at 1030, 1430 and 2330) on same frequencies as English General Service, varying with season and time of day among $15,105,9505,5990 \mathrm{kHz}$. The 0130 En glish is preceded by Japanese at 0100, and is followed by Spanish at 0230-0300.

Korea, South. R. Korea, unlike R. Pyongyang, broadcasts more Korean than English our way. At 0230, 0430, 1600 , and 2030 on 11,850 and 9640.

Libya. The People's Revolutionary Broadcasting is in Arabic only, most of the day on $15,100,11,700$, or 9500 kHz .

Netherlands. Both R. Nederland relays, in Madagascar and Bonaire, come in better here than in Holland itself. Daily Dutch from Bonaire for North America is at 0030-0120 on 6165 and 6020 kHz ; 0430-0520 on 9590 and 6165 kHz . And on Sundays only at 2130-2220 from Holland on 9715 kHz ; from Bonaire 2230-2320 on 15,320 and $15,180 \mathrm{kHz}$. Spanish: 0330-0420 on 9590 and 6165 kHz .

Norway. Check R. Norway's English schedule. The preceding hour on Sundays, and the entire 90 minutes on weekdays is in Norwegian, with some English music announcements.

Poland. R. Warsaw insists on not specifying what times are Polish and what are English to North America-but usually the Polish is at 0230-0300 and 0330-0400 on the same frequencies. Most of them are inaudible except at midsummer.

Portugal. Another country with more native language than English. Por-
tuguese is at 0100-0300 and 03300500 on the same frequencies as English.

Romania. R. Bucharest until recently had Yiddish to North America. But now, on same channels as English, it's only Romanian at 2300-2400 and 0230-0300.

Spain. RTVE employs a different and larger frequency net for Spanish than for English. A two-hour service is aired thrice, at 2300, 0100, and 0300. Prime channels are $11,945,11,775$, 9630,9360 , and 6120 kHz . Also, for seamen in the "northwest Atlantic," from 2145

Sweden. R. Sweden has Swedish at 2330, following English at 2300; at 0100 and 0200 on the same frequency as 0030 English; at 1430 on the same as 1400 English. French is at 0230 on a single frequency. Some of the Spanish broadcasts for "Latin America" are on the same beam as eastern North America: at 0000 and 0130 , it is the same as 0030 English; and at 0300, it is the same as 0230 French. This Spanish crossing on the way to Central America holds true for many other European stations.

Switzerland. Each Swiss language gets exactly the same time as English, on the same frequencies in subsequent half hours: 0245 in German, 0315 in French, 0345 in Italian; 0500 in Italian, 0530 in German, 0600 in French; 1345 in German, 1415 in French, 1445 in Italian, and Spanish at 0215.

Taiwan. VOFC has standard Chinese at 2040-2140 on the same channels as 2140 English, and surrounding the 0100 English are standard Chinese at 0000 and Cantonese at 0200-0300, on mostly same channeis.

Thailand. Has perpetually inaudible North American service in English at $0415-0515$ around $11,905 \mathrm{kHz}$, and caps this with French to North America at 0520-0550.

USSR. R. Moscow's North American service in English is only the tip of the iceberg. There are several different Russian services. SW relays of the "Mayak" second-program home service can be heard throughout the day and evening. R. Rodina (Homeland) at 2330-0030 and 0200-0300. A distinct Golos Rodiny at 0130-0200. "Allantika," for seamen in the Allantic, at 1300-1400; Pacific Ocean Radio Station at 0700-0800 and 1930-2030. R. Kiev has Ukrainian at 2200-2300 and 0330-0400. R. Vilnius is in Lithuanian at 0100-0130. Most of Yerevan's nondaily transmission at 0300-0330 is in Armenian.

Vatican. Radio follows 0100 English with French at 0115-0130, conflicting in time with Rome's other shortwave station.

If you'd just like to hear broadcasts in a certain language, the Voice of America may be the answer. Reception is best in the central states.

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by Leslie Solomon

## HIGH-RESOLUTION GRAPHICS

$\mathbf{N}^{\circ}$O MATTER how many alphanumeric characters your terminal can put on the screen, there are times when those characters are no substitute for graphics. However, when you examine the field, you will find that many low-cost "graphics" displays are simply nonalphanumeric characters resident within the character generator used, or they are limited to simple bar graphs. There are boards and systems that have some graphics capability, but the graphics are generally small "blocks" rather than discrete dots, so their overall effect is somewhat coarse.

Recently we had an opportunity to try out a new graphics board, the ALT256*2 Graphic Display Interface (\$395 assembled and tested) from Matrox Electronic Systems, POB 56, Ahuntsic Stn., Montreal, Canada H3L 3N5 (Tel: 514-481-6838), and available through most computer stores. This S-100 bus board can display 256 lines with 256 individually addressed dots on each line, thus allowing for excellent resolution. Each dot can be made black, white, or, if three boards are used, any color. The system can work with as many as eight boards but in this case, direct manipulation of the color byte is required as explained in the manual. Access time is $3.4-\mu \mathrm{s}$ per dot, and the entire screen can be erased with a single instruction. The board can use external sync, and can be strapped for USA or European TV standards for noninterlaced display.


Graphics using the Matrox interface, shown at upper right.

The ALT-256*2 board contains 65k $\times$ 1 of RAM, and individual 8 -bit registers for the $X$ and $Y$ coordinates and the data for each dot. Thus, it is easy to specify an individual dot anywhere on the screen, and determine whether that particular dot should be black or white (or color if you use extra boards). The screen can be erased black or white depending on the data sent to the input.
The software provided with the ALT-256*2 consists of a very detailed manual, and a pair of paper tapes. One tape contains the MTX GRAPH software that is configured as a series of callable subroutines while the second tape contains a demonstration program that uses the MTX GRAPH to create a continuous action graphics display. Total memory required is about 1 k .

Six of the seven subroutines in the MTX GRAPH package are: INITG (initializes the system to standard defaults); PAGE (erases the entire screen); CURSOR (allows positioning the cursor at point $\mathrm{X}, \mathrm{Y}$ ); DOT (sets the point defined by the cursor); LINE (creates a line of dots between the current cursor position and the point $X, Y$ ); and CHAR (displays an ASCII character at the current cursor position). In the latter routine, the cursor is left at the next character position. Certain control characters are then used to select color, fixed or proportional character spacing, and dot size. The last routine is called ANIMAT, which produces a pause until the start of the next vertical blanking period, and is used in animation routines.

After plugging the ALT-256*2 into our S-100 computer, we loaded both tapes, and following the manual, started the demonstration program. Because we had only one board, and a monochrome monitor, we had no experience with color operation.

Comments on Use. The demonstration is impressive. It starts with a set of large characters on screen, followed by an interesting "lace curtain" effect that

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creeps across the screen. Suitable pauses are built-into the program so that you can study the various patterns on screen. Then follows a series of fan-like displays that illustrate line-calling routines, and some small (but very clear) alphanumerics. After a few more demonstrations of the various calling routines, a little man is made to "walk" across the screen, with swinging arms, to demonstrate the animation portion.

The manual contains a considerable amount of information and includes a complete listing of both the demonstration program and the MTX GRAPH package.

We have been playing with some 8080 machine language programs to "call" the various graphics subroutines and learn how best to use the high-resolution graphics. The results have been surprisingly good. We are also writing some BASIC programs using FILL and CALL commands to access the graphics software, also with successful results. One future project is a joystick interface for some really adventurous picture creations.

If high-resolution graphics appeals to you, drop into your local computer store and take a look at this new system from Matrox

280 Things. North Star Computers, Inc., 2465 Fourth St., Berkeley, CA 94710 (Tel: 415-549-0858) has intro-


The North Star 16k RAM board for use with 8080 or $Z 80$.
duced a couple of $Z 80$ goodies. The first is a $4-\mathrm{MHz} \mathrm{Z80}$ board compatible with an S-100 bus, called the ZPB. It is available for $\$ 199$ in kit form or $\$ 259$ fully assembled. Features include auto-jump startup, vectored interrupts, operation with or without front panel, and space for 1 k of 2708 EPROM. The EPROM option is $\$ 49$ for the kit and $\$ 69$ assembled.

The other $Z 80$ device is a 16 k RAM board (S-100) for use with either 8080 or Z80 and will operate at full speed (no wait states) at 4 MHz . The RAM's are 200 ns , and on-board refresh is provided. Bank switching capability is provided
and board addressing is switch-selectable in two 8 k sections. An important feature is the availability of a parity check option. The 16 k board is $\$ 399$ in kit form, and $\$ 459$ fully assembled. The parity option is $\$ 39$ kit and $\$ 59$ assembled.

PROMing. Oliver Audio Engineering, Inc., 676 West Wilson Ave., Glendale, CA 91203 (TeI: 213-240-0080), which brought us the first low-cost paper-tape reader, now introduces its PP-2708/16 PROM programmer (\$249 as a kit, \$295 assembled and tested). The new programmer plugs directly into any 2708 or TMS-2716 socket, and the PROM to be programmed is mounted in its zero-insertion-force socket. The data is dumped over the eight lower address lines using the OAE interface. No additional power supplies are required and all timing and control sequences are au-


The OAE PP-2708/16 PROM programmer plugs into any 2708 or TMS-2716 socket.
tomatically handled by the programmer.
Because of this simplicity, only a short software routine is required. A five-foot flat ribbon cable interconnects the programmer with the PROM socket.

Bus Stop. If you have an Altair/S-100 bus system, or are planning to get one, there are a couple of new motherboards of which you should be aware. The first is from Vector Electronic Co., Inc., 12460 Gladstone Ave., Sylmar, CA 91342 (Tel: 213-365-9661). Their Model 8803 costing $\$ 29.95$ accommodates 11 plug-in boards, and can have passive or active bus termination. One slot position may be used to interface the motherboard for system expansion. Twelve tantalum capacitors are included to suppress transients on the various power supply lines. Ground and +5 -volt traces are rated at 10 amperes while the $\pm 12-$ volt busses are rated at 7 amperes.

The second S-100 bus motherboard is from Thinker-Toys, 1201 10th St., Berkeley, CA 94710 (Tel: 415-5277548) and costs $\$ 76$. Called the Wunderbuss, this new motherboard features full shielding of the signal paths, and active termination of all data lines. Signal isolation is achieved by a cross-coupled system of ground lines interlaced between signal lines. The motherboard with 10 edge connectors is available for $\$ 120$, and with 20 edge connectors the price is $\$ 154$.

PROM/RAM Board. Now available from Vector Graphic, Inc., 790 Hampshire Road, Westlake Village, CA 91361 (Tel: 805-497-6853) is a new S-100 board that occupies two independently addressable 8 k blocks, has 1 k of RAM on board, and a capacity of up to $12 k$ of 2708 EPROM's. Complete addressing flexibility is provided, and video or diskoperating system boards can be nested in the 3 k of unused space. MWRITE logic and jump-on-reset allow operation without a front panel. A 24 -command
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PROM monitor is available to interface with most I/O boards. Price is $\$ 135$ in kit form, $\$ 175$ assembled.

Print an Apple. If you have an Apple II computer, then you should be aware that Microproducts, 1024 17th St, Hermosa Beach, CA 90265 (Tel: 213-374-1673), has introduced its PCB that interfaces the Apple II with the SWTP PR-40 Printer. The board, at $\$ 49.95$ assembled, plugs into the Apple II and comes with an interconnecting ca-
ble. A cassette with operating software is also provided. The printer prints one line at a time when the return key is depressed. The printer subroutine can also be called in a BASIC program for usual printing.

New I/O Port. Since getting stuff into and out of a computer is somewhat of a necessity, and since most hobbyists now use more than one I/O device, mul-ti-porting is becoming very important. Dajen Electronics, 7214 Springleaf Ct .,


A Microproducts board interfaces an Apple II with an SWTP PR-40.


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Citrus Heights, CA 95610 (Tel: 916-723-1050) has one answer in their System Central Interface (SCl) board selling for $\$ 285$ in kit form and $\$ 345$ assembled and tested. This S-100 bus board provides a serial port with RS-232 and 20/60-mA current loops; baud rates from 45 to 9600 ; three independent 8 -bit parallel ports that can be programmed bitwise for input and latched output; biphase (Tarbell) cassette port; onboard relays for control of two recorders; and three status lines to control an automatic tape deck. Also included are 256 bytes of RAM for stack and buffer storage; a 2708 programmer; space for three 2708 s; and $2 k$ system monitor program, having 18 commands. All IC's are socketed, and the board will work with $4-\mathrm{MHz}$ Z80 systems. All output connectors, cable and plugs for recorder are provided.

1802 Items. Netronics, 333 Litchfield Rd., New Milford, CT 06776 (Tel: 203-354-9375), the source for the Elf-II, has announced a 4 k RAM board for the Elf-II bus at $\$ 89.95$ plus $\$ 3$ postage/ handling. Using 2102's and requiring $500-\mathrm{mA}$ from the 5 -volt line, the memory is buffered and decoded with page selection in 4 k blocks anywhere in memory. There is an on-board regulator and three-state outputs. The board will preserve the 256 bytes of the original.

This same firm also has an Elf-II prototype board ( $\$ 17$ plus $\$ 1$ postage/handling) that has room for 32 IC's in a mix of $14,16,20,22,24$, or 40 pins. Wirewrap or solder pencil connections can be used. Like the memory board, this board has gold-plated Elf-II bus connector fingers, and has provisions for a 5volt regulator.

The third item announced by Netronics is an outboard Elf-II power supply that provides $\pm 8$ volts at 5 amperes, unregulated, and $\pm 16$ volts at 1 ampere, also unregulated. Price is $\$ 34.95$ plus $\$ 3$ for postage and handling.

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8080 Software Contest. College and university students and faculty are eligible to enter a software contest announced by Intel's Insite program library. Entries must be written in Intel assembly language or PL/M and must include a source listing and test program to assure program validity, and a source paper tape or diskette. More than $\$ 28,000$ worth of microcomputer development equipment, plus numerous $\$ 100$ memberships in the Intel User's Library, will be awarded to colleges and
universities submitting the best programs by June 30, 1978. Monthly winners will receive Intel PROMPT 48 or PROMPT $80 / 85 \mathrm{mi}-$ crocomputer design aid systems. The grand prize will be an Intellec 888 Microcomputer Development Center, including 64k RAM, du-al-drive one-megabyte diskette system, CRT console, and software. Programs will be judged for their originality, documentation, creativity and applicability to microprocessors. Both individuals and teams may enter. For entry forms and details, write: Insite Library Contest, Intel Corp., Microcomputer Div., 3065 Bowers Ave., Santa Clara, CA 95051.

8080/280 Cassette Operating Sys. tem. The ZAPS Cassette Operating System includes a Z80 assembler, text editor, inmemory file system, labelled cassette tape storage system and other utilities. Runs in 14 k of inemory, including buffers and 1 k for symbol table, on most 8080 and Z 80 systems. The assembler processes the Zilog mnemon-

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High-Level Language for 6502. FCL65E, a high-level language similar to DEC's FOCAL, is available for 6502-based microcomputer systems. The 6.5 k program offers 8 -to- 9 -digit accuracy, 8-level priority interrupt handling, string variables and functions. Commands include Ask, Comment, Continue, Do, Erase, For, Go, If, Modify, On, Quit, Return, Set, Type and Write, plus several commands called by symbols. The interpreter permits editing, corrections of current line, debugging and error detection, plus functions for absolute value, integer, integer rounding, random number, I/O device functions, ASCI/decimal and decimal/ASCII conversions, terminal echo suppression, memory examine and deposit, user subroutines, string comparisons, and others. FCL65E is available for TIM- and KIM-based systems. All system-specific I/O calls are located in a zero page block for easy modifica-
tion to other 6502 systems. A complete listing is $\$ 35$, a mini-manual is $\$ 6$ and a 104 -page user's manual is $\$ 12$. Hex or binary paper tape or hex dump is available for $\$ 17$. The program is also available on KIM cassettes in Hypertape ( 6 X speed) format at $\$ 19$. At regular speed, KIM cassettes are $\$ 23.50$, and $3 X$ speed cassettes are $\$ 20.25$. Write: The 6502 Program Exchange, 2910 Moana Ln., Reno, NV 89509

6800 Text Editing $\&$ Processing Sys. tems. The editing system is line- and con-tent-oriented for easier assembly-language program development and document preparation. Edit directives include Append, Change, Copy, Move, Delete, Insert, Overlay, Print, and Replace-plus pointer-addressing and string searches Other commands display output-format and up to 20 tab-stop settings, and display or suppress line numbers. Renumbering is also available, as are paper-tape or cassette save, write, read and gap (null string) commands. The Text Processing System, a companion program, is used to format edited files for printing or display. Options include paging, titling, page numbering, paragraphing, spacing and right-margin justification. The 6800 Text Editing System is $\$ 23.50$; the Text Processing System (which can also
be used independently) is $\$ 32$, with both in complete manual form with source listing. KCstandard cassettes are $\$ 6.95$ each, for either program. Write: Technical System Consultants, Inc., Box 2574, W Lafayette, IN 47906.

PDP-8 Simulator for 8080. The Simul8tor permits 8080 computers to run PDP- 8 programs, which are widely and inexpensively available. (However, speed limitations may prectude the use of such programs as PDP-8 BASIC and FOCAL, and the availability of DP-8 software on paper tape rather than cassette may be intolerable to some users.) Simul8tor is available on Intel-format paper tape or Tarbell cassette for $\$ 20$. Write: The Amide Corp., Box 600. Sag Harbor. NY 11963

6800 Disassembler. Available in punched paper tape (MIKBUG format) with assembly listing, object code and instructions, this program will print an assembly listing of the object code of any program stored in memory. It operates in less than 1.8 k bytes, beginning at 0800 hex (2k decimal). \$12.95. Write: Software Exchange, 2681 Peterboro, W. Bloomfield, MI 48033.

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## CHARGE INDICATOR

A LED indicator can be added to the charging stand for your rechargeable soldering iron to indicate whether or not the charging current is actually flowing. The circuit shown here will cause a negligible change in the normal charging current if your charger is a simple halfwave rectifier (as most are). This is especially important if you have a "quick charge" iron.


The LED will glow only when the iron is in the stand and making contact with the charger's output pins. Diodes D1 through D3 must be able to handle the maximum charging current. My "quick charge" iron draws about 440 mA maximum, so 1 N4002's are suitable. A standard TIL-32 or similar LED can be used. Resistor R1 is selected to limit LED current to a safe value. My charger has a 3-volt secondary, so I used a 330ohm, $1 / 2$-watt resistor.

When the iron is in its stand, diodes D1 and D3 allow full current to flow through the iron's batteries for one half of the ac cycle. During the other half cycle. only the low LED current flows through LED1, R1, the iron and $D 2$.

Most chargers have enough room inside to accommodate the diodes and resistor. The LED should be mounted in a hole drilled in the charger housing so as to be plainly visible when the unit is operating. Also, take care to observe polarities when connecting the new components. Diode D4 (same type as D1 through D3) is needed only if the transformer's output voltage exceeds the reverse voltage rating of the LED.-K.L. Kingston, Lafayette, $\mathbb{N}$.

## REMOVING ROSIN

While desoldering some DIP IC's, I discovered that removing the rosin from between the pins was very tedious. I tried cotton swabs, but the cotton came off the stick and left me with a bigger mess than I had started with. Happily, I discovered a solution. I took the discarded swab sticks and cut them at a $45^{\circ}$ angle at one end. The point can be easily maneuvered between DIP pins, and the flat edge can be used for cleaning larger areas. As the end gets mangled, simply cut it off and start again. Toluene, available in most hardware stores, makes a good rosin solvent.-Rebecca S. Peutz, Olympia, WA.

## RECONDITIONING NUT DRIVERS

The hex socket of an inexpensive, hand held nut driver can become worn with age, allowing slippage. But most are made with deep enough sockets to allow resurfacing. Grind down the socket past the rounded edges, but be careful not to remove so much metal that the socket becomes 100 shallow to accommodate the average hex nut.-Joseph Smolski

## PANEL MOUNTING LED'S

Mounting LED's on a panel can be a problem. Although they can always be wedged into grommets, this tends to block off much of the light output. But there's another approach you can take when using the larger "dome" LED's with a shoulder molded around their bases (MV-5000 series and similar). Drop the leads through a pair of adjacent holes in a scrap of $0.1^{\prime \prime}(2.54-\mathrm{mm})$ perforated board. (Enlarge one hole slightly if necessary.) Drill out a hole on either side for mounting hardware. Then drill a hole in the panel to pass the dome but not the shoulder, and two holes for mounting hardware. Insert the LED and perf board from the rear and secure with No. 2 machine hardware. A fancier mount can be made from a discarded nylon banana jack. Remove the innards and cut off most of the bushing, leaving just enough thread to hold the nut firmly. Cement the LED in the top of the jack with leads passing through the body. If the leads are flexible, wedge a bit of cork or rubber into the bottom of the assembly to keep the leads from shorting. Don't forget to identify polarity first! Then mount the assembly onto the panel.Parke S. Barnard


## Electronics Library

## IC OP-AMP COOKBOOK

by Walter G. Jung
Basic op amp theory is covered in this work, as well as practical circuit applications. Over 250 circuits are presented. The book is organized into three parts: an introduction to the IC op amp with discussion of general considerations; practical circuit applications; and two appendices of manufacturers' reference material. Circuits such as audio preamps, power amps, active filters, log amplifiers, function generators, and current-differencing amplifiers for general- (709, 741, and 101) and special-purpose IC's are included.
Published by Howard W. Sams \& Co., 4300 W. 62nd St., Indianapolis, IN 46206. 592 pages. \$12.95, soft cover.

## talk-back TV

## by Richard Keith

In the not-too-distant future, viewers will be able to "talk back" to their TV receivers and get results. In fact, there are already a number of locations around the U.S. where cable TV is offering a host of talk-back services that range from electronic mail to meter reading. In anticipating the time when talk-back TV will be an everyday reality, this book discusses the history of the medium and details the possibilities it has to offer to the user. It reveals how talk-back TV is already providing a variety of services to the modern family and includes detailed descriptions of a number of currently operating systems. It also discusses the potential the new medium will have for intrusion on privacy.
Published by Tab Books, Blue Ridge Summit, PA 17214. 238 pages. \$9.95 hard cover: $\$ 5.95$ soft cover.

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tions between ideal and real waveforms are illustrated, and the basic waveform-number relationships are explained without the use of higher mathematics.
Published by Reston Publishing Co., Inc., P.O. Box 547. Reston, VA 22090. Hard cover. 200 pages. $\$ 15.95$

## MICROPROCESSORS: FUNDAMENTALS \& APPLICATIONS

Edited by Wen C. Lin

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| 7476 | 276-1813 | 596 |
| 7485 7486 | 276-1826 | 1.19 496 |
| ${ }_{7490}$ | 276-1808 | 796 |
| 7492 | 276-1819 | 695 |
| 74123 | 276.1817 | 996 |
| 74145 | 276-1828 | 1.19 |
| 74150 | 276-1829 | 1.39 |
| 74154 | 276-1834 | 1.29 |
| 74192 | 276.1831 | 1.19 |
| 74193 | 276.1820 | 1.19 |
| 74194 74196 | - 276 27-1832 | 1.19 <br> 1.29 |

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| ${ }^{74} \mathrm{C} 00$ | 276-2301 | 49\% |
| $74 \mathrm{C02}$ | 276-2302 | 496 |
| $74 \mathrm{C04}$ | 276.2303 | 495 |
| $74 \mathrm{C08}$ | 276-2305 | 496 |
| 74 C 74 | 276 -2310 | 896 |
| ${ }_{744} \mathbf{C 7 6}$ | 276.2312 | 894 |
| $74 \mathrm{C90}$ | 276-2315 | 1.49 |
| 74 C 192 | 276.2321 | 1.69 |
| 74C193 | 276 -2322 | 1.69 |
| 4001 | 276-2401 | 496 |
| 4011 | 276 -2411 | 496 |
| 4013 | 276-2:13 | 898 |
| 4017 | 276-2417 | 1.49 |
| 4020 | 276-2420 | 1.49 |
| 4027 | 276-2427 | ${ }^{894}$ |
| 4049 | 276-2449 | 694 |
| 4050 | 276-2450 | 696 |
| 4514 | $\begin{array}{r}276-2447 \\ \hline 76-2490\end{array}$ | 1.69 1.49 |

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| :---: | :---: | :---: |
| 301 CN | 276-017 | 496 |
| 324 N 339 N | 276.1711 276.1712 | 1.49 1.49 |
| 386 CN | 276-1731 | 999 |
| ${ }_{555 \mathrm{CN}}$ | 276-1723 | 798 |
| 556 CN | 276-1728 | 1.39 |
| 566 CN | 276-1724 | 1.69 |
| 567 CN | 276-1721 | 1.99 |
| ${ }^{723 C N}$ | 276-1740 | 699 |
| 741 CN | 276-007 | 496 |
| ${ }^{741 \mathrm{H}}$ | 276-010 | 496 |
| 3900 N 3909 N | - 276.1705 | ${ }_{996}^{996}$ |
| $391+\mathrm{N}$ | 276-1706 | 1.99 |
| 4558 CN | 276-038 | 796 |
| 75491 | 276-1701 | 996 |
| 75492 | 276-1702 | 994 |
| 7805 | 276-1770 | 1.29 |
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| 272-1402 | 0.22 | 394 | 272-1408 | 3.3 | 456 |
| 272-1403 | 0.33 | 394 | 272-1409 | 4.7 | 496 |
| 272-1404 | 0.47 | 394 | 272-1410 | 6.8 | 496 |
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(Continued from page 116)
Dumont oscilloscope, type 767. Service manual and horizontal time-base plug in. Larry Watson, Rt. 1. Box 180, Gadsden, AL 35901

Telrad Frequency Standard Model 18A. Manufactured by Fred E. Garner Co., Chicago. Circuitry needed. A.H. Ellis. 19 McClure Avenue. Brentford, Ontario, Can. N3R $4 L 7$.

RCA TV model 8-PT-7030. Schematic and service data. P.J. Hughes. 1337 Weber Dr., Cleanwater, FL 33516.

Standard Kolisman uht converter model B, senal \#59277. Schematic Lynnie Gregory, Rt 11, Box 211, Statesville, NC 28677.

Hickok model 610A TV.FM alignment generator. Tracy LaVere. Box 589 Garden Grove, CA 92642.

Hallicrafters model S-53A shortwave receiver. Schematic. R.C. Hartman. 2208 Bayberry St., Virginia Beach, VA 23451

Conar Model 220 tube tester Tube chart needed. Bob Swanberg. Box 96. Chelsea, Mi 48118.

Tektronix model 511-AD oscilloscope. Manual and schematic. J.O Dickinson. 1408 Monmouth Court West, Richmond, VA 23233.

Technics model RS- 1030 US reel-to-reel tape deck. Service manual or schematic. Wallace Bonham. Box 94 Cape Girardeau, MO 63701.

Ross Electronics Corp cassette tape recorder, model \#8282, serial \#11404. Parts and schematic diagrams. A.T Most, 4 Lanningan Drive, Lawrenceville. NJ 08648

Webster WT-2 walkie-talkie. Need PTT switch or source of supply. Lamar Schwalke, P.O. Box 799. Florissant, MO 63033.

Hallicrafters Model S-36A. Operation manual or any available information. Chris Colson, 60 Deemaven Place. Pleasant Hill, CA 94523

Westinghouse model WR 226 3-band receiver. Schematic or senvice manual. Arthur Crough. RD 2, Bloody Pond Road, Lake George. NY 12845

Hallicrafters model S-20-R "Sky Champion" receiver. All available information. J Creasy, 824 Perry Highway, Apt. 15. Pittsburgh. PA 15229.

Roberts model 770X tape recorder. Service manual andior schematic. Robent Fisher. 6536-141h St., Rio Linda. CA 95673.

Bell \& Howell (DeVries Institute) Model 34 oscilloscope Schemalic and transformer voltage outputs or cross reference for replacement. Glenn T. Ozalan, 2220 NW Aspen Portland. OR 97210

Western Electric 25C amplifier Specifications, schematic and parts tist W.Q. Cochran, 910 Townshipline, Chalfont, PA 18914.

Burroughs model C3100 electronic calculator. Schematic and repair information needed C. Kline, 16 Old Farm Road Barrington, IL 60010

Allied Radio, Knight Star Roamer. Schematic and alignment procedure. M.F. Mattern. 1111 Warburton, Santa Clara, CA 95050.

Meisner phono disk recorder model \#9-1065. The Langevin Company phono disk recorder, type 14A. Owners/ operation manual and schematic for both units. D. Testa, Box 9064, Newark, NJ 07104.

Graymark model 506B two-band receiver. Schematic and manual. Henry M Cantor. 21 Friendly Court, Babylon. NY 11702

Hazeltine Neutrodyne receiver. Schematic or any available information. LeRoy Sampson. RR1, Radcliffe, IA 50230.

Superior instruments tube tester model TV-11. Operating instructions. James Lewis, Box 6211. Tulsa. OK 74106.

Symphonic model TPS-5050, mini-TV. Need source of 3inch picture tubes. Nicolas Dominquez, Ave. Pte., Tecamachaico No 16. Mexico City. Mexico ZPIO.

Sears Model 9169 television receiver. Technical manual needed. W.A. Greenwood. 48 Third St, Leominster. MA 01453
troubleshooting information, parts list. Walter Mesko, Box 210. Lake George, NY 12845.

Hallicrafters model S-38B shortwave receiver. Schematic and any available information. Ray Genest, 3175 Boul. Neilson. App. 3 Ste-Foy, Quebec 10, Can. G1W 2V7.

Morse color TV Model 7000 . Service manual and parts list. J.C. Morall, 1000 Cottage Pl., Baldwin, NY 11510.

Webster-Chicago Model 80 wire recorder. Schematics and parts list. Owner's manual Mike Carey. Box 361. Highway 20. Madison, AL 35758.

Philco radio model 37-116. Graybar radio model 320, need chassis for both. Philco radio model \#95, need cabinet. John Yeprad, Box 1457, Studio City, CA 91504.

Benson-Lehner Corp. digitizer "Pigmi" model Need guide books and schmatics W.E. Kelsey, Otter Pond Rd. New London, NH 03257.

Zenith Radio, model 770, chassis \#10022C4. Schematic and parts list. Ervin Thorson, RR5, Box 84 B, Martinsville, IN 45151.

Edin Co., Inc. Electrocardiograph model 8023 serial \#4526. Schematic or any source of information. Russell H. Miller. Jr. 3609 A N. Front St., Harrisburg, PA 17110.

Fatterson model PR-15 communications receiver. Schematics and service manual. E.W. Clede, 6811 Spring Forest. San Antonio. TX 78249.

Grundig TK 46 tape recorder. Owner's manual and schematic. A. Haddad, 125 Jameson Ave., \#508, Toronto, On tario, M6K $2 \times 3$ Can.

National HRO-60 General Coverage Communications Receiver. Operation manual R. Dennis Gibbs. 9214 Venetian Way. Richmond, VA 23229.

RCA regulated power supply model WP-33P. Schematic and service information. Roger A. Leone, 136 Della Circle. Valle Jo, CA 94590

Satellite S20B. Schematic and operation manual. Michael Unger, 183 S. Detroit St., Los Angeles. CA 90036.

Calbest Electronics. slereoplex model 6040 stereo receiv er. Schematic and owner's manual. R.S. Dabe. 13400 Els worth St. \#39. Edgemont, CA 92508.

US Army R-5 receiver SN 20. Any information George W Anderson. 5317 Valonia SI. Fair Oaks. CA 95628.

ACA model 6033 superheterodyne AM and shortwave receiver. Schematic or any available information. James Bias, Rt. 3. Box 175. Ridgeland. SC 29936.

Precision apparatus model \#612 tube tester. Up-to-date roll. Ron Stanford, 8428 San Antonio Ave., South Gate. CA 30280.

Allied Radio Knight kit capacitor checker, kit \#680. Superior Instrument Company mullimeter model \#670-A. Jackson Dynamic tube tester model 715. Manuals and schemat"cs. W. L. Simpson. 370 Beagle Lane. Redding. CA 96001.

Hickok model 610A Universal Television FM alignment signal generator. Operation manual and schematic. Thomas E Phillips. 404 E. Main St., West Newton, PA 15089.

RCA test oscillator WR-67A. Manual, schematic. W. Beale. Rt 1, Box 262A. Mechanicsville, MD 20659

Omnitec model 701 acoustic coupler Schematic. Ivan Berger. 215 W. 78 th St.. New York. NY 10024.

Precision Apparatus signal generator model E-200-C. Operating manual and schematics. Dan Williams, RD1, Stuyvesant, NY 12173.

Mitchell Industries, vhf navigator aircraft radio. Manuals, schematic. Dick Mayrand, 7 Maplewood, Rochester. NH 03867.

Weston vhf sweep generator model 984 . Operating manual and schematic. Ron Patton, Country Ests Pk-9B, Pratt, KS 67124.

Webcor 2712-1 tape recorder. Schematic and service data Alfred E Jordan. 897 E Vine St., Salt Lake City, UT 84107.

Cathode-Ray oscillograph. Parts list and manual. Arthur Thompson, 26441204 SE. Kent. WA 98031

Triadex Muse electronic music computer. Schematic and service manual. Robert Stek, 19 Maytield Rd., Regina, Saskatchewan, Can.


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Wurlitzer Juke Box mode 1700 F. Service manual. Clift Nadiger, 8073 Wonderlund Blvd., Redding, CA 96001.

Omnitec Model 800. portable terminal. Schematic. Stephen Karon. 601 Light St., Baltimore, MD 21230.

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# mL_ECHMRONTICS WTOIRTEID News Highlights in Brief 

## Creating Signature Profiles

The use of a person's signature as a means of identification may be furthered by the development at Sandia Laboratories of a special pen and tablet which sense three supposedly unique characteristics of a person's signature. In the pen are two piezoelectric himorph bender elements, bonded along adjacent quadrants of a slightly flexible aluminum shaft. As a signature is written. one of these transducers generates a horizontal acceleration signal while the other detects vertical acceleration. A third set of signals is generated by pressuresensitive transducers in the writing tablet. The three out put signals, which have peak amplitudes on the order of one volt on each axis and require no biasing current or external power supply, can be processed by a computer and compared with a file record retrieved from the computer's memory. The signature verification system is expected to be useful in banks. department stores. and other institutions where personnel identification is required.

## Ham Radio and Storm Spotting

The National Weather Service, which already has more than 200 weather stations and 51 storm detection radars linked by an extensive radio network, has also turned to Amateur Radio Service operators to fill some of the gaps in the reporting nets. Hams are already functioning efficiently as severe storm spotters and communicators for the NWS in the Tulsa. Ohla.. and Dallas-Fort Worth and Waco. Texas areas. Cooperation with hams in other tornado-prone areas of the U.S. is heing developed. Interested amateurs in a particular area are contacted by the local Weather Service official and the spotting and communications needs are explained. Frequencies to be used (usually in the 2-meter band) are determined. and the system activation procedure is established. Persons interested in cooperating should request copies of the publication "Amateur Radio and the National Weather Service" to: Headquarters. Southern Region, National Weather Service, Rm. IOE09. Federal Office Bldg. . Fort Worth. TX 76102.

## FCC CB Slide Show

The Federal Communications Commission has produced a 10 -minute cartoon slide-and-sound show about Citizens Band rules. The title is " $10-4$ Uncle Charlie." and it was developed to explain, in an entertaining format. the importance of the rules to the thousands of new operators who go on the air every month. The program includes 72 slides, a 10 -minute audio tape cassette, a script and a question-and-answer sheet. Clubs, schools. and others interested in CB can purchase the show package for $\$ 15$ (checks payable to the National Archives Trust Fund) from: National Audiovisual, Center General Services Administration. Order Section. Washington. DC 20409.

## Record Awards for Good Sound

Record awards are based primarily on musical con-tent-only secondarily on sound. The new Audio Excellence Record Awards. sponsered by Audio-Technica, are based primarily on sound. Winner in the rock/pop division was Stevie Wonder"s "Songs in the Key of Life" (Tamla 13-340C2). In the classical division, the winner was "Caruso-A Legendary Performer" (RCA CRM1-1749). What garnered this sound-oriented award for a record based on masters more than 50 years old was a computer process that eliminated much of the original discs' surface noise and the resonances of the recording horns, giving Caruso's voice a more modern recorded quality. Among the factors for which the records were cited were: cleanliness of sound, instrumental and vocal balance, emphasis. dynamic range. frequency response. low noise. and stereo separation as well as dispersion.

## Electronic Blackboard Transmits Data

The Bell System has developed a new blackboard. which. with the proper electronic equipment, can be used to transmit handwriting over phone lines for display on video monitors at distant locations. Used with a portable conference telephone equipped with microphones and a loudspeaker, the electronic blackboard can bring an instructor's entire presentation across the campus. The board consists of a rigid back layer and a front layer that is a thin sheet of black Mylar stretched tightly over the frame. The inner surfaces of both layers are electrically conductive and act as $\mathrm{X}-\mathrm{Y}$ lines. Pressing the board with a piece of chalk registers a point at the intersection of the two axes. When the chalk is moved, a memory unit retains the dots so they blend logether to form a line on the remote monitor. To erase. a device, normally resting on the control unit, is lifted so that the dot-by-dot process is reversed and the image is erased wherever pressure is applied.

## Electronics in the Postal Service

If the U.S. Postal Service is to serve the public adequately and be competitive with other communications systems in the country, it is going to have to start using more electronic techniques. Such was the advice of a recent study by a Commission on Postal Service in its report to the Congress and the President. For example. the commission noted that it is now technologically possible, though not yet economically feasible, to transmit more than 50 percent of the entire volume of mail by electronic means. This is partially due to the fact that about 80 percent of first-class mail today is business-related-invoices, bills, payments, etc. The Electronic Funds Transfer systems are expected to handle 1.9 billion pieces of mail by 1980 and 6.56 billion by 1985 according to a study made by Arthur D. Little. Inc. Similarly, the Treasury Department is expanding its system of direct deposit of payments in many categories.



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